

Smoke Detector Flash MCU with Calendar

# BA45F5740/BA45F5740-2

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# **Table of Contents**

Features	7
CPU Features	7
Peripheral Features	7
Calendar Features	8
General Description	8
Selection Table	9
Block Diagram	9
Pin Assignment	10
Pin Description	11
Absolute Maximum Ratings	14
D.C. Characteristics	
Operating Voltage Characteristics	
Operating Current Characteristics	
Standby Current Characteristics	15
A.C. Characteristics	
High Speed Internal Oscillator Frequency Accuracy	
Low Speed Internal Oscillator Characteristics – LIRC	16
Operating Frequency Characteristic Curves	17
System Start Up Time Characteristics	17
Input/Output Characteristics	18
Memory Characteristics	19
LVD & LVR Electrical Characteristics	19
A/D Converter Electrical Characteristics	20
Internal Reference Voltage Characteristics	20
Sink Current Generator Electrical Characteristics	
Operational Amplifier Electrical Characteristics	22
For Smoke Detector AFE	
Calendar Electrical Characteristics	24
D.C. Characteristics	24
A.C. Characteristics	25
Power-on Reset Characteristics	25
System Architecture	
Clocking and Pipelining	26
Program Counter	
Stack	
Arithmetic and Logic Unit – ALU	28



Flash Program Memory	
Structure	29
Special Vectors	
Look-up Table	
Table Program Example	
In Circuit Programming – ICP	
On Chip Debug Support – OCDS	31
Data Memory	
Structure	
Data Memory Addressing	
General Purpose Data Memory	
Special Purpose Data Memory	
Special Function Register Description	
Indirect Addressing Registers – IAR0, IAR1, IAR2	35
Memory Pointers – MP0, MP1L, MP1H, MP2L, MP2H	
Accumulator – ACC	
Program Counter Low Register – PCL	
Look-up Table Registers – TBLP, TBHP, TBLH	
Status Register – STATUS	
EEPROM Data Memory	
EEPROM Data Memory Structure	
EEPROM Registers	
Reading Data from the EEPROM	41
Writing Data to the EEPROM	41
Write Protection	41
EEPROM Interrupt	41
Programming Considerations	41
Oscillators	
Oscillator Overview	43
System Clock Configurations	43
Internal High Speed RC Oscillator – HIRC	
Internal 32kHz Oscillator – LIRC	44
Operating Modes and System Clocks	
System Clocks	
System Operation Modes	45
Control Registers	
Operating Mode Switching	
Standby Current Considerations	
Wake-up	
Watchdog Timer	
Watchdog Timer Clock Source	
Watchdog Timer Control Register	
Watchdog Timer Operation	



