



A/D Flash MCU with EEPROM

HT66F0195

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Features

CPU Features

- Operating Voltage
 - ♦ $f_{SYS}=8\text{MHz}$: 2.2V~5.5V
 - ♦ $f_{SYS}=12\text{MHz}$: 2.7V~5.5V
 - ♦ $f_{SYS}=16\text{MHz}$: 3.3V~5.5V
- Up to 0.25 μs instruction cycle with 16MHz system clock at $V_{DD}=5\text{V}$
- Power down and wake-up functions to reduce power consumption
- Oscillators
 - ♦ External High Speed Crystal – HXT
 - ♦ Internal High Speed RC – HIRC
 - ♦ External Low Speed 32.768kHz Crystal – LXT
 - ♦ Internal 32kHz RC – LIRC
- Multi-mode operation: NORMAL, SLOW, IDLE and SLEEP
- Fully integrated oscillators require no external components
- All instructions executed in one or two instruction cycles
- Table read instructions
- 61 powerful instructions
- 8-level subroutine nesting
- Bit manipulation instruction

Peripheral Features

- Flash Program Memory: 8K \times 16
- RAM Data Memory: 512 \times 8
- True EEPROM Memory: 128 \times 8
- Watchdog Timer function
- 26 bidirectional I/O lines
- Two external interrupt lines shared with I/O pins
- Programmable I/O port source current for LED applications
- Multiple Timer Modules for time measure, input capture, compare match output, PWM output function or single pulse output function
- Serial Interfaces Module – SIM for SPI or I²C
- Fully-duplex Universal Asynchronous Receiver and Transmitter Interface – UART
- Software controlled 6-SCOM/SSEG and 18-SSEG lines LCD driver with 1/3 bias
- 12 external channels 12-bit resolution A/D converter
- One comparator function
- Dual Time-Base function for generation of fixed time interrupt signals
- Low voltage reset function
- Low voltage detect function
- Flash program memory can be re-programmed up to 10,000 times
- Flash program memory data retention > 10 years
- True EEPROM data memory can be re-programmed up to 100,000 times
- True EEPROM data memory data retention > 10 years
- Package types: 24-pin SSOP, 28-pin SSOP

General Description

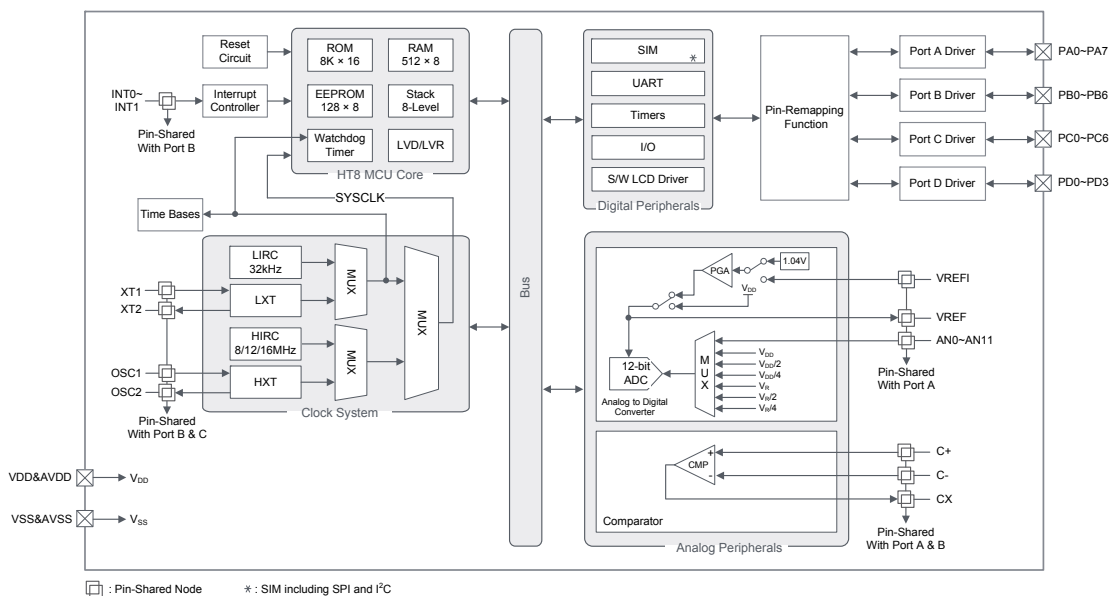
The device is a Flash Memory type 8-bit high performance RISC architecture microcontroller. Offering users the convenience of Flash Memory multi-programming features, the device also includes a wide range of functions and features. Other memory includes an area of RAM Data Memory as well as an area of true EEPROM memory for storage of non-volatile data such as serial numbers, calibration data etc.

Analog features include a multi-channel 12-bit A/D converter and a comparator function. Multiple and extremely flexible Timer Modules provide timing, pulse generation and PWM generation functions. Communication with the outside world is catered for by including fully integrated SPI, I²C, and UART interface functions, three popular interfaces which provide designers with a means of easy communication with external peripheral hardware. Protective features such as an internal Watchdog Timer, Low Voltage Reset and Low Voltage Detector coupled with excellent noise immunity and ESD protection ensure that reliable operation is maintained in hostile electrical environments.

A full choice of external and internal low and high oscillator functions are provided including fully integrated system oscillators which require no external components for their implementation. The ability to operate and switch dynamically between a range of operating modes using different clock sources gives users the ability to optimise microcontroller operation and minimise power consumption.

The inclusion of flexible I/O programming features, Time-Base functions along with many other features ensure that the device will find excellent use in applications such as electronic metering, environmental monitoring, handheld instruments, household appliances, electronically controlled tools, motor driving in addition to many others.

Block Diagram



Pin Assignment

| | | | | | |
|-----------------------------------|------|--|----|--|-------------------------|
| VSS&AVSS | □ 1 | | 24 | | VDD&AVSS |
| PC0/SSEG19/OSC1 | □ 2 | | 23 | | PB0/INT0/SSEG18/AN0/XT1 |
| PC1/SSEG20/OSC2 | □ 3 | | 22 | | PB1/INT1/SSEG17/AN1/XT2 |
| PC2/[SDO]/SSEG0/SCOM0 | □ 4 | | 21 | | PB2/STCK/SSEG16/AN2 |
| PA0/STP/ICPDA/OCSDA | □ 5 | | 20 | | PA4/PTCK/SSEG15/AN3 |
| PC4/SDI/SDA/SSEG22 | □ 6 | | 19 | | PD2/TX/SSEG13 |
| PC5/SCK/SCL/SSEG1/SCOM1 | □ 7 | | 18 | | PD1/RX/SSEG12 |
| PA1/[SDO]/SSEG2/SCOM2 | □ 8 | | 17 | | PA5/SSEG10/AN4/VREF1 |
| PA2/ICPCK/OCDSCK | □ 9 | | 16 | | PA6/CTCK/SSEG9/AN5/VREF |
| PA3/[SDI/SDA]/CX/SSEG3/SCOM3/AN11 | □ 10 | | 15 | | PA7/PTP/SSEG8/AN6 |
| PB6/[SCK/SCL]/C+/SSEG4/SCOM4/AN10 | □ 11 | | 14 | | PB3/[TX]/CTP/SSEG7/AN7 |
| PB5/[SCS]/C-/SSEG5/SCOM5/AN9 | □ 12 | | 13 | | PB4/[RX]/CLO/SSEG6/AN8 |

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| | | | | | |
|-----------------------------------|------|--|----|--|-------------------------|
| VSS&AVSS | □ 1 | | 28 | | VDD&AVDD |
| PC0/SSEG19/OSC1 | □ 2 | | 27 | | PB0/INT0/SSEG18/AN0/XT1 |
| PC1/SSEG20/OSC2 | □ 3 | | 26 | | PB1/INT1/SSEG17/AN1/XT2 |
| PC2/[SDO]/SSEG0/SCOM0 | □ 4 | | 25 | | PB2/STCK/SSEG16/AN2 |
| PA0/STP/ICPDA/OCSDA | □ 5 | | 24 | | PA4/PTCK/SSEG15/AN3 |
| PC3/SDO/SSEG21 | □ 6 | | 23 | | PD3/SSEG14 |
| PC4/SDI/SDA/SSEG22 | □ 7 | | 22 | | PD2/TX/SSEG13 |
| PC5/SCK/SCL/SSEG1/SCOM1 | □ 8 | | 21 | | PD1/RX/SSEG12 |
| PC6/[SCS]/SSEG23 | □ 9 | | 20 | | PD0/SSEG11 |
| PA1/[SDO]/SSEG2/SCOM2 | □ 10 | | 19 | | PA5/SSEG10/AN4/VREF1 |
| PA2/ICPCK/OCDSCK | □ 11 | | 18 | | PA6/CTCK/SSEG9/AN5/VREF |
| PA3/[SDI/SDA]/CX/SSEG3/SCOM3/AN11 | □ 12 | | 17 | | PA7/PTP/SSEG8/AN6 |
| PB6/[SCK/SCL]/C+/SSEG4/SCOM4/AN10 | □ 13 | | 16 | | PB3/[TX]/CTP/SSEG7/AN7 |
| PB5/[SCS]/C-/SSEG5/SCOM5/AN9 | □ 14 | | 15 | | PB4/[RX]/CLO/SSEG6/AN8 |

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- Note: 1. If the pin-shared pin functions have multiple outputs simultaneously, pin names at the right side of the “/” sign can be used for higher priority.
2. VDD&AVDD means the VDD and AVDD are double bonding; VSS&AVSS means the VSS and AVSS are double bonding.
3. The OCSDA and OCDSCK pins are used as the OCDS dedicated pins and only available for the HT66V0195 device which is the OCDS EV chip of the HT66F0195.
4. For the less pin count package type there will be unbounded pins which should be properly configured to avoid unwanted power consumption resulting from floating input conditions. Refer to the “Standby Current Considerations” and “Input/Output Ports” sections.

Pin Description

The function of each pin is listed in the following table, however the details behind how each pin is configured is contained in other sections of the datasheet. As the Pin Description table shows the situation for the package with the most pins, not all pins in the tables will be available on smaller package sizes.

| Pin Name | Function | OPT | I/T | O/T | Description |
|--|----------|------------------|-----|------|--|
| PA0/STP/ICPDA/OCDSDA | PA0 | PAWU PAPU | ST | CMOS | General purpose I/O. Register enabled pull-up and wake-up. |
| | STP | TMPC | ST | CMOS | STM input/output |
| | ICPDA | — | ST | CMOS | ICP Data/Address pin |
| | OCDSDA | — | ST | CMOS | OCDS Data/Address pin, for EV chip only. |
| PA1/[SDO]/SSEG2/SCOM2 | PA1 | PAWU PAPU | ST | CMOS | General purpose I/O. Register enabled pull-up and wake-up. |
| | [SDO] | SIMC0 IFS | — | CMOS | SPI data output |
| | SSEG2 | SLCDC0 SLCDC1 | — | SSEG | Software controlled LCD segment output |
| | SCOM2 | SLCDC0 SLCDC1 | — | SCOM | Software controlled LCD common output |
| PA2/ICPCK/OCDSCK | PA2 | PAWU PAPU | ST | CMOS | General purpose I/O. Register enabled pull-up and wake-up. |
| | ICPCK | — | ST | CMOS | ICP Clock pin |
| | OCDSCK | — | ST | — | OCDS Clock pin, for EV chip only. |
| PA3/[SDI]/SDA]/CX/SSEG3/ SCOM3/AN11 | PA3 | PAWU PAPU | ST | CMOS | General purpose I/O. Register enabled pull-up and wake-up. |
| | [SDI] | SIMC0 IFS | ST | — | SPI data input |
| | [SDA] | SIMC0 IFS | ST | NMOS | I ² C address/data line |
| | CX | CPC | — | CMOS | Comparator output |
| | SSEG3 | SLCDC0 SLCDC1 | — | SSEG | Software controlled LCD segment output |
| | SCOM3 | SLCDC0 SLCDC1 | — | SCOM | Software controlled LCD common output |
| | AN11 | ACERH | AN | — | A/D Converter analog input |
| PA4/PTCK/SSEG15/AN3 | PA4 | PAWU PAPU | ST | CMOS | General purpose I/O. Register enabled pull-up and wake-up. |
| | PTCK | PTMC0 | ST | — | PTM clock input |
| | SSEG15 | SLCDC3 | — | SSEG | Software controlled LCD segment output |
| | AN3 | ACERL | AN | — | A/D Converter analog input |
| PA5/SSEG10/AN4/VREFI | PA5 | PAWU PAPU | ST | CMOS | General purpose I/O. Register enabled pull-up and wake-up. |
| | SSEG10 | SLCDC2 | — | SSEG | Software controlled LCD segment output |
| | AN4 | ACERL | AN | — | A/D Converter analog input |
| | VREFI | SADC2 | AN | — | A/D Converter PGA voltage input |
| PA6/CTCK/SSEG9/AN5/ VREF | PA6 | PAWU PAPU | ST | CMOS | General purpose I/O. Register enabled pull-up and wake-up. |
| | CTCK | CTMC0 | ST | — | CTM clock input |
| | SSEG9 | SLCDC2 | — | SSEG | Software controlled LCD segment output |
| | AN5 | ACERL | AN | — | A/D Converter analog input |
| | VREF | SADC2 | — | AN | A/D Converter reference voltage output |

| Pin Name | Function | OPT | I/T | O/T | Description |
|----------------------------------|-------------------------|-----------------------|-----|------|--|
| PA7/PTP/SSEG8/AN6 | PA7 | PAWU PAPU | ST | CMOS | General purpose I/O. Register enabled pull-up and wake-up. |
| | PTP | TMPC | ST | CMOS | PTM input/output |
| | SSEG8 | SLCDC2 | — | SSEG | Software controlled LCD segment output |
| | AN6 | ACERL | AN | — | A/D Converter analog input |
| PB0/INT0/SSEG18/AN0/XT1 | PB0 | PBPU | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | INT0 | INTEG INTC0 | ST | — | External Interrupt 0 |
| | SSEG18 | SLCDC3 | — | SSEG | Software controlled LCD segment output |
| | AN0 | ACERL | AN | — | A/D Converter analog input |
| | XT1 | CO | LXT | — | LXT oscillator pin |
| PB1/INT1/SSEG17/AN1/XT2 | PB1 | PBPU | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | INT1 | INTEG INTC2 | ST | — | External Interrupt 1 |
| | SSEG17 | SLCDC3 | — | SSEG | Software controlled LCD segment output |
| | AN1 | ACERL | AN | — | A/D Converter analog input |
| | XT2 | CO | — | LXT | LXT oscillator pin |
| PB2/STCK/SSEG16/AN2 | PB2 | PBPU | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | STCK | STMC0 | ST | — | STM clock input |
| | SSEG16 | SLCDC3 | — | SSEG | Software controlled LCD segment output |
| | AN2 | ACERL | AN | — | A/D Converter analog input |
| PB3/[TX]/CTP/SSEG7/AN7 | PB3 | PBPU | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | [TX] | UCR1 UCR2 IFS | — | CMOS | UART TX serial data output |
| | CTP | TMPC | — | CMOS | CTM output |
| | SSEG7 | SLCDC2 | — | SSEG | Software controlled LCD segment output |
| | AN7 | ACERL | AN | — | A/D Converter analog input |
| PB4/[RX]/CLO/SSEG6/AN8 | PB4 | PBPU | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | [RX] | UCR1 UCR2 IFS | ST | — | UART RX serial data input |
| | CLO | TMPC | — | CMOS | System clock output |
| | SSEG6 | SLCDC2 | — | SSEG | Software controlled LCD segment output |
| | AN8 | ACERH | AN | — | A/D Converter analog input |
| PB5/[SCS]/C-/SSEG5/ SCOM5/AN9 | PB5 | PBPU | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | $\overline{\text{SCS}}$ | SIMC0 SIMC2 IFS | ST | CMOS | SPI slave select |
| | C- | CPC | AN | — | Comparator negative input |
| | SSEG5 | SLCDC1 | — | SSEG | Software controlled LCD segment output |
| | SCOM5 | SLCDC1 | — | SCOM | Software controlled LCD common output |
| | AN9 | ACERH | AN | — | A/D Converter analog input |

| Pin Name | Function | OPT | I/T | O/T | Description |
|-----------------------------------|------------------|-----------------|-----|------|--|
| PB6/[SCK/SCL]/C+/SSEG4/SCOM4/AN10 | PB6 | PBPU | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | [SCK] | SIMC0 IFS | ST | CMOS | SPI serial clock |
| | [SCL] | SIMC0 IFS | ST | NMOS | I ² C clock line |
| | C+ | CPC | AN | — | Comparator positive input |
| | SSEG4 | SLCDC1 | — | SSEG | Software controlled LCD segment output |
| | SCOM4 | SLCDC1 | — | SCOM | Software controlled LCD common output |
| | AN10 | ACERH | AN | — | A/D Converter analog input |
| PC0/SSEG19/OSC1 | PC0 | PCPU | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | SSEG19 | SLCDC3 | — | SSEG | Software controlled LCD segment output |
| | OSC1 | CO | HXT | — | HXT oscillator pin |
| PC1/SSEG20/OSC2 | PC1 | PCPU | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | SSEG20 | SLCDC3 | — | SSEG | Software controlled LCD segment output |
| | OSC2 | CO | — | HXT | HXT oscillator pin |
| PC2/[SDO]/SSEG0/SCOM0 | PC2 | PCPU | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | [SDO] | SIMC0 IFS | — | CMOS | SPI data output |
| | SSEG0 | SLCDC0 SLCDC1 | — | SSEG | Software controlled LCD segment output |
| | SCOM0 | SLCDC0 SLCDC1 | — | SCOM | Software controlled LCD common output |
| PC3/SDO/SSEG21 | PC3 | PCPU | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | SDO | SIMC0 IFS | — | CMOS | SPI data output |
| | SSEG21 | SLCDC3 | — | SSEG | Software controlled LCD segment output |
| PC4/SDI/SDA/SSEG22 | PC4 | PCPU | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | SDI | SIMC0 IFS | ST | — | SPI data input |
| | SDA | SIMC0 IFS | ST | NMOS | I ² C data line |
| | SSEG22 | SLCDC4 | — | SSEG | Software controlled LCD segment output |
| PC5/SCK/SCL/SSEG1/SCOM1 | PC5 | PCPU | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | SCK | SIMC0 IFS | ST | CMOS | SPI serial clock |
| | SCL | SIMC0 IFS | ST | NMOS | I ² C clock line |
| | SSEG1 | SLCDC0 SLCDC1 | — | SSEG | Software controlled LCD segment output |
| | SCOM1 | SLCDC0 SLCDC1 | — | SCOM | Software controlled LCD common output |
| PC6/ \overline{SCS} /SSEG23 | PC6 | PCPU | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | \overline{SCS} | SIMC0 SIMC2 IFS | ST | CMOS | SPI slave select |
| | SSEG23 | SLCDC4 | — | SSEG | Software controlled LCD segment output |
| PD0/SSEG11 | PD0 | PDPUP | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | SSEG11 | SLCDC2 | — | SSEG | Software controlled LCD segment output |

| Pin Name | Function | OPT | I/T | O/T | Description |
|---------------|----------|---------------------|-----|------|--|
| PD1/RX/SSEG12 | PD1 | PDPUP | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | RX | UCR1 UCR2 IFS | ST | — | UART RX serial data input |
| | SSEG12 | SLCDC2 | — | SSEG | Software controlled LCD segment output |
| PD2/TX/SSEG13 | PD2 | PDPUP | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | TX | UCR1 UCR2 IFS | — | CMOS | UART TX serial data output |
| | SSEG13 | SLCDC2 | — | SSEG | Software controlled LCD segment output |
| PD3/SSEG14 | PD3 | PDPUP | ST | CMOS | General purpose I/O. Register enabled pull-up. |
| | SSEG14 | SLCDC3 | — | SSEG | Software controlled LCD segment output |
| VDD&AVDD * | VDD | — | PWR | — | Positive power supply |
| | AVDD | — | PWR | — | Analog positive power supply |
| VSS&AVSS * | VSS | — | PWR | — | Negative power supply, ground. |
| | AVSS | — | PWR | — | Analog negative power supply, ground. |

Legend: I/T: Input type;

O/T: Output type;

OPT: Optional by configuration option (CO) or register option;

CO: Configuration option;

PWR: Power;

AN: Analog signal;

ST: Schmitt Trigger input;

CMOS: CMOS output;

NMOS; NMOS output;

SSEG: Software controlled LCD SEG;

SCOM: Software controlled LCD COM;

HXT: High frequency crystal oscillator;

LXT: Low frequency crystal oscillator

*: The AVDD pin is internal bonded together with the VDD pin while the AVSS pin is internally bonded together with the VSS pin.

Absolute Maximum Ratings

| | |
|-------------------------------|----------------------------------|
| Supply Voltage | $V_{SS}-0.3V$ to $V_{SS}+6.0V$ |
| Input Voltage | $V_{SS}-0.3V$ to $V_{DD}+0.3V$ |
| Storage Temperature..... | $-50^{\circ}C$ to $125^{\circ}C$ |
| Operating Temperature..... | $-40^{\circ}C$ to $85^{\circ}C$ |
| I_{OH} Total | -80mA |
| I_{OL} Total | 80mA |
| Total Power Dissipation | 500mW |

Note: These are stress ratings only. Stresses exceeding the range specified under “Absolute Maximum Ratings” may cause substantial damage to the device. Functional operation of the devices at other conditions beyond those listed in the specification is not implied and prolonged exposure to extreme conditions may affect device reliability.

D.C. Characteristics

Ta=-40°C~85°C

| Symbol | Parameter | Test Conditions | | Min. | Typ. | Max. | Unit | |
|------------------------------------|----------------------------------|-------------------------------|--|--|------|--------------------|------|----|
| | | V _{DD} | Conditions | | | | | |
| V _{DD} | Operating Voltage (HXT) | — | f _{sys} =f _{HXT} =8MHz | 2.2 | — | 5.5 | V | |
| | | | f _{sys} =f _{HXT} =12MHz | 2.7 | — | 5.5 | | |
| | | | f _{sys} =f _{HXT} =16MHz | 3.3 | — | 5.5 | | |
| | Operating Voltage (HIRC) | — | f _{sys} =f _{HIRC} =8MHz | 2.2 | — | 5.5 | V | |
| | | | f _{sys} =f _{HIRC} =12MHz | 2.7 | — | 5.5 | | |
| | | | f _{sys} =f _{HIRC} =16MHz | 3.3 | — | 5.5 | | |
| I _{DD} | Operating Current (HXT) | 3V | f _{sys} =f _H =f _{HXT} =8MHz, No load, all peripherals off | — | 1.0 | 1.5 | mA | |
| | | | 5V | — | 2.0 | 3.0 | | |
| | | 3V | f _{sys} =f _H =f _{HXT} =12MHz, No load, all peripherals off | — | 1.5 | 2.75 | mA | |
| | | | 5V | — | 3.0 | 4.5 | | |
| | | 5V | f _{sys} =f _H =f _{HXT} =16MHz, no load, all peripherals off | — | 4.5 | 7.0 | mA | |
| | | | | | | | | |
| | Operating Current (HIRC) | 3V | f _{sys} =f _H =f _{HIRC} =8MHz, No load, all peripherals off | — | 1.0 | 1.5 | mA | |
| | | | 5V | — | 2.0 | 3.0 | | |
| | | 3V | f _{sys} =f _H =f _{HIRC} =12MHz, No load, all peripherals off | — | 1.5 | 2.75 | mA | |
| | | | 5V | — | 3.0 | 4.5 | | |
| | | 5V | f _{sys} =f _H =f _{HIRC} =16MHz, No load, all peripherals off | — | 4.5 | 7.0 | mA | |
| | | | | | | | | |
| | Operating Current (LXT) | 3V | f _{sys} =f _{SUB} =f _{LXT} =32.768kHz, No load, all peripherals off | — | 10 | 20 | μA | |
| | | | 5V | — | 30 | 50 | | |
| | Operating Current (LIRC) | 3V | f _{sys} =f _{SUB} =f _{LIRC} =32kHz, No load, all peripherals off | — | 10 | 20 | μA | |
| | | | 5V | — | 30 | 50 | | |
| | I _{STB} | Standby Current (SLEEP0 Mode) | 3V | No load, all peripherals off, WDT off | — | — | 1 | μA |
| | | | | 5V | — | — | 2 | |
| Standby Current (SLEEP1 Mode) | | 3V | No load, all peripherals off, WDT on | — | — | 3 | μA | |
| | | | 5V | — | — | 5 | | |
| Standby Current (IDLE0 Mode) | | 3V | No load, all peripherals off, f _{SUB} on | — | 3 | 5 | μA | |
| | | | 5V | — | 5 | 10 | | |
| Standby Current (IDLE1 Mode, HXT) | | 3V | f _{sys} =f _{HXT} =8MHz, f _{SUB} on, No load, all peripherals off | — | 0.5 | 1 | mA | |
| | | | 5V | — | 1 | 2 | | |
| | | 3V | f _{sys} =f _{HXT} =12MHz, f _{SUB} on, No load, all peripherals off | — | 0.6 | 1.2 | mA | |
| | | | 5V | — | 1.2 | 2.4 | | |
| | | 5V | f _{sys} =f _{HXT} =16MHz, f _{SUB} on, No load, all peripherals off | — | 2.0 | 4.0 | mA | |
| | | | | | | | | |
| Standby Current (IDLE1 Mode, HIRC) | | 3V | f _{sys} =f _{HIRC} =8MHz, f _{SUB} on, No load, all peripherals off | — | 0.8 | 1.6 | mA | |
| | | | 5V | — | 1.0 | 2.0 | | |
| | | 3V | f _{sys} =f _{HIRC} =12MHz, f _{SUB} on, No load, all peripherals off | — | 1.2 | 2.4 | mA | |
| | | | 5V | — | 1.5 | 3.0 | | |
| | | 5V | f _{sys} =f _{HIRC} =16MHz, f _{SUB} on, No load, all peripherals off | — | 2.0 | 4.0 | mA | |
| | | | | | | | | |
| V _{IL} | Input Low Voltage for I/O Ports | 5V | — | 0 | — | 1.5 | V | |
| | | — | — | 0 | — | 0.2V _{DD} | | |
| V _{IH} | Input High Voltage for I/O Ports | 5V | — | 3.5 | — | 5.0 | V | |
| | | — | — | 0.8V _{DD} | — | V _{DD} | | |

| Symbol | Parameter | Test Conditions | | Min. | Typ. | Max. | Unit |
|-----------------|---|-----------------|---|-------|------|------|------|
| | | V _{DD} | Conditions | | | | |
| I _{OL} | Sink Current for I/O Ports | 3V | V _{OL} =0.1V _{DD} | 16 | 32 | — | mA |
| | | 5V | | 32 | 64 | — | |
| I _{OH} | Source Current for I/O Ports | 3V | V _{OH} =0.9V _{DD} , SLEDCn[m+1, m]=00, n=0 or 1; m=0, 2, 4 or 6 | -1.0 | -2.0 | — | mA |
| | | 5V | | -2.0 | -4.0 | — | |
| | | 3V | V _{OH} =0.9V _{DD} , SLEDCn[m+1, m]=01, n=0 or 1; m=0, 2, 4 or 6 | -1.75 | -3.5 | — | mA |
| | | 5V | | -3.5 | -7.0 | — | |
| | | 3V | V _{OH} =0.9V _{DD} , SLEDCn[m+1, m]=10, n=0 or 1; m=0, 2, 4 or 6 | -2.5 | -5.0 | — | mA |
| | | 5V | | -5.0 | -10 | — | |
| 3V | V _{OH} =0.9V _{DD} , SLEDCn[m+1, m]=11, n=0 or 1; m=0, 2, 4 or 6 | -5.5 | -11 | — | mA | | |
| 5V | | -11 | -22 | — | | | |
| R _{PH} | Pull-high Resistance for I/O Ports | 3V | — | 20 | 60 | 100 | kΩ |
| | | 5V | — | 10 | 30 | 50 | |

A.C. Characteristics

T_a=-40°C~85°C, unless otherwise specify

| Symbol | Parameter | Test Condition | | Min. | Typ. | Max. | Unit |
|--|--|---|--|--------|------|----------------|-------------------|
| | | V _{DD} | Condition | | | | |
| f _{sys} | System Clock (HXT) | 2.2V~5.5V | f _{sys} =f _{HXT} =8MHz | 0.4 | — | 8 | MHz |
| | | 2.7V~5.5V | f _{sys} =f _{HXT} =12MHz | 0.4 | — | 12 | |
| | | 3.3V~5.5V | f _{sys} =f _{HXT} =16MHz | 0.4 | — | 16 | |
| | System Clock (HIRC) | 2.2V~5.5V | f _{sys} =f _{HIRC} =8MHz | — | 8 | — | MHz |
| | | 2.7V~5.5V | f _{sys} =f _{HIRC} =12MHz | — | 12 | — | |
| | | 3.3V~5.5V | f _{sys} =f _{HIRC} =16MHz | — | 16 | — | |
| System Clock (LXT) | 2.2~5.5V | f _{sys} =f _{LXT} =32.768kHz | — | 32.768 | — | kHz | |
| System Clock (LIRC) | 2.2~5.5V | f _{sys} =f _{LIRC} =32kHz | — | 32 | — | kHz | |
| f _{LIRC} | Low Speed Internal RC oscillator (LIRC) | 5V | T _a =25°C | -2% | 32 | +2% | kHz |
| | | 2.2V~5.5V | T _a =-40°C to 85°C | -10% | 32 | +10% | |
| t _{START} | LIRC Start Up Time | — | — | — | — | 100 | μs |
| t _{TCK} | xTCK pin Minimum Input Pulse Width | — | — | 0.3 | — | — | μs |
| t _{INT} | Interrupt Pin Minimum Input Pulse Width | — | — | 10 | — | — | μs |
| t _{SST} | System Start-up Timer Period (Wake-up from Condition where f _{sys} is off) | — | f _{sys} =f _{HXT} ~f _{HXT} /64 | 128 | — | — | t _{HXT} |
| | | — | f _{sys} =f _{HIRC} ~f _{HIRC} /64 | 16 | — | — | t _{HIRC} |
| | | — | f _{sys} =f _{LXT} | 128 | — | — | t _{LXT} |
| | | — | f _{sys} =f _{LIRC} | 2 | — | — | t _{LIRC} |
| | System Start-up Timer Period (Wake-up from Condition where f _{sys} is on) | — | f _{sys} =f _H ~f _H /64, f _H =f _{HXT} or f _{HIRC} | 2 | — | — | t _H |
| | | — | f _{sys} =f _{LXT} or f _{LIRC} | 2 | — | — | t _{SUB} |
| System Start-up Timer Period (WDT Time-out Hardware Cold Reset) | — | — | 0 | — | — | t _H | |
| t _{RSTD} | System Reset Delay Time (Power-on Reset, LVR Hardware Reset, LVR/WDC Software Reset) | — | — | 25 | 50 | 100 | ms |
| | System Reset Delay Time (WDT Time-out Hardware Cold Reset) | — | — | 8.3 | 16.7 | 33.3 | ms |
| t _{EEERD} | EEPROM Read Time | — | — | — | — | 4 | t _{sys} |
| t _{EEWR} | EEPROM Write Time | — | — | — | 4 | 6 | ms |

Note: t_{sys}=1/f_{sys}

HIRC Electrical Characteristics

During the program writing operation the writer will trim the HIRC oscillator at a user selected HIRC frequency and user selected voltage of either 3V or 5V.

| Symbol | Parameter | Test Conditions | | Min. | Typ. | Max. | Unit |
|-------------------|-------------------------------------|-----------------|------------|-------|------|-------|------|
| | | V _{DD} | Temp. | | | | |
| f _{HIRC} | 8MHz Writer Trimmed HIRC Frequency | 3V/5V | 25°C | -1% | 8 | +1% | MHz |
| | | | -40°C~85°C | -2% | 8 | +2% | |
| | | 2.2V~5.5V | 25°C | -2.5% | 8 | +2.5% | |
| | | | -40°C~85°C | -3% | 8 | +3% | |
| | 12MHz Writer Trimmed HIRC Frequency | 3V/5V | 25°C | -1% | 12 | +1% | MHz |
| | | | -40°C~85°C | -2% | 12 | +2% | |
| | | 2.7V~5.5V | 25°C | -2.5% | 12 | +2.5% | |
| | | | -40°C~85°C | -3% | 12 | +3% | |
| | 16MHz Writer Trimmed HIRC Frequency | 5V | 25°C | -1% | 16 | +1% | MHz |
| | | | -40°C~85°C | -2% | 16 | +2% | |
| | | 3.3V~5.5V | 25°C | -2.5% | 16 | +2.5% | |
| | | | -40°C~85°C | -3% | 16 | +3% | |

- Note: 1. The 3V/5V values for V_{DD} are provided as these are the two selectable fixed voltages at which the HIRC frequency is trimmed by the writer.
2. The row below the 3V/5V trim voltage row is provided to show the values for the full V_{DD} range operating voltage. It is recommended that the trim voltage is fixed at 3V for application voltage ranges from 2.2V to 3.6V and fixed at 5V for application voltage ranges from 3.3V to 5.5V.
3. The minimum and maximum tolerance values provided in the table are for the frequency at which the writer trims the HIRC oscillator.

A/D Converter Electrical Characteristics

T_a=-40°C~85°C

| Symbol | Parameter | Test Conditions | | Min. | Typ. | Max. | Unit |
|--------------------|---|-----------------|--|------|------|------------------|-------------------|
| | | V _{DD} | Conditions | | | | |
| V _{DD} | Operating Voltage | — | — | 2.2 | — | 5.5 | V |
| V _{ADI} | Input Voltage | — | — | 0 | — | V _{REF} | V |
| V _{REF} | Reference Voltage | — | — | 2 | — | V _{DD} | V |
| DNL | Differential Non-linearity | 3V | V _{REF} =V _{DD} , t _{ADCK} =0.5μs | -3 | — | +3 | LSB |
| | | | | | | | |
| | | 3V | V _{REF} =V _{DD} , t _{ADCK} =10μs | | | | |
| | | | | | | | |
| INL | Integral Non-linearity | 3V | V _{REF} =V _{DD} , t _{ADCK} =0.5μs | -4 | — | +4 | LSB |
| | | | | | | | |
| | | 3V | V _{REF} =V _{DD} , t _{ADCK} =10μs | | | | |
| | | | | | | | |
| I _{ADC} | Additional Current for A/D Converter Enable | 3V | No load, t _{ADCK} =0.5μs | — | 1.0 | 2.0 | mA |
| | | 5V | | — | 1.5 | 3.0 | |
| t _{ADCK} | Clock Period | — | — | 0.5 | — | 10 | μs |
| t _{ADC} | Conversion Time (Includes A/D Sample and Hold Time) | — | — | — | 16 | — | t _{ADCK} |
| t _{ON2ST} | A/D Converter On-to-Start Time | — | — | 4 | — | — | μs |

LVD/LVR Electrical Characteristics

 $T_a = -40^{\circ}\text{C} \sim 85^{\circ}\text{C}$

| Symbol | Parameter | Test Conditions | | Min. | Typ. | Max. | Unit |
|-------------------|--|-----------------|--|------|------|------|------|
| | | V _{DD} | Conditions | | | | |
| V _{LVR} | Low Voltage Reset Voltage | — | LVR Enable, voltage select 2.1V | -5% | 2.1 | +5% | V |
| | | | LVR Enable, voltage select 2.55V | | 2.55 | | |
| | | | LVR Enable, voltage select 3.15V | | 3.15 | | |
| | | | LVR Enable, voltage select 3.8V | | 3.8 | | |
| V _{LVD} | Low Voltage Detector Voltage | — | LVD Enable, voltage select 2.0V | -5% | 2.0 | +5% | V |
| | | | LVD Enable, voltage select 2.2V | | 2.2 | | |
| | | | LVD Enable, voltage select 2.4V | | 2.4 | | |
| | | | LVD Enable, voltage select 2.7V | | 2.7 | | |
| | | | LVD Enable, voltage select 3.0V | | 3.0 | | |
| | | | LVD Enable, voltage select 3.3V | | 3.3 | | |
| | | | LVD Enable, voltage select 3.6V | | 3.6 | | |
| | | | LVD Enable, voltage select 4.0V | | 4.0 | | |
| I _{OP} | Operating Current | 5V | LVD enable, LVR enable, VBGEN=0 | — | 20 | 25 | μA |
| | | 5V | LVD enable, LVR enable, VBGEN=1 | — | 180 | 200 | |
| t _{LVDS} | LVDO Stable Time | — | For LVR enable, VBGEN=0, LVD off → on | — | — | 15 | μs |
| | | — | For LVR disable, VBGEN=0, LVD off → on | — | — | 150 | |
| t _{LVR} | Minimum Low Voltage Width to Reset | — | — | 120 | 240 | 480 | μs |
| t _{LVD} | Minimum Low Voltage Width to Interrupt | — | — | 60 | 120 | 240 | μs |

Internal Bandgap Reference Voltage Electrical Characteristics

 $T_a = -40^{\circ}\text{C} \sim 85^{\circ}\text{C}$

| Symbol | Parameter | Test Conditions | | Min. | Typ. | Max. | Unit |
|------------------|-------------------------------------|-----------------|------------|------|------|------|------|
| | | V _{DD} | Conditions | | | | |
| V _{BG} | Bandgap Reference Voltage | — | — | -3% | 1.04 | +3% | V |
| t _{BGS} | V _{BG} Turn On Stable Time | — | No load | — | — | 150 | μs |

Note: The V_{BG} voltage is used as the A/D converter PGA input.

Comparator Electrical Characteristics

 $T_a = -40^{\circ}\text{C} \sim 85^{\circ}\text{C}$

| Symbol | Parameter | Test Conditions | | Min. | Typ. | Max. | Unit |
|-------------------|--|-----------------|------------|-----------------|------|----------------------|------|
| | | V _{DD} | Conditions | | | | |
| V _{DD} | Operating Voltage | — | — | 2.2 | — | 5.5 | V |
| I _{COMP} | Additional Current for Comparator Enable | 3V | — | — | 37 | 56 | μA |
| | | 5V | — | — | 130 | 200 | |
| V _{OS} | Input Offset Voltage | 3V | — | -10 | — | 10 | mV |
| | | 5V | — | -10 | — | 10 | |
| V _{CM} | Common Mode Voltage Range | — | — | V _{SS} | — | V _{DD} -1.4 | V |
| A _{OL} | Open Loop Gain | 3V | — | 60 | 80 | — | dB |
| | | 5V | — | 60 | 80 | — | |

| Symbol | Parameter | Test Conditions | | Min. | Typ. | Max. | Unit |
|------------------|---------------|-----------------|----------------------|------|------|------|------|
| | | V _{DD} | Conditions | | | | |
| V _{HYS} | Hysteresis | 3V | CHYEN=0 | 0 | 0 | 5 | mV |
| | | 5V | | 0 | 0 | 5 | |
| | | 3V | CHYEN=1 | 40 | 60 | 80 | mV |
| | | 5V | | 40 | 60 | 80 | |
| t _{RP} | Response Time | 3V | With 100mV overdrive | — | 370 | 560 | ns |
| | | 5V | | — | 370 | 560 | |

Note: All measurement is under the condition where the comparator positive input voltage is equal to (V_{DD}-1.4)/2 and remains constant.

Software Controlled LCD Driver Electrical Characteristics

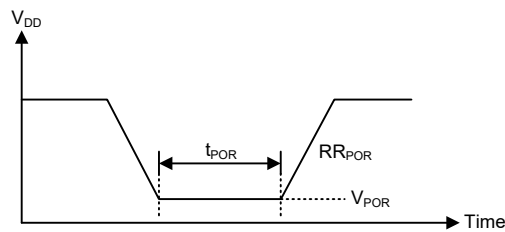
T_a=-40°C~85°C

| Symbol | Parameter | Test Conditions | | Min. | Typ. | Max. | Unit |
|--------------------|---|-----------------|--------------|-------|------|-------|-----------------|
| | | V _{DD} | Conditions | | | | |
| I _{BIAS} | Bias Current | 5V | ISEL[1:0]=00 | 4.2 | 8.3 | 13 | μA |
| | | | ISEL[1:0]=01 | 8.3 | 16.7 | 25 | |
| | | | ISEL[1:0]=10 | 25 | 50 | 75 | |
| | | | ISEL[1:0]=11 | 50 | 100 | 150 | |
| V _{LCD_H} | V _{DD} ×2/3 Voltage for LCD SCOM/SSEG Output | 2.2V~5.5V | No load | 0.645 | 0.67 | 0.695 | V _{DD} |
| V _{LCD_L} | V _{DD} ×1/3 Voltage for LCD SCOM/SSEG Output | 2.2V~5.5V | No load | 0.305 | 0.33 | 0.355 | V _{DD} |

Power-on Reset Characteristics

T_a=25°C

| Symbol | Parameter | Test Conditions | | Min. | Typ. | Max. | Unit |
|-------------------|---|-----------------|------------|-------|------|------|------|
| | | V _{DD} | Conditions | | | | |
| V _{POR} | V _{DD} Start Voltage to Ensure Power-on Reset | — | — | — | — | 100 | mV |
| RR _{POR} | V _{DD} Rising Rate to Ensure Power-on Reset | — | — | 0.035 | — | — | V/ms |
| t _{POR} | Minimum Time for V _{DD} Stays at V _{POR} to Ensure Power-on Reset | — | — | 1 | — | — | ms |

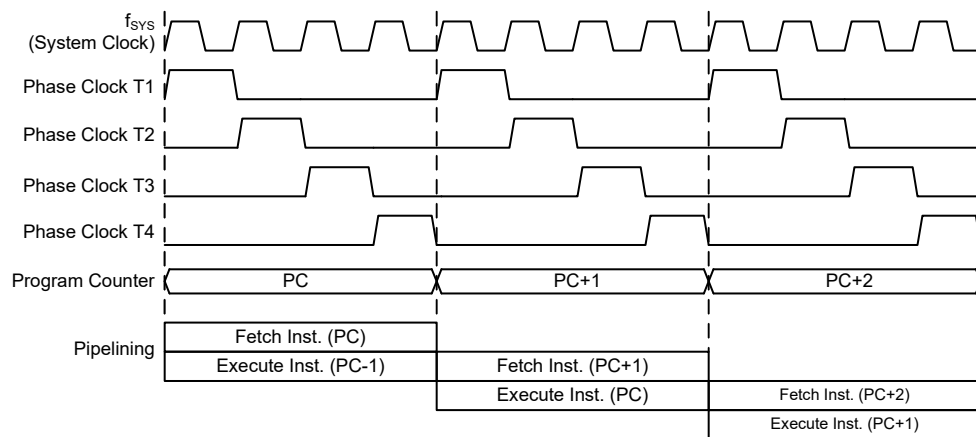


System Architecture

A key factor in the high-performance features of the Holtek range of microcontrollers is attributed to their internal system architecture. The device takes advantage of the usual features found within RISC microcontrollers providing increased speed of operation and enhanced performance. The pipelining scheme is implemented in such a way that instruction fetching and instruction execution are overlapped, hence instructions are effectively executed in one cycle, with the exception of branch or call instructions which need one more cycle. An 8-bit wide ALU is used in practically all instruction set operations, which carries out arithmetic operations, logic operations, rotation, increment, decrement, branch decisions, etc. The internal data path is simplified by moving data through the Accumulator and the ALU. Certain internal registers are implemented in the Data Memory and can be directly or indirectly addressed. The simple addressing methods of these registers along with additional architectural features ensure that a minimum of external components is required to provide a functional I/O and A/D control system with maximum reliability and flexibility. This makes the device suitable for low-cost, high-volume production for controller applications.

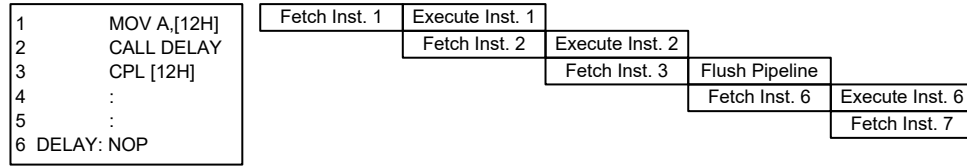
Clocking and Pipelining

The main system clock, derived from either an HXT, LXT, HIRC or LIRC oscillator is subdivided into four internally generated non-overlapping clocks, T1~T4. The Program Counter is incremented at the beginning of the T1 clock during which time a new instruction is fetched. The remaining T2~T4 clocks carry out the decoding and execution functions. In this way, one T1~T4 clock cycle forms one instruction cycle. Although the fetching and execution of instructions takes place in consecutive instruction cycles, the pipelining structure of the microcontroller ensures that instructions are effectively executed in one instruction cycle. The exception to this are instructions where the contents of the Program Counter are changed, such as subroutine calls or jumps, in which case the instruction will take one more instruction cycle to execute.



System Clocking and Pipelining

For instructions involving branches, such as jump or call instructions, two machine cycles are required to complete instruction execution. An extra cycle is required as the program takes one cycle to first obtain the actual jump or call address and then another cycle to actually execute the branch. The requirement for this extra cycle should be taken into account by programmers in timing sensitive applications.



Instruction Fetching

Program Counter

During program execution, the Program Counter is used to keep track of the address of the next instruction to be executed. It is automatically incremented by one each time an instruction is executed except for instructions, such as “JMP” or “CALL” that demands a jump to a non-consecutive Program Memory address. Only the lower 8 bits, known as the Program Counter Low Register, are directly addressable by the application program.

When executing instructions requiring jumps to non-consecutive addresses such as a jump instruction, a subroutine call, interrupt or reset, etc., the microcontroller manages program control by loading the required address into the Program Counter. For conditional skip instructions, once the condition has been met, the next instruction, which has already been fetched during the present instruction execution, is discarded and a dummy cycle takes its place while the correct instruction is obtained.

| Program Counter | |
|---------------------------|--------------|
| Program Counter High Byte | PCL Register |
| PC12~PC8 | PCL7~PCL0 |

Program Counter

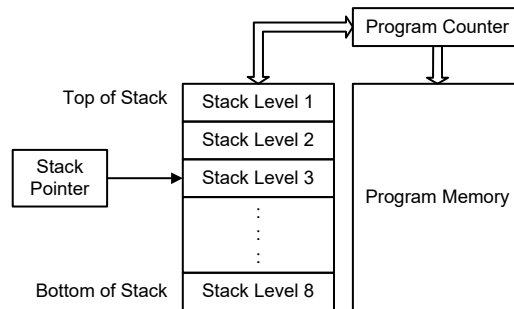
The lower byte of the Program Counter, known as the Program Counter Low register or PCL, is available for program control and is a readable and writeable register. By transferring data directly into this register, a short program jump can be executed directly. However, as only this low byte is available for manipulation, the jumps are limited to the present page of memory that is 256 locations. When such program jumps are executed it should also be noted that a dummy cycle will be inserted. Manipulating the PCL register may cause program branching, so an extra cycle is needed to pre-fetch.

Stack

This is a special part of the memory which is used to save the contents of the Program Counter only. The stack is organized into multiple levels and neither part of the data nor part of the program space, and is neither readable nor writeable. The activated level is indexed by the Stack Pointer, and is neither readable nor writeable. At a subroutine call or interrupt acknowledge signal, the contents of the Program Counter are pushed onto the stack. At the end of a subroutine or an interrupt routine, signaled by a return instruction, RET or RETI, the Program Counter is restored to its previous value from the stack. After a device reset, the Stack Pointer will point to the top of the stack.

If the stack is full and an enabled interrupt takes place, the interrupt request flag will be recorded but the acknowledge signal will be inhibited. When the Stack Pointer is decremented, by RET or RETI, the interrupt will be serviced. This feature prevents stack overflow allowing the programmer to use the structure more easily. However, when the stack is full, a CALL subroutine instruction can still be executed which will result in a stack overflow. Precautions should be taken to avoid such cases which might cause unpredictable program branching.

If the stack is overflow, the first Program Counter save in the stack will be lost.



Arithmetic and Logic Unit – ALU

The arithmetic-logic unit or ALU is a critical area of the microcontroller that carries out arithmetic and logic operations of the instruction set. Connected to the main microcontroller data bus, the ALU receives related instruction codes and performs the required arithmetic or logical operations after which the result will be placed in the specified register. As these ALU calculation or operations may result in carry, borrow or other status changes, the status register will be correspondingly updated to reflect these changes. The ALU supports the following functions:

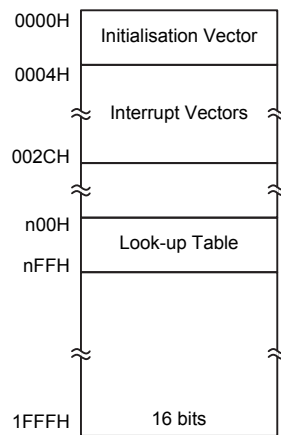
- Arithmetic operations: ADD, ADDM, ADC, ADCM, SUB, SUBM, SBC, SBCM, DAA
- Logic operations: AND, OR, XOR, ANDM, ORM, XORM, CPL, CPLA
- Rotation: RRA, RR, RRCA, RRC, RLA, RL, RLCA, RLC
- Increment and Decrement: INCA, INC, DECA, DEC
- Branch decision: JMP, SZ, SZA, SNZ, SIZ, SDZ, SIZA, SDZA, CALL, RET, RETI

Flash Program Memory

The Program Memory is the location where the user code or program is stored. For this device the Program Memory is Flash type, which means it can be programmed and re-programmed a large number of times, allowing the user the convenience of code modification on the same device. By using the appropriate programming tools, the Flash device offer users the flexibility to conveniently debug and develop their applications while also offering a means of field programming and updating.

Structure

The Program Memory has a capacity of 8K×16 bits. The Program Memory is addressed by the Program Counter and also contains data, table information and interrupt entries. Table data, which can be arranged in any location within the Program Memory, is addressed by a separate table pointer register.



Program Memory Structure

Special Vectors

Within the Program Memory, certain locations are reserved for the reset and interrupts. The location 0000H is reserved for use by the device reset for program initialisation. After a device reset is initiated, the program will jump to this location and begin execution.

Look-up Table

Any location within the Program Memory can be defined as a look-up table where programmers can store fixed data. To use the look-up table, the table pointer must first be configured by placing the address of the look up data to be retrieved in the table pointer registers, TBLP and TBHP. These registers define the total address of the look-up table.

After setting up the table pointer, the table data can be retrieved from the Program Memory using the “TABRD [m]” or “TABRDL[m]” instructions respectively. When the instruction is executed, the lower order table byte from the Program Memory will be transferred to the user defined Data Memory register [m] as specified in the instruction. The higher order table data byte from the Program Memory will be transferred to the TBLH special register. Any unused bits in this transferred higher order byte will be read as “0”.

The accompanying diagram illustrates the addressing data flow of the look-up table.

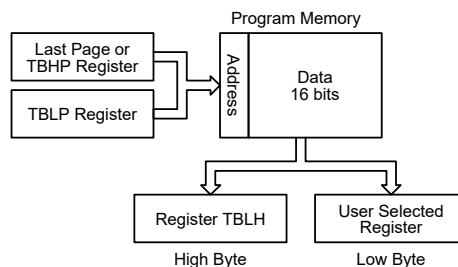


Table Program Example

The following example shows how the table pointer and table data is defined and retrieved from the microcontroller. This example uses raw table data located in the Program Memory which is stored there using the ORG statement. The value at this ORG statement is “1F00H” which refers to the start address of the last page within the 8K words Program Memory of the device. The table pointer low byte register is set here to have an initial value of “06H”. This will ensure that the first data read from the data table will be at the Program Memory address “1F06H” or 6 locations after the start of the last page. Note that the value for the table pointer is referenced to the specific address pointed by the TBLP and TBHP registers if the “TABRD [m]” instruction is being used. The high byte of the table data which in this case is equal to zero will be transferred to the TBLH register automatically when the “TABRD [m]” instruction is executed.

Because the TBLH register is a read-only register and cannot be restored, care should be taken to ensure its protection if both the main routine and Interrupt Service Routine use table read instructions. If using the table read instructions, the Interrupt Service Routines may change the value of the TBLH and subsequently cause errors if used again by the main routine. As a rule it is recommended that simultaneous use of the table read instructions should be avoided. However, in situations where simultaneous use cannot be avoided, the interrupts should be disabled prior to the execution of any main routine table-read instructions. Note that all table related instructions require two instruction cycles to complete their operation.

Table Read Program Example

```

tempreg1 db ? ; temporary register #1
tempreg2 db ? ; temporary register #2
:
:
mov a,06h      ; initialise low table pointer - note that this address is referenced
mov tblp,a    ; to the last page or the page that tbhp pointed
mov a,1Fh     ; initialise high table pointer
mov tbhp,a
:
:
tabrd tempreg1 ; transfers value in table referenced by table pointer,
               ; data at program memory address "1F06H" transferred to tempreg1 and TBLH
dec tblp      ; reduce value of table pointer by one
tabrd tempreg2 ; transfers value in table referenced by table pointer,
               ; data at program memory address "1F05H" transferred to tempreg2 and TBLH
               ; in this example the data "1AH" is transferred to tempreg1 and data "0FH" to
               ; register tempreg2
               ; the value "00H" will be transferred to the high byte register TBLH
:
:
org 1F00h     ; sets initial address of program memory
dc 00Ah, 00Bh, 00Ch, 00Dh, 00Eh, 00Fh, 01Ah, 01Bh
:

```

In Circuit Programming – ICP

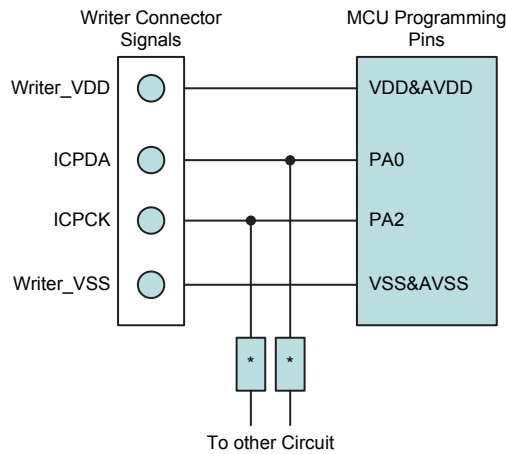
The provision of Flash type Program Memory provides the user with a means of convenient and easy upgrades and modifications to their programs on the same device.

As an additional convenience, Holtek has provided a means of programming the microcontroller in-circuit using a 4-pin interface. This provides manufacturers with the possibility of manufacturing their circuit boards complete with a programmed or un-programmed microcontroller, and then programming or upgrading the program at a later stage. This enables product manufacturers to easily keep their manufactured products supplied with the latest program releases without removal and re-insertion of the device.

| Holtek Writer Pins | MCU Programming Pins | Pin Description |
|--------------------|----------------------|---------------------------------|
| ICPDA | PA0 | Programming Serial Data/Address |
| ICPCK | PA2 | Programming Clock |
| VDD | VDD&AVDD | Power Supply |
| VSS | VSS&AVSS | Ground |

The Program Memory can be programmed serially in-circuit using this 4-wire interface. Data is downloaded and uploaded serially on a single pin with an additional line for the clock. Two additional lines are required for the power supply. The technical details regarding the in-circuit programming of the device are beyond the scope of this document and will be supplied in supplementary literature.

During the programming process, the user can take care of the ICPDA and ICPCK pins for data and clock programming purposes to ensure that no other outputs are connected to these two pins.



Note: * may be resistor or capacitor. The resistance of * must be greater than 1kΩ or the capacitance of * must be less than 1nF.

On-Chip Debug Support – OCDS

There is an EV chip named HT66V0195 which is used to emulate the real MCU device named HT66F0195. The EV chip device also provides an “On-Chip Debug” function to debug the real MCU device during the development process. The EV chip and the real MCU device are almost functionally compatible except for “On-Chip Debug” function. Users can use the EV chip device to emulate the real chip device behavior by connecting the OCSDSA and OCDSCCK pins to the Holtek HT-IDE development tools. The OCSDSA pin is the OCDS Data/Address input/output pin while the OCDSCCK pin is the OCDS clock input pin. When users use the EV chip for debugging, other functions which are shared with the OCSDSA and OCDSCCK pins in the device will have no effect in the EV chip. For more detailed OCDS information, refer to the corresponding document named “Holtek e-Link for 8-bit MCU OCDS User’s Guide”.

| Holtek e-Link Pins | EV Chip Pins | Pin Description |
|--------------------|--------------|---|
| OCSDSA | OCSDSA | On-chip Debug Support Data/Address input/output |
| OCDSCCK | OCDSCCK | On-chip Debug Support Clock input |
| VDD | VDD&AVDD | Power Supply |
| VSS | VSS&AVSS | Ground |

Data Memory

The Data Memory is a volatile area of 8-bit wide RAM internal memory and is the location where temporary information is stored.

Structure

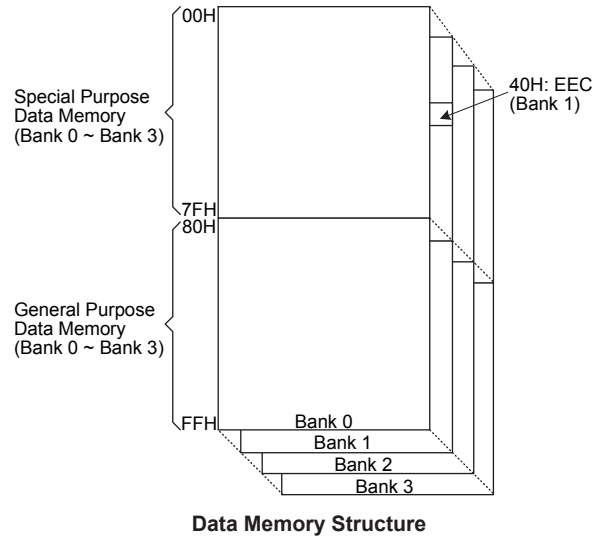
Categorised into two types, the first of these is an area of RAM, known as the Special Function Data Memory. These registers have fixed locations and are necessary for correct operation of the device. Many of these registers can be read from and written to directly under program control, however, some remain protected from user manipulation. The second area of Data Memory is known as the General Purpose Data Memory, which is reserved for general purpose use. All locations within this area are read and write accessible under program control.

The overall Data Memory is subdivided into four banks, which are implemented in 8-bit wide Memory. The Special Purpose Data Memory registers are accessible in all banks, with the exception of the EEC register at address 40H, which is only accessible in Bank 1. Switching between the different Data Memory banks is achieved by properly setting the Bank Pointer to the correct value.

The start address of the Data Memory for the device is the address 00H. The address range of the Special Purpose Data Memory for the device is from 00H to 7FH while the address range of the General Purpose Data Memory is from 80H to FFH.

| Special Purpose Data Memory | General Purpose Data Memory | |
|-----------------------------|-----------------------------|--|
| Located Bank | Capacity | Bank: Address |
| 0~3 | 512×8 | 0: 80H~FFH 1: 80H~FFH 2: 80H~FFH 3: 80H~FFH |

Data Memory Summary



General Purpose Data Memory

All microcontroller programs require an area of read/write memory where temporary data can be stored and retrieved for use later. It is this area of RAM memory that is known as General Purpose Data Memory. This area of Data Memory is fully accessible by the user programming for both reading and writing operations. By using the bit operation instructions individual bits can be set or reset under program control giving the user a large range of flexibility for bit manipulation in the Data Memory.