Reset Functions       .54         Reset Initial Conditions       .56         Input/Output Ports       .59         Pull-high Resistors       .59         Port A Wake-up       .60         I/O Port Control Registers       .60         Source Current Selection       .60         Sink Current Generator       .61         Programming Considerations       .62         I/O Pin Structures       .67         Programming Considerations       .67         Timer Modules – TM       .68         Introduction       .68         TM Operation       .68         TM Operation       .68         TM Clock Source       .68         TM Interrupts       .68         TM External Pins       .69         Programming Considerations       .70         Standard Type TM – STM       .71         Standard Type TM – STM       .71         Standard Type TM Operation       .71         Standard Type TM – PTM       .85         Periodic Type TM Perstion       .86         Periodic Type TM O
Input/Output Ports         59           Pull-high Resistors         59           Port A Wake-up         60           I/O Port Control Registers         60           Source Current Selection         60           Sink Current Generator         61           Pin-shared Functions         62           I/O Pin Structures         67           Programming Considerations         67           Timer Modules – TM         68           Introduction         68           TM Operation         68           TM Clock Source         68           TM Interrupts         68           TM External Pins.         69           Programming Considerations         70           Standard Type TM – STM         71           Standard Type TM – STM         71           Standard Type TM Persition         71           Standard Type TM Persition         71           Standard Type TM Persition         75           Periodic Type TM – PTM         85           Periodic Type TM Persition         85           Periodic TM Operation         85           Periodic TM Operation         85           Periodic Type TM Register Description         86
Pull-high Resistors       59         Port A Wake-up       60         I/O Port Control Registers       60         Source Current Selection       60         Sink Current Generator       61         Pin-shared Functions       62         I/O Pin Structures       67         Programming Considerations       67         Timer Modules – TM       68         Introduction       68         TM Operation       68         TM Operation       68         TM Clock Source       68         TM Interrupts       68         TM External Pins       69         Programming Considerations       70         Standard Type TM – STM       71         Standard Type TM – STM       71         Standard Type TM Peration       75         Periodic Type TM – PTM       85         Periodic TM Operation Modes       75         Periodic TM Operation Modes       75         Periodic TM Operation       86         Periodic Type TM Peration       86         Periodic Type TM Operation Modes       91         Smoke Detector AFE       103         Smoke Detector AFE Registers       103
Port A Wake-up       60         I/O Port Control Registers       60         Source Current Selection       60         Sink Current Generator       61         Pin-shared Functions       62         I/O Pin Structures       67         Programming Considerations       67         Timer Modules – TM       68         Introduction       68         TM Operation       68         TM Clock Source       68         TM Clock Source       68         TM External Pins       69         Programming Considerations       70         Standard Type TM – STM       71         Standard Type TM Segister Description       71         Standard Type TM Operation Modes       75         Periodic Type TM – PTM       85         Periodic Type TM Perstion       85         Periodic Type TM Operation Modes       91         Smoke Detector AFE       103         Smoke Detector AFE       103
I/O Port Control Registers60Source Current Selection60Sink Current Generator61Pin-shared Functions62I/O Pin Structures67Programming Considerations67Timer Modules – TM68Introduction68TM Operation68TM Clock Source68TM Interrupts68TM External Pins69Programming Considerations70Standard Type TM – STM71Standard Type TM Register Description71Standard Type TM Operation75Periodic Type TM – PTM85Periodic Type TM Register Description85Periodic Type TM Register Description85Periodic Type TM Register Description85Periodic Type TM Register Description86Periodic Type TM Register Description85Periodic Type TM Register Description85Periodic Type TM Register Description85Periodic Type TM Register Description86Periodic Type TM Register Description86 <t< td=""></t<>
Source Current Selection60Sink Current Generator61Pin-shared Functions62I/O Pin Structures67Programming Considerations67Timer Modules – TM68Introduction68TM Operation68TM Operation68TM Clock Source68TM Interrupts69Programming Considerations70Standard Type TM – STM71Standard Type TM Register Description71Standard Type TM Operation75Periodic Type TM – PTM85Periodic Type TM Register Description85Periodic Type TM Operation85Periodic Type TM Operation85Periodic Type TM Register Description85Periodic Type TM Register Description86Periodic Type TM Register Description81Smoke Detector AFE103Smoke Detector AFE Registers103
Sink Current Generator       61         Pin-shared Functions       62         I/O Pin Structures       67         Programming Considerations       67         Timer Modules – TM       68         Introduction       68         TM Operation       68         TM Clock Source       68         TM Interrupts       68         TM External Pins       69         Programming Considerations       70         Standard Type TM – STM       71         Standard Type TM Register Description       71         Standard Type TM Operation Modes       75         Periodic Type TM – PTM       85         Periodic Type TM Register Description       85         Periodic Type TM Register Description       85         Periodic Type TM Register Description       85         Periodic Type TM Operation Modes       75         Standard Type TM Operation Modes       75         Periodic Type TM PERION       85         Periodic Type TM Register Description       86         Periodic Type TM Register Descrip
Pin-shared Functions62I/O Pin Structures67Programming Considerations67Timer Modules – TM68Introduction68TM Operation68TM Clock Source68TM Interrupts68TM External Pins69Programming Considerations70Standard Type TM – STM71Standard Type TM Register Description71Standard Type TM Operation Modes75Periodic Type TM – PTM85Periodic Type TM Register Description85Periodic Type TM Register Description86Periodic Type TM Register Description85Periodic Type TM Operation85Periodic Type TM Register Description81Periodic Type TM Register Description81Smoke Detector AFE103Smoke Detector AFE Registers103
I/O Pin Structures.67Programming Considerations67Timer Modules – TM68Introduction68TM Operation68TM Operation68TM Clock Source.68TM Interrupts.68TM External Pins.69Programming Considerations.70Standard Type TM – STM71Standard Type TM – STM71Standard Type TM Periodic Type TM Operation Modes75Periodic Type TM – PTM.85Periodic Type TM Register Description85Periodic Type TM Register Description86Periodic Type TM Operation85Periodic Type TM Register Description86Periodic Type TM Register Description81Periodic Type TM PERION85Periodic Type TM Register Description86Periodic Type TM Register Description<
Programming Considerations67Timer Modules – TM68Introduction68TM Operation68TM Operation68TM Clock Source68TM Interrupts68TM External Pins69Programming Considerations70Standard Type TM – STM71Standard Type TM Partier71Standard Type TM Periodic Type TM Register Description71Standard Type TM Operation Modes75Periodic Type TM – PTM85Periodic Type TM Register Description86Periodic Type TM Register Description85Periodic Type TM Register Description85Periodic Type TM Register Description86Periodic Type TM PERE103Smoke Detector AFE103
Timer Modules – TM       68         Introduction       68         TM Operation       68         TM Operation       68         TM Clock Source       68         TM Interrupts       68         TM External Pins       69         Programming Considerations       70         Standard Type TM – STM       71         Standard Type TM Register Description       71         Standard Type TM Register Description       71         Standard Type TM – PTM       85         Periodic Type TM – PTM       85         Periodic Type TM Register Description       86         Periodic Type TM Register Description       86
Introduction68TM Operation68TM Clock Source68TM Interrupts68TM External Pins69Programming Considerations70Standard Type TM – STM71Standard Type TM Register Description71Standard Type TM Operation Modes75Periodic Type TM – PTM85Periodic Type TM Register Description85Periodic Type TM Register Description85Periodic Type TM Register Description85Periodic Type TM Register Description86Periodic Type TM Register Description81Smoke Detector AFE103Smoke Detector AFE Registers103
TM Operation       68         TM Clock Source       68         TM Interrupts       68         TM External Pins       69         Programming Considerations       70         Standard Type TM – STM       71         Standard Type TM Register Description       71         Standard Type TM Register Description       71         Standard Type TM Operation Modes       75         Periodic Type TM – PTM       85         Periodic TM Operation       85         Periodic TM Operation       85         Periodic Type TM Register Description       86         Periodic Type TM Operation       85         Periodic Type TM Operation       85         Periodic Type TM Register Description       86         Periodic Type TM Register Description       86         Periodic Type TM Register Description       86         Periodic Type TM Operating Modes       91         Smoke Detector AFE       103         Smoke Detector AFE Registers       103