Special Purpose Data Memory

This area of Data Memory is where registers, necessary for the correct operation of the microcontroller, are stored. Most of the registers are both readable and writeable but some are protected and are readable only, the details of which are located under the relevant Special Function Register section. Note that for locations that are unused, any read instruction to these addresses will return the value "00H".

| Bank 0~3 | | Bank 0,2,3 : Bank 1 | |
|----------|--------|---------------------|------------|
| 00H | IAR0 | 40H | EEC |
| 01H | MP0 | 41H | PC |
| 02H | IAR1 | 42H | PCC |
| 03H | MP1 | 43H | PCPU |
| 04H | BP | 44H | ACERL |
| 05H | ACC | 45H | SIMC0 |
| 06H | PCL | 46H | SIMC1 |
| 07H | TBLP | 47H | SIMD |
| 08H | TBLH | 48H | SIMC2/SIMA |
| 09H | TBHP | 49H | SIMTOC |
| 0AH | STATUS | 4AH | SLCDC0 |
| 0BH | SMOD | 4BH | SLCDC1 |
| 0CH | LVDC | 4CH | SLCDC2 |
| 0DH | INTEG | 4DH | SLCDC3 |
| 0EH | INTC0 | 4EH | SLCDC4 |
| 0FH | INTC1 | 4FH | SLEDC0 |
| 10H | INTC2 | 50H | SLEDC1 |
| 11H | MF10 | 51H | IFS |
| 12H | MF11 | 52H | PD |
| 13H | MF12 | 53H | PDC |
| 14H | PA | 54H | PDPU |
| 15H | PAC | 55H | USR |
| 16H | PAPU | 56H | UCR1 |
| 17H | PAWU | 57H | UCR2 |
| 18H | ACERH | 58H | BRG |
| 19H | TMPC | 59H | TXR_RXR |
| 1AH | WDTC | 5AH | |
| 1BH | TBC | | |
| 1CH | CTRL | | |
| 1DH | LVRC | | |
| 1EH | EEA | | |
| 1FH | EED | | |
| 20H | SADOL | | |
| 21H | SADOH | | |
| 22H | SADC0 | | |
| 23H | SADC1 | | |
| 24H | SADC2 | | |
| 25H | PB | | |
| 26H | PBC | | |
| 27H | PBPU | | |
| 28H | CTMC0 | | |
| 29H | CTMC1 | | |
| 2AH | CTMDL | | |
| 2BH | CTMDH | | |
| 2CH | CTMAL | | |
| 2DH | CTMAH | | |
| 2EH | CTMRP | | |
| 2FH | STMC0 | | |
| 30H | STMC1 | | |
| 31H | STMDL | | |
| 32H | STMDH | | |
| 33H | STMAL | | |
| 34H | STMAH | | |
| 35H | STMRP | | |
| 36H | | | |
| 37H | PTMC0 | | |
| 38H | PTMC1 | | |
| 39H | PTMDL | | |
| 3AH | PTMDH | | |
| 3BH | PTMAL | | |
| 3CH | PTMAH | | |
| 3DH | PTMRPL | | |
| 3EH | PTMRPH | | |
| 3FH | CPC | 7FH | |

☐ : Unused, read as 00H

Special Purpose Data Memory

Special Function Register Description

Most of the Special Function Register details will be described in the relevant functional section, however several registers require a separate description in this section.

Indirect Addressing Registers – IAR0, IAR1

The Indirect Addressing Registers, IAR0 and IAR1, although having their locations in normal RAM register space, do not actually physically exist as normal registers. The method of indirect addressing for RAM data manipulation uses these Indirect Addressing Registers and Memory Pointers, in contrast to direct memory addressing, where the actual memory address is specified. Actions on the IAR0 and IAR1 registers will result in no actual read or write operation to these registers but rather to the memory location specified by their corresponding Memory Pointers, MP0 or MP1. Acting as a pair, IAR0 and MP0 can together access data from Bank 0 while the IAR1 register together with the MP1 register can access data from any Data Memory Bank. As the Indirect Addressing Registers are not physically implemented, reading the Indirect Addressing Registers will return a result of “00H” and writing to the registers will result in no operation.

Memory Pointers – MP0, MP1

Two Memory Pointers, known as MP0 and MP1 are provided. These Memory Pointers are physically implemented in the Data Memory and can be manipulated in the same way as normal registers providing a convenient way with which to address and track data. When any operation to the relevant Indirect Addressing Registers is carried out, the actual address that the microcontroller is directed to is the address specified by the related Memory Pointer. MP0, together with Indirect Addressing Register, IAR0, are used to access data from Bank 0, while MP1 together with IAR1 are used to access data from all data banks according to the BP register. Direct Addressing can only be used with Bank 0, all other Banks must be addressed indirectly using MP1 and IAR1.

The following example shows how to clear a section of four Data Memory locations already defined as locations adres1 to adres4.

Indirect Addressing Program Example

```
data .section 'data'
adres1 db ?
adres2 db ?
adres3 db ?
adres4 db ?
block db ?
code .section at 0 'code'
org 00h
start:
    mov a,04h          ; set size of block
    mov block,a
    mov a,offset adres1 ; Accumulator loaded with first RAM address
    mov mp0,a         ; set memory pointer with first RAM address
loop:
    clr IAR0          ; clear the data at address defined by mp0
    inc mp0           ; increment memory pointer
    sdz block         ; check if last memory location has been cleared
    jmp loop
continue:
```

The important point to note here is that in the example shown above, no reference is made to specific Data Memory addresses.

Bank Pointer – BP

The Data Memory is divided into four banks, Banks 0~3. Selecting the required Data Memory area is achieved using the Bank Pointer. Bits 1~0 of the Bank Pointer are used to select Data Memory Banks 0~3. The Data Memory is initialised to Bank 0 after a reset, except for a WDT time-out reset in the IDLE or SLEEP Mode, in which case, the Data Memory bank remains unaffected. It should be noted that the Special Function Data Memory is not affected by the bank selection, which means that the Special Function Registers can be accessed within any bank. Directly addressing the Data Memory will always result in Bank 0 being accessed irrespective of the value of the Bank Pointer. Accessing data from banks other than Bank 0 must be implemented using the indirect addressing.

• BP Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---|---|---|---|---|---|-------|-------|
| Name | — | — | — | — | — | — | DMBP1 | DMBP0 |
| R/W | — | — | — | — | — | — | R/W | R/W |
| POR | — | — | — | — | — | — | 0 | 0 |

Bit 7~2 Unimplemented, read as “0”

Bit 1~0 **DMBP1~DMBP0**: Data Memory Bank selection
 00: Bank 0
 01: Bank 1
 10: Bank 2
 11: Bank 3

Accumulator – ACC

The Accumulator is central to the operation of any microcontroller and is closely related with operations carried out by the ALU. The Accumulator is the place where all intermediate results from the ALU are stored. Without the Accumulator it would be necessary to write the result of each calculation or logical operation such as addition, subtraction, shift, etc., to the Data Memory resulting in higher programming and timing overheads. Data transfer operations usually involve the temporary storage function of the Accumulator; for example, when transferring data between one user-defined register and another, it is necessary to do this by passing the data through the Accumulator as no direct transfer between two registers is permitted.

Program Counter Low Register – PCL

To provide additional program control functions, the low byte of the Program Counter is made accessible to programmers by locating it within the Special Purpose area of the Data Memory. By manipulating this register, direct jumps to other program locations are easily implemented. Loading a value directly into this PCL register will cause a jump to the specified Program Memory location, however, as the register is only 8-bit wide, only jumps within the current Program Memory page are permitted. When such operations are used, note that a dummy cycle will be inserted.

Look-up Table Registers – TBLP, TBHP, TBLH

These three special function registers are used to control operation of the look-up table which is stored in the Program Memory. TBLP and TBHP are the table pointers and indicate the location where the table data is located. Their value must be set before any table read commands are executed. Their value can be changed, for example using the “INC” or “DEC” instructions, allowing for easy table data pointing and reading. TBLH is the location where the high order byte of the table data is stored after a table read data instruction has been executed. Note that the lower order table data byte is transferred to a user defined location.

Status Register – STATUS

This 8-bit register contains the zero flag (Z), carry flag (C), auxiliary carry flag (AC), overflow flag (OV), power down flag (PDF), and watchdog time-out flag (TO). These arithmetic/logical operation and system management flags are used to record the status and operation of the microcontroller.

With the exception of the TO and PDF flags, bits in the status register can be altered by instructions like most other registers. Any data written into the status register will not change the TO or PDF flag. In addition, operations related to the status register may give different results due to the different instruction operations. The TO flag can be affected only by a system power-up, a WDT time-out or by executing the “CLR WDT” or “HALT” instruction. The PDF flag is affected only by executing the “HALT” or “CLR WDT” instruction or during a system power-up.

The Z, OV, AC, and C flags generally reflect the status of the latest operations.

- **C** is set if an operation results in a carry during an addition operation or if a borrow does not take place during a subtraction operation; otherwise C is cleared. C is also affected by a rotate through carry instruction.
- **AC** is set if an operation results in a carry out of the low nibbles in addition, or no borrow from the high nibble into the low nibble in subtraction; otherwise AC is cleared.
- **Z** is set if the result of an arithmetic or logical operation is zero; otherwise Z is cleared.
- **OV** is set if an operation results in a carry into the highest-order bit but not a carry out of the highest-order bit, or vice versa; otherwise OV is cleared.
- **PDF** is cleared by a system power-up or executing the “CLR WDT” instruction. PDF is set by executing the “HALT” instruction.
- **TO** is cleared by a system power-up or executing the “CLR WDT” or “HALT” instruction. TO is set by a WDT time-out.

In addition, on entering an interrupt sequence or executing a subroutine call, the status register will not be pushed onto the stack automatically. If the contents of the status registers are important and if the subroutine can corrupt the status register, precautions must be taken to correctly save it.

• STATUS Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---|---|----|-----|-----|-----|-----|-----|
| Name | — | — | TO | PDF | OV | Z | AC | C |
| R/W | — | — | R | R | R/W | R/W | R/W | R/W |
| POR | — | — | 0 | 0 | x | x | x | x |

“x”: unknown

Bit 7~6 Unimplemented, read as “0”

Bit 5 **TO**: Watchdog Time-Out flag
 0: After power up or executing the “CLR WDT” or “HALT” instruction
 1: A watchdog time-out occurred.

Bit 4 **PDF**: Power down flag
 0: After power up or executing the “CLR WDT” instruction
 1: By executing the “HALT” instruction

Bit 3 **OV**: Overflow flag
 0: No overflow
 1: An operation results in a carry into the highest-order bit but not a carry out of the highest-order bit or vice versa.

Bit 2 **Z**: Zero flag
 0: The result of an arithmetic or logical operation is not zero
 1: The result of an arithmetic or logical operation is zero

- Bit 1 **AC:** Auxiliary flag
 0: No auxiliary carry
 1: An operation results in a carry out of the low nibbles in addition, or no borrow from the high nibble into the low nibble in subtraction
- Bit 0 **C:** Carry flag
 0: No carry-out
 1: An operation results in a carry during an addition operation or if a borrow does not take place during a subtraction operation
 The C flag is also affected by a rotate through carry instruction.

EEPROM Data Memory

This device contains an area of internal EEPROM Data Memory. EEPROM is a non-volatile form of re-programmable memory, with data retention even when its power supply is removed. By incorporating this kind of data memory, a whole new host of application possibilities are made available to the designer. The availability of EEPROM storage allows information such as product identification numbers, calibration values, specific user data, system setup data or other product information to be stored directly within the product microcontroller. The process of reading and writing data to the EEPROM memory has been reduced to a very trivial affair.

EEPROM Data Memory Structure

The EEPROM Data Memory capacity is 128×8 bits for the device. Unlike the Program Memory and RAM Data Memory, the EEPROM Data Memory is not directly mapped into memory space and is therefore not directly addressable in the same way as the other types of memory. Read and Write operations to the EEPROM are carried out in single byte operations using an address and a data register in Bank 0 and a single control register in Bank 1.

EEPROM Registers

Three registers control the overall operation of the internal EEPROM Data Memory. These are the address register, EEA, the data register, EED and a single control register, EEC. As both the EEA and EED registers are located in Bank 0, they can be directly accessed in the same way as any other Special Function Register. The EEC register however, being located in Bank 1, can only be read from or written to indirectly using the MP1 Memory Pointer and Indirect Addressing Register, IAR1. Because the EEC control register is located at address 40H in Bank 1, the MP1 Memory Pointer must first be set to the value 40H and the Bank Pointer register, BP, set to the value, 01H, before any operations on the EEC register are executed.

| Register Name | Bit | | | | | | | |
|---------------|-----|------|------|------|------|------|------|------|
| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| EEA | — | EEA6 | EEA5 | EEA4 | EEA3 | EEA2 | EEA1 | EEA0 |
| EED | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| EEC | — | — | — | — | WREN | WR | RDEN | RD |

EEPROM Register List

• EEA Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---|------|------|------|------|------|------|------|
| Name | — | EEA6 | EEA5 | EEA4 | EEA3 | EEA2 | EEA1 | EEA0 |
| R/W | — | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | — | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

- Bit 7 Unimplemented, read as “0”
 Bit 6~0 **EEA6~EEA0:** Data EEPROM address bit 6 ~ bit 0

• **EED Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-----|-----|-----|-----|-----|-----|-----|-----|
| Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~0 **D7~D0**: Data EEPROM data bit 7 ~ bit 0

• **EEC Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---|---|---|---|------|-----|------|-----|
| Name | — | — | — | — | WREN | WR | RDEN | RD |
| R/W | — | — | — | — | R/W | R/W | R/W | R/W |
| POR | — | — | — | — | 0 | 0 | 0 | 0 |

Bit 7~4 Unimplemented, read as “0”

Bit 3 **WREN**: Data EEPROM Write Enable
 0: Disable
 1: Enable

This is the Data EEPROM Write Enable Bit which must be set high before Data EEPROM write operations are carried out. Clearing this bit to zero will inhibit Data EEPROM write operations.

Bit 2 **WR**: EEPROM Write Control
 0: Write cycle has finished
 1: Activate a write cycle

This is the Data EEPROM Write Control Bit and when set high by the application program will activate a write cycle. This bit will be automatically reset to zero by the hardware after the write cycle has finished. Setting this bit high will have no effect if the WREN has not first been set high.

Bit 1 **RDEN**: Data EEPROM Read Enable
 0: Disable
 1: Enable

This is the Data EEPROM Read Enable Bit which must be set high before Data EEPROM read operations are carried out. Clearing this bit to zero will inhibit Data EEPROM read operations.

Bit 0 **RD**: EEPROM Read Control
 0: Read cycle has finished
 1: Activate a read cycle

This is the Data EEPROM Read Control Bit and when set high by the application program will activate a read cycle. This bit will be automatically reset to zero by the hardware after the read cycle has finished. Setting this bit high will have no effect if the RDEN has not first been set high.

- Note: 1. The WREN, WR, RDEN and RD cannot be set high at the same time in one instruction. The WR and RD cannot be set high at the same time.
 2. Ensure that the f_{SUB} clock is stable before executing the write operation.
 3. Ensure that the write operation is totally complete before changing the contents of the EEPROM related registers.

Reading Data from the EEPROM

To read data from the EEPROM, the EEPROM address of the data to be read must first be placed in the EEA register. Then the read enable bit, RDEN, in the EEC register must be set high to enable the read function. If the RD bit in the EEC register is now set high, a read cycle will be initiated. Setting the RD bit high will not initiate a read operation if the RDEN bit has not been set. When the read cycle terminates, the RD bit will be automatically cleared to zero, after which the data can be read from the EED register. The data will remain in the EED register until another read or write operation is executed. The application program can poll the RD bit to determine when the data is valid for reading.

Writing Data to the EEPROM

To write data to the EEPROM, the EEPROM address of the data to be written must first be placed in the EEA register and the data placed in the EED register. To initiate a write cycle, the write enable bit, WREN, in the EEC register must first be set high to enable the write function. After this, the WR bit in the EEC register must be immediately set high to initiate a write cycle. These two instructions must be executed in two consecutive instruction cycles. The global interrupt bit EMI should also first be cleared before implementing any write operations, and then set again after the write cycle has started. Note that setting the WR bit high will not initiate a write cycle if the WREN bit has not been set. As the EEPROM write cycle is controlled using an internal timer whose operation is asynchronous to microcontroller system clock, a certain time will elapse before the data will have been written into the EEPROM. Detecting when the write cycle has finished can be implemented either by polling the WR bit in the EEC register or by using the EEPROM interrupt. When the write cycle terminates, the WR bit will be automatically cleared to zero by the microcontroller, informing the user that the data has been written to the EEPROM. The application program can therefore poll the WR bit to determine when the write cycle has ended.

Write Protection

Protection against inadvertent write operation is provided in several ways. After the device is powered-on the Write Enable bit in the control register will be cleared preventing any write operations. Also at power-on the Bank Pointer register, BP, will be reset to zero, which means that Data Memory Bank 0 will be selected. As the EEPROM control register is located in Bank 1, this adds a further measure of protection against spurious write operations. During normal program operation, ensuring that the Write Enable bit in the control register is cleared will safeguard against incorrect write operations.

EEPROM Interrupt

The EEPROM write interrupt is generated when an EEPROM write cycle has ended. The EEPROM interrupt must first be enabled by setting the DEE bit in the relevant interrupt register. However, as the EEPROM interrupt is contained within a Multi-function Interrupt, the associated multi-function interrupt enable bit must also be set. When an EEPROM write cycle ends, the DEF request flag and its associated multi-function interrupt request flag will both be set high. If the global, EEPROM and multi-function interrupts are enabled and the stack is not full, a jump to the associated Multi-function Interrupt vector will take place. When the interrupt is serviced the Multi-function interrupt flag will be automatically reset, the EEPROM interrupt flag will be automatically reset.

Programming Considerations

Care must be taken that data is not inadvertently written to the EEPROM. Protection can be enhanced by ensuring that the Write Enable bit is normally cleared to zero when not writing. Also the Bank Pointer register, BP, could be normally cleared to zero as this would inhibit access to Bank 1 where the EEPROM control register exists. Although certainly not necessary, consideration might be given in the application program to the checking of the validity of new write data by a simple read back process.

When writing data the WR bit must be set high immediately after the WREN bit has been set high, to ensure the write cycle executes correctly. The global interrupt bit EMI should also be cleared before a write cycle is executed and then re-enabled after the write cycle starts. Note that the device should not enter the IDLE or SLEEP mode until the EEPROM read or write operation is totally complete. Otherwise, the EEPROM read or write operation will fail.

Programming Examples

Reading data from the EEPROM – polling method

```
MOV A, EEPROM_ADRES      ; user defined address
MOV EEA, A
MOV A, 40H                ; set memory pointer MP1
MOV MP1, A                ; MP1 points to EEC register
MOV A, 01H                ; set bank pointer BP
MOV BP, A
SET IAR1.1                ; set RDEN bit, enable read operations
SET IAR1.0                ; start Read Cycle - set RD bit
BACK:
SZ IAR1.0                 ; check for read cycle end
JMP BACK
CLR IAR1                  ; disable EEPROM read if no more read operations are required
CLR BP
MOV A, EED                 ; move read data to register
MOV READ_DATA, A
```

Note: For each read operation, the address register should be re-specified followed by setting the RD bit high to activate a read cycle even if the target address is consecutive.

Writing Data to the EEPROM – polling method

```
MOV A, EEPROM_ADRES      ; user defined address
MOV EEA, A
MOV A, EEPROM_DATA       ; user defined data
MOV EED, A
MOV A, 40H                ; set memory pointer MP1
MOV MP1, A                ; MP1 points to EEC register
MOV A, 01H                ; set bank pointer BP
MOV BP, A
CLR EMI
SET IAR1.3                ; set WREN bit, enable write operations
SET IAR1.2                ; start Write Cycle - set WR bit - executed immediately
                        ; after set WREN bit

SET EMI
BACK:
SZ IAR1.2                 ; check for write cycle end
JMP BACK
CLR BP
```

Oscillators

Various oscillator options offer the user a wide range of functions according to their various application requirements. The flexible features of the oscillator functions ensure that the best optimisation can be achieved in terms of speed and power saving. Oscillator selections and operation are selected through a combination of configuration options and relevant control registers.

Oscillator Overview

In addition to being the source of the main system clock the oscillators also provide clock sources for the Watchdog Timer and Time Base Interrupts. External oscillators requiring some external components as well as fully integrated internal oscillators, requiring no external components, are provided to form a range of both fast and slow system oscillators. All oscillator type selections are implemented through configuration options. The higher frequency oscillator provides higher performance but carry with it the disadvantage of higher power requirements, while the opposite is of course true for the lower frequency oscillator. With the capability of dynamically switching between fast and slow system clock, the device has the flexibility to optimize the performance/power ratio, a feature especially important in power sensitive portable applications.

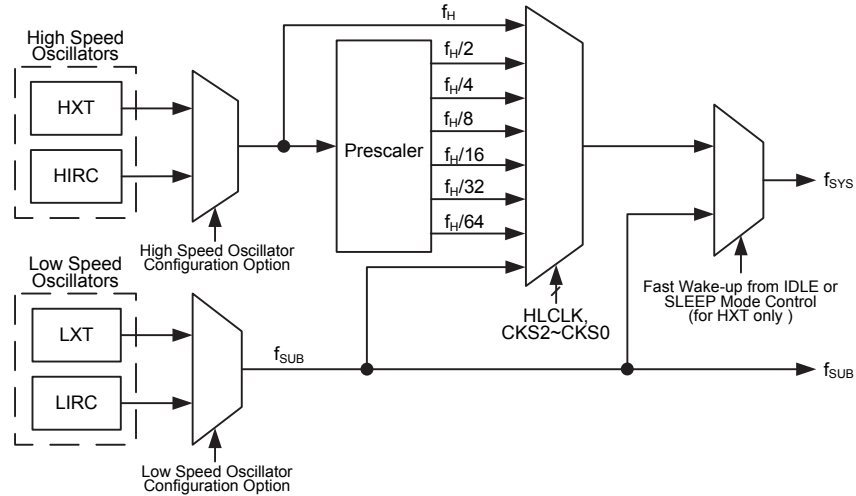
| Type | Name | Frequency | Pins |
|-----------------------------|------|--------------|-----------|
| External High Speed Crystal | HXT | 400kHz~16MHz | OSC1/OSC2 |
| Internal High Speed RC | HIRC | 8/12/16MHz | — |
| External Low Speed Crystal | LXT | 32.768kHz | XT1/XT2 |
| Internal Low Speed RC | LIRC | 32kHz | — |

Oscillator Types

System Clock Configurations

There are four methods of generating the system clock, two high speed oscillators and two low speed oscillators. The high speed oscillators are the external crystal/ceramic oscillator, HXT, and the internal 8/12/16MHz RC oscillator, HIRC. The two low speed oscillators are the internal 32kHz RC oscillator, LIRC, and the external 32.768kHz crystal oscillator, LXT. Selecting whether the low or high speed oscillator is used as the system oscillator is implemented using the HLCLK bit and CKS2~CKS0 bits in the SMOD register and the system clock can be dynamically selected.

The actual source clock used for each of the high and low speed oscillators is chosen via configuration options. The frequency of the slow speed or high speed system clock is also determined using the HLCLK bit and CKS2~CKS0 bits in the SMOD register. Note that two oscillator selections must be made namely one high speed and one low speed system oscillators. It is not possible to choose a no-oscillator selection for either the high or low speed oscillator.

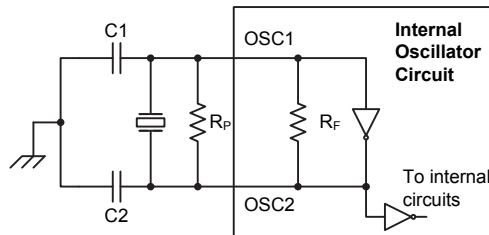


System Clock Configurations

External Crystal/Ceramic Oscillator – HXT

The External Crystal/Ceramic System Oscillator is one of the high frequency oscillators. For most crystal oscillator configurations, the simple connection of a crystal across OSC1 and OSC2 will create the necessary phase shift and feedback for oscillation, without requiring external capacitors. However, for some crystal types and frequencies, to ensure oscillation, it may be necessary to add two small value capacitors, C1 and C2. Using a ceramic resonator will usually require two small value capacitors, C1 and C2, to be connected as shown for oscillation to occur. The values of C1 and C2 should be selected in consultation with the crystal or resonator manufacturer’s specification.

For oscillator stability and to minimise the effects of noise and crosstalk, it is important to ensure that the crystal and any associated resistors and capacitors along with interconnecting lines are all located as close to the MCU as possible.



- Note: 1. R_P is normally not required. C1 and C2 are required.
 2. Although not shown OSC1/OSC2 pins have a parasitic capacitance of around 7pF.

Crystal/Resonator Oscillator – HXT

| Crystal Oscillator C1 and C2 Values | | |
|---|-------|-------|
| Crystal Frequency | C1 | C2 |
| 16MHz | 0pF | 0pF |
| 12MHz | 0pF | 0pF |
| 8MHz | 0pF | 0pF |
| 6MHz | 0pF | 0pF |
| 4MHz | 0pF | 0pF |
| 1MHz | 100pF | 100pF |
| Note: C1 and C2 values are for guidance only. | | |

Crystal Recommended Capacitor Values

High Speed Internal RC Oscillator – HIRC

The high speed internal RC oscillator is a fully integrated system oscillator requiring no external components. The internal RC oscillator has three fixed frequencies of 8MHz, 12MHz and 16MHz, selected via configuration option. Device trimming during the manufacturing process and the inclusion of internal frequency compensation circuits are used to ensure that the influence of the power supply voltage, temperature and process variations on the oscillation frequency are minimised. Note that if this internal system clock option is selected, it requires no external pins for its operation.

External 32.768kHz Crystal Oscillator – LXT

The External 32.768kHz Crystal System Oscillator is one of the low frequency oscillator choices, which is selected via configuration option. This clock source has a fixed frequency of 32.768kHz and requires a 32.768kHz crystal to be connected between pins XT1 and XT2. The external resistor and capacitor components connected to the 32.768kHz crystal are necessary to provide oscillation. For applications where precise frequencies are essential, these components may be required to provide frequency compensation due to different crystal manufacturing tolerances. During power-up there is a time delay associated with the LXT oscillator waiting for it to start-up.

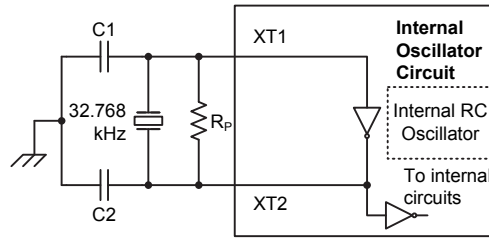
When the microcontroller enters the SLEEP or IDLE Mode, the system clock is switched off to stop microcontroller activity and to conserve power. However, in many microcontroller applications it may be necessary to keep the internal timers operational even when the microcontroller is in the SLEEP or IDLE Mode. To do this, another clock, independent of the system clock, must be provided.

However, for some crystals, to ensure oscillation and accurate frequency generation, it is necessary to add two small value external capacitors, C1 and C2. The exact values of C1 and C2 should be selected in consultation with the crystal or resonator manufacturer's specification. The external parallel feedback resistor, R_p , is required.

The pin-shared software control bits determine if the XT1/XT2 pins are used for the LXT oscillator or as I/O or other pin-shared functional pins.

- If the LXT oscillator is not used for any clock source, the XT1/XT2 pins can be used as normal I/O or other pin-shared functional pins.
- If the LXT oscillator is used for any clock source, the 32.768kHz crystal should be connected to the XT1/XT2 pins.

For oscillator stability and to minimise the effects of noise and crosstalk, it is important to ensure that the crystal and any associated resistors and capacitors along with interconnecting lines are all located as close to the MCU as possible.



Note: 1. R_p , C1 and C2 are required.
 2. Although not shown XT1/XT2 pins have a parasitic capacitance of around 7pF.

External LXT Oscillator

| LXT Oscillator C1 and C2 Values | | |
|---|------|------|
| Crystal Frequency | C1 | C2 |
| 32.768kHz | 10pF | 10pF |
| Note: 1. C1 and C2 values are for guidance only. 2. $R_p=5M\sim 10M\Omega$ is recommended. | | |

32.768kHz Crystal Recommended Capacitor Values

LXT Oscillator Low Power Function

The LXT oscillator can function in one of two modes, the Quick Start Mode and the Low Power Mode. The mode selection is executed using the LXTLP bit in the TBC register.

| LXTLP | LXT Operating Mode |
|-------|--------------------|
| 0 | Quick Start |
| 1 | Low-Power |

After power on, the LXTLP bit will be automatically cleared to zero ensuring that the LXT oscillator is in the Quick Start operating mode. In the Quick Start Mode the LXT oscillator will power up and stabilise quickly. However, after the LXT oscillator has fully powered up it can be placed into the Low-power mode by setting the LXTLP bit high. The oscillator will continue to run but with reduced current consumption, as the higher current consumption is only required during the LXT oscillator start-up. In power sensitive applications, such as battery applications, where power consumption must be kept to a minimum, it is therefore recommended that the application program sets the LXTLP bit high about 2 seconds after power-on. It should be noted that, no matter what condition the LXTLP bit is set to, the LXT oscillator will always function normally and the only difference is that it will take more time to start up if in the Low-power mode.

Internal 32kHz Oscillator – LIRC

The internal 32kHz System Oscillator is one of the low frequency oscillator choices, which is selected via configuration option. It is a fully integrated RC oscillator with a typical frequency of 32kHz, requiring no external components for its implementation. Device trimming during the manufacturing process and the inclusion of internal frequency compensation circuits are used to ensure that the influence of the power supply voltage, temperature and process variations on the oscillation frequency are minimised.

Supplementary Oscillators

The low speed oscillators, in addition to providing a system clock source are also used to provide a clock source to other device functions, such as the Watchdog Timer and the Time Base Interrupts, etc.

Operating Modes and System Clocks

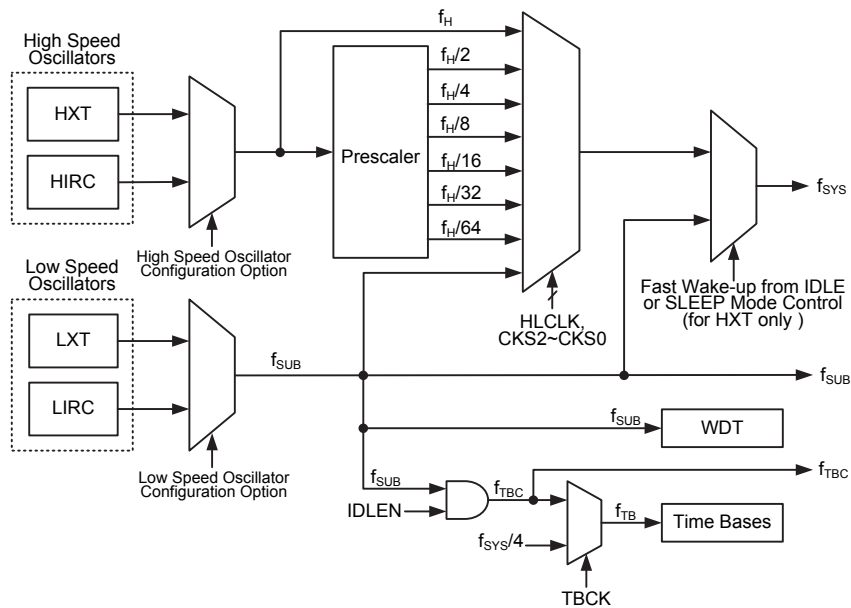
Present day applications require that their microcontrollers have high performance but often still demand that they consume as little power as possible, conflicting requirements that are especially true in battery powered portable applications. The fast clocks required for high performance will by their nature increase current consumption and of course vice-versa, lower speed clocks reduce current consumption. As Holtek has provided the device with both high and low speed clock sources and the means to switch between them dynamically, the user can optimise the operation of their microcontroller to achieve the best performance/power ratio.

System Clocks

The device has many different clock sources for both the CPU and peripheral function operation. By providing the user with a wide range of clock selections using configuration options and register programming, a clock system can be configured to obtain maximum application performance.

The main system clock, can come from either a high frequency f_H or low frequency f_{SUB} source, and is selected using the HLCLK bit and CKS2~CKS0 bits in the SMOD register. The high speed system clock can be sourced from either an HXT or HIRC oscillator, selected via a configuration option. The low speed system clock source can be sourced from internal clock f_{SUB} . If f_{SUB} is selected then it can be sourced by either the LXT or LIRC oscillator, selected via a configuration option. The other choice, which is a divided version of the high speed system oscillator has a range of $f_H/2 \sim f_H/64$.

There are two additional internal clocks for the peripheral circuits, the substitute clock, f_{SUB} , and the Time Base clock, f_{TBC} . Each of these internal clocks is sourced by either the LXT or LIRC oscillator, selected via a configuration option. The f_{SUB} clock is used to provide a substitute clock for the microcontroller just after a wake-up has occurred to enable faster wake-up times.



Device Clock Configurations

Note: When the system clock source f_{SYS} is switched to f_{SUB} from f_H , the high speed oscillator will be stopped to conserve the power. Thus there is no $f_H \sim f_H/64$ clock sources for peripheral circuit to use.

System Operation Modes

There are six different modes of operation for the microcontroller, each one with its own special characteristics and which can be chosen according to the specific performance and power requirements of the application. There are two modes allowing normal operation of the microcontroller, the NORMAL Mode and SLOW Mode. The remaining four modes, the SLEEP0, SLEEP1, IDLE0 and IDLE1 Mode, are used when the microcontroller CPU is switched off to conserve power.

| Operation Mode | Description | | | |
|----------------|-------------|------------------------------------|------------------|------------------|
| | CPU | f _{sys} | f _{sub} | f _{rbc} |
| NORMAL | On | f _H ~f _H /64 | On | On |
| SLOW | On | f _{sub} | On | On |
| IDLE0 | Off | Off | On | On |
| IDLE1 | Off | On | On | On |
| SLEEP0 | Off | Off | Off | Off |
| SLEEP1 | Off | Off | On | Off |

NORMAL Mode

As the name suggests this is one of the main operating modes where the microcontroller has all of its functions operational and where the system clock is provided by one of the high speed oscillators. This mode operates allowing the microcontroller to operate normally with a clock source will come from one of the high speed oscillators, either the HXT or HIRC oscillator. The high speed oscillator will however first be divided by a ratio ranging from 1 to 64, the actual ratio being selected by the CKS2~CKS0 and HLCLK bits in the SMOD register. Although a high speed oscillator is used, running the microcontroller at a divided clock ratio reduces the operating current.

SLOW Mode

This is also a mode where the microcontroller operates normally although now with a slower speed clock source. The clock source used will be from one of the slow speed oscillators, either the LXT or LIRC oscillator. Running the microcontroller in this mode allows it to run with much lower operating current. In the SLOW mode, the f_H clock is off.

SLEEP0 Mode

The SLEEP0 Mode is entered when an HALT instruction is executed and when the IDLEN bit in the SMOD register is low. In the SLEEP0 mode the CPU will be stopped, the f_{sub} clock will also be stopped and the Watchdog Timer function is disabled. In this mode, the LVDEN must be cleared to "0". If the LVDEN bit is set to "1", the system will not enter the SLEEP0 Mode.

SLEEP1 Mode

The SLEEP1 Mode is entered when an HALT instruction is executed and when the IDLEN bit in the SMOD register is low. In the SLEEP1 mode the CPU will be stopped. However the f_{sub} clock will continue to operate if the Watchdog Timer function is enabled.

IDLE0 Mode

The IDLE0 Mode is entered when a HALT instruction is executed and when the IDLEN bit in the SMOD register is high and the FSYSON bit in the CTRL register is low. In the IDLE0 Mode the system oscillator will be inhibited from driving the CPU but some peripheral functions will remain operational such as the Watchdog Timer and TMs. In the IDLE0 Mode, the system oscillator will be stopped.

IDLE1 Mode

The IDLE1 Mode is entered when an HALT instruction is executed and when the IDLEN bit in the SMOD register is high and the FSYSN bit in the CTRL register is high. In the IDLE1 Mode the system oscillator will be inhibited from driving the CPU but may continue to provide a clock source to keep some peripheral functions operational such as the Watchdog Timer and TMs. In the IDLE1 Mode, the system oscillator will continue to run, and this system oscillator may be a high speed or low speed system oscillator.

Control Registers

The SMOD register and the FSYSN bit in the CTRL register are used for overall control of the internal clocks within the device.

• SMOD Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|------|------|------|-------|-----|-----|-------|-------|
| Name | CKS2 | CKS1 | CKS0 | FSTEN | LTO | HTO | IDLEN | HLCLK |
| R/W | R/W | R/W | R/W | R/W | R | R | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 |

Bit 7~5 **CKS2~CKS0:** System clock selection when HLCLK is “0”

- 000: f_{SUB}
- 001: f_{SUB}
- 010: $f_H/64$
- 011: $f_H/32$
- 100: $f_H/16$
- 101: $f_H/8$
- 110: $f_H/4$
- 111: $f_H/2$

These three bits are used to select which clock is used as the system clock source. In addition to the system clock source, which can be either the LXT or LIRC oscillator, a divided version of the high speed system oscillator can also be chosen as the system clock source.

Bit 4 **FSTEN:** Fast Wake-up Control (only for HXT)

- 0: Disable
- 1: Enable

This is the Fast Wake-up Control bit which determines if the f_{SUB} clock source is initially used after the device wakes up. When the bit is high, the f_{SUB} clock source can be used as a temporary system clock to provide a faster wake up time as the f_{SUB} clock is available.

Bit 3 **LTO:** Low speed system oscillator ready flag

- 0: Not ready
- 1: Ready

This is the low speed system oscillator ready flag which indicates when the low speed system oscillator is stable after power on reset or a wake-up has occurred. The flag will be low when in the SLEEP0 Mode but after a wake-up has occurred, the flag will change to a high level after 128 clock cycles if the LXT oscillator is used and 1~2 clock cycles if the LIRC oscillator is used.

- Bit 2 **HTO**: High speed system oscillator ready flag
 0: Not ready
 1: Ready
 This is the high speed system oscillator ready flag which indicates when the high speed system oscillator is stable. This flag is cleared to “0” by hardware when the device is powered on and then changes to a high level after the high speed system oscillator is stable.
 Therefore this flag will always be read as “1” by the application program after device power-on. The flag will be low when in the SLEEP or IDLE0 Mode but after a wake-up has occurred, the flag will change to a high level after 128 clock cycles if the HXT oscillator is used and after 15~16 clock cycles if the HIRC oscillator is used.
- Bit 1 **IDLEN**: IDLE mode control
 0: Disable
 1: Enable
 This is the IDLE mode control bit and determines what happens when the HALT instruction is executed. If this bit is high, when a HALT instruction is executed, the device will enter the IDLE mode. In the IDLE mode the CPU will stop running but the system clock will continue to keep the peripheral functions operational, if the FSYSON bit is high. If the FSYSON bit is low, the CPU and the system clock will all stop in IDLE0 mode. If the bit is low, the device will enter the SLEEP mode when a HALT instruction is executed.
- Bit 0 **HLCLK**: System clock selection
 0: $f_H/2 \sim f_H/64$ or f_{SUB}
 1: f_H
 This bit is used to select if the f_H clock or the $f_H/2 \sim f_H/64$ or f_{SUB} clock is used as the system clock. When the bit is high the f_H clock will be selected and if low the $f_H/2 \sim f_H/64$ or f_{SUB} clock will be selected. When system clock switches from the f_H clock to the f_{SUB} clock and the f_H clock will be automatically switched off to conserve power.

• **CTRL Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|--------|---|---|---|---|------|-----|-----|
| Name | FSYSON | — | — | — | — | LVRF | LRF | WRF |
| R/W | R/W | — | — | — | — | R/W | R/W | R/W |
| POR | 0 | — | — | — | — | x | 0 | 0 |

“x”: unknown

- Bit 7 **FSYSON**: f_{SYS} Control in IDLE Mode
 0: Disable
 1: Enable
 This bit is used to control whether the system clock is switched on or not in the IDLE Mode. If this bit is cleared to “0”, the system clock will be switched off in the IDLE Mode. However, the system clock will be switched on in the IDLE Mode when the FSYSON bit is set to “1”.
- Bit 6~3 Unimplemented, read as “0”
- Bit 2 **LVRF**: LVR function reset flag
 Refer to the Low Voltage Reset section.
- Bit 1 **LRF**: LVR control register software reset flag
 Refer to the Low Voltage Reset section.
- Bit 0 **WRF**: WDT control register software reset flag
 Refer to the Watchdog Timer Control Register section.

Fast Wake-up

To minimise power consumption the device can enter the SLEEP or IDLE0 Mode, where the system clock source to the device will be stopped. However when the device is woken up again, it can take a considerable time for the original system oscillator to restart, stabilise and allow normal operation to resume. To ensure the device is up and running as fast as possible a Fast Wake-up function is provided, which allows f_{SUB} , namely either the LXT or LIRC oscillator, to act as a temporary clock to first drive the system until the original system oscillator has stabilised. As the clock source for the Fast Wake-up function is f_{SUB} , the Fast Wake-up function is only available in the SLEEP1 and IDLE0 modes. When the device is woken up from the SLEEP0 mode, the Fast Wake-up function has no effect because the f_{SUB} clock is stopped. The Fast Wake-up enable/disable function is controlled using the FSTEN bit in the SMOD register.

If the HXT oscillator is selected as the NORMAL Mode system clock, and if the Fast Wake-up function is enabled, then it will take one to two t_{SUB} clock cycles of the LIRC or LXT oscillator for the system to wake-up. The system will then initially run under the f_{SUB} clock source until 512 HXT clock cycles have elapsed, at which point the HTO flag will switch high and the system will switch over to operating from the HXT oscillator.

If the HIRC oscillator or LIRC oscillator is used as the system oscillator then it will take 15~16 clock cycles of the HIRC or 1~2 cycles of the LIRC to wake up the system from the SLEEP or IDLE0 Mode. The Fast Wake-up bit, FSTEN will have no effect in these cases.

| System Oscillator | FSTEN Bit | Wake-up Time (SLEEP0 Mode) | Wake-up Time (SLEEP1 Mode) | Wake-up Time (IDLE0 Mode) | Wake-up Time (IDLE1 Mode) |
|-------------------|-----------|----------------------------|--|---------------------------|---------------------------|
| HXT | 0 | 128 HXT cycles | 128 HXT cycles | | 1~2 HXT cycles |
| | 1 | 128 HXT cycles | 1~2 f_{SUB} cycles (System runs with f_{SUB} first for 512 HXT cycles and then switches over to run with the HXT clock) | | 1~2 HXT cycles |
| HIRC | x | 15~16 HIRC cycles | 15~16 HIRC cycles | | 1~2 HIRC cycles |
| LIRC | x | 1~2 LIRC cycles | 1~2 LIRC cycles | | 1~2 LIRC cycles |
| LXT | x | 128 LXT cycles | 1~2 LXT cycles | | 1~2 LXT cycles |

"x": Don't care

Wake-up Times

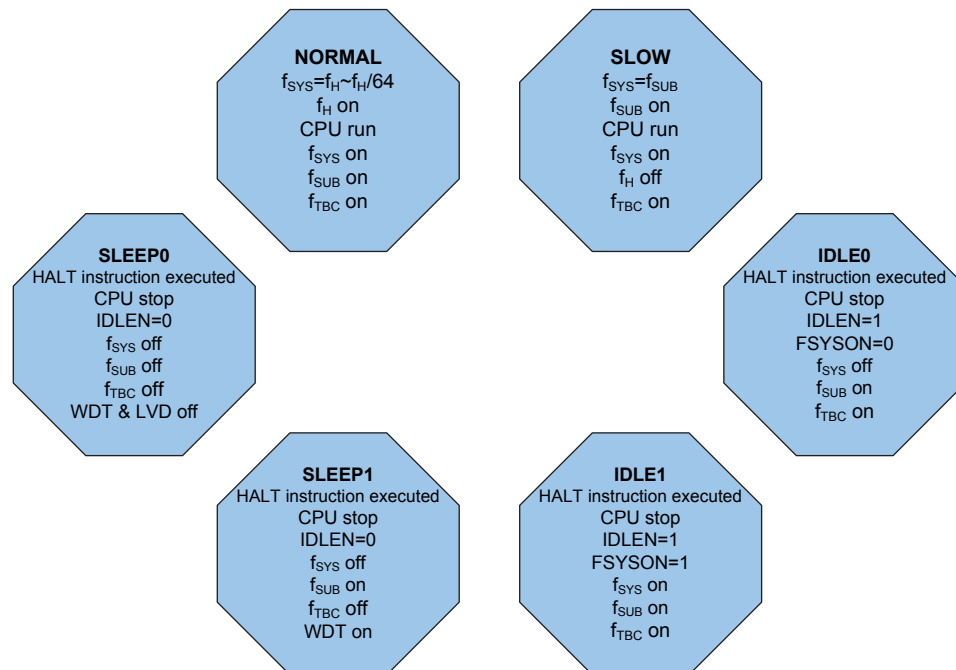
Note that if the Watchdog Timer is disabled, which means that the LXT and LIRC are both off, then there will be no Fast Wake-up function available when the device wakes up from the SLEEP0 Mode.

Operating Mode Switching

The device can switch between operating modes dynamically allowing the user to select the best performance/power ratio for the present task in hand. In this way microcontroller operations that do not require high performance can be executed using slower clocks thus requiring less operating current and prolonging battery life in portable applications.

In simple terms, mode switching between the NORMAL Mode and SLOW Mode is executed using the HLCLK bit and CKS2~CKS0 bits in the SMOD register while mode switching from the NORMAL/SLOW Modes to the SLEEP/IDLE Modes is executed via the HALT instruction. When an HALT instruction is executed, whether the device enters the IDLE Mode or the SLEEP Mode is determined by the condition of the IDLEN bit in the SMOD register and the FSYSON bit in the CTRL register.

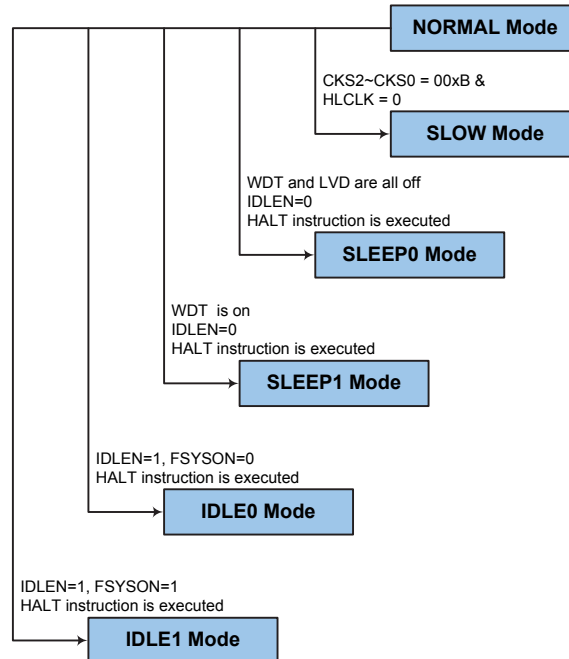
When the HLCLK bit switches to a low level, which implies that clock source is switched from the high speed clock, f_H , to the clock source, $f_H/2 \sim f_H/64$ or f_{SUB} . If the clock is from the f_{SUB} , the high speed clock source will stop running to conserve power. When this happens, it must be noted that the $f_H/16$ and $f_H/64$ internal clock sources will also stop running, which may affect the operation of other internal functions such as the TMs. The accompanying chart shows what happens when the device moves between the various operating modes.



NORMAL Mode to SLOW Mode Switching

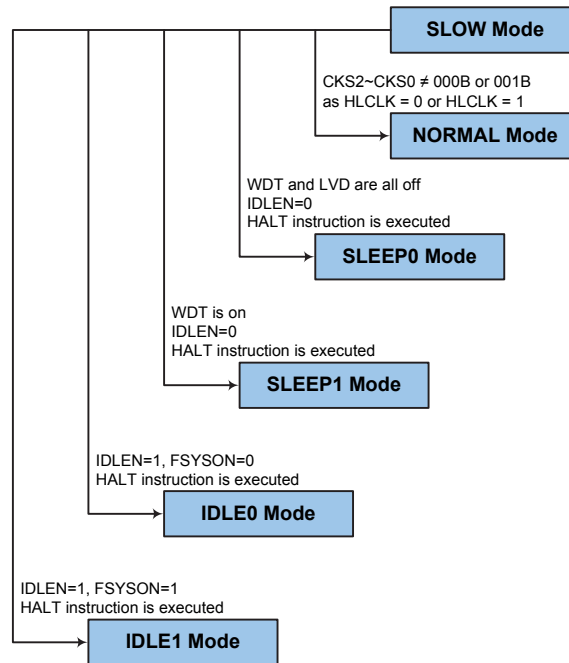
When running in the NORMAL Mode, which uses the high speed system oscillator, and therefore consumes more power, the system clock can switch to run in the SLOW Mode by setting the HLCLK bit to “0” and setting the CKS2~CKS0 bits to “000” or “001” in the SMOD register. This will then use the low speed system oscillator which will consume less power. Users may decide to do this for certain operations which do not require high performance and can subsequently reduce power consumption.

The SLOW Mode system clock is sourced from the LXT or LIRC oscillator determined by the configuration option and therefore requires the selected oscillator to be stable before full mode switching occurs. This is monitored using the LTO bit in the SMOD register.



SLOW Mode to NORMAL Mode Switching

In SLOW mode the system uses either the LXT or LIRC low speed system oscillator. To switch back to the NORMAL Mode, where the high speed system oscillator is used, the HLCLK bit should be set to “1” or HLCLK bit is “0”, but the CKS2~CKS0 field is set to “010”, “011”, “100”, “101”, “110” or “111”. As a certain amount of time will be required for high frequency clock to stabilise, the status of the HTO bit is checked. The amount of time required for high speed system oscillator stabilization depends upon which high speed system oscillator type is used.



Entering the SLEEP0 Mode

There is only one way for the device to enter the SLEEP0 Mode and that is to execute the “HALT” instruction in the application program with the IDLEN bit in the SMOD register equal to “0” and the WDT and LVD both off. When this instruction is executed under the conditions described above, the following will occur:

- The system clock, WDT clock and Time Base clock will be stopped and the application program will stop at the “HALT” instruction.
- The Data Memory contents and registers will maintain their present condition.
- The WDT will be cleared and stopped no matter if the WDT clock source originates from the LXT or LIRC oscillator.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag PDF will be set, and WDT timeout flag TO will be cleared.

Entering the SLEEP1 Mode

There is only one way for the device to enter the SLEEP1 Mode and that is to execute the “HALT” instruction in the application program with the IDLEN bit in the SMOD register equal to “0” and the WDT on. When this instruction is executed under the conditions described above, the following will occur:

- The system clock and Time Base clock will be stopped and the application program will stop at the “HALT” instruction, but the WDT will remain with the clock source coming from the f_{SUB} clock.
- The Data Memory contents and registers will maintain their present condition.
- The WDT will be cleared and resume counting if the WDT function is enabled.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag PDF will be set, and WDT timeout flag TO will be cleared.

Entering the IDLE0 Mode

There is only one way for the device to enter the IDLE0 Mode and that is to execute the “HALT” instruction in the application program with the IDLEN bit in the SMOD register equal to “1” and the FSYSN bit in the CTRL register equal to “0”. When this instruction is executed under the conditions described above, the following will occur:

- The system clock will be stopped and the application program will stop at the “HALT” instruction, but the f_{TBC} and f_{SUB} clocks will be on.
- The Data Memory contents and registers will maintain their present condition.
- The WDT will be cleared and resume counting if the WDT function is enabled.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag PDF will be set, and WDT timeout flag TO will be cleared.

Entering the IDLE1 Mode

There is only one way for the device to enter the IDLE1 Mode and that is to execute the “HALT” instruction in the application program with the IDLEN bit in the SMOD register equal to “1” and the FSYSOEN bit in the CTRL register equal to “1”. When this instruction is executed under the conditions described above, the following will occur:

- The system clock, f_{TBC} and f_{SUB} clocks will be on but the application program will stop at the “HALT” instruction.
- The Data Memory contents and registers will maintain their present condition.
- The WDT will be cleared and resume counting if the WDT function is enabled.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag PDF will be set, and WDT timeout flag TO will be cleared.

Standby Current Considerations

As the main reason for entering the SLEEP or IDLE Mode is to keep the current consumption of the device to as low a value as possible, perhaps only in the order of several micro-amps except in the IDLE1 Mode, there are other considerations which must also be taken into account by the circuit designer if the power consumption is to be minimised. Special attention must be made to the I/O pins on the device. All high-impedance input pins must be connected to either a fixed high or low level as any floating input pins could create internal oscillations and result in increased current consumption. These must either be set as outputs or if set as inputs must have pull-high resistors connected.

Care must also be taken with the loads, which are connected to I/O pins, which are set as outputs. These should be placed in a condition in which minimum current is drawn or connected only to external circuits that do not draw current, such as other CMOS inputs. Also note that additional standby current will also be required if the LXT or LIRC oscillator has enabled via configuration option.

In the IDLE1 Mode the system oscillator is on, if the peripheral function clock source is derived from the high speed oscillator, the additional standby current will also be perhaps in the order of several hundred micro-amps.

Wake-up

To minimise power consumption the device can enter the SLEEP or any IDLE Mode, where the CPU will be switched off. However, when the device is woken up again, it will take a considerable time for the original system oscillator to restart, stabilise and allow normal operation to resume.

After the system enters the SLEEP or IDLE Mode, it can be woken up from one of various sources listed as follows:

- An external falling edge on Port A
- A system interrupt
- A WDT overflow

When the device executes the “HALT” instruction, it will enter the IDLE or SLEEP mode and the PDF flag will be set to 1. The PDF flag will be cleared to 0 if the device experiences a system power-up or executes the clear Watchdog Timer instruction. If the system is woken up by a WDT overflow, a Watchdog Timer reset will be initiated and the TO flag will be set to 1. The TO flag is set if a WDT time-out occurs and causes a wake-up that only resets the Program Counter and Stack Pointer, other flags remain in their original status.

Each pin on Port A can be set using the PAWU register to permit a negative transition on the pin to wake up the system. When a Port A pin wake-up occurs, the program will resume execution at the instruction following the “HALT” instruction. If the system is woken up by an interrupt, then two possible situations may occur. The first is where the related interrupt is disabled or the interrupt is enabled but the stack is full, in which case the program will resume execution at the instruction following the “HALT” instruction. In this situation, the interrupt which woke up the device will not be immediately serviced, but will rather be serviced later when the related interrupt is finally enabled or when a stack level becomes free. The other situation is where the related interrupt is enabled and the stack is not full, in which case the regular interrupt response takes place. If an interrupt request flag is set high before entering the SLEEP or IDLE Mode, the wake-up function of the related interrupt will be disabled.

Programming Considerations

The high speed and low speed oscillators both use the same SST counter. For example, if the system is woken up from the SLEEP0 Mode and both the HIRC and LXT oscillators need to start-up from an off state. The LXT oscillator uses the SST counter after the HIRC oscillator has finished its SST period.

- If the device is woken up from the SLEEP0 Mode to the NORMAL Mode, the high speed system oscillator needs an SST period. The device will execute first instruction after HTO is “1”. At this time, the LXT oscillator may not be stability if f_{SUB} is from the LXT oscillator. The same situation occurs in the power-on state. The LXT oscillator is not ready yet when the first instruction is executed.
- If the device is woken up from the SLEEP1 Mode to NORMAL Mode, and the system clock source is from the HXT oscillator and FSTEN is “1”, the system clock can be switched to the LIRC oscillator after wake up.
- There are peripheral functions, such as WDT and TMs, for which the f_{SYS} is used. If the system clock source is switched from f_H to f_{SUB} , the clock source to the peripheral functions mentioned above will change accordingly.
- The on/off condition of f_{SUB} and f_S depends upon whether the WDT is enabled or disabled as the WDT clock source is selected from f_{SUB} .

Watchdog Timer

The Watchdog Timer is provided to prevent program malfunctions or sequences from jumping to unknown locations, due to certain uncontrollable external events such as electrical noise.

Watchdog Timer Clock Source

The Watchdog Timer clock source is provided by the internal clock, f_{SUB} , which is in turn supplied by the LIRC or LXT oscillator. The LXT oscillator is supplied by an external 32.768kHz crystal. The LIRC internal oscillator has an approximate frequency of 32kHz and this specified internal clock period can vary with V_{DD} , temperature and process variations. The Watchdog Timer source clock is then subdivided by a ratio of 2^8 to 2^{18} to give longer timeouts, the actual value being chosen using the WS2~WS0 bits in the WDTC register.

Watchdog Timer Control Register

A single register, WDTC, controls the required timeout period as well as the enable/disable operation. This register controls the overall operation of the Watchdog Timer.

• WDTC Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-----|-----|-----|-----|-----|-----|-----|-----|
| Name | WE4 | WE3 | WE2 | WE1 | WE0 | WS2 | WS1 | WS0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 1 | 0 | 1 | 0 | 0 | 1 | 1 |

Bit 7~3 **WE4~WE0**: WDT function software control

10101: Disable

01010: Enable

Other values: Reset MCU

When these bits are changed to any other values due to environmental noise the microcontroller will be reset; this reset operation will be activated after 2~3 LIRC clock cycles and the WRF bit in the CTRL register will be set high.

Bit 2~0 **WS2~WS0**: WDT time-out period selection

000: $2^8/f_{SUB}$

001: $2^{10}/f_{SUB}$

010: $2^{12}/f_{SUB}$

011: $2^{14}/f_{SUB}$

100: $2^{15}/f_{SUB}$

101: $2^{16}/f_{SUB}$

110: $2^{17}/f_{SUB}$

111: $2^{18}/f_{SUB}$

These three bits determine the division ratio of the Watchdog Timer source clock, which in turn determines the timeout period.

• **CTRL Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|--------|---|---|---|---|------|-----|-----|
| Name | FSYSON | — | — | — | — | LVRF | LRF | WRF |
| R/W | R/W | — | — | — | — | R/W | R/W | R/W |
| POR | 0 | — | — | — | — | x | 0 | 0 |

“x”: unknown

- Bit 7 **FSYSON**: f_{SYS} Control in IDLE Mode
Refer to the Operating Modes and System Clocks section.
- Bit 6~3 Unimplemented, read as “0”
- Bit 2 **LVRF**: LVR function reset flag
Refer to the Low Voltage Reset section.
- Bit 1 **LRF**: LVR control register software reset flag
Refer to the Low Voltage Reset section.
- Bit 0 **WRF**: WDT control register software reset flag
0: Not occurred
1: Occurred

This bit is set to 1 by the WDT control register software reset and cleared to zero by the application program. Note that this bit can only be cleared to zero by the application program.

Watchdog Timer Operation

The Watchdog Timer operates by providing a device reset when its timer overflows. This means that in the application program and during normal operation the user has to strategically clear the Watchdog Timer before it overflows to prevent the Watchdog Timer from executing a reset. This is done using the clear watchdog instruction. If the program malfunctions for whatever reason, jumps to an unknown location, or enters an endless loop, the clear instruction will not be executed in the correct manner, in which case the Watchdog Timer will overflow and reset the device. There are five bits, WE4~WE0, in the WDTC register to offer additional enable/disable and reset control of the Watchdog Timer. The WDT function will be disabled when the WE4~WE0 bits are set to a value of 10101B. The WDT function will be enabled if the WE4~WE0 bits value is equal to 01010B. If the WE4~WE0 bits are set to any other values, other than 01010B and 10101B, it will reset the device after 2~3 LIRC clock cycles. After power on these bits will have the value of 01010B.