TM Clock Source68TM Interrupts68TM External Pins69Programming Considerations70Standard Type TM – STM71Standard Type TM Register Description71Standard Type TM Register Description71Standard Type TM Operation Modes75Periodic Type TM – PTM85Periodic Type TM Register Description85Periodic Type TM Operation85Periodic Type TM Register Description86Periodic Type TM Register Description86Smoke Detector AFE103Smoke Detector AFE Registers103
TM Interrupts68TM External Pins69Programming Considerations70Standard Type TM – STM71Standard TM Operation71Standard Type TM Register Description71Standard Type TM Operation Modes75Periodic Type TM – PTM85Periodic Type TM Register Description85Periodic Type TM Register Description85Periodic Type TM Register Description85Periodic Type TM Register Description86Periodic Type TM Operating Modes91Smoke Detector AFE103Smoke Detector AFE Registers103
TM External Pins69Programming Considerations70Standard Type TM – STM71Standard TM Operation71Standard Type TM Register Description71Standard Type TM Operation Modes75Periodic Type TM - PTM85Periodic TM Operation85Periodic Type TM Register Description86Periodic Type TM Register Description86Periodic Type TM Operating Modes91Smoke Detector AFE103Smoke Detector AFE Registers103
Programming Considerations70Standard Type TM – STM71Standard TM Operation71Standard Type TM Register Description71Standard Type TM Operation Modes75Periodic Type TM – PTM85Periodic TM Operation85Periodic Type TM Register Description86Periodic Type TM Operating Modes91Smoke Detector AFE103Smoke Detector AFE Registers103
Standard Type TM – STM       71         Standard TM Operation       71         Standard Type TM Register Description       71         Standard Type TM Operation Modes       75         Periodic Type TM – PTM       85         Periodic Type TM Register Description       85         Periodic Type TM Register Description       86         Periodic Type TM Register Description       86         Periodic Type TM Register Description       86         Smoke Detector AFE       103         Smoke Detector AFE Registers       103
Standard TM Operation       71         Standard Type TM Register Description       71         Standard Type TM Operation Modes       75         Periodic Type TM – PTM       85         Periodic TM Operation       85         Periodic Type TM Register Description       86         Periodic Type TM Register Description       86         Periodic Type TM Register Description       86         Periodic Type TM Operating Modes       91         Smoke Detector AFE       103         Smoke Detector AFE Registers       103
Standard Type TM Register Description       71         Standard Type TM Operation Modes       75         Periodic Type TM – PTM       85         Periodic TM Operation       85         Periodic Type TM Register Description       86         Periodic Type TM Register Description       86         Periodic Type TM Operating Modes       91         Smoke Detector AFE       103         Smoke Detector AFE Registers       103
Standard Type TM Operation Modes       75         Periodic Type TM – PTM       85         Periodic TM Operation       85         Periodic Type TM Register Description       86         Periodic Type TM Operating Modes       91         Smoke Detector AFE       103         Smoke Detector AFE Registers       103
Periodic Type TM – PTM.       85         Periodic TM Operation       85         Periodic Type TM Register Description       86         Periodic Type TM Operating Modes       91         Smoke Detector AFE       103         Smoke Detector AFE Registers       103
Periodic TM Operation       .85         Periodic Type TM Register Description       .86         Periodic Type TM Operating Modes       .91         Smoke Detector AFE       .103         Smoke Detector AFE Registers       .103
Periodic TM Operation       .85         Periodic Type TM Register Description       .86         Periodic Type TM Operating Modes       .91         Smoke Detector AFE       .03         Smoke Detector AFE Registers       .103
Periodic Type TM Operating Modes
Smoke Detector AFE       103         Smoke Detector AFE Registers       103
Smoke Detector AFE Registers
Smoke Detector AFE Registers
Analog to Digital Convertor 407
A/D Converter Overview
A/D Converter Register Description
A/D Converter Operation
A/D Converter Reference Voltage 111
A/D Converter Input Signals
Conversion Rate and Timing Diagram
Summary of A/D Conversion Steps
Summary of A/D Conversion Steps
Summary of A/D Conversion Steps
Summary of A/D Conversion Steps113Programming Considerations114A/D Conversion Function114
Summary of A/D Conversion Steps113Programming Considerations114A/D Conversion Function114A/D Conversion Programming Examples115