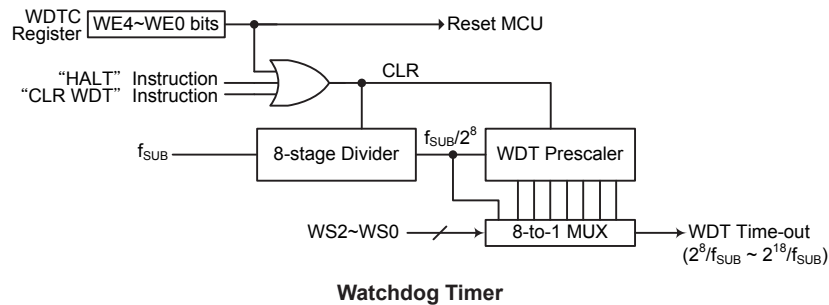
| WE4~WE0 Bits | WDT Function |
|-----------------|--------------|
| 10101B | Disable |
| 01010B | Enable |
| Any other value | Reset MCU |

Watchdog Timer Enable/Disable Control

Under normal program operation, a Watchdog Timer time-out will initialise a device reset and set the status bit TO. However, if the system is in the SLEEP or IDLE Mode, when a Watchdog Timer time-out occurs, the TO bit in the status register will be set and only the Program Counter and Stack Pointer will be reset. Three methods can be adopted to clear the contents of the Watchdog Timer. The first is a WDTC software reset, which means a certain value except 01010B and 10101B written into the WE4~WE0 bit filed, the second is using the Watchdog Timer software clear instruction and the third is via a HALT instruction.

There is only one method of using software instruction to clear the Watchdog Timer. That is to use the single “CLR WDT” instruction to clear the WDT.

The maximum time out period is when the 2^{18} division ratio is selected. As an example, with a 32kHz LIRC oscillator as its source clock, this will give a maximum watchdog period of around 8 seconds for the 2^{18} division ratio, and a minimum timeout of 8ms for the 2^8 division ratio.



Reset and Initialisation

A reset function is a fundamental part of any microcontroller ensuring that the device can be set to some predetermined condition irrespective of outside parameters. The most important reset condition is after power is first applied to the microcontroller. In this case, internal circuitry will ensure that the microcontroller, after a short delay, will be in a well-defined state and ready to execute the first program instruction. After this power-on reset, certain important internal registers will be set to defined states before the program commences. One of these registers is the Program Counter, which will be reset to zero forcing the microcontroller to begin program execution from the lowest Program Memory address.

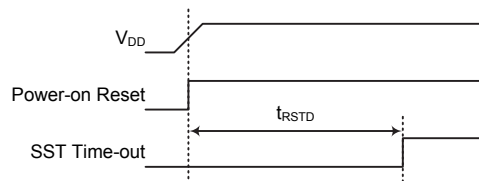
In addition to the power-on reset, another reset exists in the form of a Low Voltage Reset, LVR, where a full reset is implemented in situations where the power supply voltage falls below a certain threshold. Another type of reset is when the Watchdog Timer overflows and resets the microcontroller. All types of reset operations result in different register conditions being setup.

Reset Functions

There are several ways in which a microcontroller reset can occur, each of which will be described as follows.

Power-on Reset

The most fundamental and unavoidable reset is the one that occurs after power is first applied to the microcontroller. As well as ensuring that the Program Memory begins execution from the first memory address, a power-on reset also ensures that certain other registers are preset to known conditions. All the I/O port and port control registers will power up in a high condition ensuring that all pins will be first set to inputs.

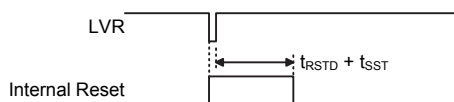


Note: t_{RSTD} is power-on delay with typical time=50ms.

Power-On Reset Timing Chart

Low Voltage Reset – LVR

The microcontroller contains a low voltage reset circuit in order to monitor the supply voltage of the device and provides an MCU reset should the value fall below a certain predefined level. The LVR function is always enabled with a specific LVR voltage V_{LVR} . If the supply voltage of the device drops to within a range of $0.9V \sim V_{LVR}$ such as might occur when changing the battery in battery powered applications, the LVR will automatically reset the device internally and the LVRF bit in the CTRL register will also be set high. For a valid LVR signal, a low supply voltage, i.e., a voltage in the range between $0.9V \sim V_{LVR}$ must exist for a time greater than that specified by t_{LVR} in the LVD/LVR characteristics. If the low supply voltage state does not exceed this value, the LVR will ignore the low supply voltage and will not perform a reset function. The actual V_{LVR} value can be selected by the LVS bits in the LVRC register. If the LVS7~LVS0 bits are changed to some different values by environmental noise, the LVR will reset the device after 2~3 LIRC clock cycles. When this happens, the LRF bit in the CTRL register will be set high. After power on the register will have the value of 01010101B. Note that the LVR function will be automatically disabled when the device enters the IDLE or SLEEP mode.



Note: t_{RSTD} is power-on delay with typical time=50ms.

Low Voltage Reset Timing Chart

• LVRC Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|------|------|------|------|------|------|------|------|
| Name | LVS7 | LVS6 | LVS5 | LVS4 | LVS3 | LVS2 | LVS1 | LVS0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 |

Bit 7~0 **LVS7~LVS0:** LVR voltage selection

01010101: 2.1V

00110011: 2.55V

10011001: 3.15V

10101010: 3.8V

Any other value: Generates MCU reset – register is reset to POR value

When an actual low voltage condition occurs, as specified by one of the four defined LVR voltage values above, an MCU reset will be generated. The reset operation will be activated after the low voltage condition keeps more than a t_{LVR} time. In this situation the register contents will remain the same after such a reset occurs.

Any register value, other than the four defined LVR values above, will also result in the generation of an MCU reset. The reset operation will be activated after 2~3 LIRC clock cycles. However in this situation the register contents will be reset to the POR value.

• CTRL Register

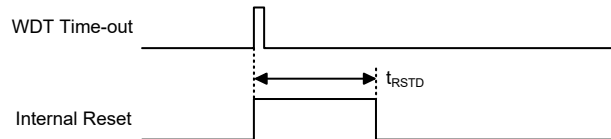
| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|--------|---|---|---|---|------|-----|-----|
| Name | FSYSON | — | — | — | — | LVRF | LRF | WRF |
| R/W | R/W | — | — | — | — | R/W | R/W | R/W |
| POR | 0 | — | — | — | — | x | 0 | 0 |

“x”: unknown

- Bit 7 **FSYSON**: f_{SYS} Control in IDLE Mode
Refer to the Operating Modes and System Clocks section.
- Bit 6~3 Unimplemented, read as “0”
- Bit 2 **LVRF**: LVR function reset flag
0: Not occur
1: Occurred
This bit is set high when a specific low voltage reset situation condition occurs. This bit can only be cleared to zero by the application program.
- Bit 1 **LRF**: LVR control register software reset flag
0: Not occur
1: Occurred
This bit is set high if the LVRC register contains any undefined LVR voltage register values. This in effect acts like a software-reset function. This bit can only be cleared to zero by the application program.
- Bit 0 **WRF**: WDT control register software reset flag
Refer to the Watchdog Timer Control Register section.

Watchdog Time-out Reset during Normal Operation

The Watchdog time-out Reset during normal operations in the NORMAL or SLOW mode is the same as the hardware Low Voltage Reset except that the Watchdog time-out flag TO will be set to “1”.

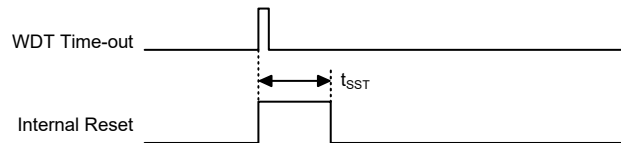


Note: t_{RSTD} is power-on delay with typical time=16.7ms.

WDT Time-out Reset during Normal Operation Timing Chart

Watchdog Time-out Reset during SLEEP or IDLE Mode

The Watchdog time-out Reset during SLEEP or IDLE Mode is a little different from other kinds of reset. Most of the conditions remain unchanged except that the Program Counter and the Stack Pointer will be cleared to “0” and the TO flag will be set to “1”. Refer to the A.C. Characteristics for t_{SST} details.



WDT Time-out Reset during SLEEP or IDLE Timing Chart

Reset Initial Conditions

The different types of reset described affect the reset flags in different ways. These flags, known as PDF and TO are located in the status register and are controlled by various microcontroller operations, such as the SLEEP or IDLE Mode function or Watchdog Timer. The reset flags are shown in the table:

| TO | PDF | Reset Conditions |
|----|-----|---|
| 0 | 0 | Power-on reset |
| u | u | LVR reset during NORMAL or SLOW Mode operation |
| 1 | u | WDT time-out reset during NORMAL or SLOW Mode operation |
| 1 | 1 | WDT time-out reset during IDLE or SLEEP Mode operation |

Note: "u" stands for unchanged

The following table indicates the way in which the various components of the microcontroller are affected after a power-on reset occurs.

| Item | Condition after Reset |
|--------------------|--|
| Program Counter | Reset to zero |
| Interrupts | All interrupts will be disabled |
| WDT, Time Bases | Cleared after reset, WDT begins counting |
| Timer Modules | Timer Module will be turned off |
| Input/Output Ports | I/O ports will be set as inputs |
| Stack Pointer | Stack Pointer will point to the top of the stack |

The different kinds of resets all affect the internal registers of the microcontroller in different ways. To ensure reliable continuation of normal program execution after a reset occurs, it is important to know what condition the microcontroller is in after a particular reset occurs. The following table describes how each type of reset affects each of the microcontroller internal registers. Note that as more than one package type exists, the table will reflect the situation for the larger package type.

| Register | Reset (Power On) | LVR Reset (Normal Operation) | WDT Time-out (Normal Operation) | WDT Time-out (IDLE or SLEEP) |
|----------|------------------|------------------------------|---------------------------------|------------------------------|
| IAR0 | xxxx xxxx | uuuu uuuu | xxxx xxxx | uuuu uuuu |
| MP0 | xxxx xxxx | xxxx xxxx | xxxx xxxx | uuuu uuuu |
| IAR1 | xxxx xxxx | uuuu uuuu | xxxx xxxx | uuuu uuuu |
| MP1 | xxxx xxxx | xxxx xxxx | xxxx xxxx | uuuu uuuu |
| BP | ---- --00 | ---- --00 | ---- --00 | ---- --uu |
| ACC | xxxx xxxx | uuuu uuuu | uuuu uuuu | uuuu uuuu |
| PCL | 0000 0000 | 0000 0000 | 0000 0000 | 0000 0000 |
| TBLP | xxxx xxxx | uuuu uuuu | uuuu uuuu | uuuu uuuu |
| TBLH | xxxx xxxx | uuuu uuuu | uuuu uuuu | uuuu uuuu |
| TBHP | ---x xxxx | ---u uuuu | ---u uuuu | ---u uuuu |
| STATUS | --00 xxxx | --uu uuuu | --1u uuuu | --11 uuuu |
| SMOD | 0000 0011 | 0000 0011 | 0000 0011 | uuuu uuuu |
| LVDC | --00 -000 | --00 -000 | --00 -000 | --uu -uuu |
| INTEG | ---- 0000 | ---- 0000 | ---- 0000 | ---- uuuu |
| INTC0 | -000 0000 | -000 0000 | -000 0000 | -uuu uuuu |
| INTC1 | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| INTC2 | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| MFIO | --00 --00 | --00 --00 | --00 --00 | --uu --uu |
| MF11 | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| MF12 | --00 --00 | --00 --00 | --00 --00 | --uu --uu |

| Register | Reset (Power On) | LVR Reset (Normal Operation) | WDT Time-out (Normal Operation) | WDT Time-out (IDLE or SLEEP) |
|----------|------------------|------------------------------|---------------------------------|------------------------------|
| PA | 1111 1111 | 1111 1111 | 1111 1111 | uuuu uuuu |
| PAC | 1111 1111 | 1111 1111 | 1111 1111 | uuuu uuuu |
| PAPU | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| PAWU | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| ACERH | ---- 0000 | ---- 0000 | ---- 0000 | ---- uuuu |
| TMPC | 0--- -000 | 0--- -000 | 0--- -000 | u--- -uuu |
| WDTC | 0101 0011 | 0101 0011 | 0101 0011 | uuuu uuuu |
| TBC | 0011 0111 | 0011 0111 | 0011 0111 | uuuu uuuu |
| CTRL | 0--- -x00 | 0--- -000 | 0--- -uuu | u--- -uuu |
| LVRC | 0101 0101 | uuuu uuuu | 0101 0101 | uuuu uuuu |
| EEA | -000 0000 | -000 0000 | -000 0000 | -uuu uuuu |
| EED | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| SADOL | xxxx ---- | xxxx ---- | xxxx ---- | uuuu ---- (ADRF=0) |
| | | | | uuuu uuuu (ADRF=1) |
| SADOH | xxxx xxxx | uuuu uuuu | uuuu uuuu | uuuu uuuu (ADRF=0) |
| | | | | ---- uuuu (ADRF=1) |
| SADC0 | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| SADC1 | 000- -000 | 000- -000 | 000- -000 | uuu- -uuu |
| SADC2 | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| PB | -111 1111 | -111 1111 | -111 1111 | -uuu uuuu |
| PBC | -111 1111 | -111 1111 | -111 1111 | -uuu uuuu |
| PBPU | -000 0000 | -000 0000 | -000 0000 | -uuu uuuu |
| CTMC0 | 0000 0--- | 0000 0--- | 0000 0--- | uuuu u--- |
| CTMC1 | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| CTMDL | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| CTMDH | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| CTMAL | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| CTMAH | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| CTMRP | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| STMC0 | 0000 0--- | 0000 0--- | 0000 0--- | uuuu u--- |
| STMC1 | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| STMDL | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| STMDH | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| STMAL | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| STMAH | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| STMAL | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| PTMC0 | 0000 0--- | 0000 0--- | 0000 0--- | uuuu u--- |
| PTMC1 | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| PTMDL | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| PTMDH | ----- --00 | ----- --00 | ----- --00 | ----- --uu |
| PTMAL | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| PTMAH | ----- --00 | ----- --00 | ----- --00 | ----- --uu |
| PTMRPL | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| PTMRPH | ----- --00 | ----- --00 | ----- --00 | ----- --uu |

| Register | Reset (Power On) | LVR Reset (Normal Operation) | WDT Time-out (Normal Operation) | WDT Time-out (IDLE or SLEEP) |
|------------|------------------|------------------------------|---------------------------------|------------------------------|
| CPC | 1000 0001 | 1000 0001 | 1000 0001 | uuuu uuuu |
| PC | -111 1111 | -111 1111 | -111 1111 | -uuu uuuu |
| PCC | -111 1111 | -111 1111 | -111 1111 | -uuu uuuu |
| PCPU | -000 0000 | -000 0000 | -000 0000 | -uuu uuuu |
| ACERL | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| SIMC0 | 111- 0000 | 111- 0000 | 111- 0000 | uuu- uuuu |
| SIMC1 | 1000 0001 | 1000 0001 | 1000 0001 | uuuu uuuu |
| SIMD | xxxx xxxx | xxxx xxxx | xxxx xxxx | uuuu uuuu |
| SIMA/SIMC2 | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| SIMTOC | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| SLCDC0 | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| SLCDC1 | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| SLCDC2 | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| SLCDC3 | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| SLCDC4 | ---- --00 | ---- --00 | ---- --00 | ---- --uu |
| SLEDC0 | 0101 0101 | 0101 0101 | 0101 0101 | uuuu uuuu |
| SLEDC1 | --01 0101 | --01 0101 | --01 0101 | --uu uuuu |
| IFS | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| PD | ---- 1111 | ---- 1111 | ---- 1111 | ---- uuuu |
| PDC | ---- 1111 | ---- 1111 | ---- 1111 | ---- uuuu |
| PDPU | ---- 0000 | ---- 0000 | ---- 0000 | ---- uuuu |
| USR | 0000 1011 | 0000 1011 | 0000 1011 | uuuu uuuu |
| UCR1 | 0000 00x0 | 0000 00x0 | 0000 00x0 | uuuu uuuu |
| UCR2 | 0000 0000 | 0000 0000 | 0000 0000 | uuuu uuuu |
| BRG | xxxx xxxx | xxxx xxxx | xxxx xxxx | uuuu uuuu |
| TXR_RXR | xxxx xxxx | xxxx xxxx | xxxx xxxx | uuuu uuuu |
| EEC | ---- 0000 | ---- 0000 | ---- 0000 | ---- uuuu |

Note: “u” stands for unchanged
“x” stands for unknown
“-” stands for unimplemented

Input/Output Ports

Holtek microcontrollers offer considerable flexibility on their I/O ports. With the input or output designation of every pin fully under user program control, pull-high selections for all ports and wake-up selections on certain pins, the user is provided with an I/O structure to meet the needs of a wide range of application possibilities.

The device provides bidirectional input/output lines labeled with port name PA~PD. These I/O ports are mapped to the RAM Data Memory with specific addresses as shown in the Special Purpose Data Memory table. All of these I/O ports can be used for input and output operations. For input operation, these ports are non-latching, which means the inputs must be ready at the T2 rising edge of instruction “MOV A, [m]”, where m denotes the port address. For output operation, all the data is latched and remains unchanged until the output latch is rewritten.

| Register Name | Bit | | | | | | | |
|---------------|-------|-------|-------|-------|-------|-------|-------|-------|
| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| PA | PA7 | PA6 | PA5 | PA4 | PA3 | PA2 | PA1 | PA0 |
| PAC | PAC7 | PAC6 | PAC5 | PAC4 | PAC3 | PAC2 | PAC1 | PAC0 |
| PAPU | PAPU7 | PAPU6 | PAPU5 | PAPU4 | PAPU3 | PAPU2 | PAPU1 | PAPU0 |
| PAWU | PAWU7 | PAWU6 | PAWU5 | PAWU4 | PAWU3 | PAWU2 | PAWU1 | PAWU0 |
| PB | — | PB6 | PB5 | PB4 | PB3 | PB2 | PB1 | PB0 |
| PBC | — | PBC6 | PBC5 | PBC4 | PBC3 | PBC2 | PBC1 | PBC0 |
| PBPU | — | PBPU6 | PBPU5 | PBPU4 | PBPU3 | PBPU2 | PBPU1 | PBPU0 |
| PC | — | PC6 | PC5 | PC4 | PC3 | PC2 | PC1 | PC0 |
| PCC | — | PCC6 | PCC5 | PCC4 | PCC3 | PCC2 | PCC1 | PCC0 |
| PCPU | — | PCPU6 | PCPU5 | PCPU4 | PCPU3 | PCPU2 | PCPU1 | PCPU0 |
| PD | — | — | — | — | PD3 | PCD2 | PD1 | PD0 |
| PDC | — | — | — | — | PDC3 | PDC2 | PDC1 | PDC0 |
| PDPU | — | — | — | — | PDPU3 | PDPU2 | PDPU1 | PDPU0 |

“—”: Unimplemented, read as “0”

I/O Logic Function Register List

Pull-high Resistors

Many product applications require pull-high resistors for their switch inputs usually requiring the use of an external resistor. To eliminate the need for these external resistors, all I/O pins, when configured as an input, except the \overline{SCS} input, have the capability of being connected to an internal pull-high resistor. These pull-high resistors are selected using the relevant pull-high control registers PAPU~PDPU, and are implemented using weak PMOS transistors.

• PxPU Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-------|-------|-------|-------|-------|-------|-------|-------|
| Name | PxPU7 | PxPU6 | PxPU5 | PxPU4 | PxPU3 | PxPU2 | PxPU1 | PxPU0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

PxPUn: I/O Port x Pin pull-high function control

0: Disable

1: Enable

The PxPUn bit is used to control the pin pull-high function. Here the “x” can be A, B, C or D. However, the actual available bits for each I/O Port may be different.

Port A Wake-up

The HALT instruction forces the microcontroller into the SLEEP or IDLE Mode which preserves power, a feature that is important for battery and other low-power applications. Various methods exist to wake-up the microcontroller, one of which is to change the logic condition on one of the Port A pins from high to low. This function is especially suitable for applications that can be woken up via external switches. Each pin on Port A can be selected individually to have this wake-up feature using the PAWU register.

• PAWU Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-------|-------|-------|-------|-------|-------|-------|-------|
| Name | PAWU7 | PAWU6 | PAWU5 | PAWU4 | PAWU3 | PAWU2 | PAWU1 | PAWU0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~0 **PAWU7~PAWU0**: PA7~PA0 wake-up function control
 0: Disable
 1: Enable

I/O Port Control Registers

Each I/O port has its own control register known as PAC~PDC, to control the input/output configuration. With this control register, each CMOS output or input can be reconfigured dynamically under software control. Each pin of the I/O ports is directly mapped to a bit in its associated port control register. For the I/O pin to function as an input, the corresponding bit of the control register must be written as a “1”. This will then allow the logic state of the input pin to be directly read by instructions. When the corresponding bit of the control register is written as a “0”, the I/O pin will be set as a CMOS output. If the pin is currently set as an output, instructions can still be used to read the output register. However, it should be noted that the program will in fact only read the status of the output data latch and not the actual logic status of the output pin.

• PxC Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|------|------|------|------|------|------|------|------|
| Name | PxC7 | PxC6 | PxC5 | PxC4 | PxC3 | PxC2 | PxC1 | PxC0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

PxCn: I/O Port x Pin type selection
 0: Output
 1: Input

The PxCn bit is used to control the pin type selection. Here the “x” can be A, B, C or D. However, the actual available bits for each I/O Port may be different.

I/O Port Source Current Control

This device supports different source current driving capability for each I/O port. With the corresponding selection register, SLEDC0 and SLEDC1, each I/O port can support four levels of the source current driving capability. Users should refer to the D.C. Characteristics section to select the desired source current for different applications.

| Register Name | Bit | | | | | | | |
|---------------|-------|-------|-------|-------|-------|-------|-------|-------|
| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| SLEDC0 | PBPS3 | PBPS2 | PBPS1 | PBPS0 | PAPS3 | PAPS2 | PAPS1 | PAPS0 |
| SLEDC1 | — | — | PDPS1 | PDPS0 | PCPS3 | PCPS2 | PCPS1 | PCPS0 |

I/O Port Source Current Selection Register List

• **SLEDC0 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-------|-------|-------|-------|-------|-------|-------|-------|
| Name | PBPS3 | PBPS2 | PBPS1 | PBPS0 | PAPS3 | PAPS2 | PAPS1 | PAPS0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 |

- Bit 7~6 **PBPS3~PBPS2**: PB6~PB4 source current selection
 00: Source current=Level 0 (Min.)
 01: Source current=Level 1
 10: Source current=Level 2
 11: Source current=Level 3 (Max.)
- Bit 5~4 **PBPS1~PBPS0**: PB3~PB0 source current selection
 00: Source current=Level 0 (Min.)
 01: Source current=Level 1
 10: Source current=Level 2
 11: Source current=Level 3 (Max.)
- Bit 3~2 **PAPS3~PAPS2**: PA7~PA4 source current selection
 00: Source current=Level 0 (Min.)
 01: Source current=Level 1
 10: Source current=Level 2
 11: Source current=Level 3 (Max.)
- Bit 1~0 **PAPS1~PAPS0**: PA3~PA0 source current selection
 00: Source current=Level 0 (Min.)
 01: Source current=Level 1
 10: Source current=Level 2
 11: Source current=Level 3 (Max.)

• **SLEDC1 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---|---|-------|-------|-------|-------|-------|-------|
| Name | — | — | PDPS1 | PDPS0 | PCPS3 | PCPS2 | PCPS1 | PCPS0 |
| R/W | — | — | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | — | — | 0 | 1 | 0 | 1 | 0 | 1 |

- Bit 7~6 Unimplemented, read as “0”
- Bit 5~4 **PDPS1~PDPS0**: PD3~PD0 source current selection
 00: Source current=Level 0 (Min.)
 01: Source current=Level 1
 10: Source current=Level 2
 11: Source current=Level 3 (Max.)
- Bit 3~2 **PCPS3~PCPS2**: PC6~PC4 source current selection
 00: Source current=Level 0 (Min.)
 01: Source current=Level 1
 10: Source current=Level 2
 11: Source current=Level 3 (Max.)
- Bit 1~0 **PCPS1~PCPS0**: PC3~PC0 source current selection
 00: Source current=Level 0 (Min.)
 01: Source current=Level 1
 10: Source current=Level 2
 11: Source current=Level 3 (Max.)

Pin-remapping Functions

The flexibility of the microcontroller range is greatly enhanced by the use of pins that have more than one function. Limited numbers of pins can force serious design constraints on designers but by supplying pins with multi-functions, many of these difficulties can be overcome. The way in which the pin function of each pin is selected is different for each function and a priority order is established where more than one pin function is selected simultaneously. Additionally there is a register, IFS, to establish certain pin functions.

The limited number of supplied pins in a package can impose restrictions on the amount of functions a certain device can contain. However by allowing the same pins to share several different functions and providing a means of function selection, a wide range of different functions can be incorporated into even relatively small package sizes. If the pin-shared pin function have multiple outputs simultaneously, its pin names at the right side of the “/” sign can be used for higher priority.

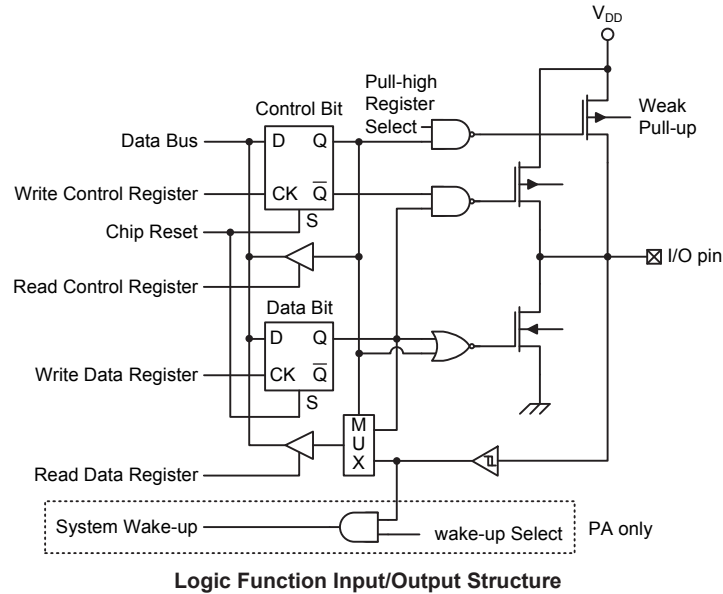
• IFS Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-------|--------|--------|-----------|-----------|--------|------|------|
| Name | RXCTL | SDOPS1 | SDOPS0 | SDI_SDAPS | SCK_SCLPS | SCSBPS | TXPS | RXPS |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

- Bit 7 **RXCTL**: RX enable control
0: Disable
1: Enable
- Bit 6~5 **SDOPS1~SDOPS0**: SDO pin-remapping selection
00: SDO on PC3
01: SDO on PA1
10: Undefined
11: SDO on PC2
- Bit 4 **SDI_SDAPS**: SDI/SDA pin-remapping selection
0: SDI/SDA on PC4
1: SDI/SDA on PA3
- Bit 3 **SCK_SCLPS**: SCK/SCL pin-remapping selection
0: SCK/SCL on PC5
1: SCK/SCL on PB6
- Bit 2 **SCSBPS**: \overline{SCS} pin-remapping selection
0: \overline{SCS} on PC6
1: \overline{SCS} on PB5
- Bit 1 **TXPS**: TX pin-remapping selection
0: TX on PD2
1: TX on PB3
- Bit 0 **RXPS**: RX pin-remapping selection
0: RX on PD1
1: RX on PB4

I/O Pin Structures

The accompanying diagram illustrates the internal structure of the I/O logic function. As the exact logical construction of the I/O pin will differ from this drawing, it is supplied as a guide only to assist with the functional understanding of the I/O logic function. The wide range of pin-shared structures does not permit all types to be shown.



Programming Considerations

Within the user program, one of the first things to consider is port initialisation. After a reset, all of the I/O data and port control registers will be set high. This means that all I/O pins will default to an input state, the level of which depends on the other connected circuitry and whether pull-high selections have been chosen. If the port control registers are then programmed to set some pins as outputs, these output pins will have an initial high output value unless the associated port data registers are first programmed. Selecting which pins are inputs and which are outputs can be achieved byte-wide by loading the correct values into the appropriate port control register or by programming individual bits in the port control register using the "SET [m].i" and "CLR [m].i" instructions. Note that when using these bit control instructions, a read-modify-write operation takes place. The microcontroller must first read in the data on the entire port, modify it to the required new bit values and then rewrite this data back to the output ports.

Port A has the additional capability of providing wake-up functions. When the device is in the SLEEP or IDLE Mode, various methods are available to wake the device up. One of these is a high to low transition of any of the Port A pins. Single or multiple pins on Port A can be set to have this function.

Timer Modules – TM

One of the most fundamental functions in any microcontroller device is the ability to control and measure time. To implement time related functions this device includes a Timer Module, abbreviated to the name TM. The TM is a multi-purpose timing unit and serves to provide operations such as Timer/Counter, Compare Match Output as well as being the functional unit for the generation of PWM signals. The TM has two individual interrupts. The addition of input and output pins for the TM ensures that users are provided with timing units with a wide and flexible range of features.

The common features of the different TM types are described here with more detailed information provided in the individual Compact, Standard and Periodic TM sections.

Introduction

The device contains three TMs and each individual TM can be categorised as a certain type, namely Compact Type TM, Standard Type TM or Periodic Type TM. Although similar in nature, the different TM types vary in their feature complexity. The common features to all of the Compact, Standard and Periodic TMs will be described in this section and the detailed operation regarding each of the TM types will be described in separate sections. The main features and differences between the three types of TMs are summarised in the accompanying table.

| TM Function | CTM | STM | PTM |
|------------------------------|----------------|----------------|----------------|
| Timer/Counter | √ | √ | √ |
| Input Capture | — | √ | √ |
| Compare Match Output | √ | √ | √ |
| PWM Channels | 1 | 1 | 1 |
| Single Pulse Output | — | 1 | 1 |
| PWM Alignment | Edge | Edge | Edge |
| PWM Adjustment Period & Duty | Duty or Period | Duty or Period | Duty or Period |

TM Function Summary

TM Operation

The different types of TM offer a diverse range of functions, from simple timing operations to PWM signal generation. The key to understanding how the TM operates is to see it in terms of a free running count-up counter whose value is then compared with the value of pre-programmed internal comparators. When the free running count-up counter has the same value as the pre-programmed comparator, known as a compare match situation, a TM interrupt signal will be generated which can clear the counter and perhaps also change the condition of the TM output pin. The internal TM counter is driven by a user selectable clock source, which can be an internal clock or an external pin.

TM Clock Source

The clock source which drives the main counter in each TM can originate from various sources. The selection of the required clock source is implemented using the xTCK2~xTCK0 bits in the xTM control registers, where “x” stands for C, S or P type TM. The clock source can be a ratio of the system clock, f_{SYS} , or the internal high clock, f_H , the f_{SUB} clock source or the external xTCK pin. The xTCK pin clock source is used to allow an external signal to drive the TM as an external clock source for event counting.

TM Interrupts

The Compact Type, Standard Type and Periodic Type TMs each have two internal interrupts, one for each of the internal comparator A or comparator P, which generate a TM interrupt when a compare match condition occurs. When a TM interrupt is generated it can be used to clear the counter and also to change the state of the TM output pin.

TM External Pins

Each of the TMs, irrespective of what type, has one TM input pin, with the label xTCK. The xTM input pin, xTCK, is essentially a clock source for the xTM and is selected using the xTCK2~xTCK0 bits in the xTMC0 register. This external TM input pin allows an external clock source to drive the internal TM. The xTCK input pin can be chosen to have either a rising or falling active edge. The STCK and PTCK pins are also used as the external trigger input pin in single pulse output mode for the STM and PTM respectively.

The TMs each have one output pin, xTP. When the TM is in the Compare Match Output Mode, these pins can be controlled by the TM to switch to a high or low level or to toggle when a compare match situation occurs. The external xTP output pin is also the pin where the TM generates the PWM output waveform. The xTP pin acts as an input when the TM is setup to operate in the Capture Input Mode.

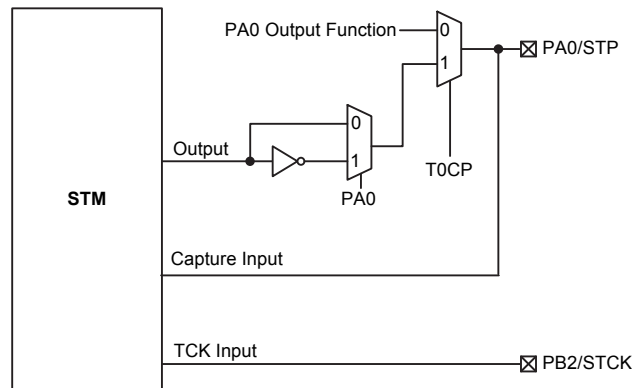
As the xTP pins are pin-shared with other functions, the xTP pin function is enabled or disabled according to the internal TM on/off control, operation mode and output control settings. When the corresponding TM configuration selects the xTP pin to be used as an output pin, the associated pin will be setup as an external TM output pin. If the TM configuration selects the xTP pin to be setup as an input pin, the input signal supplied on the associated pin can be derived from an external signal and other pin-shared output function. If the TM configuration determines that the xTP pin function is not used, the associated pin will be controlled by other pin-shared functions. The details of the external pins for each TM type and device are provided in the accompanying table.

| STM | PTM | CTM | Register |
|------------|-----------|-----------|-------------|
| STCK0; STP | PTCK; PTP | CTCK; CTP | xTMC0; TMPC |

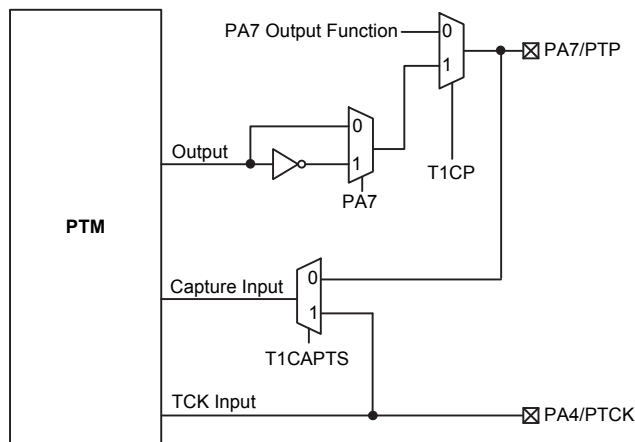
TM External Pins

TM Input/Output Pin Control Register

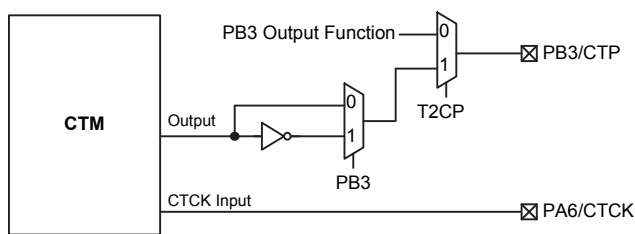
Selecting to have a TM input/output or whether to retain its other shared function is implemented using one register, with a single bit in the register corresponding to a TM input/output pin. Setting the bit high will setup the corresponding pin as a TM input/output, if reset to zero the pin will return its original other functions.



STM Function Pin Control Block Diagram



PTM Function Pin Control Block Diagram



CTM Function Pin Control Block Diagram

- Note: 1. The I/O register data bits shown are used for TM output inversion control.
2. In the Capture Input Mode, the TM pin control register must never enable more than one TM input.

• **TMPC Register**

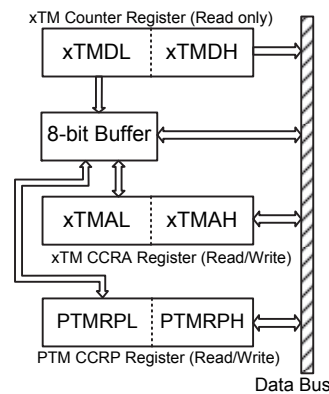
| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|------|---|---|---|---|------|------|------|
| Name | CLOP | — | — | — | — | T2CP | T1CP | T0CP |
| R/W | R/W | — | — | — | — | R/W | R/W | R/W |
| POR | 0 | — | — | — | — | 0 | 0 | 0 |

- Bit 7 **CLOP**: CLO pin control
0: Disable
1: Enable
- Bit 6~3 Unimplemented, read as “0”
- Bit 2 **T2CP**: CTP pin control
0: Disable
1: Enable
- Bit 1 **T1CP**: PTP pin control
0: Disable
1: Enable
- Bit 0 **T0CP**: STP pin control
0: Disable
1: Enable

Programming Considerations

The TM Counter Registers and the Capture/Compare CCRA and CCRP registers, all have a low and high byte structure. The high bytes can be directly accessed, but as the low bytes can only be accessed via an internal 8-bit buffer, reading or writing to these register pairs must be carried out in a specific way. The important point to note is that data transfer to and from the 8-bit buffer and its related low byte only takes place when a write or read operation to its corresponding high byte is executed.

As the CCRA and CCRP registers are implemented in the way shown in the following diagram and accessing these register pairs is carried out in a specific way as described above, it is recommended to use the “MOV” instruction to access the CCRA and CCRP low byte registers, named xTMAL and PTMRPL, using the following access procedures. Accessing the CCRA or CCRP low byte registers without following these access procedures will result in unpredictable values.



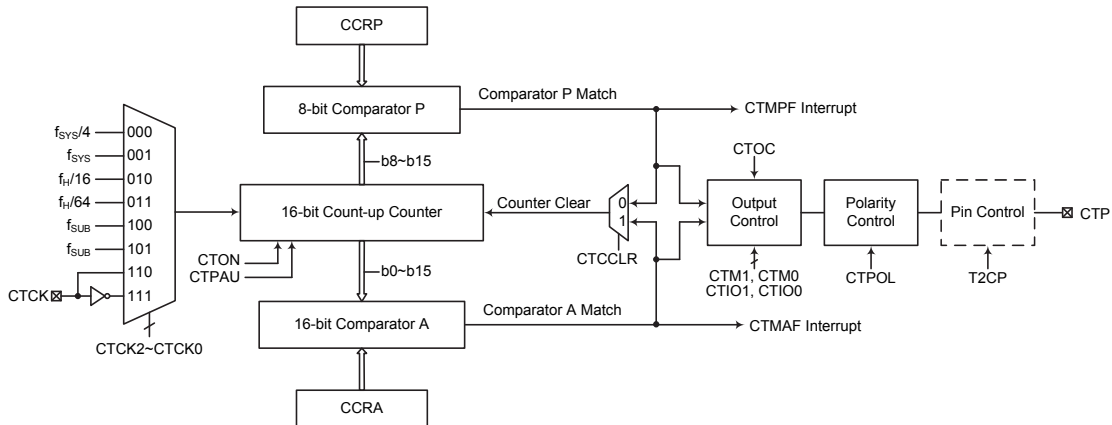
The following steps show the read and write procedures:

- Writing Data to CCRA or CCRP
 - ♦ Step 1. Write data to Low Byte xTMAL or PTMRPL
 - Note that here data is only written to the 8-bit buffer.
 - ♦ Step 2. Write data to High Byte xTMAH or PTMRPH
 - Here data is written directly to the high byte registers and simultaneously data is latched from the 8-bit buffer to the Low Byte registers.
- Reading Data from the Counter Registers and CCRA or CCRP
 - ♦ Step 1. Read data from the High Byte xTMDH, xTMAH or PTMRPH
 - Here data is read directly from the High Byte registers and simultaneously data is latched from the Low Byte register into the 8-bit buffer.
 - ♦ Step 2. Read data from the Low Byte xTMDL, xTMAL or PTMRPL
 - This step reads data from the 8-bit buffer.

Compact Type TM – CTM

Although the simplest form of the three TM types, the Compact TM type still contains three operating modes, which are Compare Match Output, Timer/Event Counter and PWM Output modes. The Compact TM can also be controlled with an external input pin and can drive one external output pin.

| CTM Core | CTM Input Pin | CTM Output Pin |
|------------|---------------|----------------|
| 16-bit CTM | CTCK | CTP |



16-bit Compact Type TM Block Diagram

Compact Type TM Operation

At its core is a 16-bit count-up counter which is driven by a user selectable internal or external clock source. There are also two internal comparators with the names, Comparator A and Comparator P. These comparators will compare the value in the counter with CCRP and CCRA registers. The CCRP is 8-bit wide whose value is compared with the highest eight bits in the counter while the CCRA is 16-bit wide and therefore compares with all counter bits.

The only way of changing the value of the 16-bit counter using the application program, is to clear the counter by changing the CTON bit from low to high. The counter will also be cleared automatically by a counter overflow or a compare match with one of its associated comparators. When these conditions occur, a CTM interrupt signal will also usually be generated. The Compact Type TM can operate in a number of different operational modes, can be driven by different clock sources including an input pin and can also control the output pin. All operating setup conditions are selected using relevant internal registers.

Compact Type TM Register Description

Overall operation of the Compact TM is controlled using a series of registers. A read only register pair exists to store the internal counter 16-bit value, while a read/write register pair exists to store the internal 16-bit CCRA value. The CTMRP register is used to store the 8-bit CCRP value. The remaining two registers are control registers which setup the different operating and control modes.

| Register Name | Bit | | | | | | | |
|---------------|-------|-------|-------|-------|-------|-------|-------|--------|
| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| CTMC0 | CTPAU | CTCK2 | CTCK1 | CTCK0 | CTON | — | — | — |
| CTMC1 | CTM1 | CTM0 | CTIO1 | CTIO0 | CTOC | CTPOL | CTDPX | CTCCLR |
| CTMDL | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| CTMDH | D15 | D14 | D13 | D12 | D11 | D10 | D9 | D8 |
| CTMAL | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| CTMAH | D15 | D14 | D13 | D12 | D11 | D10 | D9 | D8 |
| CTMRP | CTRP7 | CTRP6 | CTRP5 | CTRP4 | CTRP3 | CTRP2 | CTRP1 | CTRP0 |

16-bit Compact TM Register List

• CTMC0 Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-------|-------|-------|-------|------|---|---|---|
| Name | CTPAU | CTCK2 | CTCK1 | CTCK0 | CTON | — | — | — |
| R/W | R/W | R/W | R/W | R/W | R/W | — | — | — |
| POR | 0 | 0 | 0 | 0 | 0 | — | — | — |

Bit 7 **CTPAU**: CTM counter pause control

0: Run
1: Pause

The counter can be paused by setting this bit high. Clearing the bit to zero restores normal counter operation. When in a Pause condition the CTM will remain powered up and continue to consume power. The counter will retain its residual value when this bit changes from low to high and resume counting from this value when the bit changes to a low value again.

Bit 6~4 **CTCK2~CTCK0**: CTM counter clock selection

000: $f_{SYS}/4$
001: f_{SYS}
010: $f_H/16$
011: $f_H/64$
100: f_{SUB}
101: f_{SUB}
110: CTCK rising edge clock
111: CTCK falling edge clock

These three bits are used to select the clock source for the CTM. The external pin clock source can be chosen to be active on the rising or falling edge. The clock source f_{SYS} is the system clock, while f_H and f_{SUB} are other internal clocks, the details of which can be found in the oscillator section.

Bit 3 **CTON**: CTM counter on/off control

0: Off
1: On

This bit controls the overall on/off function of the CTM. Setting the bit high enables the counter to run while clearing the bit disables the CTM. Clearing this bit to zero will stop the counter from counting and turn off the CTM which will reduce its power consumption. When the bit changes state from low to high the internal counter value will be reset to zero, however when the bit changes from high to low, the internal counter will retain its residual value until the bit returns high again. If the CTM is in

the Compare Match Output Mode or the PWM Output Mode then the CTM output pin will be reset to its initial condition, as specified by the CTOC bit, when the CTON bit changes from low to high.

Bit 2~0 Unimplemented, read as “0”

• **CTMC1 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|------|------|-------|-------|------|-------|-------|--------|
| Name | CTM1 | CTM0 | CTIO1 | CTIO0 | CTOC | CTPOL | CTDPX | CTCCLR |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~6 **CTM1~CTM0**: CTM operating mode selection

- 00: Compare Match Output Mode
- 01: Undefined
- 10: PWM Output Mode
- 11: Timer/Counter Mode

These bits set the required operating mode for the CTM. To ensure reliable operation the CTM should be switched off before any changes are made to the CTM1 and CTM0 bits. In the Timer/Counter Mode, the CTM output pin control must be disabled.

Bit 5~4 **CTIO1~CTIO0**: CTM output function selection

Compare Match Output Mode

- 00: No change
- 01: Output low
- 10: Output high
- 11: Toggle output

PWM Output Mode

- 00: PWM output inactive state
- 01: PWM output active state
- 10: PWM output
- 11: Undefined

Timer/Counter Mode

Unused

These two bits are used to determine how the CTM output pin changes state when a certain condition is reached. The function that these bits select depends upon in which mode the CTM is running.

In the Compare Match Output Mode, the CTIO1 and CTIO0 bits determine how the CTM output pin changes state when a compare match occurs from the Comparator A. The CTM output pin can be set to switch high, switch low or to toggle its present state when a compare match occurs from the Comparator A. When the bits are both zero, then no change will take place on the output. The initial value of the CTM output pin should be configured using the CTOC bit in the CTMC1 register. Note that the output level requested by the CTIO1 and CTIO0 bits must be different from the initial value setup using the CTOC bit otherwise no change will occur on the CTM output pin when a compare match occurs. After the CTM output pin changes state, it can be reset to its initial level by changing the level of the CTON bit from low to high.

In the PWM Output Mode, the CTIO1 and CTIO0 bits determine how the CTM output pin changes state when a certain compare match condition occurs. The PWM output function is modified by changing these two bits. It is necessary to only change the values of the CTIO1 and CTIO0 bits only after the CTM has been switched off. Unpredictable PWM outputs will occur if the CTIO1 and CTIO0 bits are changed when the CTM is running.

- Bit 3** **CTOC:** CTM CTP output control
 Compare Match Output Mode
 0: Initial low
 1: Initial high
 PWM Output Mode
 0: Active low
 1: Active high
- This is the output control bit for the CTM output pin. Its operation depends upon whether CTM is being used in the Compare Match Output Mode or in the PWM Output Mode. It has no effect if the CTM is in the Timer/Counter Mode. In the Compare Match Output Mode it determines the logic level of the CTM output pin before a compare match occurs. In the PWM Output Mode it determines if the PWM signal is active high or active low.
- Bit 2** **CTPOL:** CTM CTP output polarity control
 0: Non-invert
 1: Invert
- This bit controls the polarity of the CTP output pin. When the bit is set high the CTM output pin will be inverted and not inverted when the bit is zero. It has no effect if the CTM is in the Timer/Counter Mode.
- Bit 1** **CTDPX:** CTM PWM duty/period control
 0: CCRP – period; CCRA – duty
 1: CCRP – duty; CCRA – period
- This bit determines which of the CCRA and CCRP registers are used for period and duty control of the PWM waveform.
- Bit 0** **CTCCLR:** CTM counter clear condition selection
 0: CTM Comparator P match
 1: CTM Comparator A match
- This bit is used to select the method which clears the counter. Remember that the Compact TM contains two comparators, Comparator A and Comparator P, either of which can be selected to clear the internal counter. With the CTCCLR bit set high, the counter will be cleared when a compare match occurs from the Comparator A. When the bit is low, the counter will be cleared when a compare match occurs from the Comparator P or with a counter overflow. A counter overflow clearing method can only be implemented if the CCRP bits are all cleared to zero. The CTCCLR bit is not used in the PWM Output Mode.

• **CTMDL Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|----|----|----|----|----|----|----|----|
| Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| R/W | R | R | R | R | R | R | R | R |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~0 **D7~D0:** CTM Counter Low Byte Register bit 7 ~ bit 0
 CTM 16-bit Counter bit 7 ~ bit 0

• **CTMDH Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-----|-----|-----|-----|-----|-----|----|----|
| Name | D15 | D14 | D13 | D12 | D11 | D10 | D9 | D8 |
| R/W | R | R | R | R | R | R | R | R |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~0 **D15~D8:** CTM Counter High Byte Register bit 7 ~ bit 0
 CTM 16-bit Counter bit 15 ~ bit 8

• **CTMAL Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-----|-----|-----|-----|-----|-----|-----|-----|
| Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~0 **D7~D0**: CTM CCRA Low Byte Register bit 7 ~ bit 0
CTM 16-bit CCRA bit 7 ~ bit 0

• **CTMAH Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-----|-----|-----|-----|-----|-----|-----|-----|
| Name | D15 | D14 | D13 | D12 | D11 | D10 | D9 | D8 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~0 **D15~D8**: CTM CCRA High Byte Register bit 7 ~ bit 0
CTM 16-bit CCRA bit 15 ~ bit 8

• **CTMRP Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-------|-------|-------|-------|-------|-------|-------|-------|
| Name | CTRP7 | CTRP6 | CTRP5 | CTRP4 | CTRP3 | CTRP2 | CTRP1 | CTRP0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~0 **CTRP7~CTRP0**: CTM CCRP 8-bit register, compared with the CTM Counter bit 15 ~ bit 8
Comparator P Match Period =
0: 65536 CTM clocks
1~255: $256 \times (1\sim255)$ CTM clocks

These eight bits are used to setup the value on the internal CCRP 8-bit register, which are then compared with the internal counter's highest eight bits. The result of this comparison can be selected to clear the internal counter if the CTCCLR bit is set to zero. Setting the CTCCLR bit to zero ensures that a compare match with the CCRP values will reset the internal counter. As the CCRP bits are only compared with the highest eight counter bits, the compare values exist in 256 clock cycle multiples. Clearing all eight bits to zero is in effect allowing the counter to overflow at its maximum value.

Compact Type TM Operating Modes

The Compact Type TM can operate in one of three operating modes, Compare Match Output Mode, PWM Output Mode or Timer/Counter Mode. The operating mode is selected using the CTM1 and CTM0 bits in the CTMC1 register.

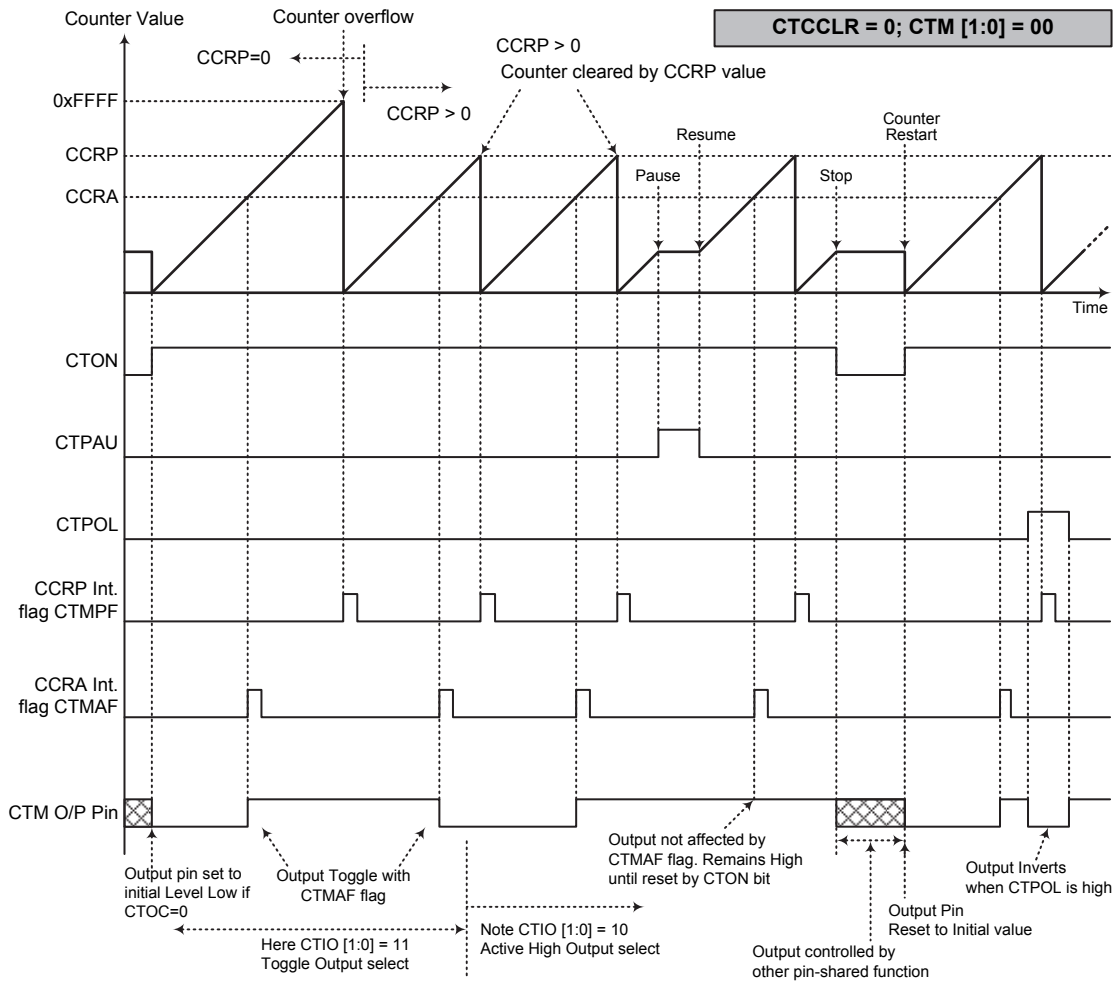
Compare Match Output Mode

To select this mode, bits CTM1 and CTM0 in the CTMC1 register, should be set to "00" respectively. In this mode once the counter is enabled and running it can be cleared by three methods. These are a counter overflow, a compare match from Comparator A and a compare match from Comparator P. When the CTCCLR bit is low, there are two ways in which the counter can be cleared. One is when a compare match occurs from Comparator P, the other is when the CCRP bits are all zero which allows the counter to overflow. Here both CTMAF and CTMPF interrupt request flags for the Comparator A and Comparator P respectively, will both be generated.

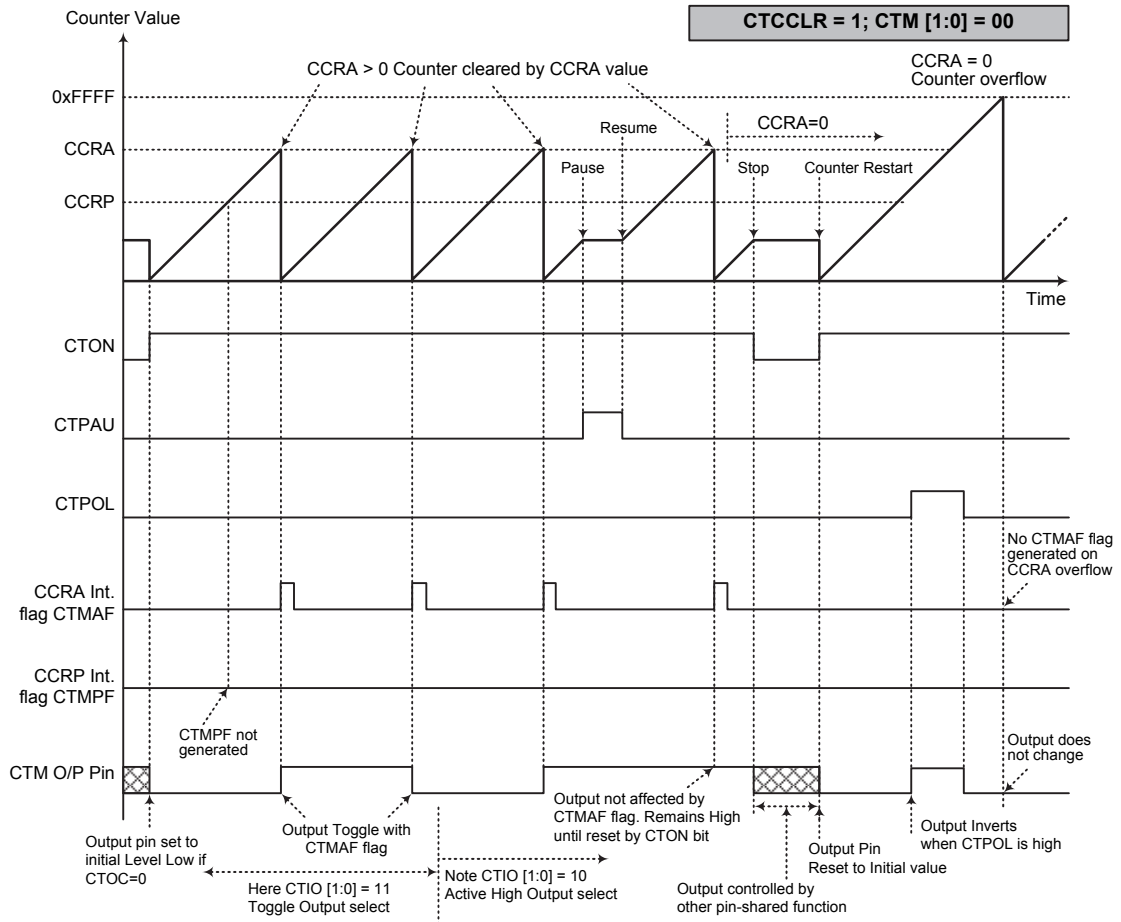
If the CTCCLR bit in the CTMC1 register is high then the counter will be cleared when a compare match occurs from Comparator A. However, here only the CTMAF interrupt request flag will be generated even if the value of the CCRP bits is less than that of the CCRA registers. Therefore when

CTCCLR is high no CTMPF interrupt request flag will be generated. If the CCRA bits are all zero, the counter will overflow when it reaches its maximum 16-bit, FFFF Hex, value, however here the CTMAF interrupt request flag will not be generated.

As the name of the mode suggests, after a comparison is made, the CTM output pin will change state. The CTM output pin condition however only changes state when a CTMAF interrupt request flag is generated after a compare match occurs from Comparator A. The CTMPF interrupt request flag, generated from a compare match occurs from Comparator P, will have no effect on the CTM output pin. The way in which the CTM output pin changes state are determined by the condition of the CTIO1 and CTIO0 bits in the CTMC1 register. The CTM output pin can be selected using the CTIO1 and CTIO0 bits to go high, to go low or to toggle from its present condition when a compare match occurs from Comparator A. The initial condition of the CTM output pin, which is setup after the CTON bit changes from low to high, is setup using the CTOC bit. Note that if the CTIO1 and CTIO0 bits are zero then no pin change will take place.



- Note: 1. With CTCCLR=0, a Comparator P match will clear the counter
 2. The CTM output pin is controlled only by the CTMAF flag
 3. The output pin is reset to its initial state by a CTON bit rising edge



Compare Match Output Mode – CTCCLR=1

- Note: 1. With CTCCLR=1, a Comparator A match will clear the counter
 2. The CTM output pin is controlled only by the CTMAF flag
 3. The output pin is reset to its initial state by a CTON bit rising edge
 4. The CTMPF flag is not generated when CTCCLR=1

Timer/Counter Mode

To select this mode, bits CTM1 and CTM0 in the CTMC1 register should be set to “11” respectively. The Timer/Counter Mode operates in an identical way to the Compare Match Output Mode generating the same interrupt flags. The exception is that in the Timer/Counter Mode the CTM output pin is not used. Therefore the above description and Timing Diagrams for the Compare Match Output Mode can be used to understand its function. As the CTM output pin is not used in this mode, the pin can be used as a normal I/O pin or other pin-shared functions.

PWM Output Mode

To select this mode, bits CTM1 and CTM0 in the CTMC1 register should be set to “10” respectively. The PWM function within the CTM is useful for applications which require functions such as motor control, heating control, illumination control etc. By providing a signal of fixed frequency but of varying duty cycle on the CTM output pin, a square wave AC waveform can be generated with varying equivalent DC RMS values.

As both the period and duty cycle of the PWM waveform can be controlled, the choice of generated waveform is extremely flexible. In the PWM Output Mode, the CTCCLR bit has no effect on the PWM operation. Both of the CCRA and CCRP registers are used to generate the PWM waveform, one register is used to clear the internal counter and thus control the PWM waveform frequency, while the other one is used to control the duty cycle. Which register is used to control either frequency or duty cycle is determined using the CTD PX bit in the CTMC1 register. The PWM waveform frequency and duty cycle can therefore be controlled by the values in the CCRA and CCRP registers.

An interrupt flag, one for each of the CCRA and CCRP, will be generated when a compare match occurs from either Comparator A or Comparator P. The CTOC bit in the CTMC1 register is used to select the required polarity of the PWM waveform while the two CTIO1 and CTIO0 bits are used to enable the PWM output or to force the CTM output pin to a fixed high or low level. The CTPOL bit is used to reverse the polarity of the PWM output waveform.

• **16-bit CTM, PWM Output Mode, Edge-aligned Mode, CTD PX=0**

| CCRP | 1~255 | 0 |
|--------|----------|-------|
| Period | CCRP×256 | 65536 |
| Duty | CCRA | |

If $f_{SYS}=8\text{MHz}$, CTM clock source is $f_{SYS}/4$, CCRP=2, CCRA=128,

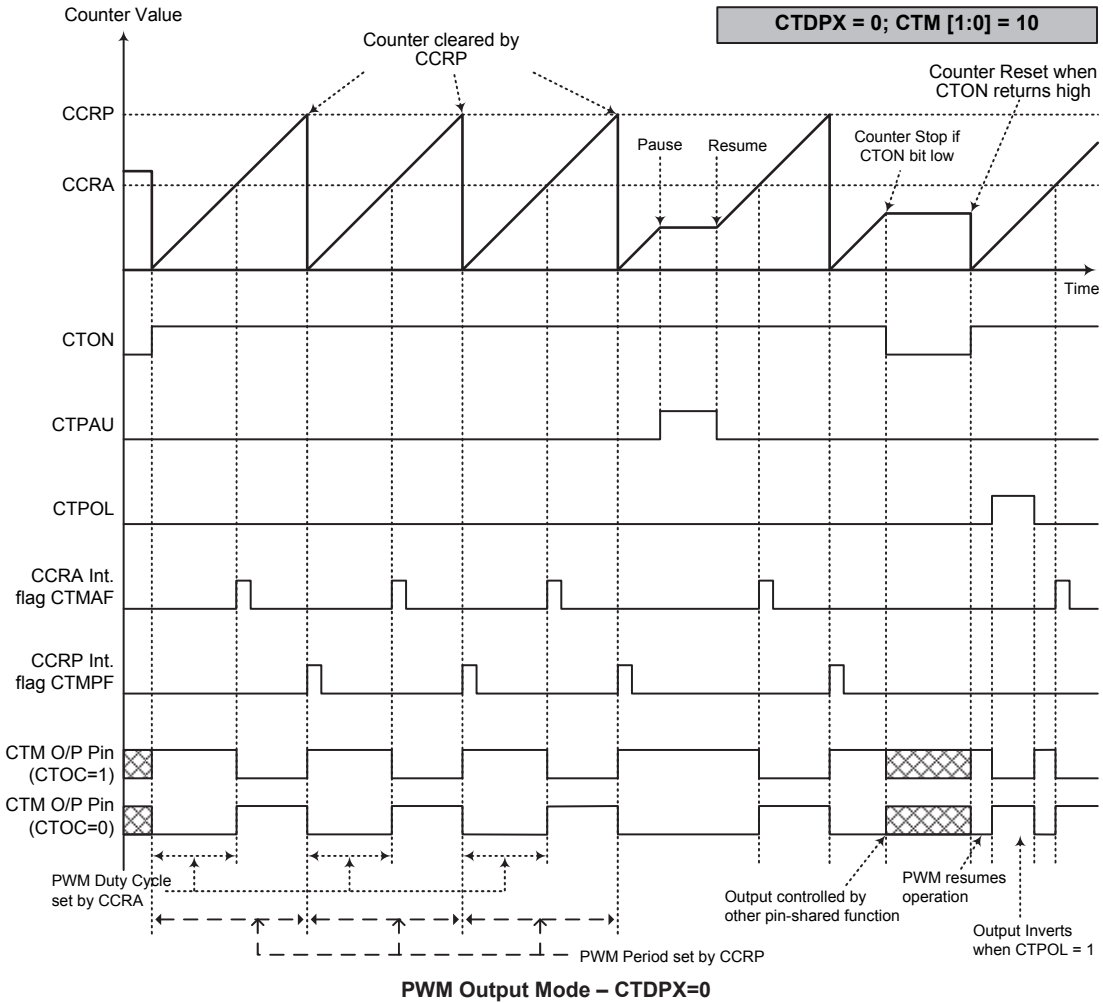
The CTM PWM output frequency= $(f_{SYS}/4)/512=f_{SYS}/2048\approx 4\text{kHz}$, duty=128/512=25%.

If the Duty value defined by the CCRA register is equal to or greater than the Period value, then the PWM output duty is 100%.

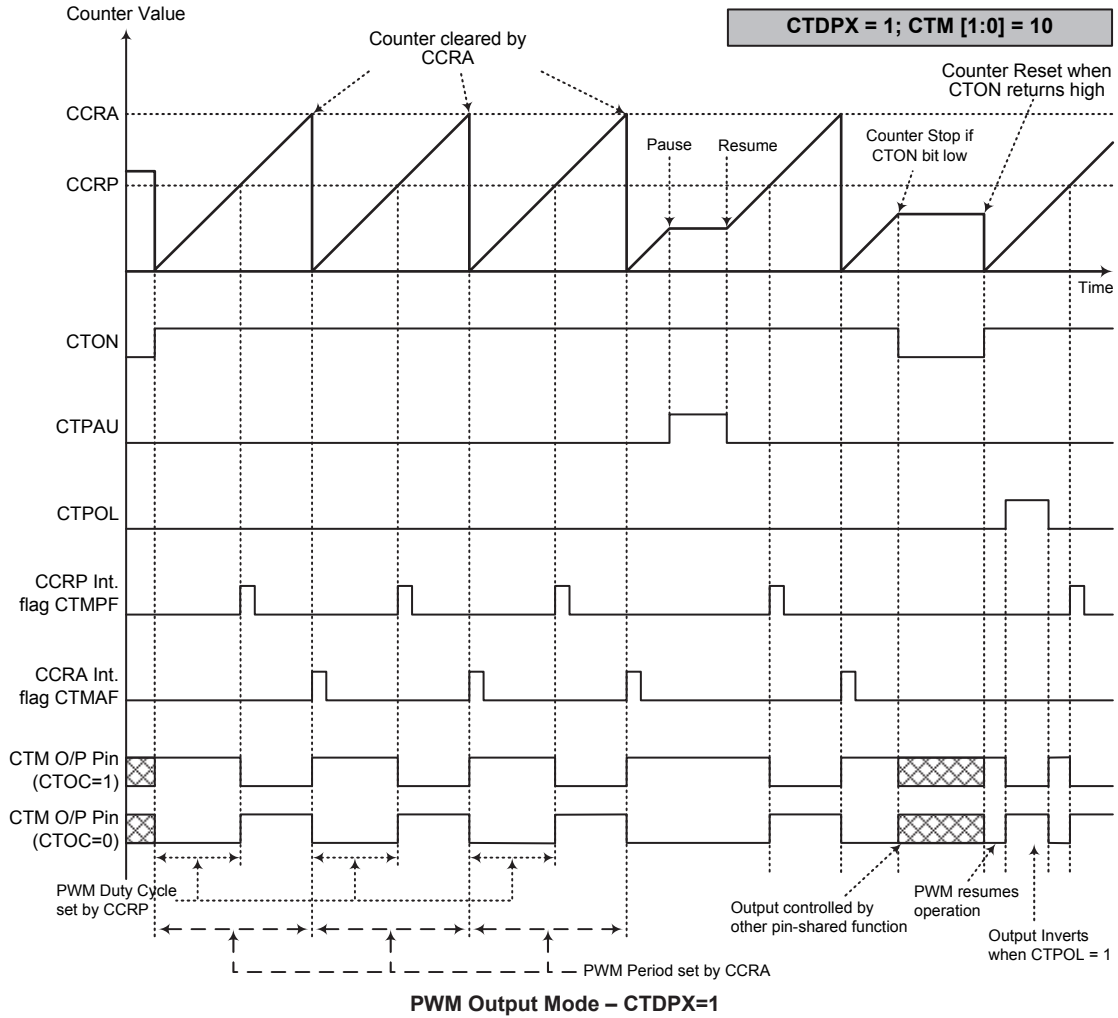
• **16-bit CTM, PWM Output Mode, Edge-aligned Mode, CTD PX=1**

| CCRP | 1~255 | 0 |
|--------|----------|-------|
| Period | CCRA | |
| Duty | CCRP×256 | 65536 |

The PWM output period is determined by the CCRA register value together with the CTM clock while the PWM duty cycle is defined by the CCRP register value.



- Note: 1. Here CTDPX=0 – Counter cleared by CCRP
 2. A counter clear sets the PWM Period
 3. The internal PWM function continues even when CTIO[1:0]=00 or 01
 4. The CTCCLR bit has no influence on PWM operation

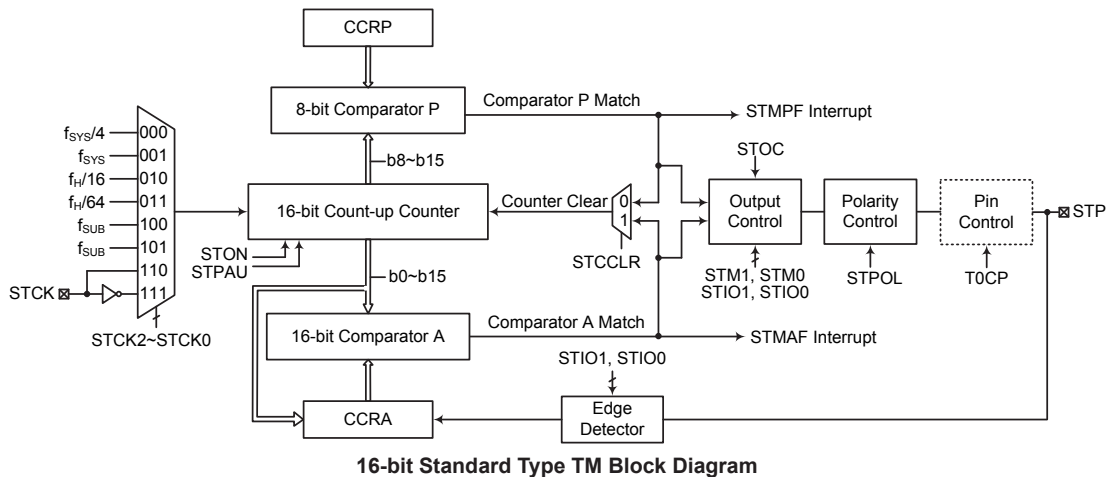


- Note: 1. Here CTDPX=1 – Counter cleared by CCRA
 2. A counter clear sets the PWM Period
 3. The internal PWM function continues even when CTIO [1:0]=00 or 01
 4. The CTCCLR bit has no influence on PWM operation

Standard Type TM – STM

The Standard Type TM contains five operating modes, which are Compare Match Output, Timer/Event Counter, Capture Input, Single Pulse Output and PWM Output modes. The Standard TM can also be controlled with two external input pins and can drive one external output pin.

| STM Core | STM Input Pin | STM Output Pin |
|------------|---------------|----------------|
| 16-bit STM | STCK, STP | STP |



Standard Type TM Operation

The size of Standard TM is 16-bit wide and its core is a 16-bit count-up counter which is driven by a user selectable internal or external clock source. There are also two internal comparators with the names, Comparator A and Comparator P. These comparators will compare the value in the counter with CCRP and CCRA registers. The CCRP comparator is 8-bit wide whose value is compared with the highest 8 bits in the counter while the CCRA is the sixteen bits and therefore compares all counter bits.

The only way of changing the value of the 16-bit counter using the application program, is to clear the counter by changing the STON bit from low to high. The counter will also be cleared automatically by a counter overflow or a compare match with one of its associated comparators. When these conditions occur, a STM interrupt signal will also usually be generated. The Standard Type TM can operate in a number of different operational modes, can be driven by different clock sources including an input pin and can also control the output pin. All operating setup conditions are selected using relevant internal registers.

Standard Type TM Register Description

Overall operation of the Standard TM is controlled using a series of registers. A read only register pair exists to store the internal counter 16-bit value, while a read/write register pair exists to store the internal 16-bit CCRA value. The STMRP register is used to store the 8-bit CCRP value. The remaining two registers are control registers which setup the different operating and control modes.

| Register Name | Bit | | | | | | | |
|---------------|-------|-------|-------|-------|-------|-------|-------|--------|
| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| STMC0 | STPAU | STCK2 | STCK1 | STCK0 | STON | — | — | — |
| STMC1 | STM1 | STM0 | STIO1 | STIO0 | STOC | STPOL | STDPX | STCCLR |
| STMDL | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| STMDH | D15 | D14 | D13 | D12 | D11 | D10 | D9 | D8 |
| STMAL | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| STMAH | D15 | D14 | D13 | D12 | D11 | D10 | D9 | D8 |
| STMRP | STRP7 | STRP6 | STRP5 | STRP4 | STRP3 | STRP2 | STRP1 | STRP0 |

16-bit Standard TM Register List

• **STMC0 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-------|-------|-------|-------|------|---|---|---|
| Name | STPAU | STCK2 | STCK1 | STCK0 | STON | — | — | — |
| R/W | R/W | R/W | R/W | R/W | R/W | — | — | — |
| POR | 0 | 0 | 0 | 0 | 0 | — | — | — |

Bit 7 **STPAU**: STM counter pause control

0: Run
1: Pause

The counter can be paused by setting this bit high. Clearing the bit to zero restores normal counter operation. When in a Pause condition the STM will remain powered up and continue to consume power. The counter will retain its residual value when this bit changes from low to high and resume counting from this value when the bit changes to a low value again.

Bit 6~4 **STCK2~STCK0**: STM counter clock selection

000: $f_{SYS}/4$
001: f_{SYS}
010: $f_H/16$
011: $f_H/64$
100: f_{SUB}
101: f_{SUB}
110: STCK rising edge clock
111: STCK falling edge clock

These three bits are used to select the clock source for the STM. The external pin clock source can be chosen to be active on the rising or falling edge. The clock source f_{SYS} is the system clock, while f_H and f_{SUB} are other internal clocks, the details of which can be found in the oscillator section.

Bit 3 **STON**: STM counter on/off control

0: Off
1: On

This bit controls the overall on/off function of the STM. Setting the bit high enables the counter to run while clearing the bit disables the STM. Clearing this bit to zero will stop the counter from counting and turn off the STM which will reduce its power consumption. When the bit changes state from low to high the internal counter value will be reset to zero, however when the bit changes from high to low, the internal counter will retain its residual value until the bit returns high again. If the STM is in the Compare Match Output Mode or PWM output Mode or Single Pulse Output Mode, then the STM output pin will be reset to its initial condition, as specified by the STOC bit, when the STON bit changes from low to high.

Bit 2~0 Unimplemented, read as “0”

• **STMC1 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|------|------|-------|-------|------|-------|-------|--------|
| Name | STM1 | STM0 | STIO1 | STIO0 | STOC | STPOL | STDPX | STCCLR |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

- Bit 7~6 **STM1~STM0:** STM operating mode selection
 00: Compare Match Output Mode
 01: Capture Input Mode
 10: PWM Output Mode or Single Pulse Output Mode
 11: Timer/Counter Mode

These bits setup the required operating mode for the STM. To ensure reliable operation the STM should be switched off before any changes are made to the STM1 and STM0 bits. In the Timer/Counter Mode, the STM output pin control must be disabled.

- Bit 5~4 **STIO1~STIO0:** STM STP pin function selection

- Compare Match Output Mode
 00: No change
 01: Output low
 10: Output high
 11: Toggle output
- PWM Output Mode/Single Pulse Output Mode
 00: PWM output inactive state
 01: PWM output active state
 10: PWM output
 11: Single Pulse Output
- Capture Input Mode
 00: Input capture at rising edge of STP
 01: Input capture at falling edge of STP
 10: Input capture at rising/falling edge of STP
 11: Input capture disabled

- Timer/Counter Mode
 Unused

These two bits are used to determine how the STM output pin changes state when a certain condition is reached. The function that these bits select depends upon in which mode the STM is running.

In the Compare Match Output Mode, the STIO1 and STIO0 bits determine how the STM output pin changes state when a compare match occurs from the Comparator A. The TM output pin can be setup to switch high, switch low or to toggle its present state when a compare match occurs from the Comparator A. When the bits are both zero, then no change will take place on the output. The initial value of the STM output pin should be setup using the STOC bit in the STMC1 register. Note that the output level requested by the STIO1 and STIO0 bits must be different from the initial value setup using the STOC bit otherwise no change will occur on the STM output pin when a compare match occurs. After the STM output pin changes state, it can be reset to its initial level by changing the level of the STON bit from low to high.

In the PWM Output Mode, the STIO1 and STIO0 bits determine how the STM output pin changes state when a certain compare match condition occurs. The PWM output function is modified by changing these two bits. It is necessary to only change the values of the STIO1 and STIO0 bits only after the STM has been switched off. Unpredictable PWM outputs will occur if the STIO1 and STIO0 bits are changed when the STM is running.

- Bit 3 STOC: STM STP output control**
 Compare Match Output Mode
 0: Initial low
 1: Initial high
 PWM Output Mode/Single Pulse Output Mode
 0: Active low
 1: Active high
 This is the output control bit for the STM output pin. Its operation depends upon whether STM is being used in the Compare Match Output Mode or in the PWM Output Mode/Single Pulse Output Mode. It has no effect if the STM is in the Timer/Counter Mode. In the Compare Match Output Mode it determines the logic level of the STM output pin before a compare match occurs. In the PWM Output Mode it determines if the PWM signal is active high or active low. In the Single Pulse Output Mode it determines the logic level of the STM output pin when the STON bit changes from low to high.
- Bit 2 STPOL: STM STP output polarity control**
 0: Non-invert
 1: Invert
 This bit controls the polarity of the STP output pin. When the bit is set high the STM output pin will be inverted and not inverted when the bit is zero. It has no effect if the STM is in the Timer/Counter Mode.
- Bit 1 STDPX: STM PWM duty/period control**
 0: CCRP – period; CCRA – duty
 1: CCRP – duty; CCRA – period
 This bit determines which of the CCRA and CCRP registers are used for period and duty control of the PWM waveform.
- Bit 0 STCCLR: STM counter clear condition selection**
 0: Comparator P match
 1: Comparator A match
 This bit is used to select the method which clears the counter. Remember that the Standard TM contains two comparators, Comparator A and Comparator P, either of which can be selected to clear the internal counter. With the STCCLR bit set high, the counter will be cleared when a compare match occurs from the Comparator A. When the bit is low, the counter will be cleared when a compare match occurs from the Comparator P or with a counter overflow. A counter overflow clearing method can only be implemented if the CCRP bits are all cleared to zero. The STCCLR bit is not used in the PWM Output Mode, Single Pulse Output Mode or Capture Input Mode.

• **STMDL Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|----|----|----|----|----|----|----|----|
| Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| R/W | R | R | R | R | R | R | R | R |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~0 **D7~D0: STM Counter Low Byte Register bit 7 ~ bit 0**
 STM 16-bit Counter bit 7 ~ bit 0

• **STMDH Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-----|-----|-----|-----|-----|-----|----|----|
| Name | D15 | D14 | D13 | D12 | D11 | D10 | D9 | D8 |
| R/W | R | R | R | R | R | R | R | R |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~0 **D15~D8: STM Counter High Byte Register bit 7 ~ bit 0**
 STM 16-bit Counter bit 15 ~ bit 8

• **STMAL Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-----|-----|-----|-----|-----|-----|-----|-----|
| Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~0 **D7~D0:** STM CCRA Low Byte Register bit 7 ~ bit 0
STM 16-bit CCRA bit 7 ~ bit 0

• **STMAH Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-----|-----|-----|-----|-----|-----|-----|-----|
| Name | D15 | D14 | D13 | D12 | D11 | D10 | D9 | D8 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~0 **D15~D8:** STM CCRA High Byte Register bit 7 ~ bit 0
STM 16-bit CCRA bit 15 ~ bit 8

• **STMRP Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-------|-------|-------|-------|-------|-------|-------|-------|
| Name | STRP7 | STRP6 | STRP5 | STRP4 | STRP3 | STRP2 | STRP1 | STRP0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~0 **STRP7~STRP0:** STM CCRP 8-bit register, compared with the STM counter bit 15 ~ bit 8
Comparator P match period =
0: 65536 STM clocks
1~255: $(1\sim255) \times 256$ STM clocks

These eight bits are used to setup the value on the internal CCRP 8-bit register, which are then compared with the internal counter's highest eight bits. The result of this comparison can be selected to clear the internal counter if the STCCLR bit is set to zero. Setting the STCCLR bit to zero ensures that a compare match with the CCRP values will reset the internal counter. As the CCRP bits are only compared with the highest eight counter bits, the compare values exist in 256 clock cycle multiples. Clearing all eight bits to zero is in effect allowing the counter to overflow at its maximum value.

Standard Type TM Operation Modes

The Standard Type TM can operate in one of five operating modes, Compare Match Output Mode, PWM Output Mode, Single Pulse Output Mode, Capture Input Mode or Timer/Counter Mode. The operating mode is selected using the STM1 and STM0 bits in the STMC1 register.

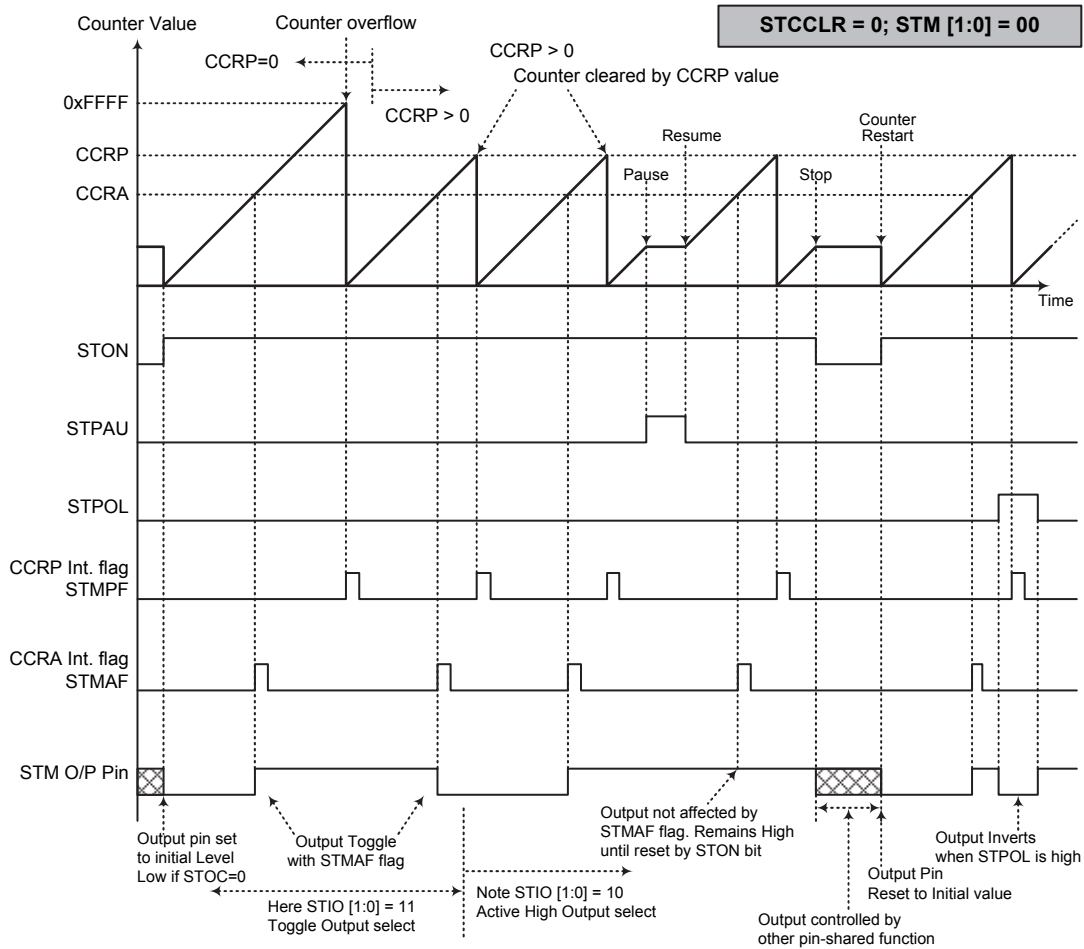
Compare Match Output Mode

To select this mode, bits STM1 and STM0 in the STMC1 register, should be set to “00” respectively. In this mode once the counter is enabled and running it can be cleared by three methods. These are a counter overflow, a compare match from Comparator A and a compare match from Comparator P. When the STCCLR bit is low, there are two ways in which the counter can be cleared. One is when a compare match from Comparator P, the other is when the CCRP bits are all zero which allows the counter to overflow. Here both STMAF and STMPF interrupt request flags for Comparator A and Comparator P respectively, will both be generated.

If the STCCLR bit in the STMC1 register is high then the counter will be cleared when a compare match occurs from Comparator A. However, here only the STMAF interrupt request flag will be generated even if the value of the CCRP bits is less than that of the CCRA registers. Therefore when STCCLR is high no STMPF interrupt request flag will be generated. In the Compare Match Output Mode, the CCRA can not be set to “0”.

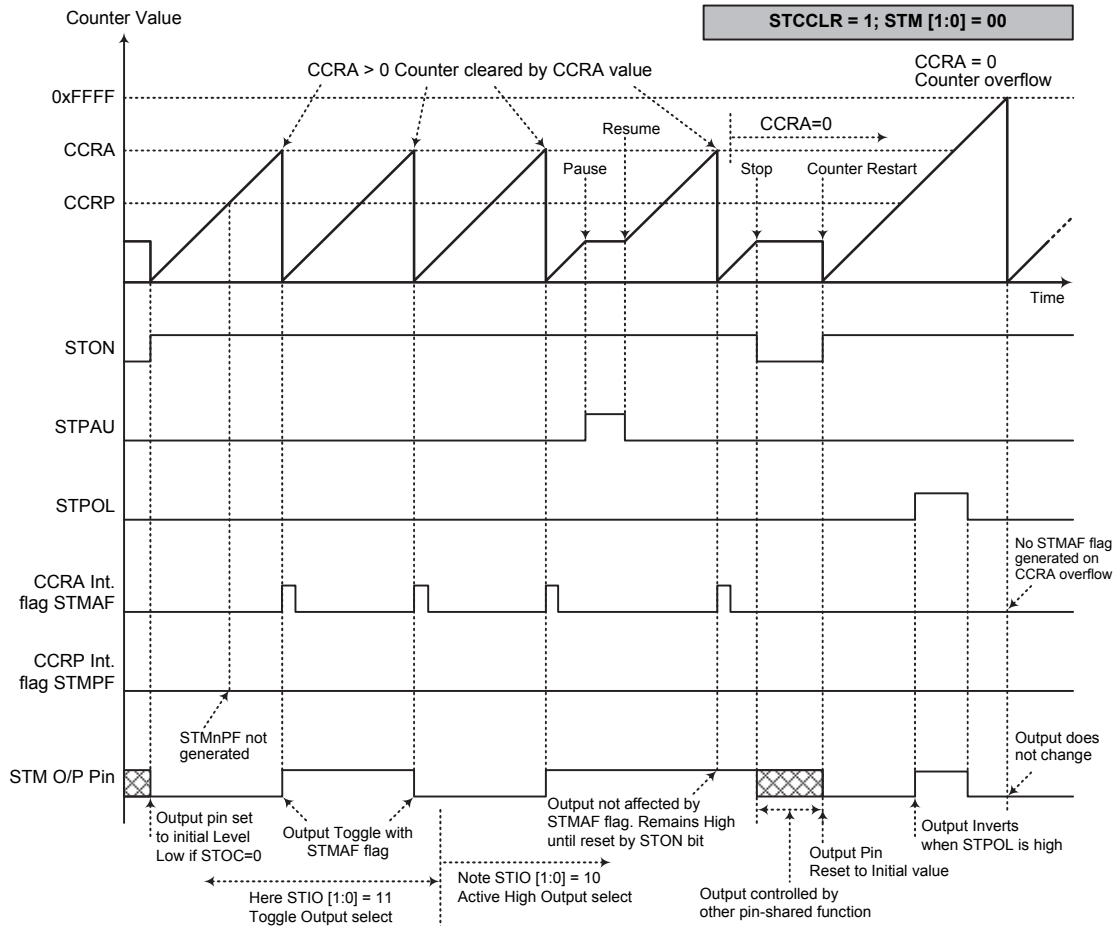
If the CCRA bits are all zero, the counter will overflow when it reaches its maximum 16-bit, FFFF Hex, value, however here the STMAF interrupt request flag will not be generated.

As the name of the mode suggests, after a comparison is made, the STM output pin, will change state. The STM output pin condition however only changes state when a STMAF interrupt request flag is generated after a compare match occurs from Comparator A. The STMPF interrupt request flag, generated from a compare match occurs from Comparator P, will have no effect on the STM output pin. The way in which the STM output pin changes state are determined by the condition of the STIO1 and STIO0 bits in the STMC1 register. The STM output pin can be selected using the STIO1 and STIO0 bits to go high, to go low or to toggle from its present condition when a compare match occurs from Comparator A. The initial condition of the STM output pin, which is setup after the STON bit changes from low to high, is setup using the STOC bit. Note that if the STIO1 and STIO0 bits are zero then no pin change will take place.



Compare Match Output Mode – STCCLR=0

- Note: 1. With STCCLR=0, a Comparator P match will clear the counter
2. The STM output pin is controlled only by the STMAF flag
3. The output pin is reset to its initial state by a STON bit rising edge



Compare Match Output Mode – STCCLR=1

- Note: 1. With $STCCLR=1$, a Comparator A match will clear the counter
 2. The STM output pin is controlled only by the STMAF flag
 3. The output pin is reset to its initial state by a STON bit rising edge
 4. The STMPF flag is not generated when $STCCLR=1$

Timer/Counter Mode

To select this mode, bits STM1 and STM0 in the STMC1 register should be set to “11” respectively. The Timer/Counter Mode operates in an identical way to the Compare Match Output Mode generating the same interrupt flags. The exception is that in the Timer/Counter Mode the STM output pin is not used. Therefore the above description and Timing Diagrams for the Compare Match Output Mode can be used to understand its function. As the STM output pin is not used in this mode, the pin can be used as a normal I/O pin or other pin-shared functions.

PWM Output Mode

To select this mode, bits STM1 and STM0 in the STMC1 register should be set to “10” respectively and also the STIO1 and STIO0 bits should be set to “10” respectively. The PWM function within the STM is useful for applications which require functions such as motor control, heating control, illumination control etc. By providing a signal of fixed frequency but of varying duty cycle on the STM output pin, a square wave AC waveform can be generated with varying equivalent DC RMS values.

As both the period and duty cycle of the PWM waveform can be controlled, the choice of generated waveform is extremely flexible. In the PWM Output Mode, the STCCLR bit has no effect as the PWM period. Both of the CCRA and CCRP registers are used to generate the PWM waveform, one register is used to clear the internal counter and thus control the PWM waveform frequency, while the other one is used to control the duty cycle. Which register is used to control either frequency or duty cycle is determined using the STDPX bit in the STMC1 register. The PWM waveform frequency and duty cycle can therefore be controlled by the values in the CCRA and CCRP registers.

An interrupt flag, one for each of the CCRA and CCRP, will be generated when a compare match occurs from either Comparator A or Comparator P. The STOC bit in the STMC1 register is used to select the required polarity of the PWM waveform while the two STIO1 and STIO0 bits are used to enable the PWM output or to force the STM output pin to a fixed high or low level. The STPOL bit is used to reverse the polarity of the PWM output waveform.

• **16-bit STM, PWM Output Mode, Edge-aligned Mode, STDPX=0**

| CCRP | 1~255 | 0 |
|--------|----------|-------|
| Period | CCRP×256 | 65536 |
| Duty | CCRA | |

If $f_{SYS}=8\text{MHz}$, STM clock source is $f_{SYS}/4$, CCRP=2 and CCRA=128,

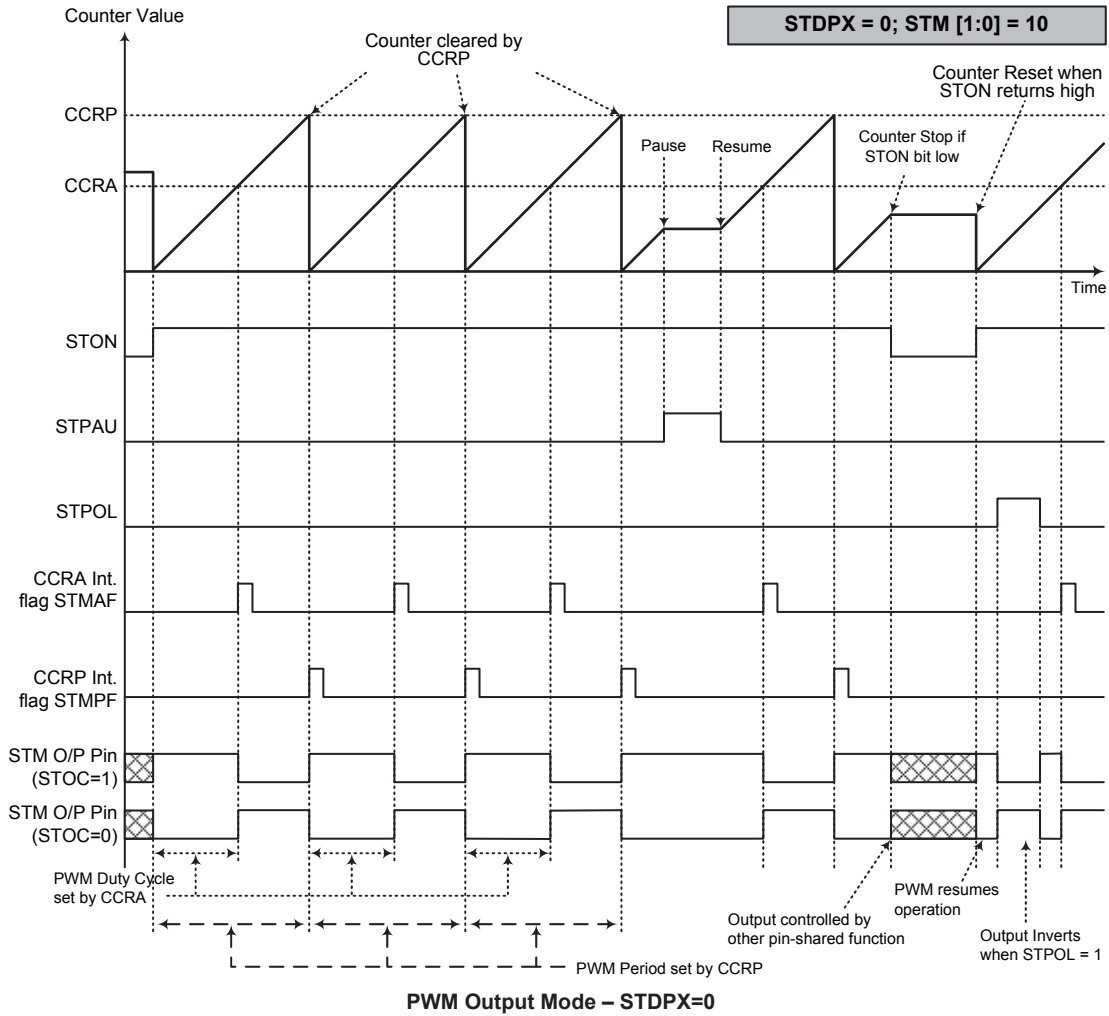
The STM PWM output frequency= $(f_{SYS}/4)/(2\times 256)=f_{SYS}/2048\approx 4\text{kHz}$, duty= $128/(2\times 256)=25\%$.

If the Duty value defined by the CCRA register is equal to or greater than the Period value, then the PWM output duty is 100%.

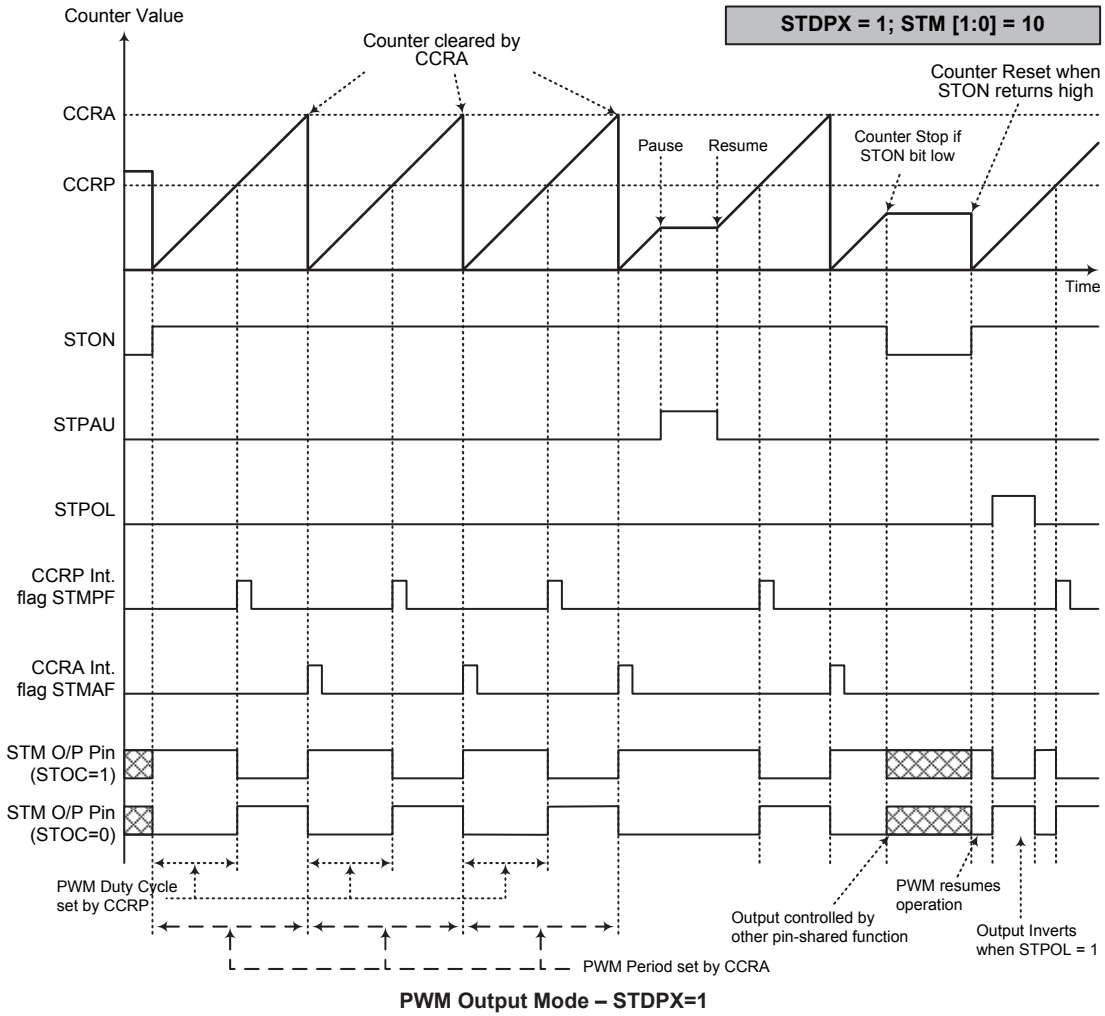
• **16-bit STM, PWM Output Mode, Edge-aligned Mode, STDPX=1**

| CCRP | 1~255 | 0 |
|--------|----------|-------|
| Period | CCRA | |
| Duty | CCRP×256 | 65536 |

The PWM output period is determined by the CCRA register value together with the STM clock while the PWM duty cycle is defined by the CCRP register value except when the CCRP value is equal to 0.



- Note: 1. Here STDPX=0 – Counter cleared by CCRP
 2. A counter clear sets the PWM Period
 3. The internal PWM function continues even when STIO[1:0]=00 or 01
 4. The STCCLR bit has no influence on PWM operation



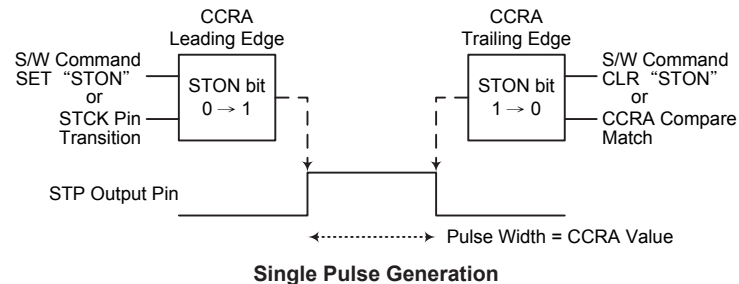
- Note: 1. Here STDPX=1 – Counter cleared by CCRA
 2. A counter clear sets the PWM Period
 3. The internal PWM function continues even when STIO [1:0]=00 or 01
 4. The STCCLR bit has no influence on PWM operation

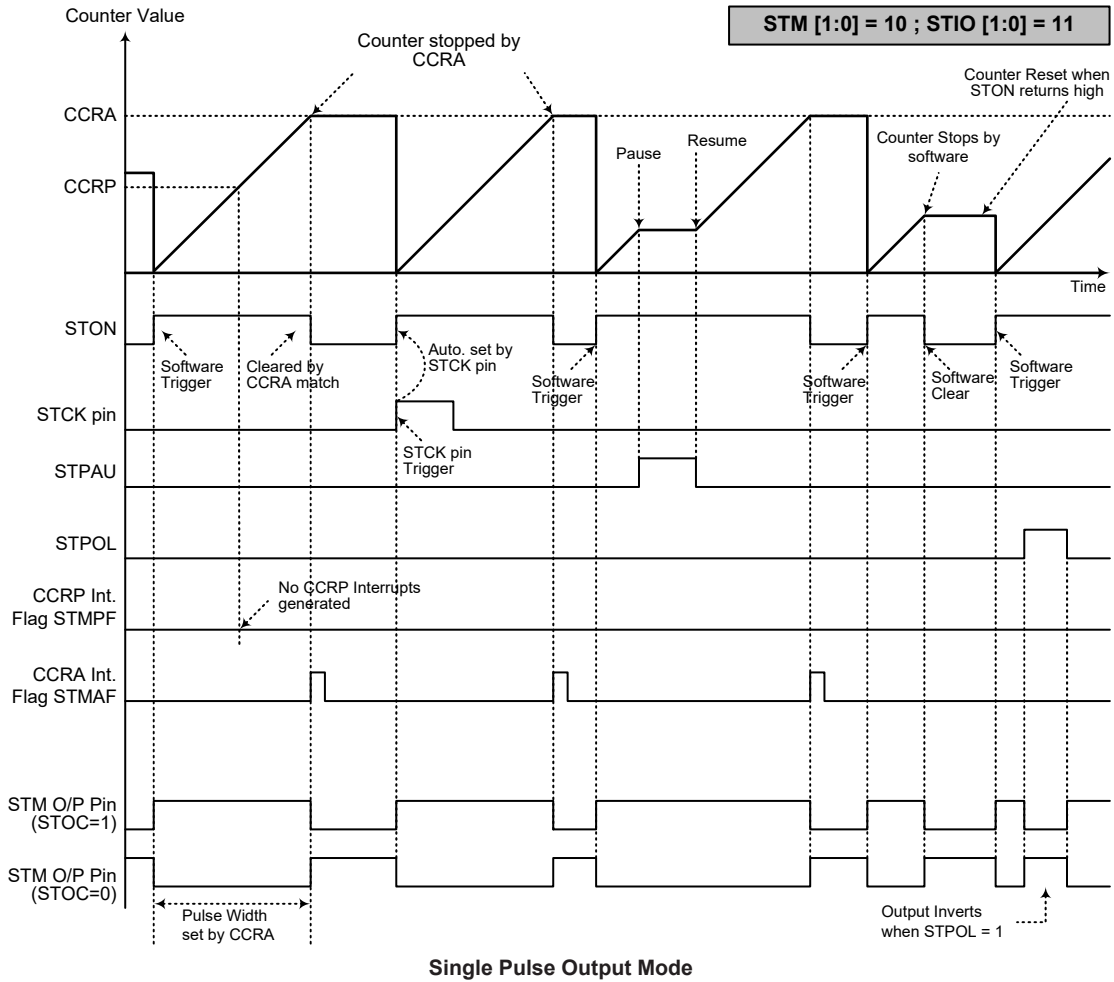
Single Pulse Output Mode

To select this mode, bits STM1 and STM0 in the STMC1 register should be set to “10” respectively and also the STIO1 and STIO0 bits should be set to “11” respectively. The Single Pulse Output Mode, as the name suggests, will generate a single shot pulse on the STM output pin.

The trigger for the pulse output leading edge is a low to high transition of the STON bit, which can be implemented using the application program. However in the Single Pulse Output Mode, the STON bit can also be made to automatically change from low to high using the external STCK pin, which will in turn initiate the Single Pulse output. When the STON bit transitions to a high level, the counter will start running and the pulse leading edge will be generated. The STON bit should remain high when the pulse is in its active state. The generated pulse trailing edge will be generated when the STON bit is cleared to zero, which can be implemented using the application program or when a compare match occurs from Comparator A.

However a compare match from Comparator A will also automatically clear the STON bit and thus generate the Single Pulse output trailing edge. In this way the CCRA value can be used to control the pulse width. A compare match from Comparator A will also generate a STM interrupt. The counter can only be reset back to zero when the STON bit changes from low to high when the counter restarts. In the Single Pulse Output Mode CCRP is not used. The STCCLR and STDPX bits are not used in this Mode.



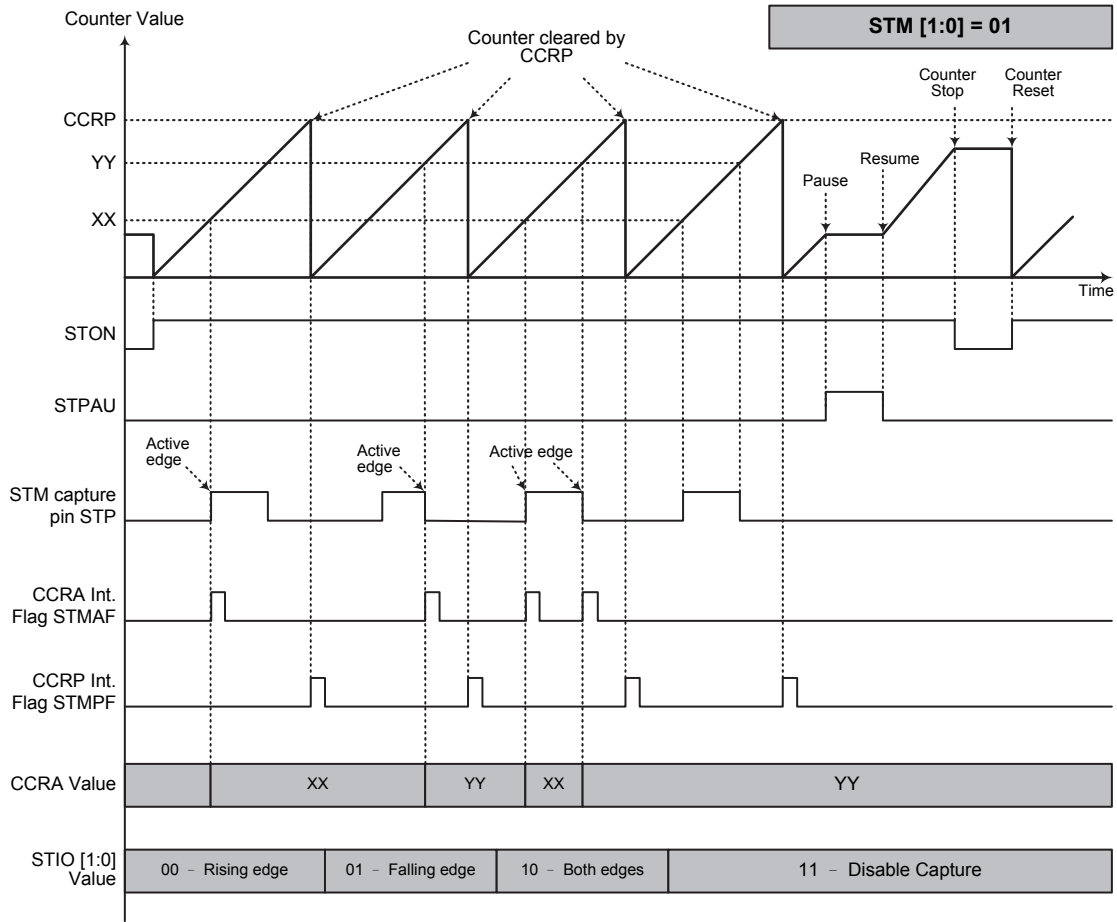


- Note: 1. Counter stopped by CCRA
 2. CCRP is not used
 3. The pulse triggered by the STCK pin or by setting the STON bit high
 4. A STCK pin active edge will automatically set the STON bit high
 5. In the Single Pulse Output Mode, STIO[1:0] must be set to "11" and can not be changed

Capture Input Mode

To select this mode bits STM1 and STM0 in the STMC1 register should be set to “01” respectively. This mode enables external signals to capture and store the present value of the internal counter and can therefore be used for applications such as pulse width measurements. The external signal is supplied on the STP pin, whose active edge can be a rising edge, a falling edge or both rising and falling edges; the active edge transition type is selected using the STIO1 and STIO0 bits in the STMC1 register. The counter is started when the STON bit changes from low to high which is initiated using the application program.

When the required edge transition appears on the STP pin the present value in the counter will be latched into the CCRA registers and a STM interrupt generated. Irrespective of what events occur on the STP pin the counter will continue to free run until the STON bit changes from high to low. When a CCRP compare match occurs the counter will reset back to zero; in this way the CCRP value can be used to control the maximum counter value. When a CCRP compare match occurs from Comparator P, a STM interrupt will also be generated. Counting the number of overflow interrupt signals from the CCRP can be a useful method in measuring long pulse widths. The STIO1 and STIO0 bits can select the active trigger edge on the STP pin to be a rising edge, falling edge or both edge types. If the STIO1 and STIO0 bits are both set high, then no capture operation will take place irrespective of what happens on the STP pin, however it must be noted that the counter will continue to run. The STCCLR and STDPX bits are not used in this mode.



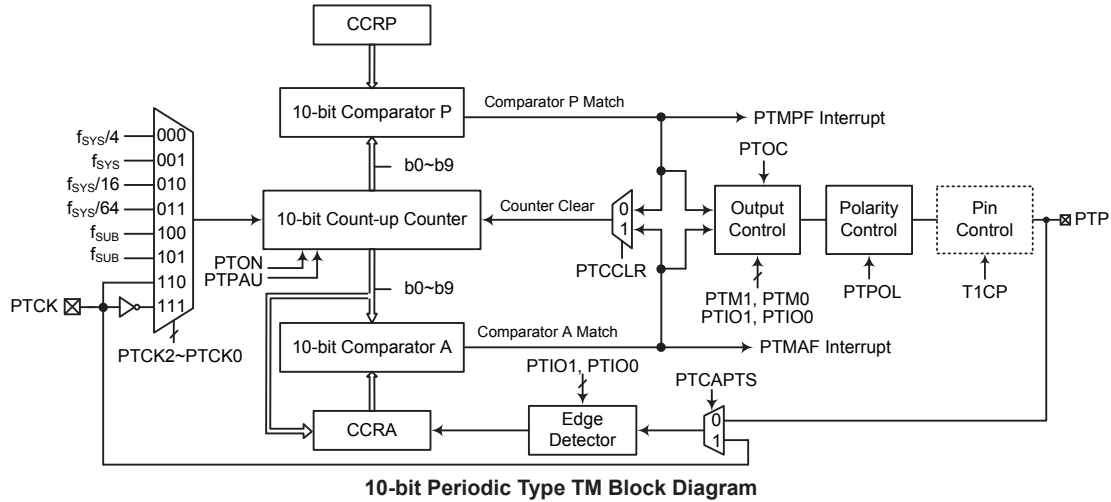
Capture Input Mode

- Note: 1. STM [1:0]=01 and active edge set by the STIO[1:0] bits
 2. A STM Capture input pin active edge transfers the counter value to CCRA
 3. STCCLR bit not used
 4. No output function – STOC and STPOL bits are not used
 5. CCRP determines the counter value and the counter has a maximum count value when CCRP is equal to zero

Periodic Type TM – PTM

The Periodic Type TM contains five operating modes, which are Compare Match Output, Timer/Event Counter, Capture Input, Single Pulse Output and PWM Output modes. The Periodic TM can also be controlled with two external input pins and can drive one external output pin.

| PTM Core | PTM Input Pin | PTM Output Pin |
|------------|---------------|----------------|
| 10-bit PTM | PTCK, PTP | PTP |



Periodic Type TM Operation

The size of Periodic TM is 10-bit wide and its core is a 10-bit count-up counter which is driven by a user selectable internal or external clock source. There are also two internal comparators with the names, Comparator A and Comparator P. These comparators will compare the value in the counter with CCRP and CCRA registers. The CCRP and CCRA comparators are 10-bit wide whose value is compared with all counter bits.

The only way of changing the value of the 10-bit counter using the application program is to clear the counter by changing the PTON bit from low to high. The counter will also be cleared automatically by a counter overflow or a compare match with one of its associated comparators. When these conditions occur, a PTM interrupt signal will also usually be generated. The Periodic Type TM can operate in a number of different operational modes, can be driven by different clock sources including an input pin and can also control two output pins. All operating setup conditions are selected using relevant internal registers.

Periodic Type TM Register Description

Overall operation of the Periodic TM is controlled using a series of registers. A read only register pair exists to store the internal counter 10-bit value, while two read/write register pairs exist to store the internal 10-bit CCRA and CCRP value. The remaining two registers are control registers which setup the different operating and control modes.

| Register Name | Bit | | | | | | | |
|---------------|-------|-------|-------|-------|-------|-------|---------|--------|
| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| PTMC0 | PTPAU | PTCK2 | PTCK1 | PTCK0 | PTON | — | — | — |
| PTMC1 | PTM1 | PTM0 | PTIO1 | PTIO0 | PTOC | PTPOL | PTCAPTS | PTCCLR |
| PTMDL | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| PTMDH | — | — | — | — | — | — | D9 | D8 |
| PTMAL | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| PTMAH | — | — | — | — | — | — | D9 | D8 |
| PTMRPL | PTRP7 | PTRP6 | PTRP5 | PTRP4 | PTRP3 | PTRP2 | PTRP1 | PTRP0 |
| PTMRPH | — | — | — | — | — | — | PTRP9 | PTRP8 |

10-bit Periodic TM Register List
• PTMC0 Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-------|-------|-------|-------|------|---|---|---|
| Name | PTPAU | PTCK2 | PTCK1 | PTCK0 | PTON | — | — | — |
| R/W | R/W | R/W | R/W | R/W | R/W | — | — | — |
| POR | 0 | 0 | 0 | 0 | 0 | — | — | — |

Bit 7 **PTPAU**: PTM counter pause control

0: Run
1: Pause

The counter can be paused by setting this bit high. Clearing the bit to zero restores normal counter operation. When in a Pause condition the PTM will remain powered up and continue to consume power. The counter will retain its residual value when this bit changes from low to high and resume counting from this value when the bit changes to a low value again.

Bit 6~4 **PTCK2~PTCK0**: PTM counter clock selection

000: $f_{SYS}/4$
001: f_{SYS}
010: $f_H/16$
011: $f_H/64$
100: f_{SUB}
101: f_{SUB}
110: PTCK rising edge clock
111: PTCK falling edge clock

These three bits are used to select the clock source for the PTM. The external pin clock source can be chosen to be active on the rising or falling edge. The clock source f_{SYS} is the system clock, while f_H and f_{SUB} are other internal clocks, the details of which can be found in the oscillator section.

Bit 3 **PTON**: PTM counter on/off control

0: Off
1: On

This bit controls the overall on/off function of the PTM. Setting the bit high enables the counter to run while clearing the bit disables the PTM. Clearing this bit to zero will stop the counter from counting and turn off the PTM which will reduce its power consumption. When the bit changes state from low to high the internal counter value will be reset to zero, however when the bit changes from high to low, the internal counter will retain its residual value until the bit returns high again. If the PTM is in the Compare Match Output Mode or PWM output Mode or Single Pulse Output Mode, then the PTM output pin will be reset to its initial condition, as specified by the PTOC bit, when the PTON bit changes from low to high.

Bit 2~0 Unimplemented, read as “0”

• **PTMC1 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|------|------|-------|-------|------|-------|---------|--------|
| Name | PTM1 | PTM0 | PTIO1 | PTIO0 | PTOC | PTPOL | PTCAPTS | PTCCLR |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~6 **PTM1~PTM0**: PTM operating mode selection
 00: Compare Match Output Mode
 01: Capture Input Mode
 10: PWM Output Mode or Single Pulse Output Mode
 11: Timer/Counter Mode

These bits setup the required operating mode for the PTM. To ensure reliable operation the PTM should be switched off before any changes are made to the PTM1 and PTM0 bits. In the Timer/Counter Mode, the PTM output pin control must be disabled.

Bit 5~4 **PTIO1~PTIO0**: PTM external pin function selection

Compare Match Output Mode
 00: No change
 01: Output low
 10: Output high
 11: Toggle output

PWM Output Mode/Single Pulse Output Mode
 00: PWM output inactive state
 01: PWM output active state
 10: PWM output
 11: Single Pulse Output

Capture Input Mode
 00: Input capture at rising edge of PTP or PTCK
 01: Input capture at falling edge of PTP or PTCK
 10: Input capture at rising/falling edge of PTP or PTCK
 11: Input capture disabled

Timer/Counter Mode
 Unused

These two bits are used to determine how the PTM output pin changes state when a certain condition is reached. The function that these bits select depends upon in which mode the PTM is running.