UART Interface	134
Calendar	
Command Byte	
R/W Signal	150
A0~A2	150
Burst Mode	151
Test Mode	151
Write Protect Register	151
Clock Halt Bit	151
12-hour/24-hour Mode	151
AM-PM Mode	
Reset and Serial Clock Control	
Data Input and Data Out	
Crystal Selection	
Operating Flowchart	
Timing Diagrams	
Low Voltage Detector – LVD	154
LVD Register	
LVD Operation	
Interrupts	
Interrupt Registers	
Interrupt Operation	
External Interrupts	
A/D Converter Interrupt	
Time Base Interrupts	
USIM Interrupt	
EEPROM Interrupt	
TM Interrupts	
Interrupt Wake-up Function	
Programming Considerations	
Application Circuits	
Stand-alone Smoke Detector Applications	164
Instruction Set	
Introduction	165
Instruction Timing	165
Moving and Transferring Data	165
Arithmetic Operations	165
Logical and Rotate Operation	166
Branches and Control Transfer	
Bit Operations	166
Table Read Operations	
Other Operations	166



167
171
187



### Features

### **CPU Features**

- Operating Voltage
  - ♦ f<sub>SYS</sub>=2MHz: 2.2V~5.5V
  - ♦ f<sub>SYS</sub>=4MHz: 2.2V~5.5V
  - ♦ f<sub>SYS</sub>=8MHz: 2.2V~5.5V
- Up to 0.5 $\mu$ s instruction cycle with 8MHz system clock at V<sub>DD</sub>=5V
- · Power down and wake-up functions to reduce power consumption
- Oscillator types
  - ♦ Internal High Speed RC HIRC
  - ♦ Internal 32kHz RC LIRC
- Multi-mode operation: FAST, SLOW, IDLE and SLEEP
- Fully integrated internal oscillators require no external components
- All instructions executed in 1~3 instruction cycles
- Table read instructions
- 115 powerful instructions
- 8-level subroutine nesting
- Bit manipulation instruction

### **Peripheral Features**

- Flash Program Memory: 4K×16
- RAM Data Memory: 256×8
- True EEPROM Memory: 64×8
- Watchdog Timer function
- Up to 13 bidirectional I/O lines
- Up to two external interrupt lines shared with I/O pins (unavailable for BA45F5740-2)
- Programmable I/O port source current for LED applications
- Sink current generator for constant current output
- Smoke Detector AFE including two operational amplifiers
- Multiple Timer Modules for time measure, input capture, compare match output, PWM output or single pulse output function
- Dual Time-Base functions for generation of fixed time interrupt signals
- Universal Serial Interface Module USIM for SPI, I<sup>2</sup>C or UART communication (SPI unavailable for BA45F5740-2)
- Up to 4 external channel 12-bit resolution A/D converter with Internal Reference Voltage VBGREF
- Low voltage reset function
- Low voltage detect function
- Package types: 16-pin NSOP, 20/24-pin SOP



#### **Calendar Features**

- Maximum input serial clock: 1MHz at  $V_{DD}$ =3V, 2MHz at  $V_{DD}$ =5V
- Operating current:
  - ♦ less than 0.7µA at 3V
  - $\blacklozenge$  less than 1.0µA at 5V
- TTL compatible
  - $V_{IH}$ : 2.0V~ $V_{DD}$ +0.3V at  $V_{DD}$ =5V
  - $V_{IL}$ : 0.3V~+0.8V at  $V_{DD}$ =5V
- · Two data transmission modes: single-byte, or burst mode
- Serial I/O transmission
- All registers store BCD format

### **General Description**

The BA45F5740/BA45F5740-2 are Flash Memory A/D type 8-bit high performance RISC architecture microcontrollers, designed especially for smoke detector with Calendar applications.

For memory features, the Flash Memory offers users the convenience of multi-programming features. Other memory includes an area of RAM Data Memory as well as an area of true EEPROM memory for storage of non-volatile data such as serial numbers, calibration data etc.

Analog features include a multi-channel Analog to Digital converter and two operational amplifiers. With regard to internal timers, the devices include multiple and extremely flexible Timer Modules providing functions for timing, pulse generation and PWM output operations. Communication with the outside world is catered for by including fully integrated SPI, I<sup>2</sup>C and UART interface functions, three popular interfaces which provide designers with a means of easy communication with external peripheral hardware. Protective features such as an internal Watchdog Timer, Low Voltage Reset and Low Voltage Detector coupled with excellent noise immunity and ESD protection ensure that reliable operation is maintained in hostile electrical environments.

The devices also include fully integrated high and low speed oscillators which can be flexibly used for different applications. The ability to operate and switch dynamically between a range of operating modes using different clock sources gives users the ability to opetimise microcontroller operation can minimum power consumption. The usual Holtek MCU features such as power down and wake-up functions, oscillator options, programmable frequency divider, etc. combine to ensure user applications require a minimum of external components.

The devices also include a Calendar, which is a serial timekeeper and provides seconds, minutes, hours, day, date, month and year information. The number of days in each month and leap years are automatically adjusted. The Calendar is designed for low power consumption and can operate in two modes: one is the 12-hour mode with an AM/PM indicator, the other is the 24-hour mode.

While the inclusion of flexible I/O programming features, Time-Base functions, Sink Current generator along with many other features ensure that the devices will find excellent use in the smoke detector with Calendar applications.



# **Selection Table**

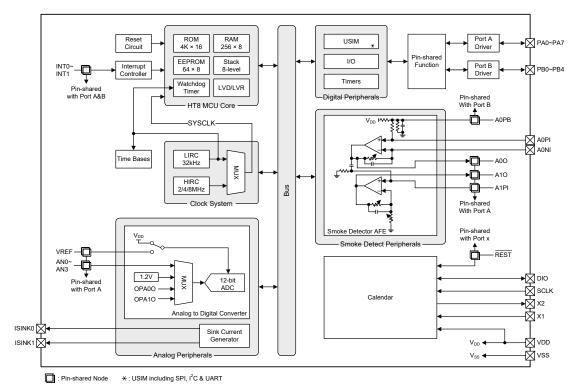
Most features are common to these devices, the main features distinguishing them are I/O count and the package type. The following table summarises the main features of each device.

Part No.	V <sub>DD</sub>	ROM	RAM	RAM True EEPROM		LVD/LVR	External Interrupt	A/D Converter
BA45F5740	2.2V~	4K×16	6 256×8	64×8	13		2	12-bit×4
BA45F5740-2	5.5V	41. 10	200^0	04^0	6	N	_	12-bit×2
Part No.	Time Modu		Smoke Detector	Calendar	USIM	Time Base	Stack	Package
BA45F5740	10-bit S		$\checkmark$		$\checkmark$	2	8	16NSOP 20/24SOP
BA45F5740-2	10-bit P	IVI×1			$\sqrt{*}$			16NSOP

Note: As devices exist in more than one package type, the table shows the situation in the largest package.

"\*": For the BA45F5740-2 device, the SPI mode in the USIM function is unavailable.

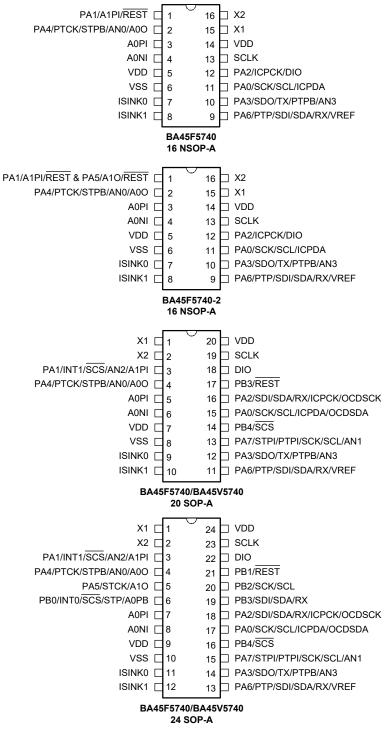
### **Block Diagram**



Note: The figure illustrates the Block Diagram of the device with maximum features, the functional, external pin and pin-shared differences between the devices are provided in the Selection Table and the Pin Assignment.



# **Pin Assignment**



Note: 1. For the BA45F5740-2, on Pin 1 there are two sets of pin functions, the PA1/A1PI/REST and PA5/A10/REST, however these I/O pins cannot be used simultaneously to control the REST pin. If the PA1/A1PI/REST function set is used, the PA5 should be configured as an input with pull-high function disabled. If the PA5/A10/REST function set is used, the PA1 should be configured as an input with pull-high function disabled.



- 2. If the pin-shared pin functions have multiple outputs simultaneously, the desired pin-shared function is determined by the corresponding software control bits.
- 3. For the unbonded lines, PA7, PB0~PB4 in the BA45F5740-2 device, PA5, PA7, PB0~PB4 in the BA45F5740 16-pin NSOP package and PA5, PB0~PB2 in the BA45F5740 20-pin SOP package, the pin status should be properly configured to avoid unwanted current consumption resulting from floating input conditions. Refer to the "Standby Current Considerations" and "Input/Output Ports" sections.
- 4. The OCDSDA and OCDSCK pins are supplied as OCDS dedicated pins and as such only available for the BA45V5740 device which is the OCDS EV chip for the BA45F5740 device.
- 5. In general calendar applications, the REST, DIO and SCLK pins should be externally connected to I/O pins respectively for normal operation. For part of package types the REST and DIO pins are pin-shared with I/O pins respectively which should be properly configured to implement calendar control, refer to the "Input/Output Ports" and "Calendar" sections.