In the Compare Match Output Mode, the PTIO1 and PTIO0 bits determine how the PTM output pin changes state when a compare match occurs from the Comparator A. The PTM output pin can be setup to switch high, switch low or to toggle its present state when a compare match occurs from the Comparator A. When the bits are both zero, then no change will take place on the output. The initial value of the PTM output pin should be setup using the PTOC bit in the PTMC1 register. Note that the output level requested by the PTIO1 and PTIO0 bits must be different from the initial value setup using the PTOC bit otherwise no change will occur on the PTM output pin when a compare match occurs. After the PTM output pin changes state, it can be reset to its initial level by changing the level of the PTON bit from low to high.

In the PWM Output Mode, the PTIO1 and PTIO0 bits determine how the PTM output pin changes state when a certain compare match condition occurs. The PTM output function is modified by changing these two bits. It is necessary to only change the values of the PTIO1 and PTIO0 bits only after the PTM has been switched off. Unpredictable PWM outputs will occur if the PTIO1 and PTIO0 bits are changed when the PTM is running.

- Bit 3 **PTOC**: PTM PTP output control
 Compare Match Output Mode
 0: Initial low
 1: Initial high
 PWM Output Mode/Single Pulse Output Mode
 0: Active low
 1: Active high
- This is the output control bit for the PTM output pin. Its operation depends upon whether PTM is being used in the Compare Match Output Mode or in the PWM Output Mode/Single Pulse Output Mode. It has no effect if the PTM is in the Timer/Counter Mode. In the Compare Match Output Mode it determines the logic level of the PTM output pin before a compare match occurs. In the PWM Output Mode it determines if the PWM signal is active high or active low. In the Single Pulse Output Mode it determines the logic level of the PTM output pin when the PTON bit changes from low to high.
- Bit 2 **PTPOL**: PTM PTP output polarity control
 0: Non-invert
 1: Invert
- This bit controls the polarity of the PTP output pin. When the bit is set high the PTM output pin will be inverted and not inverted when the bit is zero. It has no effect if the PTM is in the Timer/Counter Mode.
- Bit 1 **PTCAPTS**: PTM capture trigger source selection
 0: From PTP pin
 1: From PTCK pin
- Bit 0 **PTCCLR**: PTM counter clear condition selection
 0: Comparator P match
 1: Comparator A match
- This bit is used to select the method which clears the counter. Remember that the Periodic TM contains two comparators, Comparator A and Comparator P, either of which can be selected to clear the internal counter. With the PTCCLR bit set high, the counter will be cleared when a compare match occurs from the Comparator A. When the bit is low, the counter will be cleared when a compare match occurs from the Comparator P or with a counter overflow. A counter overflow clearing method can only be implemented if the CCRP bits are all cleared to zero. The PTCCLR bit is not used in the PWM Output, Single Pulse Output or Capture Input Mode.

• **PTMDL Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|----|----|----|----|----|----|----|----|
| Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| R/W | R | R | R | R | R | R | R | R |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~0 **D7~D0**: PTM Counter Low Byte Register bit 7 ~ bit 0
 PTM 10-bit Counter bit 7 ~ bit 0

• **PTMDH Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---|---|---|---|---|---|----|----|
| Name | — | — | — | — | — | — | D9 | D8 |
| R/W | — | — | — | — | — | — | R | R |
| POR | — | — | — | — | — | — | 0 | 0 |

Bit 7~2 Unimplemented, read as “0”
 Bit 1~0 **D9~D8**: PTM Counter High Byte Register bit 1 ~ bit 0
 PTM 10-bit Counter bit 9 ~ bit 8

• **PTMAL Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-----|-----|-----|-----|-----|-----|-----|-----|
| Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~0 **D7~D0**: PTM CCRA Low Byte Register bit 7 ~ bit 0
 PTM 10-bit CCRA bit 7 ~ bit 0

• **PTMAH Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---|---|---|---|---|---|-----|-----|
| Name | — | — | — | — | — | — | D9 | D8 |
| R/W | — | — | — | — | — | — | R/W | R/W |
| POR | — | — | — | — | — | — | 0 | 0 |

Bit 7~2 Unimplemented, read as “0”
 Bit 1~0 **D9~D8**: PTM CCRA High Byte Register bit 1 ~ bit 0
 PTM 10-bit CCRA bit 9 ~ bit 8

• **PTMRPL Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-------|-------|-------|-------|-------|-------|-------|-------|
| Name | PTRP7 | PTRP6 | PTRP5 | PTRP4 | PTRP3 | PTRP2 | PTRP1 | PTRP0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~0 **PTRP7~PTRP0**: PTM CCRP Low Byte Register bit 7 ~ bit 0
 PTM 10-bit CCRP bit 7 ~ bit 0

• **PTMRPH Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---|---|---|---|---|---|-------|-------|
| Name | — | — | — | — | — | — | PTRP9 | PTRP8 |
| R/W | — | — | — | — | — | — | R/W | R/W |
| POR | — | — | — | — | — | — | 0 | 0 |

Bit 7~2 Unimplemented, read as “0”
 Bit 1~0 **PTRP9~PTRP8**: PTM CCRP High Byte Register bit 1 ~ bit 0
 PTM 10-bit CCRP bit 9 ~ bit 8

Periodic Type TM Operation Modes

The Periodic Type TM can operate in one of five operating modes, Compare Match Output Mode, PWM Output Mode, Single Pulse Output Mode, Capture Input Mode or Timer/Counter Mode. The operating mode is selected using the PTM1 and PTM0 bits in the PTMC1 register.

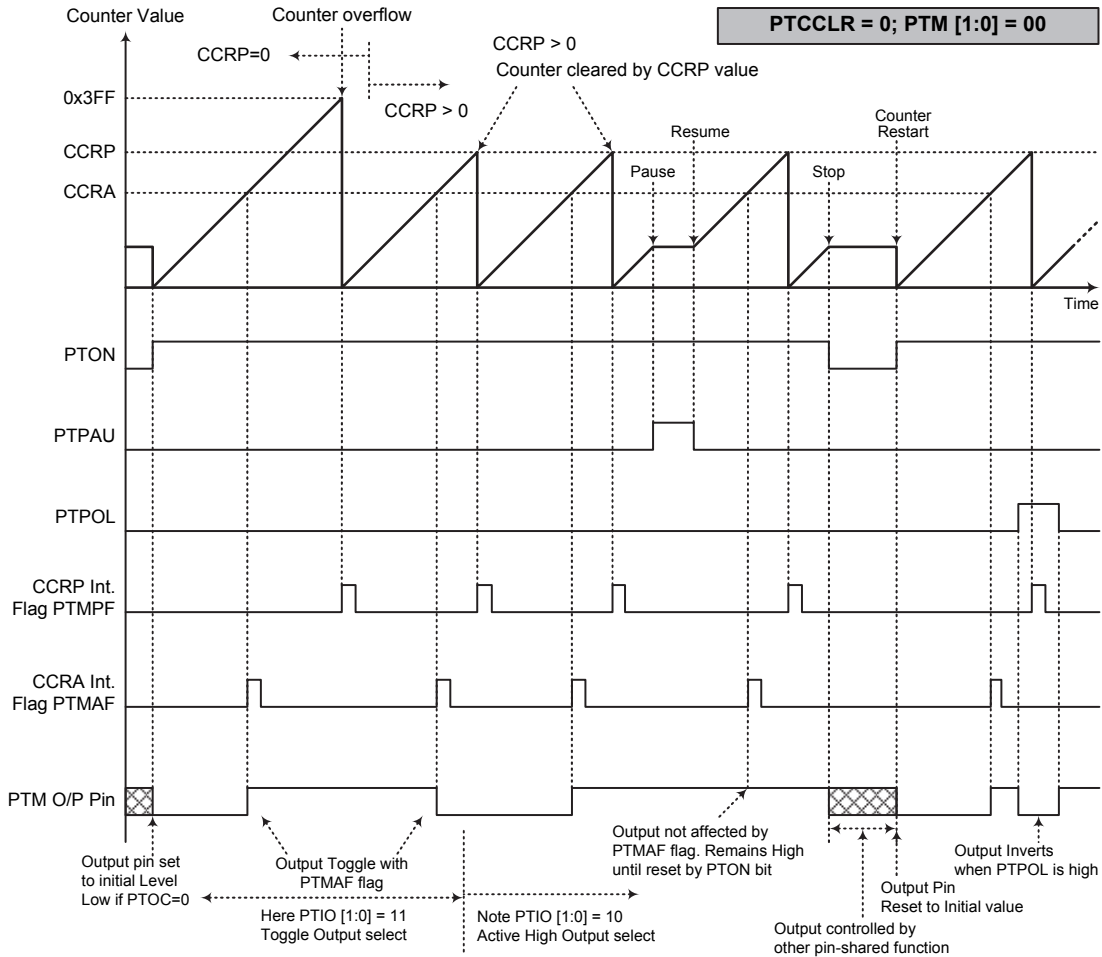
Compare Match Output Mode

To select this mode, bits PTM1 and PTM0 in the PTMC1 register, should be set to “00” respectively. In this mode once the counter is enabled and running it can be cleared by three methods. These are a counter overflow, a compare match from Comparator A and a compare match from Comparator P. When the PTCCLR bit is low, there are two ways in which the counter can be cleared. One is when a compare match from Comparator P, the other is when the CCRP bits are all zero which allows the counter to overflow. Here both PTMAF and PTMPF interrupt request flags for Comparator A and Comparator P respectively, will both be generated.

If the PTCCLR bit in the PTMC1 register is high then the counter will be cleared when a compare match occurs from Comparator A. However, here only the PTMAF interrupt request flag will be generated even if the value of the CCRP bits is less than that of the CCRA registers. Therefore when PTCCLR is high no PTMPF interrupt request flag will be generated. In the Compare Match Output Mode, the CCRA can not be set to “0”.

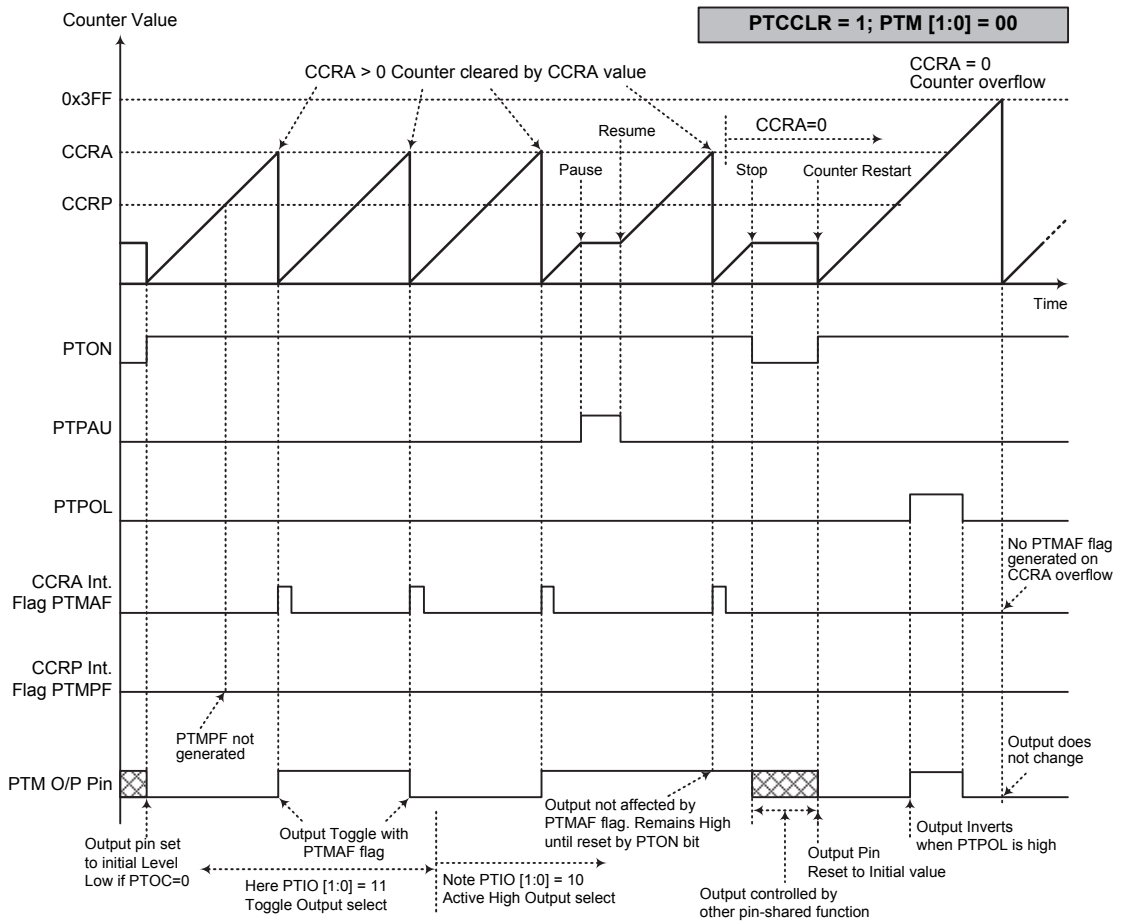
If the CCRA bits are all zero, the counter will overflow when it reaches its maximum 10-bit, 3FF Hex, value, however here the PTMAF interrupt request flag will not be generated.

As the name of the mode suggests, after a comparison is made, the PTM output pin will change state. The PTM output pin condition however only changes state when a PTMAF interrupt request flag is generated after a compare match occurs from Comparator A. The PTMPF interrupt request flag, generated from a compare match occurs from Comparator P, will have no effect on the PTM output pin. The way in which the PTM output pin changes state are determined by the condition of the PTIO1 and PTIO0 bits in the PTMC1 register. The PTM output pin can be selected using the PTIO1 and PTIO0 bits to go high, to go low or to toggle from its present condition when a compare match occurs from Comparator A. The initial condition of the PTM output pin, which is setup after the PTON bit changes from low to high, is setup using the PTOC bit. Note that if the PTIO1 and PTIO0 bits are zero then no pin change will take place.



Compare Match Output Mode – PTCCCLR=0

- Note: 1. With PTCCCLR=0, a Comparator P match will clear the counter
 2. The PTM output pin is controlled only by the PTMAF flag
 3. The output pin is reset to its initial state by a PTON bit rising edge



Compare Match Output Mode – PTCCLR=1

- Note: 1. With PTCCLR=1, a Comparator A match will clear the counter
2. The PTM output pin is controlled only by the PTMAF flag
3. The output pin is reset to its initial state by a PTON bit rising edge
4. A PTMPF flag is not generated when PTCCLR=1

Timer/Counter Mode

To select this mode, bits PTM1 and PTM0 in the PTMC1 register should be set to “11” respectively. The Timer/Counter Mode operates in an identical way to the Compare Match Output Mode generating the same interrupt flags. The exception is that in the Timer/Counter Mode the PTM output pin is not used. Therefore the above description and Timing Diagrams for the Compare Match Output Mode can be used to understand its function. As the PTM output pin is not used in this mode, the pin can be used as a normal I/O pin or other pin-shared functions.

PWM Output Mode

To select this mode, bits PTM1 and PTM0 in the PTMC1 register should be set to “10” respectively and also the PTIO1 and PTIO0 bits should be set to “10” respectively. The PWM function within the PTM is useful for applications which require functions such as motor control, heating control, illumination control, etc. By providing a signal of fixed frequency but of varying duty cycle on the PTM output pin, a square wave AC waveform can be generated with varying equivalent DC RMS values.

As both the period and duty cycle of the PWM waveform can be controlled, the choice of generated waveform is extremely flexible. In the PWM Output Mode, the PTCCLR bit has no effect as the PWM period. Both of the CCRP and CCRA registers are used to generate the PWM waveform, one register is used to clear the internal counter and thus control the PWM waveform frequency, while the other one is used to control the duty cycle. The PWM waveform frequency and duty cycle can therefore be controlled by the values in the CCRA and CCRP registers.

An interrupt flag, one for each of the CCRA and CCRP, will be generated when a compare match occurs from either Comparator A or Comparator P. The PTOC bit in the PTMC1 register is used to select the required polarity of the PWM waveform while the two PTIO1 and PTIO0 bits are used to enable the PWM output or to force the PTM output pin to a fixed high or low level. The PTPOL bit is used to reverse the polarity of the PWM output waveform.

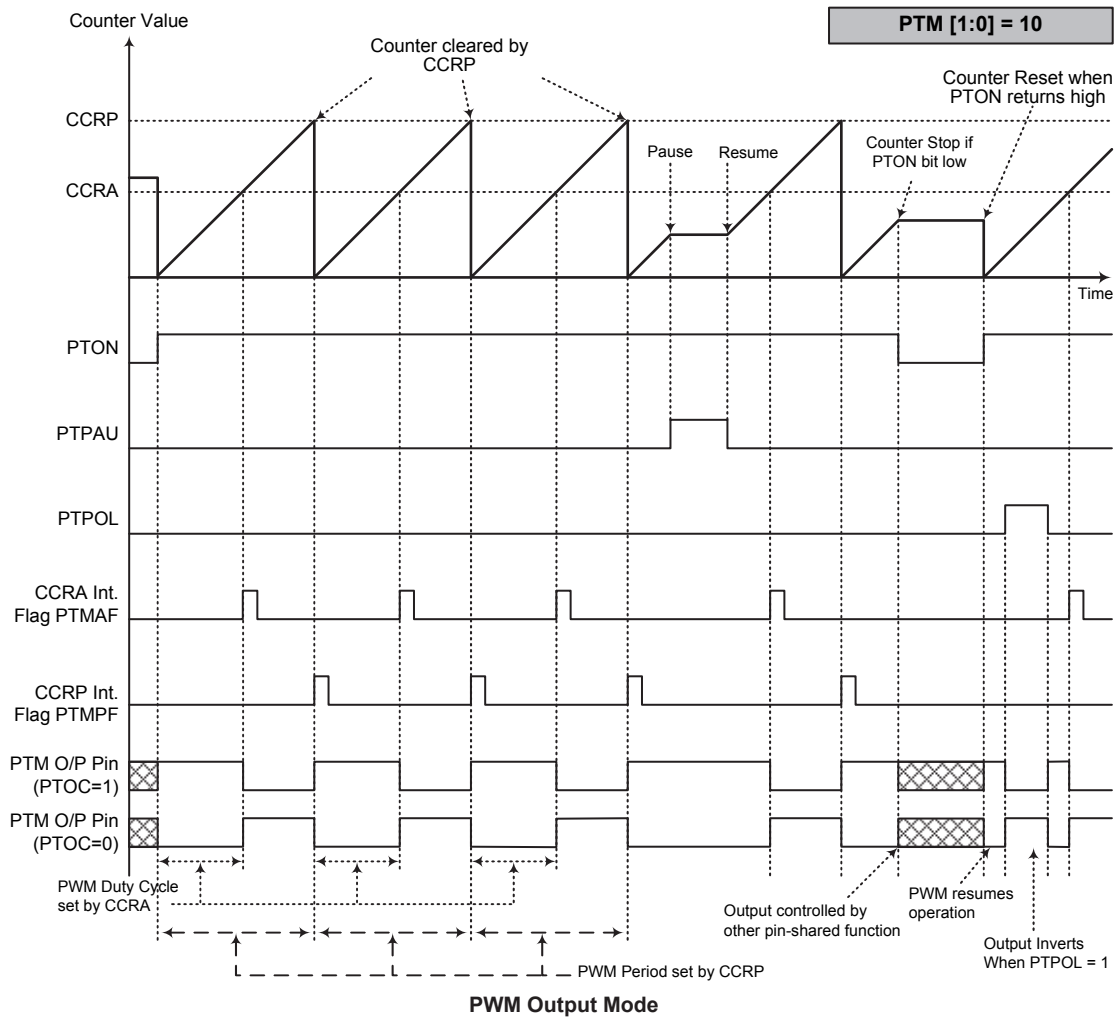
• 10-bit PWM Output Mode, Edge-aligned Mode

| CCRP | 1~1023 | 0 |
|--------|--------|------|
| Period | 1~1023 | 1024 |
| Duty | CCRA | |

If $f_{SYS}=8\text{MHz}$, PTM clock source select $f_{SYS}/4$, $\text{CCRP}=512$ and $\text{CCRA}=128$,

The PTM PWM output frequency= $(f_{SYS}/4)/512=f_{SYS}/2048\approx 4\text{kHz}$, $\text{duty}=128/512=25\%$,

If the Duty value defined by the CCRA register is equal to or greater than the Period value, then the PWM output duty is 100%.



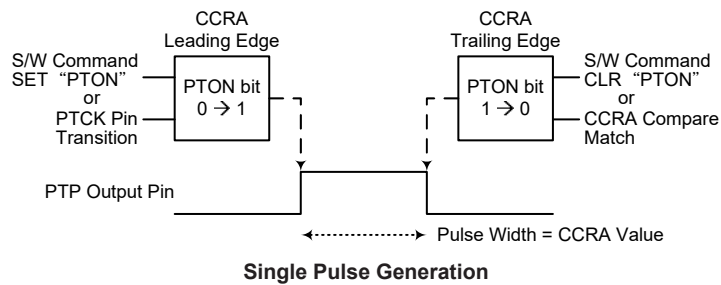
- Note:
1. The counter is cleared by CCRP
 2. A counter clear sets the PWM Period
 3. The internal PWM function continues running even when PTIO[1:0]=00 or 01
 4. The PTCCLR bit has no influence on PWM operation

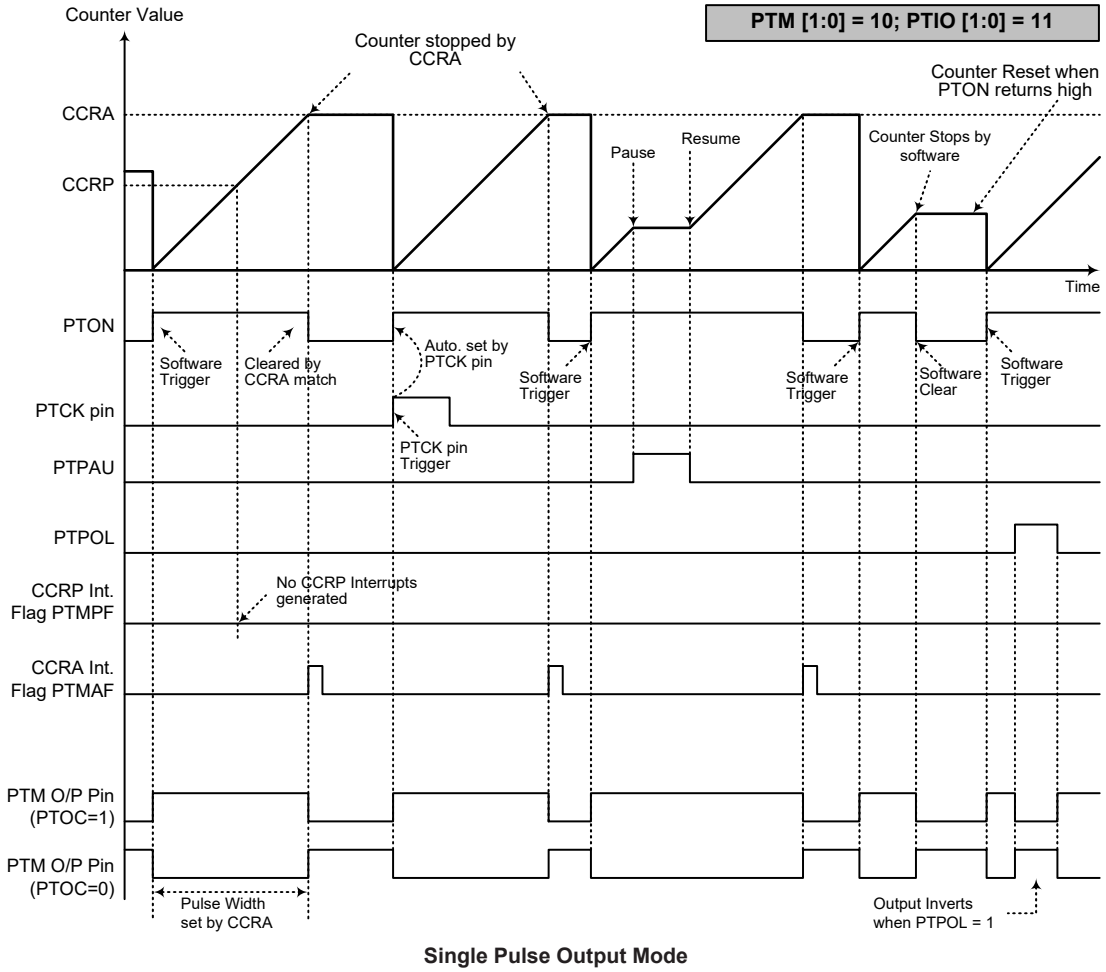
Single Pulse Output Mode

To select this mode, bits PTM1 and PTM0 in the PTMC1 register should be set to “10” respectively and also the PTIO1 and PTIO0 bits should be set to “11” respectively. The Single Pulse Output Mode, as the name suggests, will generate a single shot pulse on the PTM output pin.

The trigger for the pulse output leading edge is a low to high transition of the PTON bit, which can be implemented using the application program. However in the Single Pulse Output Mode, the PTON bit can also be made to automatically change from low to high using the external PTCK pin, which will in turn initiate the Single Pulse output. When the PTON bit transitions to a high level, the counter will start running and the pulse leading edge will be generated. The PTON bit should remain high when the pulse is in its active state. The generated pulse trailing edge will be generated when the PTON bit is cleared to zero, which can be implemented using the application program or when a compare match occurs from Comparator A.

However a compare match from Comparator A will also automatically clear the PTON bit and thus generate the Single Pulse output trailing edge. In this way the CCRA value can be used to control the pulse width. A compare match from Comparator A will also generate a PTM interrupt. The counter can only be reset back to zero when the PTON bit changes from low to high when the counter restarts. In the Single Pulse Output Mode CCRP is not used. The PTCCLR is not used in this Mode.





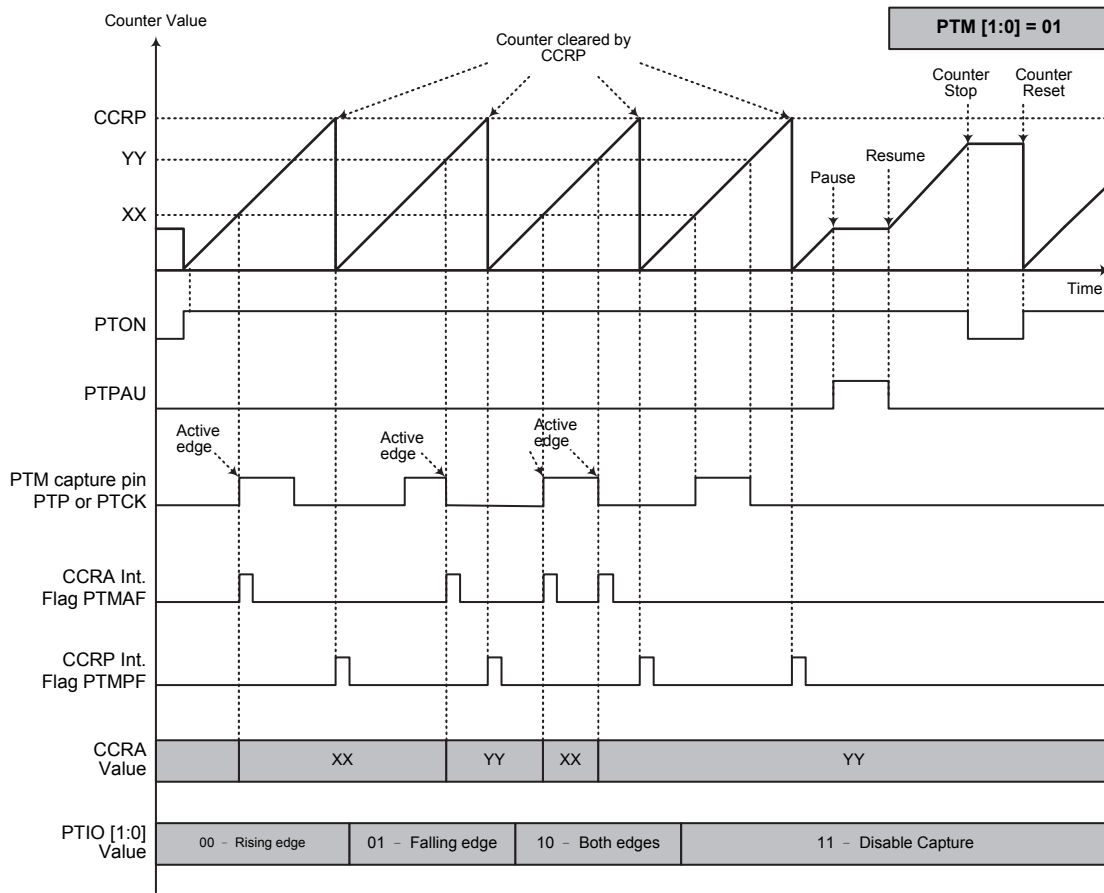
- Note: 1. Counter stopped by CCRA
 2. CCRP is not used
 3. The pulse triggered by the PTCK pin or by setting the PTON bit high
 4. A PTCK pin active edge will automatically set the PTON bit high
 5. In the Single Pulse Output Mode, PTIO[1:0] must be set to "11" and can not be changed

Capture Input Mode

To select this mode bits PTM1 and PTM0 in the PTMC1 register should be set to “01” respectively. This mode enables external signals to capture and store the present value of the internal counter and can therefore be used for applications such as pulse width measurements. The external signal is supplied on the PTP or PTCK pin, selected by the PTCAPTS bit in the PTMC1 register. The input pin active edge can be either a rising edge, a falling edge or both rising and falling edges; the active edge transition type is selected using the PTIO1 and PTIO0 bits in the PTMC1 register. The counter is started when the PTON bit changes from low to high which is initiated using the application program.

When the required edge transition appears on the PTP or PTCK pin the present value in the counter will be latched into the CCRA registers and a PTM interrupt generated. Irrespective of what events occur on the PTP or PTCK pin the counter will continue to free run until the PTON bit changes from high to low. When a CCRP compare match occurs the counter will reset back to zero; in this way the CCRP value can be used to control the maximum counter value. When a CCRP compare match occurs from Comparator P, a PTM interrupt will also be generated. Counting the number of overflow interrupt signals from the CCRP can be a useful method in measuring long pulse widths. The PTIO1 and PTIO0 bits can select the active trigger edge on the PTP or PTCK pin to be a rising edge, falling edge or both edge types. If the PTIO1 and PTIO0 bits are both set high, then no capture operation will take place irrespective of what happens on the PTP or PTCK pin, however it must be noted that the counter will continue to run.

As the PTP or PTCK pin is pin shared with other functions, care must be taken if the PTM is in the Capture Input Mode. This is because if the pin is setup as an output, then any transitions on this pin may cause an input capture operation to be executed. The PTCCLR, PTOC and PTPOL bits are not used in this mode.



Capture Input Mode

- Note: 1. PTM[1:0]=01 and active edge set by the PTIO[1:0] bits
 2. A PTM Capture input pin active edge transfers the counter value to CCRA
 3. PTCLLR bit not used
 4. No output function – PTOC and PTPOL bits are not used
 5. CCRP determines the counter value and the counter has a maximum count value when CCRP is equal to zero.

Analog to Digital Converter – ADC

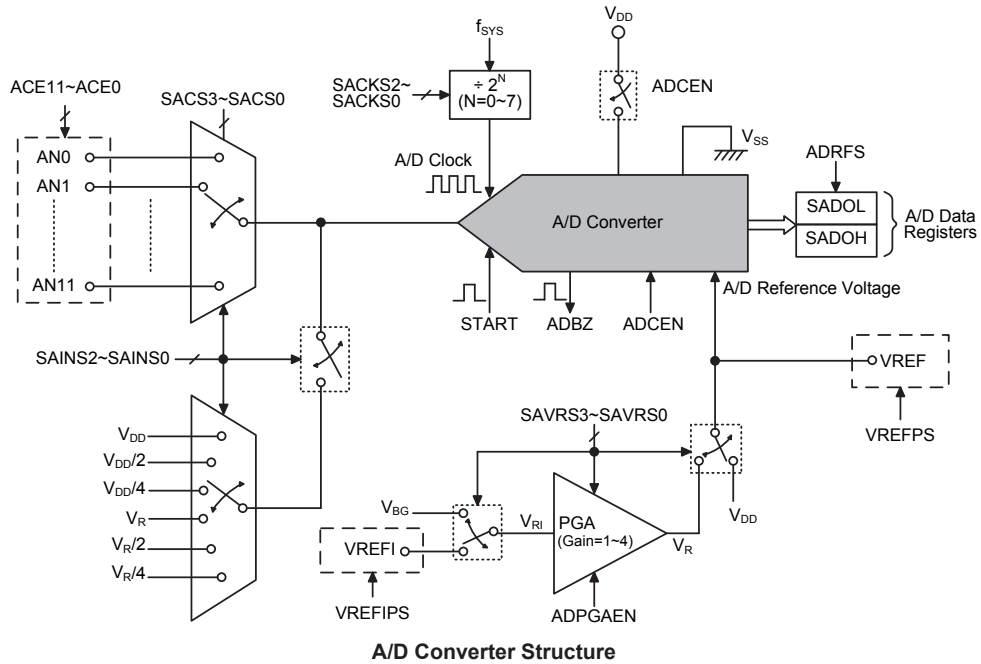
The need to interface to real world analog signals is a common requirement for many electronic systems. However, to properly process these signals by a microcontroller, they must first be converted into digital signals by A/D converters. By integrating the A/D conversion electronic circuitry into the microcontroller, the need for external components is reduced significantly with the corresponding follow-on benefits of lower costs and reduced component space requirements.

A/D Converter Overview

The device contains a multi-channel analog to digital converter which can directly interface to external analog signals, such as that from sensors or other control signals and convert these signals directly into a 12-bit digital value. It also can convert the internal signals, such as the internal reference voltage, into a 12-bit digital value. The external or internal analog signal to be converted is determined by the SAINS and SACS bit fields. Note that when the internal analog signal is to be converted using the SAINS bit field, the external channel analog input will be automatically be switched off. More detailed information about the A/D converter input signal is described in the “A/D Converter Control Registers” and “A/D Converter Input Signals” sections respectively.

| External Input Channels | Internal Analog Signals | A/D Signal Select |
|-------------------------|---|----------------------------|
| 12: AN0~AN11 | 6: V_{DD} , $V_{DD}/2$, $V_{DD}/4$, V_R , $V_R/2$, $V_R/4$ | SAINS2~SAINS0, SACS3~SACS0 |

The accompanying block diagram shows the overall internal structure of the A/D converter, together with its associated registers.



A/D Converter Register Description

Overall operation of the A/D converter is controlled using seven registers. A read only register pair exists to store the A/D converter data 12-bit value. One register pair, ACERL and ACERH, is used to configure the external analog input pin function. The remaining three registers are control registers which setup the operating and control function of the A/D converter.

| Register Name | Bit | | | | | | | |
|----------------|---------|--------|---------|--------|--------|--------|--------|--------|
| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| SADOL(ADRFs=0) | D3 | D2 | D1 | D0 | — | — | — | — |
| SADOL(ADRFs=1) | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| SADOH(ADRFs=0) | D11 | D10 | D9 | D8 | D7 | D6 | D5 | D4 |
| SADOH(ADRFs=1) | — | — | — | — | D11 | D10 | D9 | D8 |
| SADC0 | START | ADBZ | ADCEN | ADRFs | SACS3 | SACS2 | SACS1 | SACS0 |
| SADC1 | SAINS2 | SAINS1 | SAINS0 | — | — | SACKS2 | SACKS1 | SACKS0 |
| SADC2 | ADPGAEN | VBGEN | VREFIPS | VREFPS | SAVRS3 | SAVRS2 | SAVRS1 | SAVRS0 |
| ACERL | ACE7 | ACE6 | ACE5 | ACE4 | ACE3 | ACE2 | ACE1 | ACE0 |
| ACERH | — | — | — | — | ACE11 | ACE10 | ACE9 | ACE8 |

A/D Converter Register List

A/D Converter Data Registers – SADOL, SADOH

As the device contains an internal 12-bit A/D converter, it requires two data registers to store the converted value. These are a high byte register, known as SADOH, and a low byte register, known as SADOL. After the conversion process takes place, these registers can be directly read by the microcontroller to obtain the digitised conversion value. As only 12 bits of the 16-bit register space is utilised, the format in which the data is stored is controlled by the ADRFS bit in the SADC0 register as shown in the accompanying table. D0~D11 are the A/D conversion result data bits. Any unused bits will be read as zero. Note that the A/D converter data register contents will keep unchanged if the A/D converter is disabled.

| ADRFs | SADOH | | | | | | | | SADOL | | | | | | | |
|-------|-------|-----|----|----|-----|-----|----|----|-------|----|----|----|----|----|----|----|
| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| 0 | D11 | D10 | D9 | D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | 0 | 0 | 0 | 0 |
| 1 | 0 | 0 | 0 | 0 | D11 | D10 | D9 | D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |

A/D Converter Data Registers

A/D Converter Control Registers – SADC0, SADC1, SADC2, ACERL, ACERH

To control the function and operation of the A/D converter, several control registers known as SADC0, SADC1, SADC2, ACERL and ACERH are provided. These 8-bit registers define functions such as the selection of which analog channel is connected to the internal A/D converter, the digitised data format, the A/D converter clock source as well as controlling the start function and monitoring the A/D converter busy status. As the device contains only one actual analog to digital converter hardware circuit, each of the external and internal analog signals must be routed to the converter. The SAINS bit field in the SADC1 register and SACS bit field in the SADC0 register are used to which analog signal derived from the external or internal signals will be connected to the A/D converter. The A/D converter also contains programmable gain amplifier, PGA, to generate the A/D converter internal reference voltage. The overall operation of the PGA is controlled using the SADC2 register.

The analog input pin function selection bits in the ACERL and ACERH registers determine which pins on I/O ports are used as analog inputs for the A/D converter input and which pins are not to be used as the A/D converter input. When the pin is selected to be an A/D converter input, its original function whether it is an I/O or other pin-shared function will be removed. In addition, any internal pull-high resistor connected to the pin will be automatically removed if the pin is selected to be an A/D converter input.

• **SADC0 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-------|------|-------|-------|-------|-------|-------|-------|
| Name | START | ADBZ | ADCEN | ADRFS | SACS3 | SACS2 | SACS1 | SACS0 |
| R/W | R/W | R | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7 **START:** Start the A/D conversion
 0 → 1 → 0: Start A/D conversion
 This bit is used to initiate an A/D conversion process. The bit is normally low but if set high and then cleared low again, the A/D converter will initiate a conversion process.

Bit 6 **ADBZ:** A/D Converter busy flag
 0: No A/D conversion is in progress
 1: A/D conversion is in progress
 This read only flag is used to indicate whether the A/D conversion is in progress or not. When the START bit is set from low to high and then to low again, the ADBZ flag will be set high to indicate that the A/D conversion is initiated. The ADBZ flag will be cleared to zero after the A/D conversion is complete.

Bit 5 **ADCEN:** A/D Converter function enable control
 0: Disable
 1: Enable
 This bit controls the A/D converter internal function. This bit should be set high to enable the A/D converter. If the bit is set low, then the A/D converter will be switched off reducing the device power consumption. When the A/D converter function is disabled, the contents of the A/D converter data register pair, SADOH and SADOL, will keep unchanged.

Bit 4 **ADRFS:** A/D Converter data format control
 0: ADC output data format → SADOH=D[11:4]; SADOL=D[3:0]
 1: ADC output data format → SADOH=D[11:8]; SADOL=D[7:0]
 This bit controls the format of the 12-bit converted A/D converter value in the two A/D converter data registers. Details are provided in the A/D converter data register section.

Bit 3~0 **SACS3~SACS0:** A/D converter external analog input channel selection
 0000: AN0
 0001: AN1
 0010: AN2
 0011: AN3
 0100: AN4
 0101: AN5
 0110: AN6
 0111: AN7
 1000: AN8
 1001: AN9
 1010: AN10
 1011: AN11
 1100~1111: Undefined, input floating

• SADC1 Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|--------|--------|--------|---|---|--------|--------|--------|
| Name | SAINS2 | SAINS1 | SAINS0 | — | — | SACKS2 | SACKS1 | SACKS0 |
| R/W | R/W | R/W | R/W | — | — | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | — | — | 0 | 0 | 0 |

Bit 7~5 **SAINS2~SAINS0**: A/D converter input signal selection
 000, 100: External signal – External analog channel input, ANn
 001: Internal signal – Internal A/D converter power supply voltage V_{DD}
 010: Internal signal – Internal A/D converter power supply voltage $V_{DD}/2$
 011: Internal signal – Internal A/D converter power supply voltage $V_{DD}/4$
 101: Internal signal – Internal reference voltage V_R
 110: Internal signal – Internal reference voltage $V_R/2$
 111: Internal signal – Internal reference voltage $V_R/4$

When the internal analog signal is selected to be converted, the external channel input signal will automatically be switched off regardless of the SACKS bit field value. The internal reference voltage can be derived from various sources selected using the SAVRS3~SAVRS0 bits in the SADC2 register.

Bit 4~3 Unimplemented, read as “0”

Bit 2~0 **SACKS2~SACKS0**: A/D conversion clock source selection
 000: f_{SYS}
 001: $f_{SYS}/2$
 010: $f_{SYS}/4$
 011: $f_{SYS}/8$
 100: $f_{SYS}/16$
 101: $f_{SYS}/32$
 110: $f_{SYS}/64$
 111: $f_{SYS}/128$

• SADC2 Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---------|-------|---------|---------|--------|--------|--------|--------|
| Name | ADPGAEN | VBGEN | VREFIPS | VREFFPS | SAVRS3 | SAVRS2 | SAVRS1 | SAVRS0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7 **ADPGAEN**: A/D converter PGA enable/disable control
 0: Disable
 1: Enable

This bit controls the internal PGA function to provide various reference voltage for the A/D converter. When the bit is set high, the internal reference voltage, V_R , can be used as the internal converted signal or reference voltage by the A/D converter. If the internal reference voltage is not used by the A/D converter, then the PGA function should be properly configured to conserve power.

Bit 6 **VBGEN**: Internal Bandgap reference voltage enable control
 0: Disable
 1: Enable

This bit controls the internal Bandgap circuit on/off function to the A/D converter. When the bit is set high, the Bandgap reference voltage can be used by the A/D converter. If the Bandgap reference voltage is not used by the A/D converter and the LVD or LVR function is disabled, then the bandgap reference circuit will be automatically switched off to conserve power. When the Bandgap reference voltage is switched on for use by the A/D converter, a time, t_{BGS} , should be allowed for the Bandgap circuit to stabilise before implementing an A/D conversion.

- Bit 5 **VREFIPS**: VREFI pin selection
 0: Disable – VREFI pin is not selected
 1: Enable – VREFI pin is selected
- Bit 4 **VREFPS**: VREF pin selection
 0: Disable – VREF pin is not selected
 1: Enable – VREF pin is selected
- Bit 3~0 **SAVRS3~SAVRS0**: A/D converter reference voltage selection
 0000: V_{DD}
 0001: V_{REFI}
 0010: $V_{REFI} \times 2$
 0011: $V_{REFI} \times 3$
 0100: $V_{REFI} \times 4$
 1001: Reserved, can not be used
 1010: $V_{BG} \times 2$
 1011: $V_{BG} \times 3$
 1100: $V_{BG} \times 4$
 1101~1111: V_{DD}

When the A/D converter reference voltage source is selected to derive from the A/D converter power supply V_{DD} or the internal V_{BG} voltage, the reference voltage which comes from the external VREFI pin will be automatically switched off.

• **ACERL Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|------|------|------|------|------|------|------|------|
| Name | ACE7 | ACE6 | ACE5 | ACE4 | ACE3 | ACE2 | ACE1 | ACE0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

- Bit 7 **ACE7**: Defines PB3 is A/D input or not
 0: Not A/D input
 1: A/D input, AN7
- Bit 6 **ACE6**: Defines PA7 is A/D input or not
 0: Not A/D input
 1: A/D input, AN6
- Bit 5 **ACE5**: Defines PA6 is A/D input or not
 0: Not A/D input
 1: A/D input, AN5
- Bit 4 **ACE4**: Defines PA5 is A/D input or not
 0: Not A/D input
 1: A/D input, AN4
- Bit 3 **ACE3**: Defines PA4 is A/D input or not
 0: Not A/D input
 1: A/D input, AN3
- Bit 2 **ACE2**: Defines PB2 is A/D input or not
 0: Not A/D input
 1: A/D input, AN2
- Bit 1 **ACE1**: Defines PB1 is A/D input or not
 0: Not A/D input
 1: A/D input, AN1
- Bit 0 **ACE0**: Defines PB0 is A/D input or not
 0: Not A/D input
 1: A/D input, AN0

• ACERH Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---|---|---|---|-------|-------|------|------|
| Name | — | — | — | — | ACE11 | ACE10 | ACE9 | ACE8 |
| R/W | — | — | — | — | R/W | R/W | R/W | R/W |
| POR | — | — | — | — | 0 | 0 | 0 | 0 |

- Bit 7~4 Unimplemented, read as “0”
- Bit 3 **ACE11**: Defines PA3 is A/D input or not
 0: Not A/D input
 1: A/D input, AN11
- Bit 2 **ACE10**: Defines PB6 is A/D input or not
 0: Not A/D input
 1: A/D input, AN10
- Bit 1 **ACE9**: Defines PB5 is A/D input or not
 0: Not A/D input
 1: A/D input, AN9
- Bit 0 **ACE8**: Defines PB4 is A/D input or not
 0: Not A/D input
 1: A/D input, AN8

A/D Converter Reference Voltage

The actual reference voltage supply to the A/D converter can be supplied from the positive power voltage, V_{DD} , an external reference source supplied on pin VREFI or an internal reference source derived from the Bandgap circuit. Then the selected reference voltage source can be amplified through a programmable gain amplifier except the voltage sourced from V_{DD} . The desired selection is made using the SAVRS3~SAVRS0 bits in the SADC2 register and relevant pin function selection bits. Note that the selected reference voltage can be output on the VREF pin which is pin-shared with other functions by setting the VREFPS bit in the SADC2 register. As the VREFI and VREF pins both are pin-shared with other functions, when the VREFI or VREF pin is selected as the reference voltage input or output pin, the VREFI or VREF pin function selection bit should first be properly configured to disable other pin-shared functions. However, if the A/D converter power supply or the internal Bandgap reference signal is selected as the reference source, the reference voltage which originates from the external VREFI pin will automatically be switched off by hardware. The internal Bandgap reference circuit should first be enabled before the V_{BG} is selected to be used. A specific start-up time is necessary for the Bandgap circuit to become stable and accurate.

Note that the analog input signal values must not be allowed to exceed the value of the selected A/D converter reference voltage.

| SAVRS[3:0] | Reference Source | Description |
|--------------------|------------------|--|
| 0000, 1101~1111 | V_{DD} | Internal A/D converter power supply voltage V_{DD} |
| 0001 | VREFI pin | External A/D converter reference voltage V_{REFI} |
| 0010 | | External A/D converter reference voltage $V_{REFI} \times 2$ |
| 0011 | | External A/D converter reference voltage $V_{REFI} \times 3$ |
| 0100 | | External A/D converter reference voltage $V_{REFI} \times 4$ |
| 1010 | Bandgap circuit | Internal Bandgap voltage $V_{BG} \times 2$ |
| 1011 | | Internal Bandgap voltage $V_{BG} \times 3$ |
| 1100 | | Internal Bandgap voltage $V_{BG} \times 4$ |

A/D Converter Reference Voltage Selection

A/D Converter Input Signals

All of the external A/D Converter analog input pins are pin-shared with the I/O pins as well as other functions. The corresponding pin-shared function selection bits in the ACERL and ACERH registers, determine whether the external input pins are setup as A/D converter analog channel inputs or whether they have other functions. If the corresponding pin is setup to be an A/D converter analog channel input, the original pin functions will be disabled. In this way, pins can be changed under program control to change their function between A/D inputs and other functions. All pull-high resistors, which are setup through register programming, will be automatically disconnected if the pins are setup as A/D inputs. Note that it is not necessary to first setup the A/D pin as an input in the port control register to enable the A/D converter input as when the relevant A/D converter input function selection bits enable an A/D converter input, the status of the port control register will be overridden.

The SAINS2~SAINS0 bits in the SADC1 register are used to determine that the analog signal to be converted comes from the external channel input or internal analog signal. The SACS3~SACS0 bits in the SADC0 register are used to determine which external channel input is selected to be converted. If the SAINS2~SAINS0 bits are set to “000” or “100”, the external channel input will be selected to be converted and the SACS bit field can determine which external channel is selected to be converted. When the SAINS field is set to the value of “x01”, “x10” or “x11”, one of the internal analog signals will be selected to be converted. The internal analog signals can be derived from the A/D converter supply power, V_{DD} , or internal reference voltage, V_R , with a specific ratio of 1, 1/2 or 1/4. If the internal analog signal is selected to be converted, the external channel signal input will automatically be switched off regardless of the SACS field value.

| SAINS[2:0] | SACS[3:0] | Input Signals | Description |
|------------|-----------|---------------|---|
| 000, 100 | 0000~1011 | AN0~AN11 | External channel analog input ANn |
| | 11xx | — | Floating, no external channel is selected |
| 001 | xxxx | V_{DD} | Internal A/D converter power supply voltage |
| 010 | xxxx | $V_{DD}/2$ | Internal A/D converter power supply voltage/2 |
| 011 | xxxx | $V_{DD}/4$ | Internal A/D converter power supply voltage/4 |
| 101 | xxxx | V_R | Internal reference voltage |
| 110 | xxxx | $V_R/2$ | Internal reference voltage/2 |
| 111 | xxxx | $V_R/4$ | Internal reference voltage/4 |

“x”: Don't care

A/D Converter Input Signal Selection

A/D Converter Operation

The START bit in the SADC0 register is used to start the A/D conversion. When the microcontroller sets this bit from low to high and then low again, an analog to digital conversion cycle will be initiated.

The ADBZ bit in the SADC0 register is used to indicate whether the analog to digital conversion process is in process or not. This bit will be automatically set to “1” by the microcontroller after an A/D conversion is successfully initiated. When the A/D conversion is complete, the ADBZ will be cleared to “0”. In addition, the corresponding A/D converter interrupt request flag will be set in the interrupt control register, and if the interrupts are enabled, an appropriate internal interrupt signal will be generated. This A/D converter internal interrupt signal will direct the program flow to the associated A/D converter internal interrupt address for processing. If the A/D converter internal interrupt is disabled, the microcontroller can be used to poll the ADBZ bit in the SADC0 register to check whether it has been cleared as an alternative method of detecting the end of an A/D conversion cycle.

The clock source for the A/D converter, which originates from the system clock f_{SYS} , can be chosen to be either f_{SYS} or a subdivided version of f_{SYS} . The division ratio value is determined by the SACKS2~SACKS0 bits in the SADC1 register. Although the A/D conversion clock source is determined by the system clock f_{SYS} , and by bits SACKS2~SACKS0, there are some limitations on the A/D conversion clock source speed that can be selected. As the recommended value of permissible A/D conversion clock period, t_{ADCK} , is from 0.5 μ s to 10 μ s, care must be taken for system clock frequencies. For example, if the system clock operates at a frequency of 8MHz, the SACKS2~SACKS0 bits should not be set to “000”, “001” or “111”. Doing so will give A/D conversion clock periods that are less than the minimum A/D conversion clock period or greater than the maximum A/D conversion clock period which may result in inaccurate A/D conversion values. Refer to the following table for examples, where values marked with an asterisk * special care must be taken.

| f_{SYS} | A/D Conversion Clock Period (t_{ADCK}) | | | | | | | |
|-----------|--|---------------------------------------|---------------------------------------|---------------------------------------|--|--|--|---|
| | SACKS[2:0] =000 (f_{SYS}) | SACKS[2:0] =001 ($f_{SYS}/2$) | SACKS[2:0] =010 ($f_{SYS}/4$) | SACKS[2:0] =011 ($f_{SYS}/8$) | SACKS[2:0] =100 ($f_{SYS}/16$) | SACKS[2:0] =101 ($f_{SYS}/32$) | SACKS[2:0] =110 ($f_{SYS}/64$) | SACKS[2:0] =111 ($f_{SYS}/128$) |
| 1MHz | 1 μ s | 2 μ s | 4 μ s | 8 μ s | 16 μ s* | 32 μ s* | 64 μ s* | 128 μ s* |
| 2MHz | 500ns | 1 μ s | 2 μ s | 4 μ s | 8 μ s | 16 μ s* | 32 μ s* | 64 μ s* |
| 4MHz | 250ns* | 500ns | 1 μ s | 2 μ s | 4 μ s | 8 μ s | 16 μ s* | 32 μ s* |
| 8MHz | 125ns* | 250ns* | 500ns | 1 μ s | 2 μ s | 4 μ s | 8 μ s | 16 μ s* |
| 12MHz | 83ns* | 167ns* | 333ns* | 667ns | 1.33 μ s | 2.67 μ s | 5.33 μ s | 10.67 μ s* |
| 16MHz | 62.5ns* | 125ns* | 250ns* | 500ns | 1 μ s | 2 μ s | 4 μ s | 8 μ s |

A/D Conversion Clock Period Examples

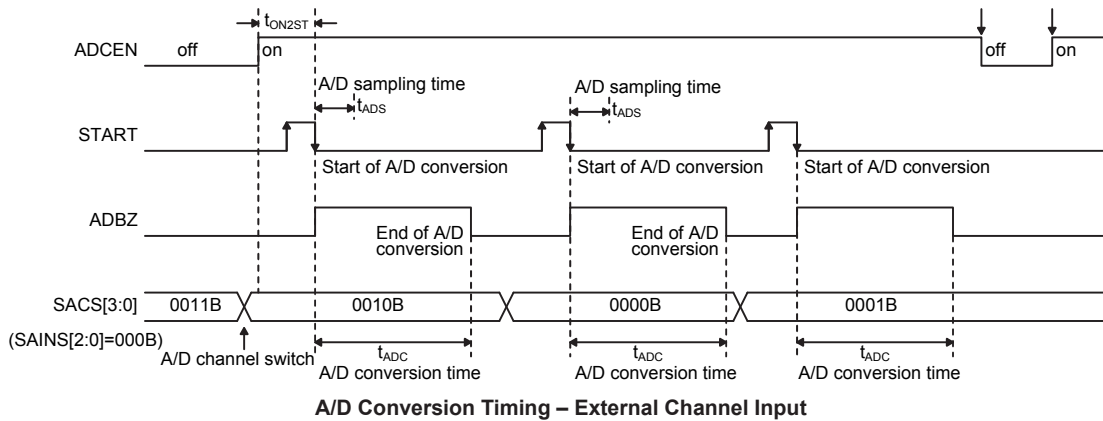
Controlling the power on/off function of the A/D conversion circuitry is implemented using the ADCEN bit in the SADC0 register. This bit must be set high to power on the A/D converter. When the ADCEN bit is set high to power on the A/D conversion internal circuitry, a certain delay as indicated in the timing diagram must be allowed before an A/D conversion is initiated. Even if no pins are selected for use as A/D converter inputs by configuring the corresponding pin control bits, if the ADCEN bit is high then some power will still be consumed. In power conscious applications it is therefore recommended that the ADCEN is set low to reduce power consumption when the A/D converter function is not being used.

A/D Conversion Rate and Timing Diagram

A complete A/D conversion contains two parts, data sampling and data conversion. The data sampling which is defined as t_{ADS} takes 4 A/D conversion clock cycles and the data conversion takes 12 A/D converter clock cycles. Therefore a total of 16 A/D conversion clock cycles for an A/D conversion which is defined as t_{ADC} are necessary.

$$\text{Maximum single A/D conversion rate} = \text{A/D conversion clock period} / 16$$

The accompanying diagram shows graphically the various stages involved in an analog to digital conversion process and its associated timing. After an A/D conversion process has been initiated by the application program, the microcontroller internal hardware will begin to carry out the conversion, during which time the program can continue with other functions. The time taken for the A/D conversion is $16t_{ADCK}$ clock cycles where t_{ADCK} is equal to the A/D conversion clock period.



Summary of A/D Conversion Steps

The following summarises the individual steps that should be executed in order to implement an A/D conversion process.

- Step 1
 Select the required A/D conversion clock by properly programming the SACKS2~SACKS0 bits in the SADC1 register.
- Step 2
 Enable the A/D converter by setting the ADCEN bit in the SADC0 register to “1”.
- Step 3
 Select which signal is to be connected to the internal A/D converter by correctly configuring the SAINS and SACS bit fields.
 Select the external channel input to be converted, go to Step 4.
 Select the internal analog signal to be converted, go to Step 5.
- Step 4
 If the A/D input signal comes from the external channel input selected by configuring the SAINS bit field, the corresponding pin should first be configured as an A/D input function by configuring the relevant pin function control bits. The desired external channel input should be selected by configuring the SACS field. After this step, go to Step 6.
- Step 5
 If the A/D input signal is selected to come from the internal analog signal, the SAINS field should be properly configured and then the external channel analog input will automatically be disconnected regardless of the SACS bit field value. After this step, go to Step 6.
- Step 6
 Select the reference voltage source by configuring the SAVRS3~SAVRS0 bits in the SADC2 register. Enable the PGA if the PGA output voltage is selected as the A/D converter reference voltage. If the PGA input signal comes from the external VREFI pin, the corresponding pin function selection bit should first be properly configured. If the A/D converter power supply voltage, internal Bandgap circuit voltage or its multiplication is selected, the reference voltage sourced from the external input pin VREFI will automatically be switched off.
- Step 7
 Select A/D converter output data format by configuring the ADRFS bit in the SADC0 register.

- Step 8
If A/D converter interrupt is used, the interrupt control registers must be correctly configured to ensure the A/D converter interrupt function is active. The master interrupt control bit, EMI, and the A/D converter interrupt bits, ADE, must both set high in advance.
- Step 9
The A/D conversion procedure can now be initialised by setting the START bit from low to high and then low again.
- Step 10
If A/D conversion is in progress, the ADBZ flag will be set high. After the A/D conversion process is completed, the ADBZ flag will go low and then output data can be read from the SADOH and SADOL registers. If the ADC interrupt is enabled and the stack is not full, data can be acquired by interrupt service program.

Note: When checking for the end of the conversion process, if the method of polling the ADBZ bit in the SADC0 register is used, the interrupt enable step above can be omitted.

Programming Considerations

During microcontroller operations where the A/D converter is not being used, the A/D conversion internal circuitry can be switched off to reduce power consumption by clearing the ADCEN bit in the SADC0 register. When this happens, the internal A/D conversion circuits will not consume power irrespective of what analog voltage is applied to their input lines. If the A/D converter input lines are used as normal I/Os, then care must be taken as if the input voltage is not at a valid logic level, then this may lead to some increase in power consumption.

A/D Conversion Function

As the device contains a 12-bit A/D converter, its full-scale converted digitised value is equal to FFFH. Since the full-scale analog input value is equal to the actual A/D converter reference voltage, V_{REF} , this gives a single bit analog input value of V_{REF} divided by 4096.