### **Pin Description**

The function of each pin is listed in the following table, however the details behind how each pin is configured is contained in other sections of the datasheet. As the Pin Description table shows the situation for the package with the most pins, not all pins in the table will be available on smaller package sizes.

Note that for the BA45F5740-2 device, the PA7 and PB0~PB4 are not available. In addition, the PA1/A1PI, PA5/A1O and  $\overline{\text{REST}}$  share the same pin location, refer to the Pin Assignment.

Pin Name	Function	OPT	I/T	O/T	Description
	PA0	PAPU PAWU PAS0	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up.
PA0/SCK/SCL/	SCK	IFS0 PAS0 SIMC0	ST	CMOS	SPI serial clock
ICPDA/OCDSDA	SCL	IFS0 PAS0 SIMC0	ST	NMOS	I²C clock line
	ICPDA		ST	CMOS	ICP data/address
	OCDSDA	_	ST	CMOS	OCDS data/address, for EV chip only
	PA1	PAPU PAWU PAS0	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up.
	INT1	PAS0 INTEG	ST	_	External Interrupt 1
PA1/INT1/SCS/AN2/A1PI	SCS	IFS0 PAS0 SIMC0 SIMC2	ST	CMOS	SPI slave device select
	AN2	PAS0	AN		A/D Converter input channel 2
	A1PI	PAS0	AN	_	SD Operational Amplifier 1 positive input



Pin Name	Function	ОРТ	I/T	O/T	Description
	PA2	PAPU PAWU PAS0	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up.
	SDI	IFS0 PAS0 SIMC0	ST	_	SPI serial data input
PA2/SDI/SDA/RX/ ICPCK/OCDSCK	SDA	IFS0 PAS0 SIMC0	ST	NMOS	I²C data line
	RX	IFS0 PAS0 SIMC0	ST	_	UART receiver pin
	ICPCK		ST		ICP clock pin
	OCDSCK		ST		OCDS clock pin, for EV chip only
	PA3	PAPU PAWU PAS0	ST	смоѕ	General purpose I/O. Register enabled pull-high and wake-up.
PA3/SDO/TX/PTPB/AN3	SDO	PAS0 SIMC0	_	CMOS	SPI serial data output
	тх	PAS0 SIMC0	_	CMOS	UART transmitter pin
	PTPB	PAS0		CMOS	PTM inverted output
	AN3	PAS0	AN		A/D Converter input channel 3
	PA4	PAPU PAWU PAS1	ST	смоѕ	General purpose I/O. Register enabled pull-high and wake-up.
PA4/PTCK/STPB/AN0/A0O	PTCK	PAS1 PTMC0 PTMC1	ST	_	PTM clock input or capture input
	STPB	PAS1	_	CMOS	STM inverted output
	AN0	PAS1	AN	—	A/D Converter input channel 0
	A0O	PAS1	_	AN	SD Operational Amplifier 0 output
	PA5	PAPU PAWU PAS1	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up.
PA5/STCK/A1O	STCK	IFS1 PAS1 STMC0	ST	_	STM clock input
	A10	PAS1	_	AN	SD Operational Amplifier 1 output
	PA6	PAPU PAWU PAS1	ST	CMOS	General purpose I/O. Register enabled pull-high and wake-up.
	PTP	PAS1	_	CMOS	PTM output
PA6/PTP/SDI/SDA/RX/	SDI	IFS0 PAS1 SIMC0	ST	_	SPI serial data input
VREF	SDA	IFS0 PAS1 SIMC0	ST	NMOS	I <sup>2</sup> C data line
	RX	IFS0 PAS1 SIMC0	ST	_	UART receiver pin
	VREF	PAS1	AN		A/D Converter external reference voltage