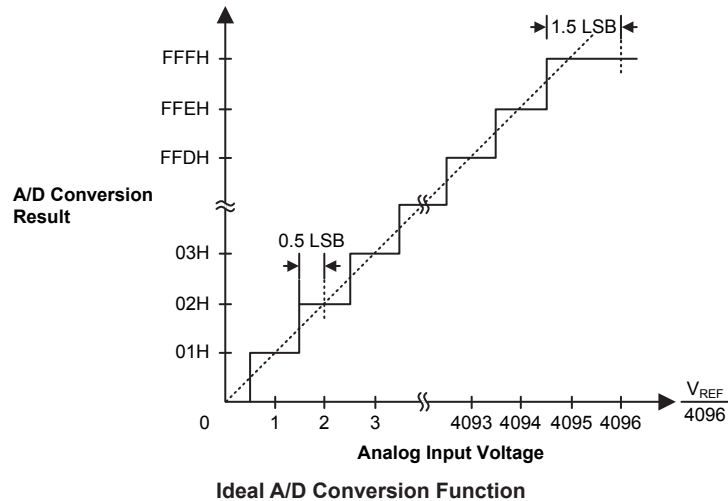
$$1 \text{ LSB} = V_{REF} \div 4096$$

The A/D Converter input voltage value can be calculated using the following equation:

$$\text{A/D converter input voltage} = \text{A/D converter output digital value} \times V_{REF} \div 4096$$

The diagram shows the ideal transfer function between the analog input value and the digitised output value for the A/D converter. Except for the digitised zero value, the subsequent digitised values will change at a point 0.5 LSB below where they would change without the offset, and the last full scale digitised value will change at a point 1.5 LSB below the V_{REF} level. Note that here the V_{REF} voltage is the actual A/D converter reference voltage determined by the SAVRS field.

Note that here the V_{REF} voltage is the actual A/D converter reference voltage determined by the SAVRS bit field.



A/D Converter Programming Examples

The following two programming examples illustrate how to setup and implement an A/D conversion. In the first example, the method of polling the ADBZ bit in the SADC0 register is used to detect when the conversion cycle is complete, whereas in the second example, the A/D converter interrupt is used to determine when the conversion is complete.

Example: using an ADBZ polling method to detect the end of conversion

```

clr ADE                ; disable ADC interrupt
mov a,03H              ; select fsys/8 as A/D clock and A/D input signal comes from
                      ; external channel

mov SADC1,a
mov a,00H              ; select VDD as A/D reference voltage source
mov SADC2,a
mov a,01h              ; setup ACERL to configure pin AN0
mov ACERL,a
mov a,20h              ; enable A/D converter and select AN0 external channel input
mov SADC0,a
:
start_conversion:
clr START              ; high pulse on start bit to initiate conversion
set START              ; reset A/D converter
clr START              ; start A/D conversion
polling_EOC:
sz ADBZ                ; poll the SADC0 register ADBZ bit to detect end of A/D conversion
jmp polling_EOC        ; continue polling
mov a,SADOL             ; read low byte conversion result value
mov SADOL_buffer,a    ; save result to user defined register
mov a,SADOH            ; read high byte conversion result value
mov SADOH_buffer,a    ; save result to user defined register
:
jmp start_conversion   ; start next A/D conversion

```

Example: using the interrupt method to detect the end of conversion

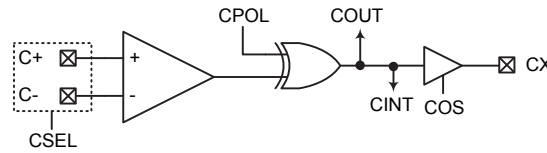
```

clr ADE          ; disable ADC interrupt
mov a,03H        ; select fsys/8 as A/D clock and A/D input signal comes from external channel
mov SADC1,a
mov a,00H        ; select VDD as A/D reference voltage source
mov SADC2,a
mov a,01h        ; setup ACERL to configure pin AN0
mov ACERL,a
mov a,20h        ; enable A/D converter and select AN0 external channel input
mov SADC0,a
:
Start_conversion:
clr START        ; high pulse on START bit to initiate conversion
set START        ; reset A/D converter
clr START        ; start A/D conversion
clr ADF          ; clear ADC interrupt request flag
set ADE          ; enable ADC interrupt
set EMI          ; enable global interrupt
:
:
; ADC interrupt service routine
ADC_ISR:
mov acc_stack,a ; save ACC to user defined memory
mov a,STATUS
mov status_stack,a ; save STATUS to user defined memory
:
:
mov a,SADOL      ; read low byte conversion result value
mov SADOL_buffer,a ; save result to user defined register
mov a,SADOH      ; read high byte conversion result value
mov SADOH_buffer,a ; save result to user defined register
:
:
EXIT_INT_ISR:
mov a,status_stack
mov STATUS,a    ; restore STATUS from user defined memory
mov a,acc_stack ; restore ACC from user defined memory
reti

```

Comparator

An analog comparator is contained within the device. The comparator function offers flexibility via their register controlled features such as power-down, polarity select, hysteresis etc. In sharing their pins with normal I/O pins the comparators do not waste precious I/O pins if there functions are otherwise unused.



Comparator

Comparator Operation

The device contains a comparator function which is used to compare two analog voltages and provide an output based on their input difference. Full control over the internal comparator is provided via the control register, CPC. The comparator output is recorded via a bit in the control register, but can also be transferred output onto a shared I/O pin. Additional comparator functions include polarity, hysteresis function and power down control.

Any pull-high resistors connected to the shared comparator input pins will be automatically disconnected when the corresponding comparator functional pins are selected. As the comparator inputs approach their switching level, some spurious output signals may be generated on the comparator output due to the slow rising or falling nature of the input signals. This can be minimised by the hysteresis function which will apply a small amount of positive feedback to the comparator. Ideally the comparator should switch at the point where the positive and negative inputs signals are at the same voltage level. However, unavoidable input offsets introduce some uncertainties here. The offset calibration function, if executed, will minimize the switching offset value.

• CPC Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|------|-----|------|------|-----|--------|--------|-------|
| Name | CSEL | CEN | CPOL | COUT | COS | CMPEG1 | CMPEG0 | CHYEN |
| R/W | R/W | R/W | R/W | R | R/W | R/W | R/W | R/W |
| POR | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 |

Bit 7 **CSEL:** Comparator pins or I/O pins selection

0: I/O pin selected

1: Comparator input pin C+ and C- selected

This is the Comparator input pin or I/O pin selection bit. If the bit is high the comparator input pins will be enabled. As a result, these two pins will lose their I/O pin functions. Any pull-high configurations associated with the comparator shared pins will also be automatically disconnected.

Bit 6 **CEN:** Comparator On/Off control

0: Off

1: On

This is the Comparator on/off control bit. If the bit is zero the comparator will be switched off and no power consumed even if analog voltages are applied to its inputs. For power sensitive applications this bit should be cleared to zero if the comparator is not used or before the device enters the SLEEP or IDLE mode.

- Bit 5 **CPOL**: Comparator output polarity control
 0: Non-invert
 1: Invert
 This is the comparator polarity control bit. If the bit is zero then the COUT bit will reflect the non-inverted output condition of the comparator. If the bit is high the comparator COUT bit will reflect the inverted output condition of the comparator.
- Bit 4 **COUT**: Comparator output bit
 CPOL=0
 0: C+ < C-
 1: C+ > C-
 CPOL=1
 0: C+ > C-
 1: C+ < C-
 This bit stores the comparator output bit. The polarity of the bit is determined by the voltages on the comparator inputs and by the condition of the CPOL bit.
- Bit 3 **COS**: Comparator output path selection
 0: CX pin – compare output can output to CX pin
 1: I/O pin – compare output is only internally used
- Bit 2~1 **CMPEG1~CMPEG0**: Comparator output interrupt edge trigger selection
 00: Rising edge – comparator interrupt will be generated if the COUT state changes from 0 to 1
 01: Falling edge – comparator interrupt will be generated if the COUT state changes from 1 to 0
 1x: Both edge – comparator interrupt will be generated if the COUT state changes from 0 to 1 or 1 to 0
- Bit 0 **CHYEN**: Comparator hysteresis function control
 0: Off
 1: On
 This is the hysteresis control bit and if set high will apply a limited amount of hysteresis to the comparator, as specified in the Comparator Electrical Characteristics table. The positive feedback induced by hysteresis reduces the effect of spurious switching near the comparator threshold.

Comparator Interrupt

The comparator possesses its own interrupt function. When the comparator output changes state, its relevant interrupt flag will be set, and if the corresponding interrupt enable bit is set, then a jump to its relevant interrupt vector will be executed. Note that it is the changing state of the COUT bit and not the output pin which generates an interrupt. If the microcontroller is in the SLEEP or IDLE Mode and the comparator is enabled, then if the external input lines cause the comparator output bit to change state, the resulting generated interrupt flag will also generate a wake-up. If it is required to disable a wake-up from occurring, then the interrupt flag should be first set high before entering the SLEEP or IDLE Mode.

Programming Considerations

If the comparator is enabled, it will remain active when the microcontroller enters the SLEEP or IDLE Mode, however as it will consume a certain amount of power, the user may wish to consider disabling it before the SLEEP or IDLE Mode is entered. As comparator pins are shared with normal I/O pins the I/O registers for these pins will be read as zero (port control register is “1”) or read as port data register value (port control register is “0”) if the comparator function is enabled.

Serial Interface Module – SIM

The device contains a Serial Interface Module, which includes both the four-line SPI interface or two-line I²C interface types, to allow an easy method of communication with external peripheral hardware. Having relatively simple communication protocols, these serial interface types allow the microcontroller to interface to external SPI or I²C based hardware such as sensors, Flash or EEPROM memory, etc. The SIM interface pins are pin-shared with other I/O pins and therefore the SIM interface functional pins must first be selected using the pin-remapping selection bits in the IFS register and the relevant control bit SIMEN bit in the SIMC0 register. As both interface types share the same pins and registers, the choice of whether the SPI or I²C type is used is made using the SIM operating mode control bits, named SIM2~SIM0, in the SIMC0 register. These pull-high resistors of the SIM pin-shared I/O pins except the \overline{SCS} pin, are selected using pull-high control registers if the SIM function is enabled and the corresponding pins are used as SIM input pins. Note that if the \overline{SCS} pin function is enabled, the corresponding internal pull-high resistor will be disconnected and cannot be controlled using the pull-high control register.

SPI Interface

The SPI interface is often used to communicate with external peripheral devices such as sensors, Flash or EEPROM memory devices, etc. Originally developed by Motorola, the four line SPI interface is a synchronous serial data interface that has a relatively simple communication protocol simplifying the programming requirements when communicating with external hardware devices.

The communication is full duplex and operates as a slave/master type, where the device can be either master or slave. Although the SPI interface specification can control multiple slave devices from a single master, the device provides only one \overline{SCS} pin. If the master needs to control multiple slave devices from a single master, the master can use I/O pin to select the slave devices.

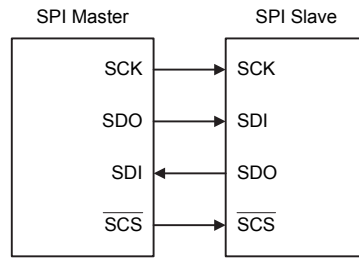
SPI Interface Operation

The SPI interface is a full duplex synchronous serial data link. It is a four line interface with pin names SDI, SDO, SCK and \overline{SCS} . Pins SDI and SDO are the Serial Data Input and Serial Data Output lines, SCK is the Serial Clock line and \overline{SCS} is the Slave Select line. As the SPI interface pins are pin-shared with normal I/O pins and with the I²C function pins, the SPI interface pins must first be selected by setting the correct bits in the SIMC0 and SIMC2 registers and the relevant pin-remapping control bits. After the desired SPI configuration has been set it can be disabled or enabled using the SIMEN bit in the SIMC0 register. Communication between devices connected to the SPI interface is carried out in a slave/master mode with all data transfer initiations being implemented by the master. The Master also controls the clock signal. As the device only contains a single \overline{SCS} pin only one slave device can be utilized. The \overline{SCS} pin is controlled by software, set CSEN bit to 1 to enable \overline{SCS} pin function, set CSEN bit to 0 the \overline{SCS} pin will be floating state.

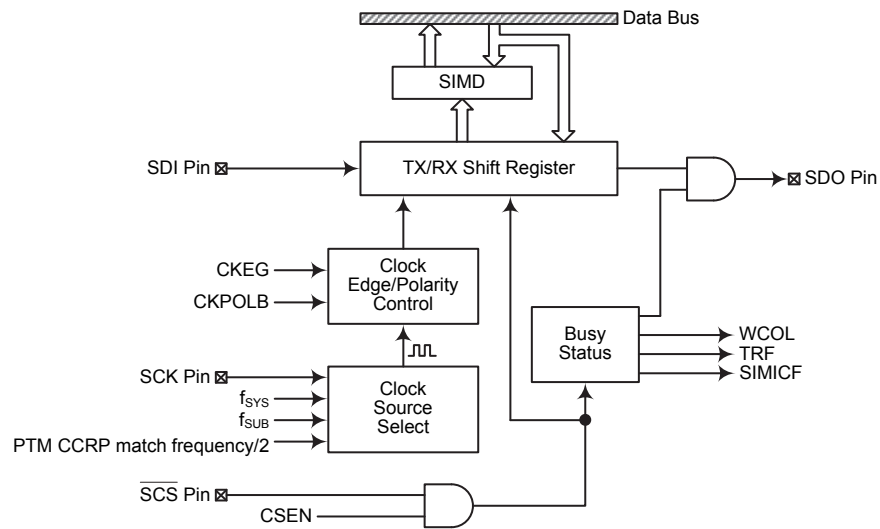
The SPI function in this device offers the following features:

- Full duplex synchronous data transfer
- Both Master and Slave modes
- LSB first or MSB first data transmission modes
- Transmission complete flag
- Rising or falling active clock edge

The status of the SPI interface pins is determined by a number of factors such as whether the device is in the master or slave mode and upon the condition of certain control bits such as CSEN and SIMEN.



SPI Master/Slave Connection



SPI Block Diagram

SPI Registers

There are three internal registers which control the overall operation of the SPI interface. These are the SIMD data register and two registers SIMC0 and SIMC2. Note that the SIMC1 register is only used by the I²C interface.

| Register Name | Bit | | | | | | | |
|---------------|------|------|--------|------|---------|---------|-------|--------|
| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| SIMC0 | SIM2 | SIM1 | SIM0 | — | SIMDEB1 | SIMDEB0 | SIMEN | SIMICF |
| SIMC2 | D7 | D6 | CKPOLB | CKEG | MLS | CSEN | WCOL | TRF |
| SIMD | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |

SPI Register List

SIMD Data Register

The SIMD register is used to store the data being transmitted and received. The same register is used by both the SPI and I²C functions. Before the device writes data to the SPI bus, the actual data to be transmitted must be placed in the SIMD register. After the data is received from the SPI bus, the device can read it from the SIMD register. Any transmission or reception of data from the SPI bus must be made via the SIMD register.

• **SIMD Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-----|-----|-----|-----|-----|-----|-----|-----|
| Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | x | x | x | x | x | x | x | x |

"x": unknown

Bit 7~0 **D7~D0**: SIM data register bit 7 ~ bit 0

SIMD Control Register

There are also two control registers for the SPI interface, SIMC0 and SIMC2. Note that the SIMC2 register also has the name SIMA which is used by the I²C function. The SIMC1 register is not used by the SPI function, only by the I²C function. The SIMC0 register is used to control the enable/disable function and to set the data transmission clock frequency. The SIMC2 register is used for other control functions such as LSB/MSB selection, write collision flag, etc.

• **SIMC0 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|------|------|------|---|---------|---------|-------|--------|
| Name | SIM2 | SIM1 | SIM0 | — | SIMDEB1 | SIMDEB0 | SIMEN | SIMICF |
| R/W | R/W | R/W | R/W | — | R/W | R/W | R/W | R/W |
| POR | 1 | 1 | 1 | — | 0 | 0 | 0 | 0 |

Bit 7~5 **SIM2~SIM0**: SIM Operating Mode Control

- 000: SPI master mode; SPI clock is $f_{SYS}/4$
- 001: SPI master mode; SPI clock is $f_{SYS}/16$
- 010: SPI master mode; SPI clock is $f_{SYS}/64$
- 011: SPI master mode; SPI clock is f_{SUB}
- 100: SPI master mode; SPI clock is PTM CCRP match frequency/2
- 101: SPI slave mode
- 110: I²C slave mode
- 111: Undefined

These bits setup the overall operating mode of the SIM function. As well as selecting if the I²C or SPI function, they are used to control the SPI Master/Slave selection and the SPI Master clock frequency. The SPI clock is a function of the system clock but can also be chosen to be sourced from PTM and f_{SUB} . If the SPI Slave Mode is selected then the clock will be supplied by an external Master device.

Bit 4 Unimplemented, read as "0"

Bit 3~2 **SIMDEB1~SIMDEB0**: I²C Debounce Time Selection

These bits are only available when the SIM is configured to operate in the I²C mode. Refer to the I²C register section.

- Bit 1 **SIMEN:** SIM Enable Control
 0: Disable
 1: Enable
- The bit is the overall on/off control for the SIM interface. When the SIMEN bit is cleared to zero to disable the SIM interface, the SDI, SDO, SCK and \overline{SCS} , or SDA and SCL lines will lose their SPI or I²C function and the SIM operating current will be reduced to a minimum value. When the bit is high the SIM interface is enabled. The SIM configuration option must have first enabled the SIM interface for this bit to be effective. If the SIM is configured to operate as an SPI interface via the SIM2~SIM0 bits, the contents of the SPI control registers will remain at the previous settings when the SIMEN bit changes from low to high and should therefore be first initialised by the application program. If the SIM is configured to operate as an I²C interface via the SIM2~SIM0 bits and the SIMEN bit changes from low to high, the contents of the I²C control bits such as HTX and TXAK will remain at the previous settings and should therefore be first initialised by the application program while the relevant I²C flags such as HCF, HAAS, HBB, SRW and RXAK will be set to their default states.
- Bit 0 **SIMICF:** SIM Incomplete Flag
 0: SIM incomplete condition not occurred
 1: SIM incomplete condition occurred
- This bit is only available when the SIM is configured to operate in an SPI slave mode. If the SPI operates in the slave mode with the SIMEN and CSEN bits both being set to 1 but the \overline{SCS} line is pulled high by the external master device before the SPI data transfer is completely finished, the SIMICF bit will be set to 1 together with the TRF bit. When this condition occurs, the corresponding interrupt will occur if the interrupt function is enabled. However, the TRF bit will not be set to 1 if the SIMICF bit is set to 1 by software application program.

• **SIMC2 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-----|-----|--------|------|-----|------|------|-----|
| Name | D7 | D6 | CKPOLB | CKEG | MLS | CSEN | WCOL | TRF |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

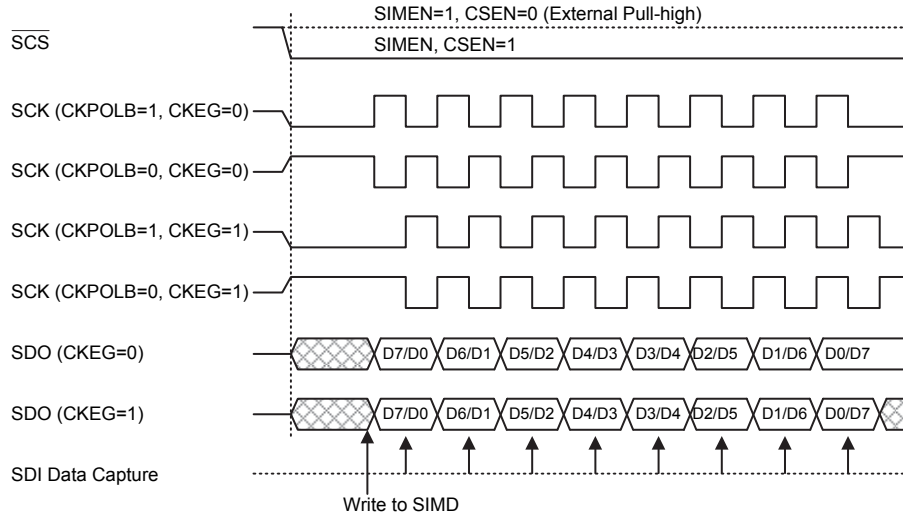
- Bit 7~6 **D7~D6:** Undefined bits
 These bits can be read or written by the application program.
- Bit 5 **CKPOLB:** SPI clock line base condition selection
 0: The SCK line will be high when the clock is inactive
 1: The SCK line will be low when the clock is inactive
- The CKPOLB bit determines the base condition of the clock line, if the bit is high, then the SCK line will be low when the clock is inactive. When the CKPOLB bit is low, then the SCK line will be high when the clock is inactive.
- Bit 4 **CKEG:** SPI SCK clock active edge type selection
 CKPOLB=0
 0: SCK is high base level and data capture at SCK rising edge
 1: SCK is high base level and data capture at SCK falling edge
 CKPOLB=1
 0: SCK is low base level and data capture at SCK falling edge
 1: SCK is low base level and data capture at SCK rising edge
- The CKEG and CKPOLB bits are used to setup the way that the clock signal outputs and inputs data on the SPI bus. These two bits must be configured before data transfer is executed otherwise an erroneous clock edge may be generated. The CKPOLB bit determines the base condition of the clock line, if the bit is high, then the SCK line will be low when the clock is inactive. When the CKPOLB bit is low, then the SCK line will be high when the clock is inactive. The CKEG bit determines active clock edge type which depends upon the condition of CKPOLB bit.

- Bit 3 **MLS:** SPI data shift order
 0: LSB first
 1: MSB first
This is the data shift select bit and is used to select how the data is transferred, either MSB or LSB first. Setting the bit high will select MSB first and low for LSB first.
- Bit 2 **CSEN:** SPI \overline{SCS} pin control
 0: Disable
 1: Enable
The CSEN bit is used as an enable/disable for the \overline{SCS} pin. If this bit is low, then the \overline{SCS} pin will be disabled and placed into a floating condition. If the bit is high, the \overline{SCS} pin will be enabled and used as a select pin.
- Bit 1 **WCOL:** SPI write collision flag
 0: No collision
 1: Collision
The WCOL flag is used to detect whether a data collision has occurred or not. If this bit is high, it means that data has been attempted to be written to the SIMD register during a data transfer operation. This writing operation will be ignored if data is being transferred. This bit can be cleared to zero by the application program.
- Bit 0 **TRF:** SPI Transmit/Receive complete flag
 0: SPI data is being transferred
 1: SPI data transfer is completed
The TRF bit is the Transmit/Receive Complete flag and is set to 1 automatically when an SPI data transfer is completed, but must be cleared to zero by the application program. It can be used to generate an interrupt.

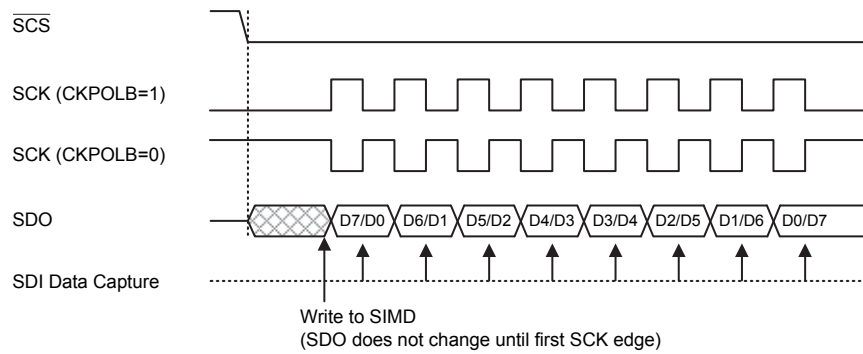
SPI Communication

After the SPI interface is enabled by setting the SIMEN bit high, then in the Master Mode, when data is written to the SIMD register, transmission/reception will begin simultaneously. When the data transfer is complete, the TRF flag will be set automatically, but must be cleared using the application program. In the Slave Mode, when the clock signal from the master has been received, any data in the SIMD register will be transmitted and any data on the SDI pin will be shifted into the SIMD register. The master should output a \overline{SCS} signal to enable the slave devices before a clock signal is provided. The slave data to be transferred should be well prepared at the appropriate moment relative to the SCK signal depending upon the configurations of the CKPOLB bit and CKEG bit. The accompanying timing diagram shows the relationship between the slave data and SCK signal for various configurations of the CKPOLB and CKEG bits.

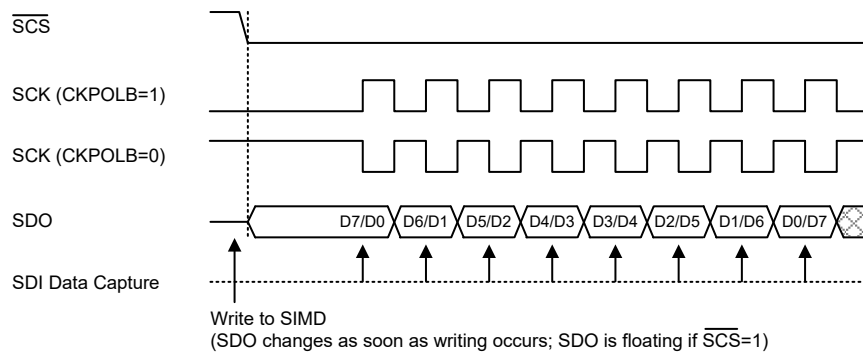
The SPI master mode will continue to function even in the IDLE Mode if the selected SPI clock source is running.



SPI Master Mode Timing

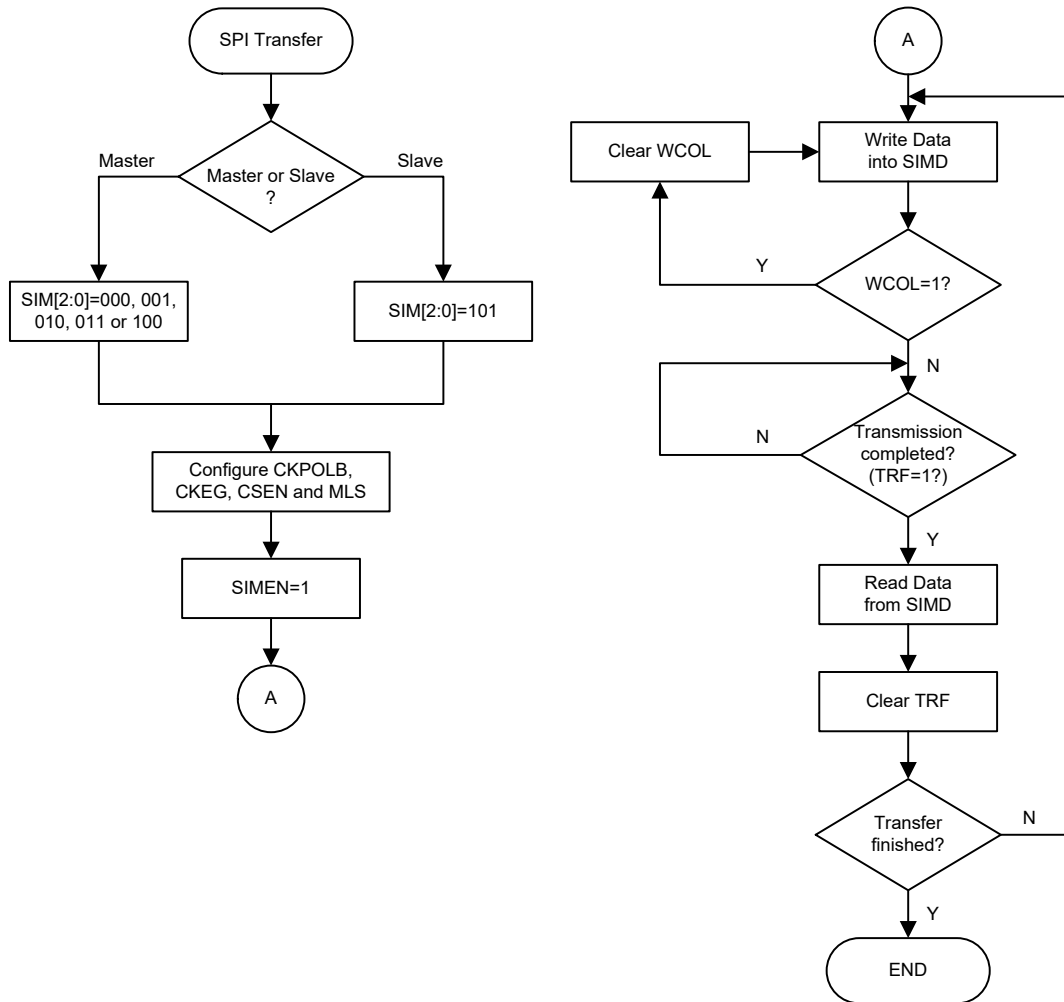


SPI Slave Mode Timing – CKEG=0



SPI Slave Mode Timing – CKEG=1

Note: For SPI slave mode, if SIMEN=1 and CSEN=0, SPI is always enabled and ignores the SCS level.



SPI Transfer Control Flow Chart

SPI Bus Enable/Disable

To enable the SPI bus, set CSEN=1 and \overline{SCS} =0, then wait for data to be written into the SIMD (TXRX buffer) register. For the Master Mode, after data has been written to the SIMD (TXRX buffer) register, then transmission or reception will start automatically. When all the data has been transferred, the TRF bit should be set. For the Slave Mode, when clock pulses are received on SCK, data in the TXRX buffer will be shifted out or data on SDI will be shifted in.

When the SPI bus is disabled, SCK, SDI, SDO and \overline{SCS} can become I/O pins or other pin-shared functions using the corresponding function control bits.

SPI Operation

All communication is carried out using the 4-line interface for either Master or Slave Mode.

The CSEN bit in the SIMC2 register controls the overall function of the SPI interface. Setting this bit high will enable the SPI interface by allowing the \overline{SCS} line to be active, which can then be used to control the SPI interface. If the CSEN bit is low, the SPI interface will be disabled and the \overline{SCS} line will be in a floating condition and can therefore not be used for control of the SPI interface. If the CSEN bit and the SIMEN bit in the SIMC0 are set high, this will place the SDI line in a floating condition and the SDO line high. If in Master Mode the SCK line will be either high or low

depending upon the clock polarity selection bit CKPOLB in the SIMC2 register. If in Slave Mode the SCK line will be in a floating condition. If the SIMEN bit is low, then the bus will be disabled and \overline{SCS} , SDI, SDO and SCK will all become I/O pins or the other functions. In the Master Mode the Master will always generate the clock signal. The clock and data transmission will be initiated after data has been written into the SIMD register. In the Slave Mode, the clock signal will be received from an external master device for both data transmission and reception. The following sequences show the order to be followed for data transfer in both Master and Slave Mode.

Master Mode

- Step 1
Select the SPI Master mode and clock source using the SIM2~SIM0 bits in the SIMC0 control register.
- Step 2
Setup the CSEN bit and setup the MLS bit to choose if the data is MSB or LSB first, this setting must be the same with the Slave devices.
- Step 3
Setup the SIMEN bit in the SIMC0 control register to enable the SPI interface.
- Step 4
For write operations: write the data to the SIMD register, which will actually place the data into the TXRX buffer. Then use the SCK and \overline{SCS} lines to output the data. After this, go to step5.
For read operations: the data transferred in on the SDI line will be stored in the TXRX buffer until all the data has been received at which point it will be latched into the SIMD register.
- Step 5
Check the WCOL bit if set high then a collision error has occurred so return to step 4. If equal to zero then go to the following step.
- Step 6
Check the TRF bit or wait for a SIM SPI serial bus interrupt.
- Step 7
Read data from the SIMD register.
- Step 8
Clear TRF.
- Step 9
Go to step 4.

Slave Mode

- Step 1
Select the SPI Slave mode using the SIM2~SIM0 bits in the SIMC0 control register
- Step 2
Setup the CSEN bit and setup the MLS bit to choose if the data is MSB or LSB first, this setting must be the same with the Master devices.
- Step 3
Setup the SIMEN bit in the SIMC0 control register to enable the SPI interface.
- Step 4
For write operations: write the data to the SIMD register, which will actually place the data into the TXRX buffer. Then wait for the master clock SCK and \overline{SCS} signal. After this, go to step5.
For read operations: the data transferred in on the SDI line will be stored in the TXRX buffer until all the data has been received at which point it will be latched into the SIMD register.

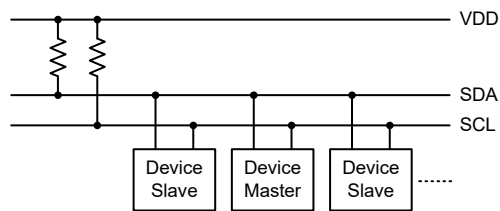
- Step 5
Check the WCOL bit if set high then a collision error has occurred so return to step 4. If equal to zero then go to the following step.
- Step 6
Check the TRF bit or wait for a SIM SPI serial bus interrupt.
- Step 7
Read data from the SIMD register.
- Step 8
Clear TRF.
- Step 9
Go to step 4.

Error Detection

The WCOL bit in the SIMC2 register is provided to indicate errors during data transfer. The bit is set by the SPI serial Interface but must be cleared by the application program. This bit indicates a data collision has occurred which happens if a write to the SIMD register takes place during a data transfer operation and will prevent the write operation from continuing.

I²C Interface

The I²C interface is used to communicate with external peripheral devices such as sensors, EEPROM memory etc. Originally developed by Philips, it is a two line low speed serial interface for synchronous serial data transfer. The advantage of only two lines for communication, relatively simple communication protocol and the ability to accommodate multiple devices on the same bus has made it an extremely popular interface type for many applications.

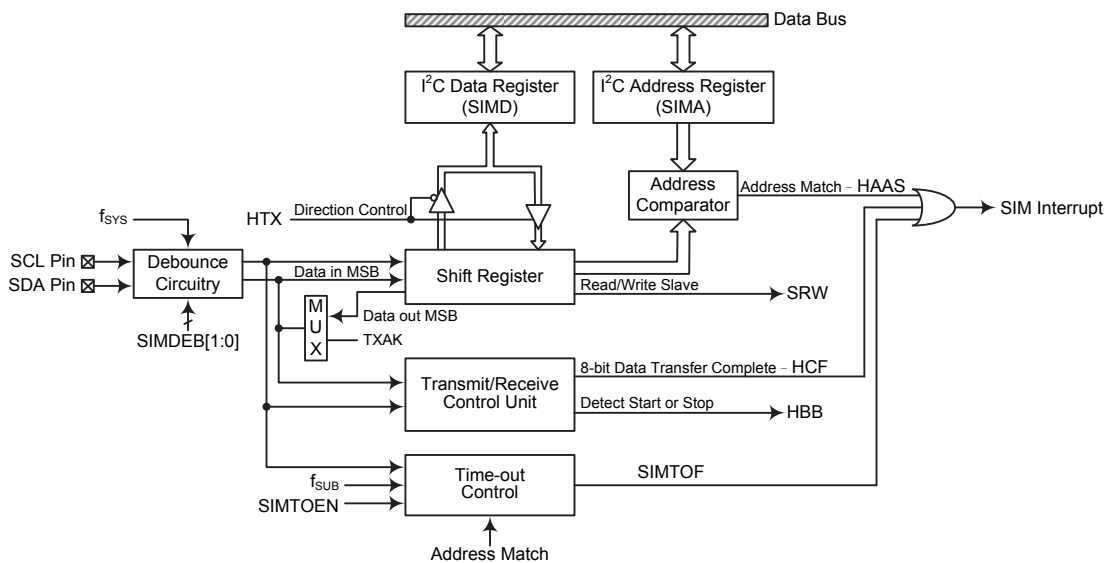


I²C Master Slave Bus Connection

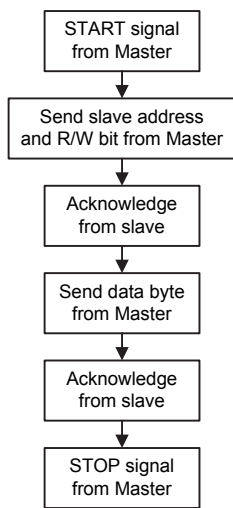
I²C interface Operation

The I²C serial interface is a two line interface, a serial data line, SDA, and serial clock line, SCL. As many devices may be connected together on the same bus, their outputs are both open drain types. For this reason it is necessary that external pull-high resistors are connected to these outputs. Note that no chip select line exists, as each device on the I²C bus is identified by a unique address which will be transmitted and received on the I²C bus.

When two devices communicate with each other on the bidirectional I²C bus, one is known as the master device and one as the slave device. Both master and slave can transmit and receive data, however, it is the master device that has overall control of the bus. For the device, which only operates in slave mode, there are two methods of transferring data on the I²C bus, the slave transmit mode and the slave receive mode.



I²C Block Diagram



I²C Interface Operation

The SIMDEB1 and SIMDEB0 bits determine the debounce time of the I²C interface. This uses the system clock to in effect add a debounce time to the external clock to reduce the possibility of glitches on the clock line causing erroneous operation. The debounce time, if selected, can be chosen to be either 2 or 4 system clocks. To achieve the required I²C data transfer speed, there exists a relationship between the system clock, f_{SYS} , and the I²C debounce time. For either the I²C Standard or Fast mode operation, users must take care of the selected system clock frequency and the configured debounce time to match the criterion shown in the following table.

| I ² C Debounce Time Selection | I ² C Standard Mode (100kHz) | I ² C Fast Mode (400kHz) |
|--|---|-------------------------------------|
| No Debounce | $f_{SYS} > 2\text{MHz}$ | $f_{SYS} > 5\text{MHz}$ |
| 2 system clock debounce | $f_{SYS} > 4\text{MHz}$ | $f_{SYS} > 10\text{MHz}$ |
| 4 system clock debounce | $f_{SYS} > 8\text{MHz}$ | $f_{SYS} > 20\text{MHz}$ |

I²C Minimum f_{SYS} Frequency Requirements

I²C Registers

There are three control registers associated with the I²C bus, SIMC0, SIMC1 and SIMTOC, one slave address register, SIMA, and one data register, SIMD. Note that the SIMA register also has the name SIMC2 which is used by the SPI function. Bit SIMEN and bits SIM2~SIM0 in register SIMC0 are used by the I²C interface.

| Register Name | Bit | | | | | | | |
|---------------|---------|--------|---------|---------|---------|---------|---------|---------|
| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| SIMC0 | SIM2 | SIM1 | SIM0 | — | SIMDEB1 | SIMDEB0 | SIMEN | SIMICF |
| SIMC1 | HCF | HAAS | HBB | HTX | TXAK | SRW | IAMWU | RXAK |
| SIMD | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| SIMA | SIMA6 | SIMA5 | SIMA4 | SIMA3 | SIMA2 | SIMA1 | SIMA0 | D0 |
| SIMTOC | SIMTOEN | SIMTOF | SIMTOS5 | SIMTOS4 | SIMTOS3 | SIMTOS2 | SIMTOS1 | SIMTOS0 |

I²C Register List

I²C Data Register

The SIMD register is used to store the data being transmitted and received. The same register is used by both the SPI and I²C functions. Before the device writes data to the I²C bus, the actual data to be transmitted must be placed in the SIMD register. After the data is received from the I²C bus, the device can read it from the SIMD register. Any transmission or reception of data from the I²C bus must be made via the SIMD register.

• SIMD Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-----|-----|-----|-----|-----|-----|-----|-----|
| Name | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | x | x | x | x | x | x | x | x |

"x": unknown

Bit 7~0 **D7~D0**: SIM data register bit 7 ~ bit 0

I²C Address Register

The SIMA register is also used by the SPI interface but has the name SIMC2. The SIMA register is the location where the 7-bit slave address of the slave device is stored. Bits 7~1 of the SIMA register define the device slave address. Bit 0 is not defined.

When a master device, which is connected to the I²C bus, sends out an address, which matches the slave address in the SIMA register, the slave device will be selected. Note that the SIMA register is the same register address as SIMC2 which is used by the SPI interface.

• SIMA Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-------|-------|-------|-------|-------|-------|-------|-----|
| Name | SIMA6 | SIMA5 | SIMA4 | SIMA3 | SIMA2 | SIMA1 | SIMA0 | D0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7~1 **SIMA6~SIMA0**: I²C slave address
SIMA6~SIMA0 is the I²C slave address bit 6~bit 0.

Bit 0 **D0**: Reserved bit, can be read or written by the application program.

I²C Control Register

There are also three control registers for the I²C interface, SIMC0, SIMC1 and SIMTOC. The register SIMC0 is used to control the enable/disable function and to set the data transmission clock frequency. The SIMC1 register contains the relevant flags which are used to indicate the I²C communication status. The SIMTOC register is used to control the I²C bus time-out function which is described in the I²C Time-out Control section.

• SIMC0 Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|------|------|------|---|---------|---------|-------|--------|
| Name | SIM2 | SIM1 | SIM0 | — | SIMDEB1 | SIMDEB0 | SIMEN | SIMICF |
| R/W | R/W | R/W | R/W | — | R/W | R/W | R/W | R/W |
| POR | 1 | 1 | 1 | — | 0 | 0 | 0 | 0 |

Bit 7~5 **SIM2~SIM0: SIM Operating Mode Control**
 000: SPI master mode; SPI clock is $f_{SYS}/4$
 001: SPI master mode; SPI clock is $f_{SYS}/16$
 010: SPI master mode; SPI clock is $f_{SYS}/64$
 011: SPI master mode; SPI clock is f_{SUB}
 100: SPI master mode; SPI clock is PTM CCRP match frequency/2
 101: SPI slave mode
 110: I²C slave mode
 111: Undefined

These bits setup the overall operating mode of the SIM function. As well as selecting if the I²C or SPI function, they are used to control the SPI Master/Slave selection and the SPI Master clock frequency. The SPI clock is a function of the system clock but can also be chosen to be sourced from PTM and f_{SUB} . If the SPI Slave Mode is selected then the clock will be supplied by an external Master device.

Bit 4 Unimplemented, read as “0”

Bit 3~2 **SIMDEB1~SIMDEB0: I²C Debounce Time Selection**
 00: No debounce
 01: 2 system clock debounce
 1x: 4 system clock debounce

These bits are used to select the I²C debounce time when the SIM is configured as the I²C interface function by setting the SIM2~SIM0 bits to “110”.

Bit 1 **SIMEN: SIM Enable Control**
 0: Disable
 1: Enable

The bit is the overall on/off control for the SIM interface. When the SIMEN bit is cleared to zero to disable the SIM interface, the SDI, SDO, SCK and \overline{SCS} , or SDA and SCL lines will lose their SPI or I²C function and the SIM operating current will be reduced to a minimum value. When the bit is high the SIM interface is enabled. The SIM configuration option must have first enabled the SIM interface for this bit to be effective. If the SIM is configured to operate as an SPI interface via the SIM2~SIM0 bits, the contents of the SPI control registers will remain at the previous settings when the SIMEN bit changes from low to high and should therefore be first initialised by the application program. If the SIM is configured to operate as an I²C interface via the SIM2~SIM0 bits and the SIMEN bit changes from low to high, the contents of the I²C control bits such as HTX and TXAK will remain at the previous settings and should therefore be first initialised by the application program while the relevant I²C flags such as HCF, HAAS, HBB, SRW and RXAK will be set to their default states.

Bit 0 **SIMICF: SIM Incomplete Flag**
 This bit is only available when the SIM is configured to operate in an SPI slave mode. Refer to the SPI register section.

• **SIMC1 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-----|------|-----|-----|------|-----|-------|------|
| Name | HCF | HAAS | HBB | HTX | TXAK | SRW | IAMWU | RXAK |
| R/W | R | R | R | R/W | R/W | R/W | R/W | R |
| POR | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 |

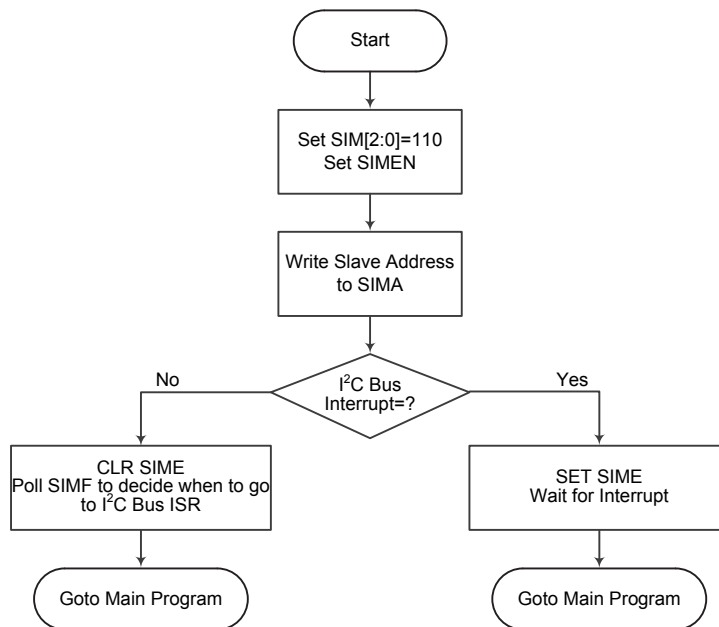
- Bit 7 HCF:** I²C Bus data transfer completion flag
 0: Data is being transferred
 1: Completion of an 8-bit data transfer
 The HCF flag is the data transfer flag. This flag will be zero when data is being transferred. Upon completion of an 8-bit data transfer the flag will go high and an interrupt will be generated.
- Bit 6 HAAS:** I²C Bus address match flag
 0: Not address match
 1: Address match
 The HAAS flag is the address match flag. This flag is used to determine if the slave device address is the same as the master transmit address. If the addresses match then this bit will be high, if there is no match then the flag will be low.
- Bit 5 HBB:** I²C Bus busy flag
 0: I²C Bus is not busy
 1: I²C Bus is busy
 The HBB flag is the I²C busy flag. This flag will be “1” when the I²C bus is busy which will occur when a START signal is detected. The flag will be set to “0” when the bus is free which will occur when a STOP signal is detected.
- Bit 4 HTX:** I²C slave device transmitter/receiver selection
 0: Slave device is the receiver
 1: Slave device is the transmitter
- Bit 3 TXAK:** I²C bus transmit acknowledge flag
 0: Slave send acknowledge flag
 1: Slave does not send acknowledge flag
 The TXAK bit is the transmit acknowledge flag. After the slave device receipt of 8-bits of data, this bit will be transmitted to the bus on the 9th clock from the slave device. The slave device must always clear the TXAK bit to “0” before further data is received.
- Bit 2 SRW:** I²C slave read/write flag
 0: Slave device should be in receive mode
 1: Slave device should be in transmit mode
 The SRW flag is the I²C Slave Read/Write flag. This flag determines whether the master device wishes to transmit or receive data from the I²C bus. When the transmitted address and slave address is match, that is when the HAAS flag is set high, the slave device will check the SRW flag to determine whether it should be in transmit mode or receive mode. If the SRW flag is high, the master is requesting to read data from the bus, so the slave device should be in transmit mode. When the SRW flag is zero, the master will write data to the bus, therefore the slave device should be in receive mode to read this data.
- Bit 1 IAMWU:** I²C Address Match Wake-Up control
 0: Disable
 1: Enable
 This bit should be set to 1 to enable the I²C address match wake up from the SLEEP or IDLE Mode. If the IAMWU bit has been set before entering either the SLEEP or IDLE mode to enable the I²C address match wake up, then this bit must be cleared to zero by the application program after wake-up to ensure correction device operation.
- Bit 0 RXAK:** I²C bus receive acknowledge flag
 0: Slave receives acknowledge flag
 1: Slave does not receive acknowledge flag

The RXAK flag is the receiver acknowledge flag. When the RXAK flag is “0”, it means that a acknowledge signal has been received at the 9th clock, after 8 bits of data have been transmitted. When the slave device in the transmit mode, the slave device checks the RXAK flag to determine if the master receiver wishes to receive the next byte. The slave transmitter will therefore continue sending out data until the RXAK flag is “1”. When this occurs, the slave transmitter will release the SDA line to allow the master to send a STOP signal to release the I²C Bus.

I²C Bus Communication

Communication on the I²C bus requires four separate steps, a START signal, a slave device address transmission, a data transmission and finally a STOP signal. When a START signal is placed on the I²C bus, all devices on the bus will receive this signal and be notified of the imminent arrival of data on the bus. The first seven bits of the data will be the slave address with the first bit being the MSB. If the address of the slave device matches that of the transmitted address, the HAAS bit in the SIMC1 register will be set and an SIM I²C interrupt will be generated. After entering the interrupt service routine, the slave device must first check the condition of the HAAS and SIMTOF bits to determine whether the interrupt source originates from an address match, 8-bit data transfer completion or I²C bus time-out occurrence. During a data transfer, note that after the 7-bit slave address has been transmitted, the following bit, which is the 8th bit, is the read/write bit whose value will be placed in the SRW bit. This bit will be checked by the slave device to determine whether to go into transmit or receive mode. Before any transfer of data to or from the I²C bus, the microcontroller must initialise the bus, the following are steps to achieve this:

- Step 1
Set the SIM2~SIM0 bits to “110” and SIMEN bit to “1” in the SIMC0 register to enable the I²C bus.
- Step 2
Write the slave address of the device to the I²C bus address register SIMA.
- Step 3
Set the SIME interrupt enable bit of the interrupt control register to enable the SIM interrupt.



I²C Bus Initialisation Flow Chart

I²C Bus Start Signal

The START signal can only be generated by the master device connected to the I²C bus and not by the slave device. This START signal will be detected by all devices connected to the I²C bus. When detected, this indicates that the I²C bus is busy and therefore the HBB bit will be set. A START condition occurs when a high to low transition on the SDA line takes place when the SCL line remains high.

I²C Slave Address

The transmission of a START signal by the master will be detected by all devices on the I²C bus. To determine which slave device the master wishes to communicate with, the address of the slave device will be sent out immediately following the START signal. All slave devices, after receiving this 7-bit address data, will compare it with their own 7-bit slave address. If the address sent out by the master matches the internal address of the microcontroller slave device, then an internal SIM I²C bus interrupt signal will be generated. The next bit following the address, which is the 8th bit, defines the read/write status and will be saved to the SRW bit of the SIMC1 register. The slave device will then transmit an acknowledge bit, which is a low level, as the 9th bit. The slave device will also set the status flag HAAS when the addresses match.

As an SIM I²C bus interrupt signal can come from three sources, when the program enters the interrupt subroutine, the HAAS and SIMTOF bits should be examined to see whether the interrupt source has come from a matching slave address, the completion of a data byte transfer or the I²C bus time-out occurrence. When a slave address is matched, the device must be placed in either the transmit mode and then write data to the SIMD register, or in the receive mode where it must implement a dummy read from the SIMD register to release the SCL line.

I²C Bus Read/Write Signal

The SRW bit in the SIMC1 register defines whether the master device wishes to read data from the I²C bus or write data to the I²C bus. The slave device should examine this bit to determine if it is to be a transmitter or a receiver. If the SRW flag is “1” then this indicates that the master device wishes to read data from the I²C bus, therefore the slave device must be setup to send data to the I²C bus as a transmitter. If the SRW flag is “0” then this indicates that the master wishes to send data to the I²C bus, therefore the slave device must be setup to read data from the I²C bus as a receiver.

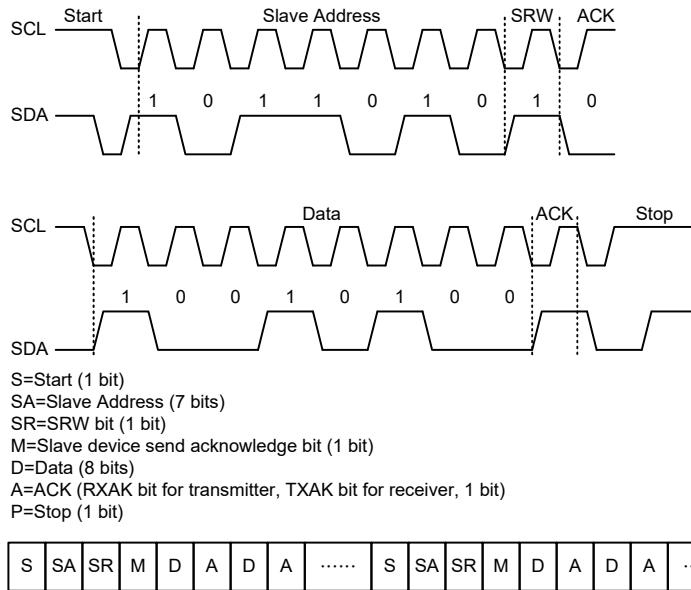
I²C Bus Slave Address Acknowledge Signal

After the master has transmitted a calling address, any slave device on the I²C bus, whose own internal address matches the calling address, must generate an acknowledge signal. The acknowledge signal will inform the master that a slave device has accepted its calling address. If no acknowledge signal is received by the master then a STOP signal must be transmitted by the master to end the communication. When the HAAS flag is high, the addresses have matched and the slave device must check the SRW flag to determine if it is to be a transmitter or a receiver. If the SRW flag is high, the slave device should be setup to be a transmitter so the HTX bit in the SIMC1 register should be set to “1”. If the SRW flag is low, then the microcontroller slave device should be setup as a receiver and the HTX bit in the SIMC1 register should be set to “0”.

I²C Bus Data and Acknowledge Signal

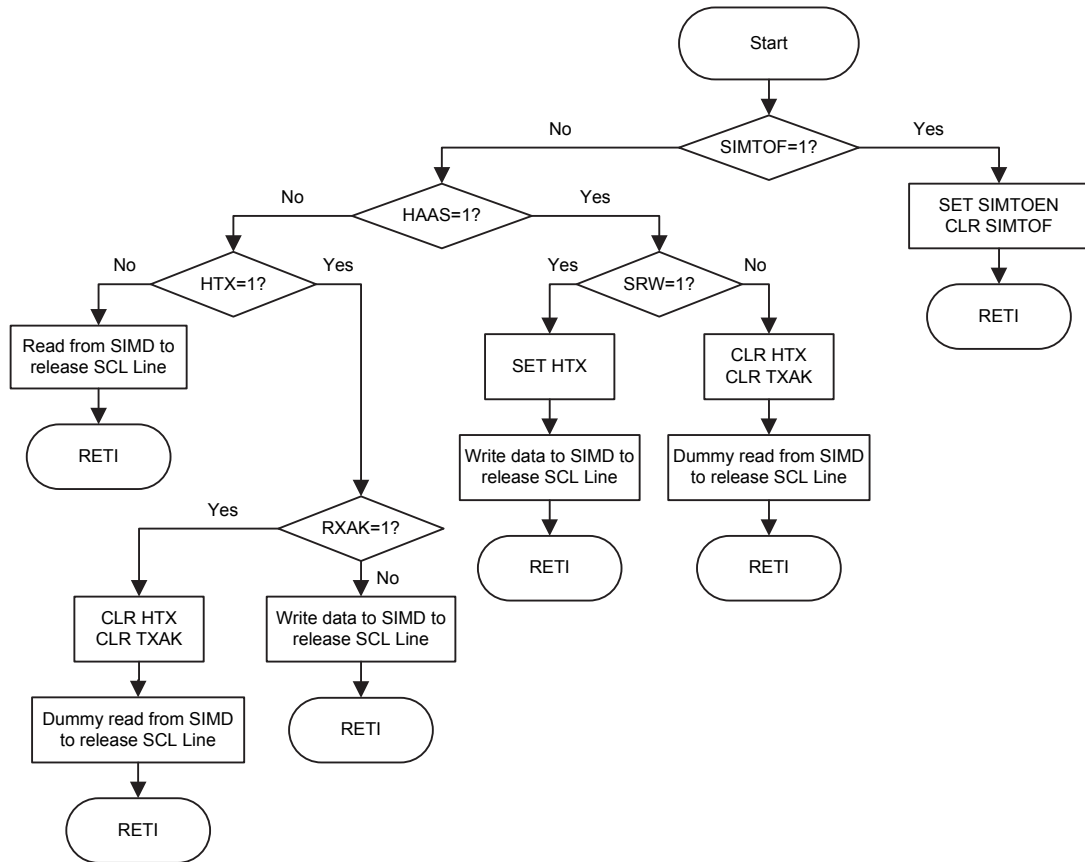
The transmitted data is 8-bits wide and is transmitted after the slave device has acknowledged receipt of its slave address. The order of serial bit transmission is the MSB first and the LSB last. After receipt of 8-bits of data, the receiver must transmit an acknowledge signal, level “0”, before it can receive the next data byte. If the slave transmitter does not receive an acknowledge bit signal from the master receiver, then the slave transmitter will release the SDA line to allow the master to send a STOP signal to release the I²C Bus. The corresponding data will be stored in the SIMD register. If setup as a transmitter, the slave device must first write the data to be transmitted into the SIMD register. If setup as a receiver, the slave device must read the transmitted data from the SIMD register.

When the slave receiver receives the data byte, it must generate an acknowledge bit, known as TXAK, on the 9th clock. The slave device, which is setup as a transmitter will check the RXAK bit in the SIMC1 register to determine if it is to send another data byte, if not then it will release the SDA line and await the receipt of a STOP signal from the master.



I²C Communication Timing Diagram

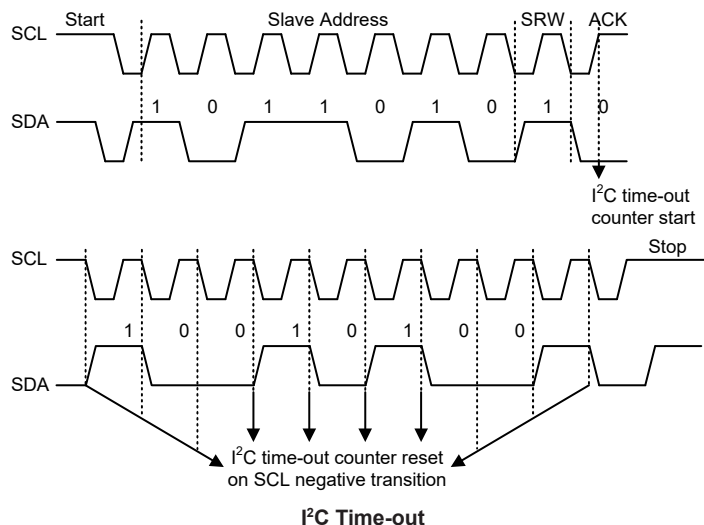
Note: When a slave address is matched, the device must be placed in either the transmit mode and then write data to the SIMD register, or in the receive mode where it must implement a dummy read from the SIMD register to release the SCL line.



I²C Bus ISR Flow Chart

I²C Time-out Control

In order to reduce the I²C lockup problem due to reception of erroneous clock sources, a time-out function is provided. If the clock source connected to the I²C bus is not received for a while, then the I²C circuitry and registers will be reset after a certain time-out period. The time-out counter starts to count on an I²C bus “START” & “address match” condition, and is cleared by an SCL falling edge. Before the next SCL falling edge arrives, if the time elapsed is greater than the time-out period specified by the SIMTOC register, then a time-out condition will occur. The time-out function will stop when an I²C “STOP” condition occurs.



When an I²C time-out counter overflow occurs, the counter will stop and the SIMTOEN bit will be cleared to zero and the SIMTOF bit will be set high to indicate that a time-out condition has occurred. The time-out condition will also generate an interrupt which uses the SIM interrupt vector. When an I²C time-out occurs, the I²C internal circuitry will be reset and the registers will be reset into the following condition:

| Registers | After I ² C Time-out |
|-------------------|---------------------------------|
| SIMD, SIMA, SIMC0 | No change |
| SIMC1 | Reset to POR condition |

I²C Register after Time-out

The SIMTOF flag can be cleared by the application program. There are 64 time-out period selections which can be selected using the SIMTOS bits in the SIMTOC register. The time-out duration is calculated by the formula: $(1-64) \times (32/f_{SUB})$. This gives a time-out period which ranges from about 1ms to 64ms.

• SIMTOC Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---------|--------|---------|---------|---------|---------|---------|---------|
| Name | SIMTOEN | SIMTOF | SIMTOS5 | SIMTOS4 | SIMTOS3 | SIMTOS2 | SIMTOS1 | SIMTOS0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7 **SIMTOEN**: SIM I²C Time-out function control
 0: Disable
 1: Enable

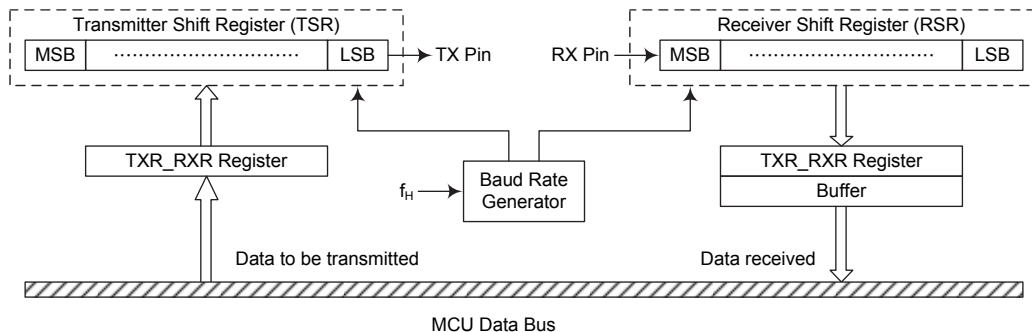
- Bit 6 **SIMTOF:** SIM I²C Time-out flag
 0: No time-out occurred
 1: Time-out occurred
 This bit is set high when time-out occurs and can only be cleared to zero by application program.
- Bit 5~0 **SIMTOS5~SIMTOS0:** SIM I²C Time-out period selection
 I²C Time-out clock source is $f_{SUB}/32$.
 I²C time-out time is equal to $(SIMTOS[5:0]+1) \times (32/f_{SUB})$.

UART Interface

The device contains an integrated full-duplex asynchronous serial communications UART interface that enables communication with external devices that contain a serial interface. The UART function has many features and can transmit and receive data serially by transferring a frame of data with eight or nine data bits per transmission as well as being able to detect errors when the data is overwritten or incorrectly framed. The UART function possesses its own internal interrupt which can be used to indicate when a reception occurs or when a transmission terminates.

The integrated UART function contains the following features:

- Full-duplex, asynchronous communication
- 8 or 9 bits character length
- Even, odd or no parity options
- One or two stop bits
- Baud rate generator with 8-bit prescaler
- Parity, framing, noise and overrun error detection
- Support for interrupt on address detect (last character bit=1)
- Separately enabled transmitter and receiver
- 2-byte Deep FIFO Receive Data Buffer
- RX pin wake-up function
- Transmit and receive interrupts
- Interrupts can be initialized by the following conditions:
 - ♦ Transmitter Empty
 - ♦ Transmitter Idle
 - ♦ Receiver Full
 - ♦ Receiver Overrun
 - ♦ Address Mode Detect



UART Data Transfer Block Diagram

UART External Pins

To communicate with an external serial interface, the internal UART has two external pins known as TX and RX. The TX and RX pins are the UART transmitter and receiver pins respectively. The TX and RX pin function should first be selected by the IFS register before the UART function is used. Along with the UARTEN bit, the TXEN and RXEN bits, if set, will setup these pins to their respective TX output and RX input conditions and disable any pull-high resistor option which may exist on the TX and RX pins. When the TX or RX pin function is disabled by clearing the UARTEN, TXEN or RXEN bit, the TX or RX pin will be set to a floating state. At this time whether the internal pull-high resistor is connected to the TX or RX pin or not is determined by the corresponding I/O pull-high function control bit.

UART Data Transfer Scheme

The above block diagram shows the overall data transfer structure arrangement for the UART. The actual data to be transmitted from the MCU is first transferred to the TXR_RXR register by the application program. The data will then be transferred to the Transmit Shift Register from where it will be shifted out, LSB first, onto the TX pin at a rate controlled by the Baud Rate Generator. Only the TXR_RXR register is mapped onto the MCU Data Memory, the Transmit Shift Register is not mapped and is therefore inaccessible to the application program.

Data to be received by the UART is accepted on the external RX pin, from where it is shifted in, LSB first, to the Receiver Shift Register at a rate controlled by the Baud Rate Generator. When the shift register is full, the data will then be transferred from the shift register to the internal TXR_RXR register, where it is buffered and can be manipulated by the application program. Only the TXR_RXR register is mapped onto the MCU Data Memory, the Receiver Shift Register is not mapped and is therefore inaccessible to the application program.

It should be noted that the actual register for data transmission and reception only exists as a single shared register in the Data Memory. This shared register known as the TXR_RXR register is used for both data transmission and data reception.

UART Status and Control Registers

There are five control registers associated with the UART function. The USR, UCR1 and UCR2 registers control the overall function of the UART, while the BRG register controls the Baud rate. The actual data to be transmitted and received on the serial interface is managed through the TXR_RXR data register.

| Register Name | Bit | | | | | | | |
|---------------|--------|-------|-------|-------|-------|-------|-------|-------|
| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| USR | PERR | NF | FERR | OERR | RIDLE | RXIF | TIDLE | TXIF |
| UCR1 | UARTEN | BNO | PREN | PRT | STOPS | TXBRK | RX8 | TX8 |
| UCR2 | TXEN | RXEN | BRGH | ADDEN | WAKE | RIE | TIIE | TEIE |
| TXR_RXR | TXRX7 | TXRX6 | TXRX5 | TXRX4 | TXRX3 | TXRX2 | TXRX1 | TXRX0 |
| BRG | BRG7 | BRG6 | BRG5 | BRG4 | BRG3 | BRG2 | BRG1 | BRG0 |

UART Register List

• TXR_RXR Register

The TXR_RXR register is the data register which is used to store the data to be transmitted on the TX pin or being received from the RX pin.

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-------|-------|-------|-------|-------|-------|-------|-------|
| Name | TXRX7 | TXRX6 | TXRX5 | TXRX4 | TXRX3 | TXRX2 | TXRX1 | TXRX0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | x | x | x | x | x | x | x | x |

"x": unknown

Bit 7~0 **TXRX7~TXRX0**: UART Transmit / Receive Data bit 7~bit 0

• USR Register

The USR register is the status register for the UART, which can be read by the program to determine the present status of the UART. All flags within the USR register are read only. Further explanation on each of the flags is given below:

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|------|----|------|------|-------|------|-------|------|
| Name | PERR | NF | FERR | OERR | RIDLE | RXIF | TIDLE | TXIF |
| R/W | R | R | R | R | R | R | R | R |
| POR | 0 | 0 | 0 | 0 | 1 | 0 | 1 | 1 |

Bit 7 **PERR**: Parity error flag
 0: No parity error is detected
 1: Parity error is detected

The PERR flag is the parity error flag. When this read only flag is "0", it indicates a parity error has not been detected. When the flag is "1", it indicates that the parity of the received word is incorrect. This error flag is applicable only if Parity mode (odd or even) is selected. The flag can also be cleared to zero by a software sequence which involves a read to the status register USR followed by an access to the TXR_RXR data register.

Bit 6 **NF**: Noise flag
 0: No noise is detected
 1: Noise is detected

The NF flag is the noise flag. When this read only flag is "0", it indicates no noise condition. When the flag is "1", it indicates that the UART has detected noise on the receiver input. The NF flag is set during the same cycle as the RXIF flag but will not be set in the case of an overrun. The NF flag can be cleared to zero by a software sequence which will involve a read to the status register USR followed by an access to the TXR_RXR data register.

- Bit 5 **FERR:** Framing error flag
 0: No framing error is detected
 1: Framing error is detected
 The FERR flag is the framing error flag. When this read only flag is “0”, it indicates that there is no framing error. When the flag is “1”, it indicates that a framing error has been detected for the current character. The flag can also be cleared to zero by a software sequence which will involve a read to the status register USR followed by an access to the TXR_RXR data register.
- Bit 4 **OERR:** Overrun error flag
 0: No overrun error is detected
 1: Overrun error is detected
 The OERR flag is the overrun error flag which indicates when the receiver buffer has overflowed. When this read only flag is “0”, it indicates that there is no overrun error. When the flag is “1”, it indicates that an overrun error occurs which will inhibit further transfers to the TXR_RXR receive data register. The flag is cleared to zero by a software sequence, which is a read to the status register USR followed by an access to the TXR_RXR data register.
- Bit 3 **RIDLE:** Receiver status
 0: Data reception is in progress (Data being received)
 1: No data reception is in progress (Receiver is idle)
 The RIDLE flag is the receiver status flag. When this read only flag is “0”, it indicates that the receiver is between the initial detection of the start bit and the completion of the stop bit. When the flag is “1”, it indicates that the receiver is idle. Between the completion of the stop bit and the detection of the next start bit, the RIDLE bit is “1” indicating that the UART receiver is idle and the RX pin stays in logic high condition.
- Bit 2 **RXIF:** Receive TXR_RXR data register status
 0: TXR_RXR data register is empty
 1: TXR_RXR data register has available data
 The RXIF flag is the receive data register status flag. When this read only flag is “0”, it indicates that the TXR_RXR read data register is empty. When the flag is “1”, it indicates that the TXR_RXR read data register contains new data. When the contents of the shift register are transferred to the TXR_RXR register, an interrupt is generated if RIE=1 in the UCR2 register. If one or more errors are detected in the received word, the appropriate receive-related flags NF, FERR, and/or PERR are set within the same clock cycle. The RXIF flag will eventually be cleared to zero when the USR register is read with RXIF set, followed by a read from the TXR_RXR register, and if the TXR_RXR register has no more new data available.
- Bit 1 **TIDLE:** Transmission idle
 0: Data transmission is in progress (Data being transmitted)
 1: No data transmission is in progress (Transmitter is idle)
 The TIDLE flag is known as the transmission complete flag. When this read only flag is “0”, it indicates that a transmission is in progress. This flag will be set high when the TXIF flag is “1” and when there is no transmit data or break character being transmitted. When TIDLE is equal to “1”, the TX pin becomes idle with the pin state in logic high condition. The TIDLE flag is cleared to zero by reading the USR register with TIDLE set and then writing to the TXR_RXR register. The flag is not generated when a data character or a break is queued and ready to be sent.
- Bit 0 **TXIF:** Transmit TXR_RXR data register status
 0: Character is not transferred to the transmit shift register
 1: Character has transferred to the transmit shift register (TXR_RXR data register is empty)
 The TXIF flag is the transmit data register empty flag. When this read only flag is “0”, it indicates that the character is not transferred to the transmitter shift register. When the flag is “1”, it indicates that the transmitter shift register has received a character from the TXR_RXR data register. The TXIF flag is cleared to zero by reading the UART status register (USR) with TXIF set and then writing to the TXR_RXR data register. Note that when the TXEN bit is set, the TXIF flag bit will also be set since the transmit data register is not yet full.

• **UCR1 Register**

The UCR1 register together with the UCR2 register are the two UART control registers that are used to set the various options for the UART function, such as overall on/off control, parity control, data transfer bit length etc. Further explanation on each of the bits is given below:

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|--------|-----|------|-----|-------|-------|-----|-----|
| Name | UARTEN | BNO | PREN | PRT | STOPS | TXBRK | RX8 | TX8 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R | W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | x | 0 |

“x”: unknown

Bit 7 **UARTEN**: UART function enable control

- 0: Disable UART. TX and RX pins are used as other pin-shared functional pins
- 1: Enable UART. TX and RX pins can function as UART pins defined by TXEN and RXEN bits

The UARTEN bit is the UART enable bit. When this bit is equal to “0”, the UART will be disabled and the RX pin as well as the TX pin will be other pin-shared functional pins. When the bit is equal to “1”, the UART will be enabled and the TX and RX pins will function as defined by the TXEN and RXEN enable control bits.

When the UART is disabled, it will empty the buffer so any character remaining in the buffer will be discarded. In addition, the value of the baud rate counter will be reset. If the UART is disabled, all error and status flags will be reset. Also the TXEN, RXEN, TXBRK, RXIF, OERR, FERR, PERR and NF bits will be cleared to zero, while the TIDLE, TXIF and RIDLE bits will be set high. Other control bits in UCR1, UCR2 and BRG registers will remain unaffected. If the UART is active and the UARTEN bit is cleared to zero, all pending transmissions and receptions will be terminated and the module will be reset as defined above. When the UART is re-enabled, it will restart in the same configuration.

Bit 6 **BNO**: Number of data transfer bits selection

- 0: 8-bit data transfer
- 1: 9-bit data transfer

This bit is used to select the data length format, which can have a choice of either 8-bit or 9-bit format. When this bit is equal to “1”, a 9-bit data length format will be selected. If the bit is equal to “0”, then an 8-bit data length format will be selected. If 9-bit data length format is selected, then bits RX8 and TX8 will be used to store the 9th bit of the received and transmitted data respectively.

Bit 5 **PREN**: Parity function enable control

- 0: Parity function is disabled
- 1: Parity function is enabled

This is the parity enable bit. When this bit is equal to “1”, the parity function will be enabled. If the bit is equal to “0”, then the parity function will be disabled.

Bit 4 **PRT**: Parity type selection bit

- 0: Even parity for parity generator
- 1: Odd parity for parity generator

This bit is the parity type selection bit. When this bit is equal to “1”, odd parity type will be selected. If the bit is equal to “0”, then even parity type will be selected.

Bit 3 **STOPS**: Number of Stop bits selection

- 0: One stop bit format is used
- 1: Two stop bits format is used

This bit determines if one or two stop bits are to be used. When this bit is equal to “1”, two stop bits are used. If this bit is equal to “0”, then only one stop bit is used.

Bit 2 **TXBRK**: Transmit break character

- 0: No break character is transmitted
- 1: Break characters transmit

The TXBRK bit is the Transmit Break Character bit. When this bit is “0”, there are no break characters and the TX pin operates normally. When the bit is “1”, there are transmit break characters and the transmitter will send logic zeros. When this bit is equal to “1”, after the buffered data has been transmitted, the transmitter output is held low for a minimum of a 13-bit length and until the TXBRK bit is reset.

- Bit 1 **RX8**: Receive data bit 8 for 9-bit data transfer format (read only)
This bit is only used if 9-bit data transfers are used, in which case this bit location will store the 9th bit of the received data known as RX8. The BNO bit is used to determine whether data transfers are in 8-bit or 9-bit format.
- Bit 0 **TX8**: Transmit data bit 8 for 9-bit data transfer format (write only)
This bit is only used if 9-bit data transfers are used, in which case this bit location will store the 9th bit of the transmitted data known as TX8. The BNO bit is used to determine whether data transfers are in 8-bit or 9-bit format.

• **UCR2 Register**

The UCR2 register is the second of the two UART control registers and serves several purposes. One of its main functions is to control the basic enable/disable operation of the UART Transmitter and Receiver as well as enabling the various UART interrupt sources. The register also serves to control the baud rate speed, receiver wake-up enable and the address detect enable. Further explanation on each of the bits is given below:

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|------|------|------|-------|------|-----|-------|------|
| Name | TXEN | RXEN | BRGH | ADDEN | WAKE | RIE | TIIIE | TEIE |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

- Bit 7 **TXEN**: UART Transmitter enabled control
 0: UART transmitter is disabled
 1: UART transmitter is enabled
 The bit named TXEN is the Transmitter Enable Bit. When this bit is equal to “0”, the transmitter will be disabled with any pending data transmissions being aborted. In addition the buffers will be reset. In this situation the TX pin will be other pin-shared functional pin.
 If the TXEN bit is equal to “1” and the UARTEN bit is also equal to “1”, the transmitter will be enabled and the TX pin will be controlled by the UART. Clearing the TXEN bit during a transmission will cause the data transmission to be aborted and will reset the transmitter. If this situation occurs, the TX pin will be other pin-shared functional pin.
- Bit 6 **RXEN**: UART Receiver enabled control
 0: UART receiver is disabled
 1: UART receiver is enabled
 The bit named RXEN is the Receiver Enable Bit. When this bit is equal to “0”, the receiver will be disabled with any pending data receptions being aborted. In addition the receive buffers will be reset. In this situation the RX pin will be other pin-shared functional pin. If the RXEN bit is equal to “1” and the UARTEN bit is also equal to “1”, the receiver will be enabled and the RX pin will be controlled by the UART. Clearing the RXEN bit during a reception will cause the data reception to be aborted and will reset the receiver. If this situation occurs, the RX pin will be other pin-shared functional pin.

- Bit 5 **BRGH**: Baud Rate speed selection
 0: Low speed baud rate
 1: High speed baud rate
The bit named BRGH selects the high or low speed mode of the Baud Rate Generator. This bit, together with the value placed in the baud rate register BRG, controls the Baud Rate of the UART. If this bit is equal to “1”, the high speed mode is selected. If the bit is equal to “0”, the low speed mode is selected.
- Bit 4 **ADDEN**: Address detect function enable control
 0: Address detect function is disabled
 1: Address detect function is enabled
The bit named ADDEN is the address detect function enable control bit. When this bit is equal to “1”, the address detect function is enabled. When it occurs, if the 8th bit, which corresponds to RX7 if BNO=0 or the 9th bit, which corresponds to RX8 if BNO=1, has a value of “1”, then the received word will be identified as an address, rather than data. If the corresponding interrupt is enabled, an interrupt request will be generated each time the received word has the address bit set, which is the 8th or 9th bit depending on the value of BNO. If the address bit known as the 8th or 9th bit of the received word is “0” with the address detect function being enabled, an interrupt will not be generated and the received data will be discarded.
- Bit 3 **WAKE**: RX pin wake-up UART function enable control
 0: RX pin wake-up UART function is disabled
 1: RX pin wake-up UART function is enabled
This bit is used to control the wake-up UART function when a falling edge on the RX pin occurs. Note that this bit is only available when the UART clock (f_{H1}) is switched off. There will be no RX pin wake-up UART function if the UART clock (f_{H1}) exists. If the WAKE bit is set to 1 as the UART clock (f_{H1}) is switched off, a UART wake-up request will be initiated when a falling edge on the RX pin occurs. When this request happens and the corresponding interrupt is enabled, an RX pin wake-up UART interrupt will be generated to inform the MCU to wake up the UART function by switching on the UART clock (f_{H1}) via the application program. Otherwise, the UART function can not resume even if there is a falling edge on the RX pin when the WAKE bit is cleared to 0.
- Bit 2 **RIE**: Receiver interrupt enable control
 0: Receiver related interrupt is disabled
 1: Receiver related interrupt is enabled
This bit enables or disables the receiver interrupt. If this bit is equal to “1” and when the receiver overrun flag OERR or receive data available flag RXIF is set, the UART interrupt request flag will be set. If this bit is equal to “0”, the UART interrupt request flag will not be influenced by the condition of the OERR or RXIF flags.
- Bit 1 **TIE**: Transmitter Idle interrupt enable control
 0: Transmitter idle interrupt is disabled
 1: Transmitter idle interrupt is enabled
This bit enables or disables the transmitter idle interrupt. If this bit is equal to “1” and when the transmitter idle flag TIDLE is set, due to a transmitter idle condition, the UART interrupt request flag will be set. If this bit is equal to “0”, the UART interrupt request flag will not be influenced by the condition of the TIDLE flag.
- Bit 0 **TEIE**: Transmitter Empty interrupt enable control
 0: Transmitter empty interrupt is disabled
 1: Transmitter empty interrupt is enabled
This bit enables or disables the transmitter empty interrupt. If this bit is equal to “1” and when the transmitter empty flag TXIF is set, due to a transmitter empty condition, the UART interrupt request flag will be set. If this bit is equal to “0”, the UART interrupt request flag will not be influenced by the condition of the TXIF flag.

• **BRG Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|------|------|------|------|------|------|------|------|
| Name | BRG7 | BRG6 | BRG5 | BRG4 | BRG3 | BRG2 | BRG1 | BRG0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | x | x | x | x | x | x | x | x |

“x”: unknown

Bit 7~0 **BRG7~BRG0:** Baud Rate values

By programming the BRGH bit in UCR2 Register which allows selection of the related formula described above and programming the required value in the BRG register, the required baud rate can be setup.

Note: Baud rate= $f_H/[64 \times (N+1)]$ if BRGH=0;

Baud rate= $f_H/[16 \times (N+1)]$ if BRGH=1.

Baud Rate Generator

To setup the speed of the serial data communication, the UART function contains its own dedicated baud rate generator. The baud rate is controlled by its own internal free running 8-bit timer, the period of which is determined by two factors. The first of these is the value placed in the baud rate register BRG and the second is the value of the BRGH bit with the control register UCR2. The BRGH bit decides if the baud rate generator is to be used in a high speed mode or low speed mode, which in turn determines the formula that is used to calculate the baud rate. The value N in the BRG register which is used in the following baud rate calculation formula determines the division factor. Note that N is the decimal value placed in the BRG register and has a range of between 0 and 255.

| UCR2 BRGH Bit | 0 | 1 |
|----------------|-------------------------|-------------------------|
| Baud Rate (BR) | $f_H/[64 \times (N+1)]$ | $f_H/[16 \times (N+1)]$ |

By programming the BRGH bit which allows selection of the related formula and programming the required value in the BRG register, the required baud rate can be setup. Note that because the actual baud rate is determined using a discrete value, N, placed in the BRG register, there will be an error associated between the actual and requested value. The following example shows how the BRG register value N and the error value can be calculated.

Calculating the Baud Rate and Error Values

For a clock frequency of 4MHz, and with BRGH cleared to zero determine the BRG register value N, the actual baud rate and the error value for a desired baud rate of 4800.

From the above table the desired baud rate $BR = f_H/[64 \times (N+1)]$

Re-arranging this equation gives $N = [f_H/(BR \times 64)] - 1$

Giving a value for $N = [4000000/(4800 \times 64)] - 1 = 12.0208$

To obtain the closest value, a decimal value of 12 should be placed into the BRG register. This gives an actual or calculated baud rate value of $BR = 4000000/[64 \times (12+1)] = 4808$

Therefore the error is equal to $(4808-4800)/4800 = 0.16\%$

UART Setup and Control

For data transfer, the UART function utilizes a non-return-to-zero, more commonly known as NRZ, format. This is composed of one start bit, eight or nine data bits, and one or two stop bits. Parity is supported by the UART hardware, and can be setup to be even, odd or no parity. For the most common data format, 8 data bits along with no parity and one stop bit, denoted as 8, N, 1, is used as the default setting, which is the setting at power-on. The number of data bits and stop bits, along with the parity, are setup by programming the corresponding BNO, PRT, PREN, and STOPS bits in the UCR1 register. The baud rate used to transmit and receive data is setup using the internal 8-bit baud rate generator, while the data is transmitted and received LSB first. Although the UART transmitter and receiver are functionally independent, they both use the same data format and baud rate. In all cases stop bits will be used for data transmission.

Enabling/Disabling the UART Interface

The basic on/off function of the internal UART function is controlled using the UARTEN bit in the UCR1 register. If the UARTEN, TXEN and RXEN bits are set, then these two UART pins will act as normal TX output pin and RX input pin respectively. If no data is being transmitted on the TX pin, then it will default to a logic high value.

Clearing the UARTEN bit will disable the TX and RX pins and allow these two pins to be used as normal I/O or other pin-shared functional pins. When the UART function is disabled the buffer will be reset to an empty condition, at the same time discarding any remaining residual data. Disabling the UART will also reset the error and status flags with bits TXEN, RXEN, TXBRK, RXIF, OERR, FERR, PERR and NF being cleared while bits TIDLE, TXIF and RIDLE will be set. The remaining control bits in the UCR1, UCR2 and BRG registers will remain unaffected. If the UARTEN bit in the UCR1 register is cleared while the UART is active, then all pending transmissions and receptions will be immediately suspended and the UART will be reset to a condition as defined above. If the UART is then subsequently re-enabled, it will restart again in the same configuration.

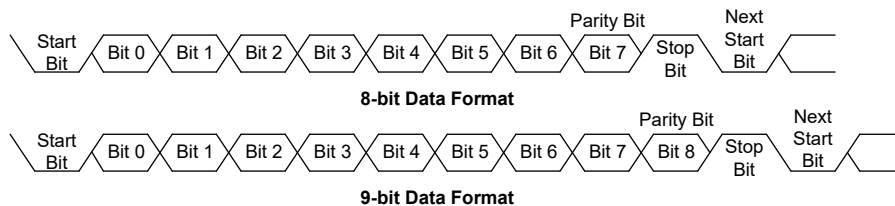
Data, Parity and Stop Bit Selection

The format of the data to be transferred is composed of various factors such as data bit length, parity on/off, parity type, address bits and the number of stop bits. These factors are determined by the setup of various bits within the UCR1 register. The BNO bit controls the number of data bits which can be set to either 8 or 9, the PRT bit controls the choice of odd or even parity, the PREN bit controls the parity on/off function and the STOPS bit decides whether one or two stop bits are to be used. The following table shows various formats for data transmission. The address bit, which is the MSB of the data byte, identifies the frame as an address character or data if the address detect function is enabled. The number of stop bits, which can be either one or two, is independent of the data length and is only used for the transmitter. There is only one stop bit for the receiver.

| Start Bit | Data Bits | Address Bit | Parity Bit | Stop Bit |
|--------------------------------------|-----------|-------------|------------|----------|
| Example of 8-bit Data Formats | | | | |
| 1 | 8 | 0 | 0 | 1 |
| 1 | 7 | 0 | 1 | 1 |
| 1 | 7 | 1 | 0 | 1 |
| Example of 9-bit Data Formats | | | | |
| 1 | 9 | 0 | 0 | 1 |
| 1 | 8 | 0 | 1 | 1 |
| 1 | 8 | 1 | 0 | 1 |

Transmitter Receiver Data Format

The following diagram shows the transmit and receive waveforms for both 8-bit and 9-bit data formats.



UART Transmitter

Data word lengths of either 8 or 9 bits can be selected by programming the BNO bit in the UCR1 register. When BNO bit is set, the word length will be set to 9 bits. In this case the 9th bit, which is the MSB, needs to be stored in the TX8 bit in the UCR1 register. At the transmitter core lies the Transmitter Shift Register, more commonly known as the TSR, whose data is obtained from the transmit data register, which is known as the TXR_RXR register. The data to be transmitted is loaded into this TXR_RXR register by the application program. The TSR register is not written to with new data until the stop bit from the previous transmission has been sent out. As soon as this stop bit has been transmitted, the TSR can then be loaded with new data from the TXR_RXR register, if it is available. It should be noted that the TSR register, unlike many other registers, is not directly mapped into the Data Memory area and as such is not available to the application program for direct read/write operations. An actual transmission of data will normally be enabled when the TXEN bit is set, but the data will not be transmitted until the TXR_RXR register has been loaded with data and the baud rate generator has defined a shift clock source. However, the transmission can also be initiated by first loading data into the TXR_RXR register, after which the TXEN bit can be set. When a transmission of data begins, the TSR is normally empty, in which case a transfer to the TXR_RXR register will result in an immediate transfer to the TSR. If during a transmission the TXEN bit is cleared, the transmission will immediately cease and the transmitter will be reset. The TX output pin will then return to the I/O or other pin-shared function.

Transmitting Data

When the UART is transmitting data, the data is shifted on the TX pin from the shift register, with the least significant bit first. In the transmit mode, the TXR_RXR register forms a buffer between the internal bus and the transmitter shift register. It should be noted that if 9-bit data format has been selected, then the MSB will be taken from the TX8 bit in the UCR1 register. The steps to initiate a data transfer can be summarized as follows:

- Make the correct selection of the BNO, PRT, PREN and STOPS bits to define the required word length, parity type and number of stop bits.
- Setup the BRG register to select the desired baud rate.
- Set the TXEN bit to ensure that the TX pin is used as a UART transmitter pin.
- Access the USR register and write the data that is to be transmitted into the TXR_RXR register. Note that this step will clear the TXIF bit.

This sequence of events can now be repeated to send additional data.

It should be noted that when TXIF=0, data will be inhibited from being written to the TXR_RXR register. Clearing the TXIF flag is always achieved using the following software sequence:

1. A USR register access
2. A TXR_RXR register write execution

The read-only TXIF flag is set by the UART hardware and if set indicates that the TXR_RXR register is empty and that other data can now be written into the TXR_RXR register without overwriting the previous data. If the TEIE bit is set then the TXIF flag will generate an interrupt.

During a data transmission, a write instruction to the TXR_RXR register will place the data into the TXR_RXR register, which will be copied to the shift register at the end of the present transmission. When there is no data transmission in progress, a write instruction to the TXR_RXR register will place the data directly into the shift register, resulting in the commencement of data transmission, and the TXIF bit being immediately set. When a frame transmission is complete, which happens after stop bits are sent or after the break frame, the TIDLE bit will be set. To clear the TIDLE bit the following software sequence is used:

1. A USR register access
2. A TXR_RXR register write execution

Note that both the TXIF and TIDLE bits are cleared by the same software sequence.

Transmit Break

If the TXBRK bit is set then break characters will be sent on the next transmission. Break character transmission consists of a start bit, followed by $13 \times N$ '0' bits and stop bits, where $N=1, 2, \text{etc.}$ If a break character is to be transmitted then the TXBRK bit must be first set by the application program, and then cleared to generate the stop bits. Transmitting a break character will not generate a transmit interrupt. Note that a break condition length is at least 13 bits long. If the TXBRK bit is continually kept at a logic high level then the transmitter circuitry will transmit continuous break characters. After the application program has cleared the TXBRK bit, the transmitter will finish transmitting the last break character and subsequently send out one or two stop bits. The automatic logic highs at the end of the last break character will ensure that the start bit of the next frame is recognized.

UART Receiver

The UART is capable of receiving word lengths of either 8 or 9 bits. If the BNO bit is set, the word length will be set to 9 bits with the MSB being stored in the RX8 bit of the UCR1 register. At the receiver core lies the Receive Serial Shift Register, commonly known as the RSR. The data which is received on the RX external input pin is sent to the data recovery block. The data recovery block operating speed is 16 times that of the baud rate, while the main receive serial shifter operates at the baud rate. After the RX pin is sampled for the stop bit, the received data in RSR is transferred to the receive data register, if the register is empty. The data which is received on the external RX input pin is sampled three times by a majority detect circuit to determine the logic level that has been placed onto the RX pin. It should be noted that the RSR register, unlike many other registers, is not directly mapped into the Data Memory area and as such is not available to the application program for direct read/write operations.

Receiving Data

When the UART receiver is receiving data, the data is serially shifted in on the external RX input pin, LSB first. In the read mode, the TXR_RXR register forms a buffer between the internal bus and the receiver shift register. The TXR_RXR register is a two byte deep FIFO data buffer, where two bytes can be held in the FIFO while a third byte can continue to be received. Note that the application program must ensure that the data is read from TXR_RXR before the third byte has been completely shifted in, otherwise this third byte will be discarded and an overrun error OERR will be subsequently indicated. The steps to initiate a data transfer can be summarized as follows:

- Make the correct selection of BNO, PRT and PREN bits to define the word length, parity type.
- Setup the BRG register to select the desired baud rate.
- Set the RXEN bit to ensure that the RX pin is used as a UART receiver pin.

At this point the receiver will be enabled which will begin to look for a start bit.

When a character is received the following sequence of events will occur:

- The RXIF bit in the USR register will be set when the TXR_RXR register has data available. There will be at most one more character available before an overrun error occurs.
- When the contents of the shift register have been transferred to the TXR_RXR register, then if the RIE bit is set, an interrupt will be generated.
- If during reception, a frame error, noise error, parity error, or an overrun error has been detected, then the error flags can be set.

The RXIF bit can be cleared using the following software sequence:

1. A USR register access
2. A TXR_RXR register read execution

Receiving Break

Any break character received by the UART will be managed as a framing error. The receiver will count and expect a certain number of bit times as specified by the values programmed into the BNO bit plus one stop bit. If the break is much longer than 13 bit times, the reception will be considered as complete after the number of bit times specified by BNO plus one stop bit. The RXIF bit is set, FERR is set, zeros are loaded into the receive data register, interrupts are generated if appropriate and the RIDLE bit is set. A break is regarded as a character that contains only zeros with the FERR flag set. If a long break signal has been detected, the receiver will regard it as a data frame including a start bit, data bits and the invalid stop bit and the FERR flag will be set. The receiver must wait for a valid stop bit before looking for the next start bit. The receiver will not make the assumption that the break condition on the line is the next start bit. The break character will be loaded into the buffer and no further data will be received until stop bits are received. It should be noted that the RIDLE read only flag will go high when the stop bits have not yet been received. The reception of a break character on the UART registers will result in the following:

- The framing error flag, FERR, will be set.
- The receive data register, TXR_RXR, will be cleared.
- The OERR, NF, PERR, RIDLE or RXIF flags will possibly be set.

Idle Status

When the receiver is reading data, which means it will be in between the detection of a start bit and the reading of a stop bit, the receiver status flag in the USR register, otherwise known as the RIDLE flag, will have a zero value. In between the reception of a stop bit and the detection of the next start bit, the RIDLE flag will have a high value, which indicates the receiver is in an idle condition.

Receiver Interrupt

The read only receive interrupt flag RXIF in the USR register is set by an edge generated by the receiver. An interrupt is generated if RIE=1, when a word is transferred from the Receive Shift Register, RSR, to the Receive Data Register, TXR_RXR. An overrun error can also generate an interrupt if RIE=1.

Managing Receiver Errors

Several types of reception errors can occur within the UART module, the following section describes the various types and how they are managed by the UART.

Overrun Error – OERR

The TXR_RXR register is composed of a two byte deep FIFO data buffer, where two bytes can be held in the FIFO register, while a third byte can continue to be received. Before this third byte has been entirely shifted in, the data should be read from the TXR_RXR register. If this is not done, the overrun error flag OERR will be consequently indicated.

In the event of an overrun error occurring, the following will happen:

- The OERR flag in the USR register will be set.
- The TXR_RXR contents will not be lost.
- The shift register will be overwritten.
- An interrupt will be generated if the RIE bit is set.

The OERR flag can be cleared by an access to the USR register followed by a read to the TXR_RXR register.

Noise Error – NF

Over-sampling is used for data recovery to identify valid incoming data and noise. If noise is detected within a frame the following will occur:

- The read only noise flag, NF, in the USR register will be set on the rising edge of the RXIF bit.
- Data will be transferred from the Shift register to the TXR_RXR register.
- No interrupt will be generated. However this bit rises at the same time as the RXIF bit which itself generates an interrupt.

Note that the NF flag is reset by a USR register read operation followed by a TXR_RXR register read operation.

Framing Error – FERR

The read only framing error flag, FERR, in the USR register, is set if a zero is detected instead of stop bits. If two stop bits are selected, both stop bits must be high; otherwise the FERR flag will be set. The FERR flag and the received data will be recorded in the USR and TXR_RXR registers respectively, and the flag is cleared in any reset.

Parity Error – PERR

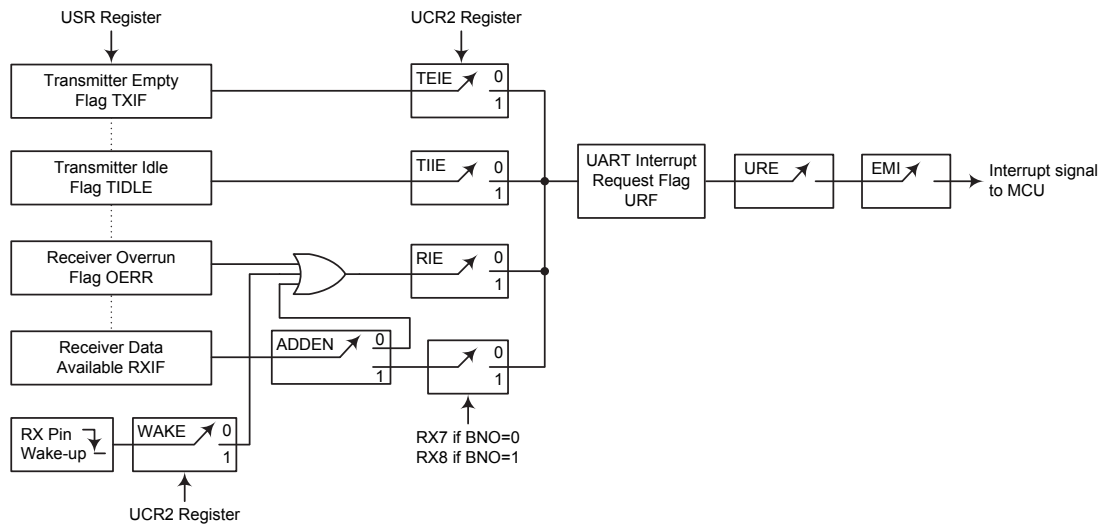
The read only parity error flag, PERR, in the USR register, is set if the parity of the received word is incorrect. This error flag is only applicable if the parity is enabled, PREN=1, and if the parity type, odd or even is selected. The read only PERR flag and the received data will be recorded in the USR and TXR_RXR registers respectively. It is cleared on any reset, it should be noted that the flags, FERR and PERR, in the USR register should first be read by the application program before reading the data word.

UART Interrupt Structure

Several individual UART conditions can generate a UART interrupt. When these conditions exist, a low pulse will be generated to get the attention of the microcontroller. These conditions are a transmitter data register empty, transmitter idle, receiver data available, receiver overrun, address detect and an RX pin wake-up. When any of these conditions are created, if the global interrupt enable bit and its corresponding interrupt control bit are enabled and the stack is not full, the program will jump to its corresponding interrupt vector where it can be serviced before returning to the main program. Four of these conditions have the corresponding USR register flags which will generate a UART interrupt if its associated interrupt enable control bit in the UCR2 register is set. The two transmitter interrupt conditions have their own corresponding enable control bits, while the two receiver interrupt conditions have a shared enable control bit. These enable bits can be used to mask out individual UART interrupt sources.

The address detect condition, which is also a UART interrupt source, does not have an associated flag, but will generate a UART interrupt when an address detect condition occurs if its function is enabled by setting the ADDEN bit in the UCR2 register. An RX pin wake-up, which is also a UART interrupt source, does not have an associated flag, but will generate a UART interrupt if the UART clock (f_{II}) source is switched off and the WAKE and RIE bits in the UCR2 register are set when a falling edge on the RX pin occurs. Note that in the event of an RX wake-up interrupt occurring, there will be a certain period of delay, commonly known as the System Start-up Time, for the oscillator to restart and stabilize before the system resumes normal operation.

Note that the USR register flags are read only and cannot be cleared or set by the application program, neither will they be cleared when the program jumps to the corresponding interrupt servicing routine, as is the case for some of the other interrupts. The flags will be cleared automatically when certain actions are taken by the UART, the details of which are given in the UART register section. The overall UART interrupt can be disabled or enabled by the related interrupt enable control bits in the interrupt control registers of the microcontroller to decide whether the interrupt requested by the UART module is masked out or allowed.



UART Interrupt Structure

Address Detect Mode

Setting the Address Detect Mode bit, ADDEN, in the UCR2 register, enables this special mode. If this bit is enabled then an additional qualifier will be placed on the generation of a Receiver Data Available interrupt, which is requested by the RXIF flag. If the ADDEN bit is enabled, then when data is available, an interrupt will only be generated, if the highest received bit has a high value. Note that the URE and EMI interrupt enable bits must also be enabled for correct interrupt generation. This highest address bit is the 9th bit if BNO=1 or the 8th bit if BNO=0. If this bit is high, then the received word will be defined as an address rather than data. A Data Available interrupt will be generated every time the last bit of the received word is set. If the ADDEN bit is not enabled, then a Receiver Data Available interrupt will be generated each time the RXIF flag is set, irrespective of the data last bit status. The address detect mode and parity enable are mutually exclusive functions. Therefore if the address detect mode is enabled, then to ensure correct operation, the parity function should be disabled by resetting the parity enable bit PREN to zero.

| ADDEN | Bit 9 if BNO=1, Bit 8 if BNO=0 | UART Interrupt Generated |
|-------|-----------------------------------|-----------------------------|
| 0 | 0 | √ |
| | 1 | √ |
| 1 | 0 | × |
| | 1 | √ |

ADDEN Bit Function

UART Power Down and Wake-up

When the UART clock, f_{H} , is switched off, the UART will cease to function. If the MCU switches off the UART clock, f_{H} , and enters the power down mode while a transmission is still in progress, then the transmission will be paused until the UART clock source derived from the microcontroller is activated. In a similar way, if the MCU switches off the UART clock f_{H} and enters the IDLE or SLEEP mode by executing the “HALT” instruction while receiving data, then the reception of data will likewise be paused. When the MCU enters the IDLE or SLEEP mode, note that the USR, UCR1, UCR2, transmit and receive registers, as well as the BRG register will not be affected. It is recommended to make sure first that the UART data transmission or reception has been finished before the microcontroller enters the IDLE or SLEEP mode.

The UART function contains a receiver RX pin wake-up function, which is enabled or disabled by the WAKE bit in the UCR2 register. If this bit, along with the UART enable bit, UARTEN, the receiver enable bit, RXEN and the receiver interrupt bit, RIE, are all set when the MCU enters the power down mode with the UART clock f_{H} being switched off, then a falling edge on the RX pin will initiate an RX pin wake-up UART interrupt. Note that as it takes certain system clock cycles after a wake-up, before normal microcontroller operation resumes, any data received during this time on the RX pin will be ignored.

For a UART wake-up interrupt to occur, in addition to the bits for the wake-up being set, the global interrupt enable bit, EMI, and the UART interrupt enable bit, URE, must be set. If the EMI and URE bits are not set then only a wake up event will occur and no interrupt will be generated. Note also that as it takes certain system clock cycles after a wake-up before normal microcontroller resumes, the UART interrupt will not be generated until after this time has elapsed.

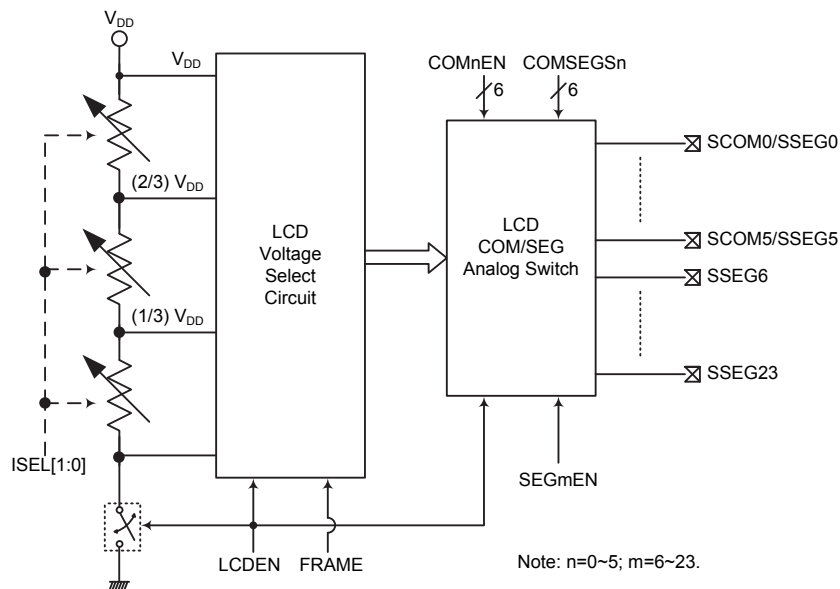
SCOM/SSEG Function for LCD

The device has the capability of driving external LCD panels. The common and segment pins for LCD driving, SCOM0~SCOM5 and SSEG0~SSEG23, are pin-shared with certain pins on the I/O ports. The LCD signals, COM and SEG, are generated using the application program.

LCD Operation

An external LCD panel can be driven using the device by configuring the I/O pins as common pins and segment pins. The LCD driver function is controlled using the LCD control registers which in addition to controlling the overall on/off function also controls the R-type bias current on the SCOM and SSEG pins. This enables the LCD COM and SEG driver to generate the necessary V_{SS} , $(1/3)V_{DD}$, $(2/3)V_{DD}$ and V_{DD} voltage levels for LCD 1/3 bias operation.

The LCDEN bit in the SLCDC0 register is the overall master control for the LCD driver. This bit is used in conjunction with the COMnEN and SEGmEN bits to select which I/O pins are used for LCD driving. Note that the corresponding Port Control register does not need to first setup the pins as outputs to enable the LCD driver operation.



Software Controlled LCD Driver Structure

LCD Frames

A cyclic LCD waveform includes two frames known as Frame 0 and Frame 1 for which the following offers a functional explanation.

Frame 0

To select Frame 0, clear the FRAME bit in the SLCDC0 register to 0.

In frame 0, the COM signal output can have a value of V_{DD} or a V_{BIAS} value of $(1/3) \times V_{DD}$. The SEG signal output can have a value of V_{SS} or a V_{BIAS} value of $(2/3) \times V_{DD}$.

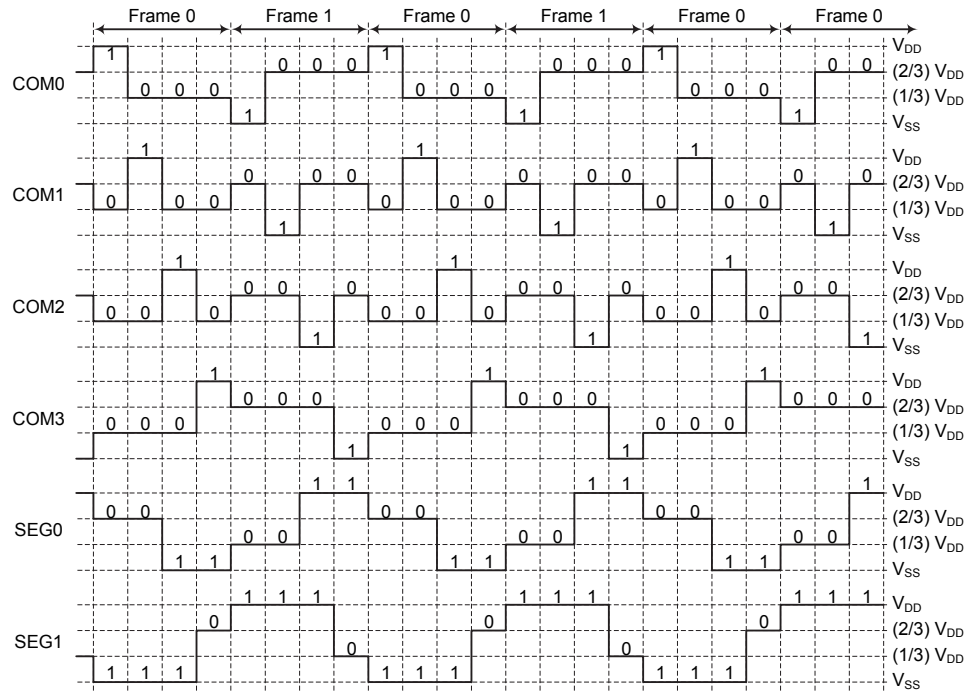
Frame 1

To select Frame 1, set the FRAME bit in the SLCDC0 register to 1.

In frame 1, the COM signal output can have a value of V_{SS} or a V_{BIAS} value of $(2/3) \times V_{DD}$. The SEG signal output can have a value of V_{DD} or a V_{BIAS} value of $(1/3) \times V_{DD}$.

The COM_n waveform is controlled by the application program using the FRAME bit in the SLCDC0 register and the corresponding pin-shared I/O data bit for the respective COM pin to determine whether the COM_n output has a value of V_{DD}, V_{SS} or V_{BIAS}. The SEG_m waveform is controlled in a similar way using the FRAME bit and the corresponding pin-shared I/O data bit for the respective SEG pin to determine whether the SEG_m output has a value of V_{DD}, V_{SS} or V_{BIAS}.

The accompanying waveform diagram shows a typical 1/3 bias LCD waveform generated using the application program together with the LCD voltage select circuit. Note that the depiction of a “1” in the diagram illustrates an illuminated LCD pixel. The COM signal polarity generated on pins SCOM0~SCOM5, whether “0” or “1”, are generated using the corresponding pin-shared I/O data register bit.



Note: The logical values shown in the above diagram are the corresponding pin-shared I/O data bit value.

1/3 Bias LCD Waveform – 4-COM & 2-SEG Application

LCD Control Registers

The LCD COM and SEG driver enables a range of selections to be provided to suit the requirement of the LCD panel which is being used. The bias resistor choice is implemented using the ISEL1 and ISEL0 bits in the SLCDC0 register. All COM and SEG pins are pin-shared with I/O pins and selected as COM and SEG pins using the corresponding pin function selection bits in the SLCDC_n registers respectively.

| Register Name | Bit | | | | | | | |
|---------------|---------|---------|----------|----------|----------|----------|----------|----------|
| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| SLCDC0 | FRAME | ISEL1 | ISEL0 | LCDEN | COM3EN | COM2EN | COM1EN | COM0EN |
| SLCDC1 | COM5EN | COM4EN | COMSEGS5 | COMSEGS4 | COMSEGS3 | COMSEGS2 | COMSEGS1 | COMSEGS0 |
| SLCDC2 | SEG13EN | SEG12EN | SEG11EN | SEG10EN | SEG9EN | SEG8EN | SEG7EN | SEG6EN |
| SLCDC3 | SEG21EN | SEG20EN | SEG19EN | SEG18EN | SEG17EN | SEG16EN | SEG15EN | SEG14EN |
| SLCDC4 | — | — | — | — | — | — | SEG23EN | SEG22EN |

LCD Driver Control Register List

• **SLCDC0 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-------|-------|-------|-------|--------|--------|--------|--------|
| Name | FRAME | ISEL1 | ISEL0 | LCDEN | COM3EN | COM2EN | COM1EN | COM0EN |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7 **FRAME:** SCOM/SSEG Output Frame selection

0: Frame 0
1: Frame 1

Bit 6~5 **ISEL1~ISEL0:** Select SCOM/SSEG typical bias current ($V_{DD}=5V$)

00: 8.3 μ A
01: 16.7 μ A
10: 50 μ A
11: 100 μ A

Bit 4 **LCDEN:** SCOM/SSEG Module enable control

0: Disable
1: Enable

The SCOMn and SSEGm lines can be enabled using COMnEN and SEGmEN if the LCDEN bit is set to 1. When the LCD bit is cleared to 0, then the SCOMn and SSEGm outputs will be fixed at a V_{SS} level.

Bit 3 **COM3EN:** SCOM3/SSEG3 or other pin function selection

0: Other pin-shared functions
1: SCOM3/SSEG3 function

Bit 2 **COM2EN:** SCOM2/SSEG2 or other pin function selection

0: Other pin-shared functions
1: SCOM2/SSEG2 function

Bit 1 **COM1EN:** SCOM1/SSEG1 or other pin function selection

0: Other pin-shared functions
1: SCOM1/SSEG1 function

Bit 0 **COM0EN:** SCOM0/SSEG0 or other pin function selection

0: Other pin-shared functions
1: SCOM0/SSEG0 function

• **SLCDC1 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|--------|--------|----------|----------|----------|----------|----------|----------|
| Name | COM5EN | COM4EN | COMSEGS5 | COMSEGS4 | COMSEGS3 | COMSEGS2 | COMSEGS1 | COMSEGS0 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7 **COM5EN:** SCOM5/SSEG5 or other pin function selection

0: Other pin-shared functions
1: SCOM5/SSEG5 function

Bit 6 **COM4EN:** SCOM4/SSEG4 or other pin function selection

0: Other pin-shared functions
1: SCOM4/SSEG4 function

Bit 5 **COMSEGS5:** SCOM5 or SSEG5 pin function selection

0: SCOM5
1: SSEG5

Bit 4 **COMSEGS4:** SCOM4 or SSEG4 pin function selection

0: SCOM4
1: SSEG4

Bit 3 **COMSEGS3:** SCOM3 or SSEG3 pin function selection

0: SCOM3
1: SSEG3

- Bit 2 **COMSEGS2:** SCOM2 or SSEG2 pin function selection
 0: SCOM2
 1: SSEG2
- Bit 1 **COMSEGS1:** SCOM1 or SSEG1 pin function selection
 0: SCOM1
 1: SSEG1
- Bit 0 **COMSEGS0:** SCOM0 or SSEG0 pin function selection
 0: SCOM0
 1: SSEG0

• **SLCDC2 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---------|---------|---------|---------|--------|--------|--------|--------|
| Name | SEG13EN | SEG12EN | SEG11EN | SEG10EN | SEG9EN | SEG8EN | SEG7EN | SEG6EN |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

- Bit 7 **SEG13EN:** SSEG13 pin function selection
 0: Other pin-shared functions
 1: SSEG13 function
- Bit 6 **SEG12EN:** SSEG12 pin function selection
 0: Other pin-shared functions
 1: SSEG12 function
- Bit 5 **SEG11EN:** SSEG11 pin function selection
 0: Other pin-shared functions
 1: SSEG11 function
- Bit 4 **SEG10EN:** SSEG10 pin function selection
 0: Other pin-shared functions
 1: SSEG10 function
- Bit 3 **SEG9EN:** SSEG9 pin function selection
 0: Other pin-shared functions
 1: SSEG9 function
- Bit 2 **SEG8EN:** SSEG8 pin function selection
 0: Other pin-shared functions
 1: SSEG8 function
- Bit 1 **SEG7EN:** SSEG7 pin function selection
 0: Other pin-shared functions
 1: SSEG7 function
- Bit 0 **SEG6EN:** SSEG6 pin function selection
 0: Other pin-shared functions
 1: SSEG6 function

• **SLCDC3 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---------|---------|---------|---------|---------|---------|---------|---------|
| Name | SEG21EN | SEG20EN | SEG19EN | SEG18EN | SEG17EN | SEG16EN | SEG15EN | SEG14EN |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

- Bit 7 **SEG21EN:** SSEG21 pin function selection
 0: Other pin-shared functions
 1: SSEG21 function
- Bit 6 **SEG20EN:** SSEG20 pin function selection
 0: Other pin-shared functions
 1: SSEG20 function

- Bit 5 **SEG19EN:** SSEG19 pin function selection
0: Other pin-shared functions
1: SSEG19 function
- Bit 4 **SEG18EN:** SSEG18 pin function selection
0: Other pin-shared functions
1: SSEG18 function
- Bit 3 **SEG17EN:** SSEG17 pin function selection
0: Other pin-shared functions
1: SSEG17 function
- Bit 2 **SEG16EN:** SSEG16 pin function selection
0: Other pin-shared functions
1: SSEG16 function
- Bit 1 **SEG15EN:** SSEG15 pin function selection
0: Other pin-shared functions
1: SSEG15 function
- Bit 0 **SEG14EN:** SSEG14 pin function selection
0: Other pin-shared functions
1: SSEG14 function

• **SLCDC4 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---|---|---|---|---|---|---------|---------|
| Name | — | — | — | — | — | — | SEG23EN | SEG22EN |
| R/W | — | — | — | — | — | — | R/W | R/W |
| POR | — | — | — | — | — | — | 0 | 0 |

- Bit 7~2 Unimplemented, read as “0”
- Bit 1 **SEG23EN:** SSEG23 pin function selection
0: Other pin-shared functions
1: SSEG23 function
- Bit 0 **SEG22EN:** SSEG22 pin function selection
0: Other pin-shared functions
1: SSEG22 function

Low Voltage Detector – LVD

The device has a Low Voltage Detector function, also known as LVD. This enabled the device to monitor the power supply voltage, V_{DD} , and provide a warning signal should it fall below a certain level. This function may be especially useful in battery applications where the supply voltage will gradually reduce as the battery ages, as it allows an early warning battery low signal to be generated. The Low Voltage Detector also has the capability of generating an interrupt signal.

LVD Register

The Low Voltage Detector function is controlled using a single register with the name LVDC. Three bits in this register, VLVD2~VLVD0, are used to select one of eight fixed voltages below which a low voltage condition will be determined. A low voltage condition is indicated when the LVDO bit is set. If the LVDO bit is low, this indicates that the V_{DD} voltage is above the preset low voltage value. The LVDEN bit is used to control the overall on/off function of the low voltage detector. Setting the bit high will enable the low voltage detector. Clearing the bit to zero will switch off the internal low voltage detector circuits. As the low voltage detector will consume a certain amount of power, it may be desirable to switch off the circuit when not in use, an important consideration in power sensitive battery powered applications.

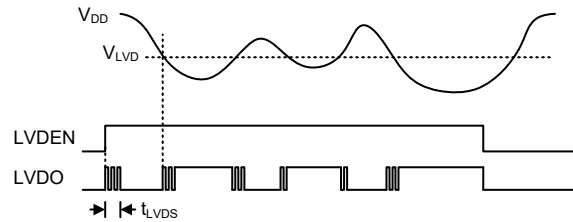
• LVDC Register

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---|---|------|-------|---|-------|-------|-------|
| Name | — | — | LVDO | LVDEN | — | VLVD2 | VLVD1 | VLVD0 |
| R/W | — | — | R | R/W | — | R/W | R/W | R/W |
| POR | — | — | 0 | 0 | — | 0 | 0 | 0 |

- Bit 7~6 Unimplemented, read as “0”
- Bit 5 **LVDO**: LVD output flag
0: No Low Voltage detected
1: Low Voltage detected
- Bit 4 **LVDEN**: Low Voltage Detector Enable control
0: Disable
1: Enable
- Bit 3 unimplemented, read as “0”
- Bit 2~0 **VLVD2~VLVD0**: LVD Voltage selection
000: 2.0V
001: 2.2V
010: 2.4V
011: 2.7V
100: 3.0V
101: 3.3V
110: 3.6V
111: 4.0V

LVD Operation

The Low Voltage Detector function operates by comparing the power supply voltage, V_{DD} , with a pre-specified voltage level stored in the LVDC register. This has a range of between 2.0V and 4.0V. When the power supply voltage, V_{DD} , falls below this pre-determined value, the LVDO bit will be set high indicating a low power supply voltage condition. When the device is powered down the low voltage detector will remain active if the LVDEN bit is high. After enabling the Low Voltage Detector, a time delay t_{LVDS} should be allowed for the circuitry to stabilise before reading the LVDO bit. Note also that as the V_{DD} voltage may rise and fall rather slowly, at the voltage nears that of V_{LVD} , there may be multiple bit LVDO transitions.



LVD Operation

The Low Voltage Detector also has its own interrupt which is contained within one of the Multi-function interrupts, providing an alternative means of low voltage detection, in addition to polling the LVDO bit. The interrupt will only be generated after a delay of t_{LVD} after the LVDO bit has been set high by a low voltage condition. In this case, the LVF interrupt request flag will be set, causing an interrupt to be generated if V_{DD} falls below the preset LVD voltage. This will cause the device to wake-up from the SLEEP or IDLE Mode, however if the Low Voltage Detector wake up function is not required then the LVF flag should be first set high before the device enters the SLEEP or IDLE Mode.

Interrupts

Interrupts are an important part of any microcontroller system. When an external event or an internal function such as a Timer Module or an A/D converter requires microcontroller attention, their corresponding interrupt will enforce a temporary suspension of the main program allowing the microcontroller to direct attention to their respective needs. The device contains several external interrupt and internal interrupt functions. The external interrupts are generated by the action of the external INT0 and INT1 pins, while the internal interrupts are generated by various internal functions such as TMs, Time Base, LVD, EEPROM, SIM, UART and the A/D converter, etc.

Interrupt Registers

Overall interrupt control, which basically means the setting of request flags when certain microcontroller conditions occur and the setting of interrupt enable bits by the application program, is controlled by a series of registers, located in the Special Purpose Data Memory, as shown in the accompanying table. The registers fall into three categories. The first is the INTC0~INTC2 registers which configure the primary interrupts, the second is the MFI0~MFI2 registers which configures the Multi-function interrupts. Finally there is an INTEG register to setup the external interrupt trigger edge type.

Each register contains a number of enable bits to enable or disable individual registers as well as interrupt flags to indicate the presence of an interrupt request. The naming convention of these follows a specific pattern. First is listed an abbreviated interrupt type, then the (optional) number of that interrupt followed by either an “E” for enable/ disable bit or “F” for request flag.

| Function | Enable Bit | Request Flag | Notes |
|------------------------|------------|--------------|----------|
| Global | EMI | — | — |
| INTn Pins | INTnE | INTnF | n=0 or 1 |
| Comparator | CPE | CPF | — |
| Multi-function | MFnE | MFnF | n=0~2 |
| A/D Converter | ADE | ADF | — |
| Time Bases | TBnE | TBnF | n=0 or 1 |
| SIM | SIME | SIMF | — |
| UART | URE | URF | — |
| LVD | LVE | LVF | — |
| EEPROM Write Operation | DEE | DEF | — |
| CTM | CTMPE | CTMPF | — |
| | CTMAE | CTMAF | — |
| STM | STMPE | STMPF | — |
| | STMAE | STMAF | — |
| PTM | PTMPE | PTMPF | — |
| | PTMAE | PTMAF | — |

Interrupt Register Bit Naming Conventions

| Register Name | Bit | | | | | | | |
|---------------|-------|-------|-------|-------|--------|--------|--------|--------|
| | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| INTEG | — | — | — | — | INT1S1 | INT1S0 | INT0S1 | INT0S0 |
| INTC0 | — | MF0F | CPF | INT0F | MF0E | CPE | INT0E | EMI |
| INTC1 | TB0F | ADF | MF2F | MF1F | TB0E | ADE | MF2E | MF1E |
| INTC2 | URF | SIMF | INT1F | TB1F | URE | SIME | INT1E | TB1E |
| MFI0 | — | — | STMAF | STMPF | — | — | STMAE | STMPE |
| MFI1 | CTMAF | CTMPF | PTMAF | PTMPF | CTMAE | CTMPE | PTMAE | PTMPE |
| MFI2 | — | — | DEF | LVF | — | — | DEE | LVE |

Interrupt Register List

• **INTEG Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---|---|---|---|--------|--------|--------|--------|
| Name | — | — | — | — | INT1S1 | INT1S0 | INT0S1 | INT0S0 |
| R/W | — | — | — | — | R/W | R/W | R/W | R/W |
| POR | — | — | — | — | 0 | 0 | 0 | 0 |

- Bit 7~4 Unimplemented, read as “0”
- Bit 3~2 **INT1S1~INT1S0**: Interrupt edge control for INT1 pin
 00: Disable
 01: Rising edge
 10: Falling edge
 11: Rising and falling edges
- Bit 1~0 **INT0S1~INT0S0**: Interrupt edge control for INT0 pin
 00: Disable
 01: Rising edge
 10: Falling edge
 11: Rising and falling edges

• **INTC0 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---|------|-----|-------|------|-----|-------|-----|
| Name | — | MF0F | CPF | INT0F | MF0E | CPE | INT0E | EMI |
| R/W | — | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | — | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

- Bit 7 Unimplemented, read as “0”
- Bit 6 **MF0F**: Multi-function 0 interrupt request flag
 0: No request
 1: Interrupt request
- Bit 5 **CPF**: Comparator interrupt request flag
 0: No request
 1: Interrupt request
- Bit 4 **INT0F**: INT0 interrupt request flag
 0: No request
 1: Interrupt request
- Bit 3 **MF0E**: Multi-function 0 interrupt control
 0: Disable
 1: Enable
- Bit 2 **CPE**: Comparator interrupt control
 0: Disable
 1: Enable
- Bit 1 **INT0E**: INT0 interrupt control
 0: Disable
 1: Enable
- Bit 0 **EMI**: Global interrupt control
 0: Disable
 1: Enable

• **INTC1 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|------|-----|------|------|------|-----|------|------|
| Name | TB0F | ADF | MF2F | MF1F | TB0E | ADE | MF2E | MF1E |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7 **TB0F**: Time Base 0 interrupt request flag

0: No request
1: Interrupt request

Bit 6 **ADF**: A/D Converter interrupt request flag

0: No request
1: Interrupt request

Bit 5 **MF2F**: Multi-function 2 interrupt request flag

0: No request
1: Interrupt request

Bit 4 **MF1F**: Multi-function 1 interrupt request flag

0: No request
1: Interrupt request

Bit 3 **TB0E**: Time Base 0 interrupt control

0: Disable
1: Enable

Bit 2 **ADE**: A/D Converter interrupt control

0: Disable
1: Enable

Bit 1 **MF2E**: Multi-function 2 interrupt control

0: Disable
1: Enable

Bit 0 **MF1E**: Multi-function 1 interrupt control

0: Disable
1: Enable

• **INTC2 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-----|------|-------|------|-----|------|-------|------|
| Name | URF | SIMF | INT1F | TB1F | URE | SIME | INT1E | TB1E |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

Bit 7 **URF**: UART interrupt request flag

0: No request
1: Interrupt request

Bit 6 **SIMF**: SIM interrupt request flag

0: No request
1: Interrupt request

Bit 5 **INT1F**: INT1 interrupt request flag

0: No request
1: Interrupt request

Bit 4 **TB1F**: Time Base 1 interrupt request flag

0: No request
1: Interrupt request

Bit 3 **URE**: UART interrupt control

0: Disable
1: Enable

- Bit 2 **SIME**: SIM interrupt control
 0: Disable
 1: Enable
- Bit 1 **INTIE**: INT1 interrupt control
 0: Disable
 1: Enable
- Bit 0 **TBIE**: Time Base 1 interrupt control
 0: Disable
 1: Enable

• **MFI0 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---|---|-------|-------|---|---|-------|-------|
| Name | — | — | STMAF | STMPF | — | — | STMAE | STMPE |
| R/W | — | — | R/W | R/W | — | — | R/W | R/W |
| POR | — | — | 0 | 0 | — | — | 0 | 0 |

- Bit 7~6 Unimplemented, read as “0”
- Bit 5 **STMAF**: STM Comparator A match interrupt request flag
 0: No request
 1: Interrupt request
- Bit 4 **STMPF**: STM Comparator P match interrupt request flag
 0: No request
 1: Interrupt request
- Bit 3~2 Unimplemented, read as “0”
- Bit 1 **STMAE**: STM Comparator A match interrupt control
 0: Disable
 1: Enable
- Bit 0 **STMPE**: STM Comparator P match interrupt control
 0: Disable
 1: Enable

• **MFI1 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|-------|-------|-------|-------|-------|-------|-------|-------|
| Name | CTMAF | CTMPF | PTMAF | PTMPF | CTMAE | CTMPE | PTMAE | PTMPE |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |

- Bit 7 **CTMAF**: CTM Comparator A match interrupt request flag
 0: No request
 1: Interrupt request
- Bit 6 **CTMPF**: CTM Comparator P match interrupt request flag
 0: No request
 1: Interrupt request
- Bit 5 **PTMAF**: PTM Comparator A match interrupt request flag
 0: No request
 1: Interrupt request
- Bit 4 **PTMPF**: PTM Comparator P match interrupt request flag
 0: No request
 1: Interrupt request
- Bit 3 **CTMAE**: CTM Comparator A match interrupt control
 0: Disable
 1: Enable

- Bit 2 **CTMPE**: CTM Comparator P match interrupt control
0: Disable
1: Enable
- Bit 1 **PTMAE**: PTM Comparator A match interrupt control
0: Disable
1: Enable
- Bit 0 **PTMPE**: PTM Comparator P match interrupt control
0: Disable
1: Enable

• **MF12 Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|---|---|-----|-----|---|---|-----|-----|
| Name | — | — | DEF | LVF | — | — | DEE | LVE |
| R/W | — | — | R/W | R/W | — | — | R/W | R/W |
| POR | — | — | 0 | 0 | — | — | 0 | 0 |

- Bit 7~6 Unimplemented, read as “0”
- Bit 5 **DEF**: Data EEPROM interrupt request flag
0: No request
1: Interrupt request
- Bit 4 **LVF**: LVD interrupt request flag
0: No request
1: Interrupt request
- Bit 3~2 Unimplemented, read as “0”
- Bit 1 **DEE**: Data EEPROM interrupt control
0: Disable
1: Enable
- Bit 0 **LVE**: LVD interrupt control
0: Disable
1: Enable

Interrupt Operation

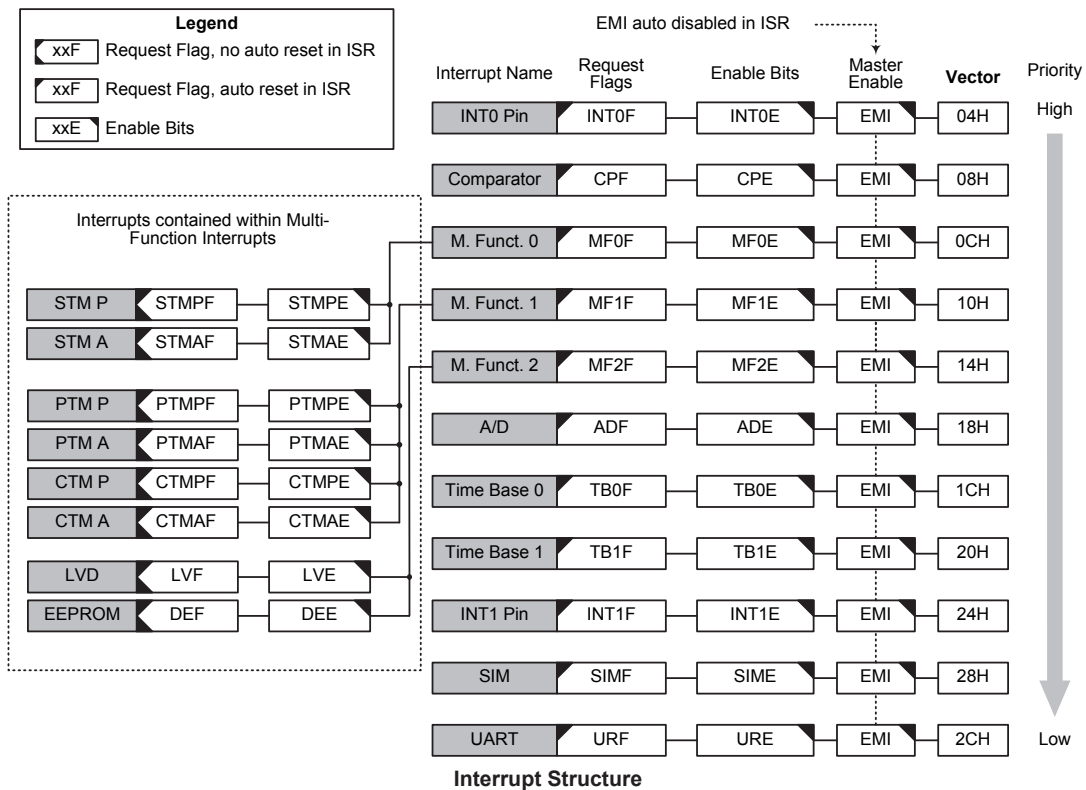
When the conditions for an interrupt event occur, such as a TM Comparator P or Comparator A match etc., the relevant interrupt request flag will be set. Whether the request flag actually generates a program jump to the relevant interrupt vector is determined by the condition of the interrupt enable bit. If the enable bit is set high then the program will jump to its relevant vector, if the enable bit is zero then although the interrupt request flag is set an actual interrupt will not be generated and the program will not jump to the relevant interrupt vector. The global interrupt enable bit, if cleared to zero, will disable all interrupts.

When an interrupt is generated, the Program Counter, which stores the address of the next instruction to be executed, will be transferred onto the stack. The Program Counter will then be loaded with a new address which will be the value of the corresponding interrupt vector. The microcontroller will then fetch its next instruction from this interrupt vector. The instruction at this vector will usually be a “JMP” which will jump to another section of program which is known as the interrupt service routine. Here is located the code to control the appropriate interrupt. The interrupt service routine must be terminated with a “RETI”, which retrieves the original Program Counter address from the stack and allows the microcontroller to continue with normal execution at the point where the interrupt occurred.

The various interrupt enable bits, together with their associated request flags, are shown in the Accompanying diagrams with their order of priority. Some interrupt sources have their own individual vector while others share the same multi-function interrupt vector. Once an interrupt subroutine is serviced, all the other interrupts will be blocked, as the global interrupt enable bit,

EMI bit will be cleared automatically. This will prevent any further interrupt nesting from occurring. However, if other interrupt requests occur during this interval, although the interrupt will not be immediately serviced, the request flag will still be recorded.

If an interrupt requires immediate servicing while the program is already in another interrupt service routine, the EMI bit should be set after entering the routine, to allow interrupt nesting. If the stack is full, the interrupt request will not be acknowledged, even if the related interrupt is enabled, until the Stack Pointer is decremented. If immediate service is desired, the stack must be prevented from becoming full. In case of simultaneous requests, the accompanying diagram shows the priority that is applied. All of the interrupt request flags when set will wake-up the device if it is in SLEEP or IDLE Mode, however to prevent a wake-up from occurring the corresponding flag should be set before the device is in SLEEP or IDLE Mode.



External Interrupt

The external interrupts are controlled by signal transitions on the pins INT0~INT1. An external interrupt request will take place when the external interrupt request flags, INT0F~INT1F, are set, which will occur when a transition, whose type is chosen by the edge select bits, appears on the external interrupt pins. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, and respective external interrupt enable bit, INT0E~INT1E, must first be set. Additionally the correct interrupt edge type must be selected using the INTEG register to enable the external interrupt function and to choose the trigger edge type. As the external interrupt pins are pin-shared with I/O pins, they can only be configured as external interrupt pins if their external interrupt enable bit in the corresponding interrupt register has been set. The pin must also be setup as an input by setting the corresponding bit in the port control register. When the interrupt is enabled, the stack is not full and the correct transition type appears on the external interrupt pin, a subroutine call to the external interrupt vector, will take place. When the interrupt is serviced, the

external interrupt request flags, INT0F~INT1F, will be automatically reset and the EMI bit will be automatically cleared to disable other interrupts. Note that any pull-high resistor selections on the external interrupt pins will remain valid even if the pin is used as an external interrupt input.

The INTEG register is used to select the type of active edge that will trigger the external interrupt. A choice of either rising or falling or both edge types can be chosen to trigger an external interrupt. Note that the INTEG register can also be used to disable the external interrupt function.

Comparator Interrupt

The comparator interrupt is controlled by the internal comparator. A comparator interrupt request will take place when the comparator interrupt request flag, CPF, is set, a situation that will occur when the comparator output bit changes state. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, and comparator interrupt enable bit, CPE, must first be set. When the interrupt is enabled, the stack is not full and the comparator inputs generate a comparator output transition, a subroutine call to the comparator interrupt vector, will take place. When the interrupt is serviced, the comparator interrupt request flag will be automatically reset and the EMI bit will be automatically cleared to disable other interrupts.

Multi-function Interrupts

Within the device there are three Multi-function interrupts. Unlike the other independent interrupts, these interrupts have no independent source, but rather are formed from other existing interrupt sources, namely the TM interrupts, LVD interrupt and EEPROM write operation interrupt.

A Multi-function interrupt request will take place when any of the Multi-function interrupt request flags MFnF are set. The Multi-function interrupt flag will be set when any of their included functions generate an interrupt request flag. To allow the program to branch to its respective interrupt vector address, when the Multi-function interrupt is enabled and the stack is not full, and any one of the interrupts contained within each of the Multi-function interrupt occurs, a subroutine call to the related Multi-function interrupt vector will take place. When the interrupt is serviced, the related Multi-function request flag will be automatically reset and the EMI bit will be automatically cleared to disable other interrupts.

However, it must be noted that, although the Multi-function Interrupt request flags will be automatically reset when the interrupt is serviced, the request flags from the original source of the Multi-function interrupt will not be automatically reset and must be manually reset by the application program.

A/D Converter Interrupt

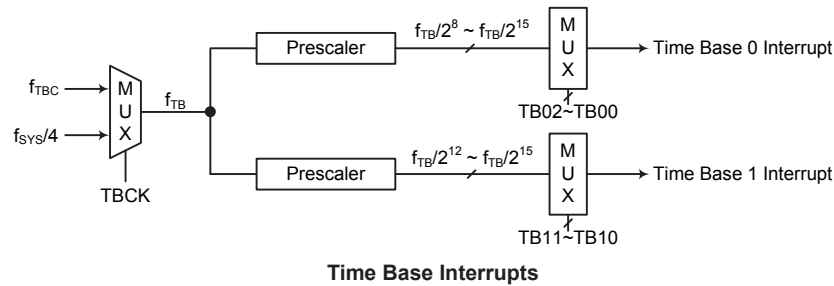
The A/D Converter Interrupt is controlled by the termination of an A/D conversion process. An A/D Converter Interrupt request will take place when the A/D Converter Interrupt request flag, ADF, is set, which occurs when the A/D conversion process finishes. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, and A/D Interrupt enable bit, ADE, must first be set. When the interrupt is enabled, the stack is not full and the A/D conversion process has ended, a subroutine call to the A/D Converter Interrupt vector, will take place. When the interrupt is serviced, the A/D Converter Interrupt flag, ADF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

Time Base Interrupts

The function of the Time Base Interrupt is to provide regular time signal in the form of an internal interrupt. They are controlled by the overflow signals from their respective timer functions. When these happen their respective interrupt request flags, TB0F or TB1F, will be set. To allow the

program to branch to their respective interrupt vector addresses, the global interrupt enable bit, EMI and Time Base enable bit, TB0E or TB1E, must first be set. When the interrupt is enabled, the stack is not full and the Time Base overflows, a subroutine call to their respective vector locations will take place. When the interrupt is serviced, the respective interrupt request flag, TB0F or TB1F, will be automatically reset and the EMI bit will be cleared to disable other interrupts.

The purpose of the Time Base Interrupt is to provide an interrupt signal at fixed time periods. Their clock sources originate from the internal clock source f_{TB} . This f_{TB} input clock passes through a divider, the division ratio of which is selected by programming the appropriate bits in the TBC register to obtain longer interrupt periods whose value ranges. The clock source that generates f_{TB} , which in turn controls the Time Base interrupt period, can originate from several different sources, as shown in the System Operating Mode section.



• **TBC Register**

| Bit | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
|------|------|------|------|------|-------|------|------|------|
| Name | TBON | TBCK | TB11 | TB10 | LXTLP | TB02 | TB01 | TB00 |
| R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W | R/W |
| POR | 0 | 0 | 1 | 1 | 0 | 1 | 1 | 1 |

- Bit 7 **TBON**: Time Base function enable control
 0: Disable
 1: Enable
- Bit 6 **TBCK**: Time Base clock source selection
 0: f_{TBC}
 1: $f_{SYS}/4$
- Bit 5~4 **TB11~TB10**: Time Base 1 time-out period selection
 00: $2^{12}/f_{TB}$
 01: $2^{13}/f_{TB}$
 10: $2^{14}/f_{TB}$
 11: $2^{15}/f_{TB}$
- Bit 3 **LXTLP**: LXT Low Power control
 0: Disable – LXT quick start-up
 1: Enable – LXT slow start-up
- Bit 2~0 **TB02~TB00**: Time Base 0 time-out period selection
 000: $2^8/f_{TB}$
 001: $2^9/f_{TB}$
 010: $2^{10}/f_{TB}$
 011: $2^{11}/f_{TB}$
 100: $2^{12}/f_{TB}$
 101: $2^{13}/f_{TB}$
 110: $2^{14}/f_{TB}$
 111: $2^{15}/f_{TB}$

Serial Interface Module Interrupt

The Serial Interface Module Interrupt, also known as the SIM interrupt, is controlled by the SPI or I²C data transfer. A SIM Interrupt request will take place when the SIM Interrupt request flag, SIMF, is set, which occurs when a byte of data has been received or transmitted by the SIM interface, an I²C slave address match or I²C bus time-out occurrence. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI and the Serial Interface Interrupt enable bit, SIME, must first be set. When the interrupt is enabled, the stack is not full and any of the above described situations occurs, a subroutine call to the respective SIM Interrupt vector, will take place. When the Serial Interface Interrupt is serviced, the SIMF flag will also be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

UART Transfer Interrupt

The UART Transfer Interrupt is controlled by several UART transfer conditions. When one of these conditions occurs, an interrupt pulse will be generated to get the attention of the microcontroller. These conditions are a transmitter data register empty, transmitter idle, receiver data available, receiver overrun, address detect and an RX pin wake-up. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, and UART Interrupt enable bit, URE, must first be set. When the interrupt is enabled, the stack is not full and any of the conditions described above occurs, a subroutine call to the UART Interrupt vector, will take place. When the interrupt is serviced, the UART Interrupt flag, URF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

LVD Interrupt

The Low Voltage Detector Interrupt is contained within the Multi-function Interrupt. An LVD Interrupt request will take place when the LVD Interrupt request flag, LVF, is set, which occurs when the Low Voltage Detector function detects a low power supply voltage. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, Low Voltage Interrupt enable bit, LVE, and associated Multi-function interrupt enable bit, must first be set. When the interrupt is enabled, the stack is not full and a low voltage condition occurs, a subroutine call to the Multi-function Interrupt vector, will take place. When the Low Voltage Interrupt is serviced, the EMI bit will be automatically cleared to disable other interrupts. However, only the Multi-function interrupt request flag will be also automatically cleared. As the LVF flag will not be automatically cleared, it has to be cleared by the application program.

EEPROM Write Interrupt

The EEPROM Write Interrupt is contained within the Multi-function Interrupt. An EEPROM Write Interrupt request will take place when the EEPROM Write Interrupt request flag, DEF, is set, which occurs when an EEPROM Write cycle ends. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, EEPROM Write Interrupt enable bit, DEE, and associated Multi-function interrupt enable bit must first be set. When the interrupt is enabled, the stack is not full and an EEPROM Write cycle ends, a subroutine call to the respective Multi-function Interrupt vector will take place. When the EEPROM Write Interrupt is serviced, the EMI bit will be automatically cleared to disable other interrupts. However, only the Multi-function interrupt request flag will be automatically cleared. As the DEF flag will not be automatically cleared, it has to be cleared by the application program.

Timer Module Interrupts

The Compact, Standard and Periodic TMs have two interrupts, one comes from the comparator A match situation and the other comes from the comparator P match situation. All of the TM interrupts are contained within the Multi-function Interrupts. For all of the TM types there are two interrupt request flags and two enable control bits. A TM interrupt request will take place when any of the TM request flags are set, a situation which occurs when a TM comparator P or A match situation happens.

To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, respective TM Interrupt enable bit, and relevant Multi-function Interrupt enable bit, MFnE, must first be set. When the interrupt is enabled, the stack is not full and a TM comparator match situation occurs, a subroutine call to the relevant Multi-function Interrupt vector locations, will take place. When the TM interrupt is serviced, the EMI bit will be automatically cleared to disable other interrupts. However, only the related MFnF flag will be automatically cleared. As the TM interrupt request flags will not be automatically cleared, they have to be cleared by the application program.

Interrupt Wake-up Function

Each of the interrupt functions has the capability of waking up the microcontroller when in the SLEEP or IDLE Mode. A wake-up is generated when an interrupt request flag changes from low to high and is independent of whether the interrupt is enabled or not. Therefore, even though the device is in the SLEEP or IDLE Mode and its system oscillator stopped, situations such as external edge transitions on the external interrupt pins, a low power supply voltage or comparator input change may cause their respective interrupt flag to be set high and consequently generate an interrupt. Care must therefore be taken if spurious wake-up situations are to be avoided. If an interrupt wake-up function is to be disabled then the corresponding interrupt request flag should be set high before the device enters the SLEEP or IDLE Mode. The interrupt enable bits have no effect on the interrupt wake-up function.

Programming Considerations

By disabling the relevant interrupt enable bits, a requested interrupt can be prevented from being serviced, however, once an interrupt request flag is set, it will remain in this condition in the interrupt register until the corresponding interrupt is serviced or until the request flag is cleared by the application program.

Where a certain interrupt is contained within a Multi-function interrupt, then when the interrupt service routine is executed, as only the Multi-function interrupt request flags, MFnF, will be automatically cleared, the individual request flag for the function needs to be cleared by the application program.

It is recommended that programs do not use the “CALL” instruction within the interrupt service subroutine. Interrupts often occur in an unpredictable manner or need to be serviced immediately. If only one stack is left and the interrupt is not well controlled, the original control sequence will be damaged once a CALL subroutine is executed in the interrupt subroutine.

Every interrupt has the capability of waking up the microcontroller when it is in SLEEP or IDLE Mode, the wake up being generated when the interrupt request flag changes from low to high. If it is required to prevent a certain interrupt from waking up the microcontroller then its respective request flag should be first set high before enter SLEEP or IDLE Mode.

As only the Program Counter is pushed onto the stack, then when the interrupt is serviced, if the contents of the accumulator, status register or other registers are altered by the interrupt service program, their contents should be saved to the memory at the beginning of the interrupt service routine.

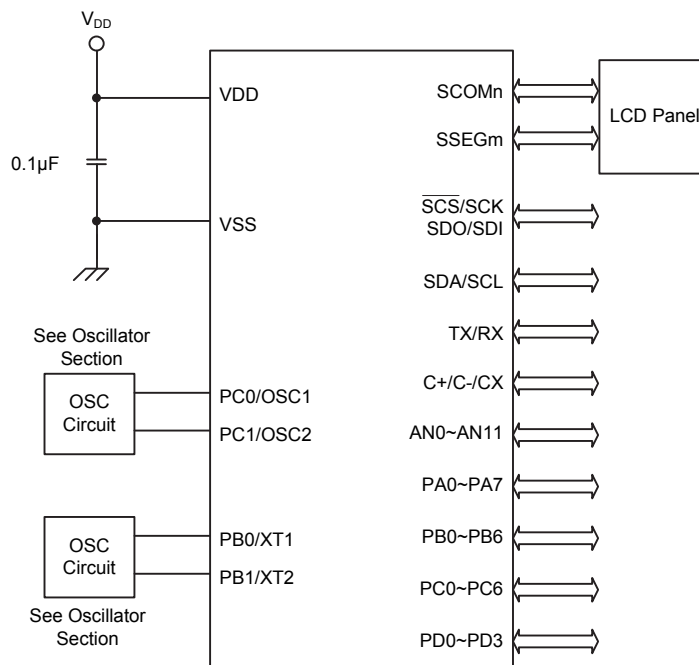
To return from an interrupt subroutine, either a RET or RETI instruction may be executed. The RETI instruction in addition to executing a return to the main program also automatically sets the EMI bit high to allow further interrupts. The RET instruction however only executes a return to the main program leaving the EMI bit in its present zero state and therefore disabling the execution of further interrupts.

Configuration Options

Configuration options refer to certain options within the MCU that are programmed into the device during the programming process. During the development process, these options are selected using the HT-IDE software development tools. As these options are programmed into the device using the hardware programming tools, once they are selected they cannot be changed later using the application program. All options must be defined for proper system function, the details of which are shown in the table.

| No. | Options |
|-----|---|
| 1 | High Speed System Oscillator Selection – f_H : HXT or HIRC |
| 2 | Low Speed System Oscillator Selection – f_{SUB} : LXT or LIRC |
| 3 | HIRC Frequency Selection – f_{HIRC} : 8MHz, 12MHz or 16MHz |

Application Circuits



Instruction Set

Introduction

Central to the successful operation of any microcontroller is its instruction set, which is a set of program instruction codes that directs the microcontroller to perform certain operations. In the case of Holtek microcontroller, a comprehensive and flexible set of over 60 instructions is provided to enable programmers to implement their application with the minimum of programming overheads.

For easier understanding of the various instruction codes, they have been subdivided into several functional groupings.

Instruction Timing

Most instructions are implemented within one instruction cycle. The exceptions to this are branch, call, or table read instructions where two instruction cycles are required. One instruction cycle is equal to 4 system clock cycles, therefore in the case of an 8MHz system oscillator, most instructions would be implemented within 0.5 μ s and branch or call instructions would be implemented within 1 μ s. Although instructions which require one more cycle to implement are generally limited to the JMP, CALL, RET, RETI and table read instructions, it is important to realize that any other instructions which involve manipulation of the Program Counter Low register or PCL will also take one more cycle to implement. As instructions which change the contents of the PCL will imply a direct jump to that new address, one more cycle will be required. Examples of such instructions would be "CLR PCL" or "MOV PCL, A". For the case of skip instructions, it must be noted that if the result of the comparison involves a skip operation then this will also take one more cycle, if no skip is involved then only one cycle is required.

Moving and Transferring Data

The transfer of data within the microcontroller program is one of the most frequently used operations. Making use of three kinds of MOV instructions, data can be transferred from registers to the Accumulator and vice-versa as well as being able to move specific immediate data directly into the Accumulator. One of the most important data transfer applications is to receive data from the input ports and transfer data to the output ports.

Arithmetic Operations

The ability to perform certain arithmetic operations and data manipulation is a necessary feature of most microcontroller applications. Within the Holtek microcontroller instruction set are a range of add and subtract instruction mnemonics to enable the necessary arithmetic to be carried out. Care must be taken to ensure correct handling of carry and borrow data when results exceed 255 for addition and less than 0 for subtraction. The increment and decrement instructions INC, INCA, DEC and DECA provide a simple means of increasing or decreasing by a value of one of the values in the destination specified.

Logical and Rotate Operation

The standard logical operations such as AND, OR, XOR and CPL all have their own instruction within the Holtek microcontroller instruction set. As with the case of most instructions involving data manipulation, data must pass through the Accumulator which may involve additional programming steps. In all logical data operations, the zero flag may be set if the result of the operation is zero. Another form of logical data manipulation comes from the rotate instructions such as RR, RL, RRC and RLC which provide a simple means of rotating one bit right or left. Different rotate instructions exist depending on program requirements. Rotate instructions are useful for serial port programming applications where data can be rotated from an internal register into the Carry bit from where it can be examined and the necessary serial bit set high or low. Another application which rotate data operations are used is to implement multiplication and division calculations.

Branches and Control Transfer

Program branching takes the form of either jumps to specified locations using the JMP instruction or to a subroutine using the CALL instruction. They differ in the sense that in the case of a subroutine call, the program must return to the instruction immediately when the subroutine has been carried out. This is done by placing a return instruction "RET" in the subroutine which will cause the program to jump back to the address right after the CALL instruction. In the case of a JMP instruction, the program simply jumps to the desired location. There is no requirement to jump back to the original jumping off point as in the case of the CALL instruction. One special and extremely useful set of branch instructions are the conditional branches. Here a decision is first made regarding the condition of a certain data memory or individual bits. Depending upon the conditions, the program will continue with the next instruction or skip over it and jump to the following instruction. These instructions are the key to decision making and branching within the program perhaps determined by the condition of certain input switches or by the condition of internal data bits.

Bit Operations

The ability to provide single bit operations on Data Memory is an extremely flexible feature of all Holtek microcontrollers. This feature is especially useful for output port bit programming where individual bits or port pins can be directly set high or low using either the "SET [m].i" or "CLR [m].i" instructions respectively. The feature removes the need for programmers to first read the 8-bit output port, manipulate the input data to ensure that other bits are not changed and then output the port with the correct new data. This read-modify-write process is taken care of automatically when these bit operation instructions are used.

Table Read Operations

Data storage is normally implemented by using registers. However, when working with large amounts of fixed data, the volume involved often makes it inconvenient to store the fixed data in the Data Memory. To overcome this problem, Holtek microcontrollers allow an area of Program Memory to be set as a table where data can be directly stored. A set of easy to use instructions provides the means by which this fixed data can be referenced and retrieved from the Program Memory.

Other Operations

In addition to the above functional instructions, a range of other instructions also exist such as the "HALT" instruction for Power-down operations and instructions to control the operation of the Watchdog Timer for reliable program operations under extreme electric or electromagnetic environments. For their relevant operations, refer to the functional related sections.

Instruction Set Summary

The following table depicts a summary of the instruction set categorised according to function and can be consulted as a basic instruction reference using the following listed conventions.

Table Conventions

- x: Bits immediate data
- m: Data Memory address
- A: Accumulator
- i: 0~7 number of bits
- addr: Program memory address

| Mnemonic | Description | Cycles | Flag Affected |
|----------------------------------|---|-------------------|---------------|
| Arithmetic | | | |
| ADD A,[m] | Add Data Memory to ACC | 1 | Z, C, AC, OV |
| ADDM A,[m] | Add ACC to Data Memory | 1 ^{Note} | Z, C, AC, OV |
| ADD A,x | Add immediate data to ACC | 1 | Z, C, AC, OV |
| ADC A,[m] | Add Data Memory to ACC with Carry | 1 | Z, C, AC, OV |
| ADCM A,[m] | Add ACC to Data memory with Carry | 1 ^{Note} | Z, C, AC, OV |
| SUB A,x | Subtract immediate data from the ACC | 1 | Z, C, AC, OV |
| SUB A,[m] | Subtract Data Memory from ACC | 1 | Z, C, AC, OV |
| SUBM A,[m] | Subtract Data Memory from ACC with result in Data Memory | 1 ^{Note} | Z, C, AC, OV |
| SBC A,[m] | Subtract Data Memory from ACC with Carry | 1 | Z, C, AC, OV |
| SBCM A,[m] | Subtract Data Memory from ACC with Carry, result in Data Memory | 1 ^{Note} | Z, C, AC, OV |
| DAA [m] | Decimal adjust ACC for Addition with result in Data Memory | 1 ^{Note} | C |
| Logic Operation | | | |
| AND A,[m] | Logical AND Data Memory to ACC | 1 | Z |
| OR A,[m] | Logical OR Data Memory to ACC | 1 | Z |
| XOR A,[m] | Logical XOR Data Memory to ACC | 1 | Z |
| ANDM A,[m] | Logical AND ACC to Data Memory | 1 ^{Note} | Z |
| ORM A,[m] | Logical OR ACC to Data Memory | 1 ^{Note} | Z |
| XORM A,[m] | Logical XOR ACC to Data Memory | 1 ^{Note} | Z |
| AND A,x | Logical AND immediate Data to ACC | 1 | Z |
| OR A,x | Logical OR immediate Data to ACC | 1 | Z |
| XOR A,x | Logical XOR immediate Data to ACC | 1 | Z |
| CPL [m] | Complement Data Memory | 1 ^{Note} | Z |
| CPLA [m] | Complement Data Memory with result in ACC | 1 | Z |
| Increment & Decrement | | | |
| INCA [m] | Increment Data Memory with result in ACC | 1 | Z |
| INC [m] | Increment Data Memory | 1 ^{Note} | Z |
| DECA [m] | Decrement Data Memory with result in ACC | 1 | Z |
| DEC [m] | Decrement Data Memory | 1 ^{Note} | Z |
| Rotate | | | |
| RRA [m] | Rotate Data Memory right with result in ACC | 1 | None |
| RR [m] | Rotate Data Memory right | 1 ^{Note} | None |
| RRCA [m] | Rotate Data Memory right through Carry with result in ACC | 1 | C |
| RRC [m] | Rotate Data Memory right through Carry | 1 ^{Note} | C |
| RLA [m] | Rotate Data Memory left with result in ACC | 1 | None |
| RL [m] | Rotate Data Memory left | 1 ^{Note} | None |
| RLCA [m] | Rotate Data Memory left through Carry with result in ACC | 1 | C |
| RLC [m] | Rotate Data Memory left through Carry | 1 ^{Note} | C |

| Mnemonic | Description | Cycles | Flag Affected |
|-----------------------------|--|-------------------|---------------|
| Data Move | | | |
| MOV A,[m] | Move Data Memory to ACC | 1 | None |
| MOV [m],A | Move ACC to Data Memory | 1 ^{Note} | None |
| MOV A,x | Move immediate data to ACC | 1 | None |
| Bit Operation | | | |
| CLR [m].i | Clear bit of Data Memory | 1 ^{Note} | None |
| SET [m].i | Set bit of Data Memory | 1 ^{Note} | None |
| Branch Operation | | | |
| JMP addr | Jump unconditionally | 2 | None |
| SZ [m] | Skip if Data Memory is zero | 1 ^{Note} | None |
| SZA [m] | Skip if Data Memory is zero with data movement to ACC | 1 ^{Note} | None |
| SZ [m].i | Skip if bit i of Data Memory is zero | 1 ^{Note} | None |
| SNZ [m].i | Skip if bit i of Data Memory is not zero | 1 ^{Note} | None |
| SIZ [m] | Skip if increment Data Memory is zero | 1 ^{Note} | None |
| SDZ [m] | Skip if decrement Data Memory is zero | 1 ^{Note} | None |
| SIZA [m] | Skip if increment Data Memory is zero with result in ACC | 1 ^{Note} | None |
| SDZA [m] | Skip if decrement Data Memory is zero with result in ACC | 1 ^{Note} | None |
| CALL addr | Subroutine call | 2 | None |
| RET | Return from subroutine | 2 | None |
| RET A,x | Return from subroutine and load immediate data to ACC | 2 | None |
| RETI | Return from interrupt | 2 | None |
| Table Read Operation | | | |
| TABRD [m] | Read table (specific page or current page) to TBLH and Data Memory | 2 ^{Note} | None |
| TABRDL [m] | Read table (last page) to TBLH and Data Memory | 2 ^{Note} | None |
| Miscellaneous | | | |
| NOP | No operation | 1 | None |
| CLR [m] | Clear Data Memory | 1 ^{Note} | None |
| SET [m] | Set Data Memory | 1 ^{Note} | None |
| CLR WDT | Clear Watchdog Timer | 1 | TO, PDF |
| SWAP [m] | Swap nibbles of Data Memory | 1 ^{Note} | None |
| SWAPA [m] | Swap nibbles of Data Memory with result in ACC | 1 | None |
| HALT | Enter power down mode | 1 | TO, PDF |

Note: 1. For skip instructions, if the result of the comparison involves a skip then two cycles are required, if no skip takes place only one cycle is required.

2. Any instruction which changes the contents of the PCL will also require 2 cycles for execution.

Instruction Definition

| | |
|-------------------|---|
| ADC A,[m] | Add Data Memory to ACC with Carry |
| Description | The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the Accumulator. |
| Operation | $ACC \leftarrow ACC + [m] + C$ |
| Affected flag(s) | OV, Z, AC, C |
| ADCM A,[m] | Add ACC to Data Memory with Carry |
| Description | The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the specified Data Memory. |
| Operation | $[m] \leftarrow ACC + [m] + C$ |
| Affected flag(s) | OV, Z, AC, C |
| ADD A,[m] | Add Data Memory to ACC |
| Description | The contents of the specified Data Memory and the Accumulator are added. The result is stored in the Accumulator. |
| Operation | $ACC \leftarrow ACC + [m]$ |
| Affected flag(s) | OV, Z, AC, C |
| ADD A,x | Add immediate data to ACC |
| Description | The contents of the Accumulator and the specified immediate data are added. The result is stored in the Accumulator. |
| Operation | $ACC \leftarrow ACC + x$ |
| Affected flag(s) | OV, Z, AC, C |
| ADDM A,[m] | Add ACC to Data Memory |
| Description | The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory. |
| Operation | $[m] \leftarrow ACC + [m]$ |
| Affected flag(s) | OV, Z, AC, C |
| AND A,[m] | Logical AND Data Memory to ACC |
| Description | Data in the Accumulator and the specified Data Memory perform a bitwise logical AND operation. The result is stored in the Accumulator. |
| Operation | $ACC \leftarrow ACC \text{ "AND" } [m]$ |
| Affected flag(s) | Z |
| AND A,x | Logical AND immediate data to ACC |
| Description | Data in the Accumulator and the specified immediate data perform a bit wise logical AND operation. The result is stored in the Accumulator. |
| Operation | $ACC \leftarrow ACC \text{ "AND" } x$ |
| Affected flag(s) | Z |
| ANDM A,[m] | Logical AND ACC to Data Memory |
| Description | Data in the specified Data Memory and the Accumulator perform a bitwise logical AND operation. The result is stored in the Data Memory. |
| Operation | $[m] \leftarrow ACC \text{ "AND" } [m]$ |
| Affected flag(s) | Z |

| | |
|------------------|--|
| CALL addr | Subroutine call |
| Description | Unconditionally calls a subroutine at the specified address. The Program Counter then increments by 1 to obtain the address of the next instruction which is then pushed onto the stack. The specified address is then loaded and the program continues execution from this new address. As this instruction requires an additional operation, it is a two cycle instruction. |
| Operation | Stack ← Program Counter + 1 Program Counter ← addr |
| Affected flag(s) | None |
| | |
| CLR [m] | Clear Data Memory |
| Description | Each bit of the specified Data Memory is cleared to 0. |
| Operation | [m] ← 00H |
| Affected flag(s) | None |
| | |
| CLR [m].i | Clear bit of Data Memory |
| Description | Bit i of the specified Data Memory is cleared to 0. |
| Operation | [m].i ← 0 |
| Affected flag(s) | None |
| | |
| CLR WDT | Clear Watchdog Timer |
| Description | The TO, PDF flags and the WDT are all cleared. |
| Operation | WDT cleared TO ← 0 PDF ← 0 |
| Affected flag(s) | TO, PDF |
| | |
| CPL [m] | Complement Data Memory |
| Description | Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa. |
| Operation | [m] ← $\overline{[m]}$ |
| Affected flag(s) | Z |
| | |
| CPLA [m] | Complement Data Memory with result in ACC |
| Description | Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa. The complemented result is stored in the Accumulator and the contents of the Data Memory remain unchanged. |
| Operation | ACC ← $\overline{[m]}$ |
| Affected flag(s) | Z |
| | |
| DAA [m] | Decimal-Adjust ACC for addition with result in Data Memory |
| Description | Convert the contents of the Accumulator value to a BCD (Binary Coded Decimal) value resulting from the previous addition of two BCD variables. If the low nibble is greater than 9 or if AC flag is set, then a value of 6 will be added to the low nibble. Otherwise the low nibble remains unchanged. If the high nibble is greater than 9 or if the C flag is set, then a value of 6 will be added to the high nibble. Essentially, the decimal conversion is performed by adding 00H, 06H, 60H or 66H depending on the Accumulator and flag conditions. Only the C flag may be affected by this instruction which indicates that if the original BCD sum is greater than 100, it allows multiple precision decimal addition. |
| Operation | [m] ← ACC + 00H or [m] ← ACC + 06H or [m] ← ACC + 60H or [m] ← ACC + 66H |
| Affected flag(s) | C |

| | |
|------------------|--|
| DEC [m] | Decrement Data Memory |
| Description | Data in the specified Data Memory is decremented by 1. |
| Operation | $[m] \leftarrow [m] - 1$ |
| Affected flag(s) | Z |
| DECA [m] | Decrement Data Memory with result in ACC |
| Description | Data in the specified Data Memory is decremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged. |
| Operation | $ACC \leftarrow [m] - 1$ |
| Affected flag(s) | Z |
| HALT | Enter power down mode |
| Description | This instruction stops the program execution and turns off the system clock. The contents of the Data Memory and registers are retained. The WDT and prescaler are cleared. The power down flag PDF is set and the WDT time-out flag TO is cleared. |
| Operation | $TO \leftarrow 0$ $PDF \leftarrow 1$ |
| Affected flag(s) | TO, PDF |
| INC [m] | Increment Data Memory |
| Description | Data in the specified Data Memory is incremented by 1. |
| Operation | $[m] \leftarrow [m] + 1$ |
| Affected flag(s) | Z |
| INCA [m] | Increment Data Memory with result in ACC |
| Description | Data in the specified Data Memory is incremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged. |
| Operation | $ACC \leftarrow [m] + 1$ |
| Affected flag(s) | Z |
| JMP addr | Jump unconditionally |
| Description | The contents of the Program Counter are replaced with the specified address. Program execution then continues from this new address. As this requires the insertion of a dummy instruction while the new address is loaded, it is a two cycle instruction. |
| Operation | Program Counter \leftarrow addr |
| Affected flag(s) | None |
| MOV A,[m] | Move Data Memory to ACC |
| Description | The contents of the specified Data Memory are copied to the Accumulator. |
| Operation | $ACC \leftarrow [m]$ |
| Affected flag(s) | None |
| MOV A,x | Move immediate data to ACC |
| Description | The immediate data specified is loaded into the Accumulator. |
| Operation | $ACC \leftarrow x$ |
| Affected flag(s) | None |
| MOV [m],A | Move ACC to Data Memory |
| Description | The contents of the Accumulator are copied to the specified Data Memory. |
| Operation | $[m] \leftarrow ACC$ |
| Affected flag(s) | None |

| | |
|------------------|--|
| NOP | No operation |
| Description | No operation is performed. Execution continues with the next instruction. |
| Operation | No operation |
| Affected flag(s) | None |
| | |
| OR A,[m] | Logical OR Data Memory to ACC |
| Description | Data in the Accumulator and the specified Data Memory perform a bitwise logical OR operation. The result is stored in the Accumulator. |
| Operation | ACC ← ACC "OR" [m] |
| Affected flag(s) | Z |
| | |
| OR A,x | Logical OR immediate data to ACC |
| Description | Data in the Accumulator and the specified immediate data perform a bitwise logical OR operation. The result is stored in the Accumulator. |
| Operation | ACC ← ACC "OR" x |
| Affected flag(s) | Z |
| | |
| ORM A,[m] | Logical OR ACC to Data Memory |
| Description | Data in the specified Data Memory and the Accumulator perform a bitwise logical OR operation. The result is stored in the Data Memory. |
| Operation | [m] ← ACC "OR" [m] |
| Affected flag(s) | Z |
| | |
| RET | Return from subroutine |
| Description | The Program Counter is restored from the stack. Program execution continues at the restored address. |
| Operation | Program Counter ← Stack |
| Affected flag(s) | None |
| | |
| RET A,x | Return from subroutine and load immediate data to ACC |
| Description | The Program Counter is restored from the stack and the Accumulator loaded with the specified immediate data. Program execution continues at the restored address. |
| Operation | Program Counter ← Stack ACC ← x |
| Affected flag(s) | None |
| | |
| RETI | Return from interrupt |
| Description | The Program Counter is restored from the stack and the interrupts are re-enabled by setting the EMI bit. EMI is the master interrupt global enable bit. If an interrupt was pending when the RETI instruction is executed, the pending Interrupt routine will be processed before returning to the main program. |
| Operation | Program Counter ← Stack EMI ← 1 |
| Affected flag(s) | None |
| | |
| RL [m] | Rotate Data Memory left |
| Description | The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0. |
| Operation | [m].(i+1) ← [m].i; (i=0~6) [m].0 ← [m].7 |
| Affected flag(s) | None |

| | |
|------------------|---|
| RLA [m] | Rotate Data Memory left with result in ACC |
| Description | The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged. |
| Operation | ACC.(i+1) ← [m].i; (i=0~6) ACC.0 ← [m].7 |
| Affected flag(s) | None |
| RLC [m] | Rotate Data Memory left through Carry |
| Description | The contents of the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into bit 0. |
| Operation | [m].(i+1) ← [m].i; (i=0~6) [m].0 ← C C ← [m].7 |
| Affected flag(s) | C |
| RLCA [m] | Rotate Data Memory left through Carry with result in ACC |
| Description | Data in the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into the bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged. |
| Operation | ACC.(i+1) ← [m].i; (i=0~6) ACC.0 ← C C ← [m].7 |
| Affected flag(s) | C |
| RR [m] | Rotate Data Memory right |
| Description | The contents of the specified Data Memory are rotated right by 1 bit with bit 0 rotated into bit 7. |
| Operation | [m].i ← [m].(i+1); (i=0~6) [m].7 ← [m].0 |
| Affected flag(s) | None |
| RRA [m] | Rotate Data Memory right with result in ACC |
| Description | Data in the specified Data Memory is rotated right by 1 bit with bit 0 rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged. |
| Operation | ACC.i ← [m].(i+1); (i=0~6) ACC.7 ← [m].0 |
| Affected flag(s) | None |
| RRC [m] | Rotate Data Memory right through Carry |
| Description | The contents of the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7. |
| Operation | [m].i ← [m].(i+1); (i=0~6) [m].7 ← C C ← [m].0 |
| Affected flag(s) | C |

| | |
|-------------------|---|
| RRCA [m] | Rotate Data Memory right through Carry with result in ACC |
| Description | Data in the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged. |
| Operation | $ACC.i \leftarrow [m].(i+1); (i=0\sim6)$ $ACC.7 \leftarrow C$ $C \leftarrow [m].0$ |
| Affected flag(s) | C |
| | |
| SBC A,[m] | Subtract Data Memory from ACC with Carry |
| Description | The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1. |
| Operation | $ACC \leftarrow ACC - [m] - \bar{C}$ |
| Affected flag(s) | OV, Z, AC, C |
| | |
| SBCM A,[m] | Subtract Data Memory from ACC with Carry and result in Data Memory |
| Description | The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1. |
| Operation | $[m] \leftarrow ACC - [m] - \bar{C}$ |
| Affected flag(s) | OV, Z, AC, C |
| | |
| SDZ [m] | Skip if decrement Data Memory is 0 |
| Description | The contents of the specified Data Memory are first decremented by 1. If the result is 0 the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction. |
| Operation | $[m] \leftarrow [m] - 1$ Skip if $[m]=0$ |
| Affected flag(s) | None |
| | |
| SDZA [m] | Skip if decrement Data Memory is zero with result in ACC |
| Description | The contents of the specified Data Memory are first decremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0, the program proceeds with the following instruction. |
| Operation | $ACC \leftarrow [m] - 1$ Skip if $ACC=0$ |
| Affected flag(s) | None |
| | |
| SET [m] | Set Data Memory |
| Description | Each bit of the specified Data Memory is set to 1. |
| Operation | $[m] \leftarrow FFH$ |
| Affected flag(s) | None |
| | |
| SET [m].i | Set bit of Data Memory |
| Description | Bit i of the specified Data Memory is set to 1. |
| Operation | $[m].i \leftarrow 1$ |
| Affected flag(s) | None |

| | |
|-------------------|--|
| SIZ [m] | Skip if increment Data Memory is 0 |
| Description | The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction. |
| Operation | $[m] \leftarrow [m] + 1$ Skip if $[m]=0$ |
| Affected flag(s) | None |
| SIZA [m] | Skip if increment Data Memory is zero with result in ACC |
| Description | The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction. |
| Operation | $ACC \leftarrow [m] + 1$ Skip if $ACC=0$ |
| Affected flag(s) | None |
| SNZ [m].i | Skip if bit i of Data Memory is not 0 |
| Description | If bit i of the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is 0 the program proceeds with the following instruction. |
| Operation | Skip if $[m].i \neq 0$ |
| Affected flag(s) | None |
| SUB A,[m] | Subtract Data Memory from ACC |
| Description | The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1. |
| Operation | $ACC \leftarrow ACC - [m]$ |
| Affected flag(s) | OV, Z, AC, C |
| SUBM A,[m] | Subtract Data Memory from ACC with result in Data Memory |
| Description | The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1. |
| Operation | $[m] \leftarrow ACC - [m]$ |
| Affected flag(s) | OV, Z, AC, C |
| SUB A,x | Subtract immediate data from ACC |
| Description | The immediate data specified by the code is subtracted from the contents of the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1. |
| Operation | $ACC \leftarrow ACC - x$ |
| Affected flag(s) | OV, Z, AC, C |
| SWAP [m] | Swap nibbles of Data Memory |
| Description | The low-order and high-order nibbles of the specified Data Memory are interchanged. |
| Operation | $[m].3\sim[m].0 \leftrightarrow [m].7\sim[m].4$ |
| Affected flag(s) | None |

| | |
|-------------------|--|
| SWAPA [m] | Swap nibbles of Data Memory with result in ACC |
| Description | The low-order and high-order nibbles of the specified Data Memory are interchanged. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged. |
| Operation | ACC.3~ACC.0 ← [m].7~[m].4 ACC.7~ACC.4 ← [m].3~[m].0 |
| Affected flag(s) | None |
| | |
| SZ [m] | Skip if Data Memory is 0 |
| Description | If the contents of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction. |
| Operation | Skip if [m]=0 |
| Affected flag(s) | None |
| | |
| SZA [m] | Skip if Data Memory is 0 with data movement to ACC |
| Description | The contents of the specified Data Memory are copied to the Accumulator. If the value is zero, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction. |
| Operation | ACC ← [m] Skip if [m]=0 |
| Affected flag(s) | None |
| | |
| SZ [m].i | Skip if bit i of Data Memory is 0 |
| Description | If bit i of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0, the program proceeds with the following instruction. |
| Operation | Skip if [m].i=0 |
| Affected flag(s) | None |
| | |
| TABRD [m] | Read table (specific page or current page) to TBLH and Data Memory |
| Description | The low byte of the program code addressed by the table pointer (TBHP and TBLP or only TBLP if no TBHP) is moved to the specified Data Memory and the high byte moved to TBLH. |
| Operation | [m] ← program code (low byte) TBLH ← program code (high byte) |
| Affected flag(s) | None |
| | |
| TABRDL [m] | Read table (last page) to TBLH and Data Memory |
| Description | The low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH. |
| Operation | [m] ← program code (low byte) TBLH ← program code (high byte) |
| Affected flag(s) | None |
| | |
| XOR A,[m] | Logical XOR Data Memory to ACC |
| Description | Data in the Accumulator and the specified Data Memory perform a bitwise logical XOR operation. The result is stored in the Accumulator. |
| Operation | ACC ← ACC "XOR" [m] |
| Affected flag(s) | Z |

| | |
|-------------------|--|
| XORM A,[m] | Logical XOR ACC to Data Memory |
| Description | Data in the specified Data Memory and the Accumulator perform a bitwise logical XOR operation. The result is stored in the Data Memory. |
| Operation | $[m] \leftarrow \text{ACC} \text{ "XOR" } [m]$ |
| Affected flag(s) | Z |
| | |
| XOR A,x | Logical XOR immediate data to ACC |
| Description | Data in the Accumulator and the specified immediate data perform a bitwise logical XOR operation. The result is stored in the Accumulator. |
| Operation | $\text{ACC} \leftarrow \text{ACC} \text{ "XOR" } x$ |
| Affected flag(s) | Z |

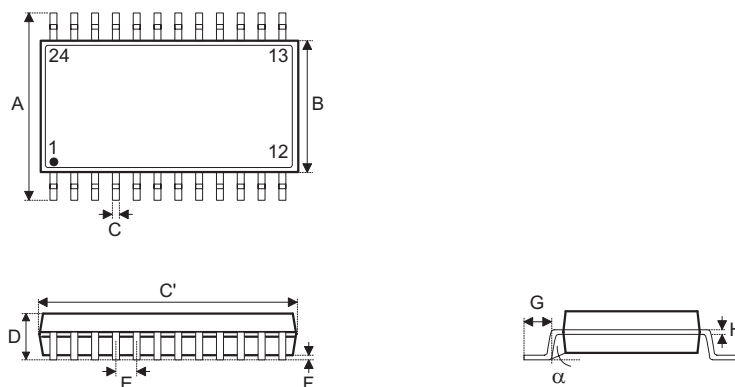
Package Information

Note that the package information provided here is for consultation purposes only. As this information may be updated at regular intervals users are reminded to consult the [Holtek website](#) for the latest version of the [Package/Carton Information](#).

Additional supplementary information with regard to packaging is listed below. Click on the relevant section to be transferred to the relevant website page.

- [Package Information \(include Outline Dimensions, Product Tape and Reel Specifications\)](#)
- [The Operation Instruction of Packing Materials](#)
- [Carton information](#)

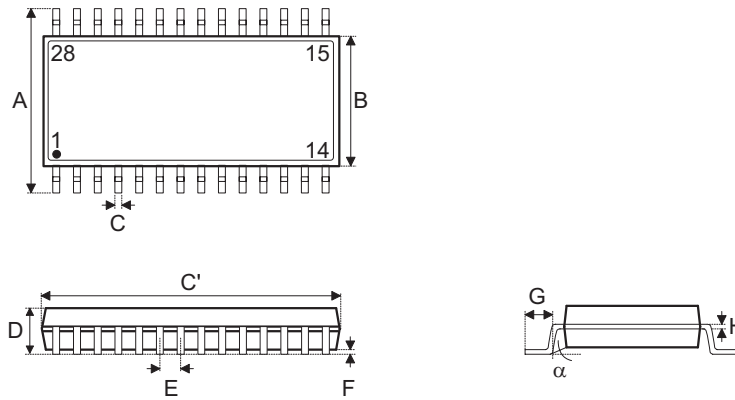
24-pin SSOP (150mil) Outline Dimensions



| Symbol | Dimensions in inch | | |
|----------|--------------------|-----------|-------|
| | Min. | Nom. | Max. |
| A | — | 0.236 BSC | — |
| B | — | 0.154 BSC | — |
| C | 0.008 | — | 0.012 |
| C' | — | 0.341 BSC | — |
| D | — | — | 0.069 |
| E | — | 0.025 BSC | — |
| F | 0.004 | — | 0.010 |
| G | 0.016 | — | 0.050 |
| H | 0.004 | — | 0.010 |
| α | 0° | — | 8° |

| Symbol | Dimensions in mm | | |
|----------|------------------|-----------|------|
| | Min. | Nom. | Max. |
| A | — | 6.000 BSC | — |
| B | — | 3.900 BSC | — |
| C | 0.20 | — | 0.30 |
| C' | — | 8.660 BSC | — |
| D | — | — | 1.75 |
| E | — | 0.635 BSC | — |
| F | 0.10 | — | 0.25 |
| G | 0.41 | — | 1.27 |
| H | 0.10 | — | 0.25 |
| α | 0° | — | 8° |

28-pin SSOP (150mil) Outline Dimensions



| Symbol | Dimensions in inch | | |
|--------|--------------------|-----------|-------|
| | Min. | Nom. | Max. |
| A | — | 0.236 BSC | — |
| B | — | 0.154 BSC | — |
| C | 0.008 | — | 0.012 |
| C' | — | 0.390 BSC | — |
| D | — | — | 0.069 |
| E | — | 0.025 BSC | — |
| F | 0.004 | — | 0.010 |
| G | 0.016 | — | 0.050 |
| H | 0.004 | — | 0.010 |
| α | 0° | — | 8° |

| Symbol | Dimensions in mm | | |
|--------|------------------|-----------|------|
| | Min. | Nom. | Max. |
| A | — | 6.0 BSC | — |
| B | — | 3.9 BSC | — |
| C | 0.20 | — | 0.30 |
| C' | — | 9.9 BSC | — |
| D | — | — | 1.75 |
| E | — | 0.635 BSC | — |
| F | 0.10 | — | 0.25 |
| G | 0.41 | — | 1.27 |
| H | 0.10 | — | 0.25 |
| α | 0° | — | 8° |

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