Pin Name	Function	OPT	I/T	O/T	Description
	PA7	PAPU PAWU PAS1	ST	смоѕ	General purpose I/O. Register enabled pull-high and wake-up.
	STPI	PAS1 STMC1	ST	_	STM capture Input
PA7/STPI/PTPI/SCK/SCL/ AN1	PTPI	IFS0 PAS1 PTMC1	ST	_	PTM capture input
ANT	SCK	IFS0 PAS1 SIMC0	ST	CMOS	SPI serial clock
	SCL	IFS0 PAS1 SIMC0	ST	NMOS	I <sup>2</sup> C clock line
	AN1	PAS1	AN		A/D Converter input channel 1
	PB0	PBPU PBS0	ST	CMOS	General purpose I/O. Register enabled pull-high.
	INT0	PBS0 INTEG	ST	_	External Interrupt 0
PB0/INT0/SCS/STP/A0PB	SCS	IFS0 PBS0 SIMC0 SIMC2	ST	смоз	SPI slave device select
	STP	PBS0		CMOS	STM output
	A0PB	PBS0	AN		SD Operational Amplifier 0 bias input
	PB1	PBPU PBS0	ST	CMOS	General purpose I/O. Register enabled pull-high.
PB1/REST*	REST	_	ST	_	Calendar Reset pin It should be controlled by the I/O pin for calendar normal operation.
	PB2	PBPU PBS0	ST	CMOS	General purpose I/O. Register enabled pull-high.
PB2/SCK/SCL	SCK	IFS0 PBS0 SIMC0	ST	CMOS	SPI serial clock
	SCL	IFS0 PBS0 SIMC0	ST	NMOS	I <sup>2</sup> C clock line
	PB3	PBPU PBS0	ST	CMOS	General purpose I/O. Register enabled pull-high.
	SDI	IFS0 PBS0 SIMC0	ST	_	SPI serial data input
PB3/SDI/SDA/RX	SDA	IFS0 PBS0 SIMC0	ST	NMOS	I <sup>2</sup> C data line
	RX	IFS0 PBS0 SIMC0	ST	_	UART receiver pin
	PB4	PBPU PBS1	ST	CMOS	General purpose I/O. Register enabled pull-high.
PB4/SCS	SCS	IFS0 PBS1 SIMC0 SIMC2	ST	смоѕ	SPI slave device select
ISINK0	ISINK0			AN	Sink current 0 source



Pin Name	Function	OPT	I/T	O/T	Description	
ISINK1	ISINK1	—	—	AN	Sink current 1 source	
A0NI	A0NI	_	AN	_	SD Operational Amplifier 0 negative input	
A0PI	A0PI	_	AN	—	SD Operational Amplifier 0 positive input	
X1	X1	_	ST	_	32768Hz crystal input pin	
X2	X2	_	—	CMOS	32768Hz crystal output pin	
DIO*	DIO	—	ST	CMOS	Calendar Data Input/Output pin It should be externally connected to an I/O pin for calendar normal operation.	
SCLK	SCLK	_	ST	_	Calendar Serial Clock pulse pin It should be externally connected to an I/O pin for calendar normal operation.	
VDD	VDD	_	PWR	—	Positive power supply	
VSS	VSS		PWR	_	Negative power supply, ground	

Legend: I/T: Input type;

O/T: Output type;

OPT: Optional by register option;

PWR: Power;

ST: Schmitt Trigger input; CMOS: CMOS output; AN: Analog signal; NMOS: NMOS output.

\*: For different package types, the REST and DIO pins maybe pin-shared with different I/O pins respectively, refer to the Pin Assignment, in which case they are directly controlled by the corresponding I/O pins for calendar normal operation.

# **Absolute Maximum Ratings**

Supply Voltage	$V_{SS}$ =0.3V to $V_{SS}$ +6.0V
Input Voltage	$V_{\text{SS}}0.3V$ to $V_{\text{DD}}\text{+-}0.3V$
Storage Temperature	50°C to 125°C
Operating Temperature	40°C to 85°C
I <sub>OL</sub> Total	
Ioн Total	80mA
Total Power Dissipation	

Note: These are stress ratings only. Stresses exceeding the range specified under "Absolute Maximum Ratings" may cause substantial damage to the devices. Functional operation of the devices at other conditions beyond those listed in the specification is not implied and prolonged exposure to extreme conditions may affect devices reliability.

### **D.C. Characteristics**

For data in the following tables, note that factors such as oscillator type, operating voltage, operating frequency, pin load conditions, temperature and program instruction type, etc., can all exert an influence on the measured values.

### **Operating Voltage Characteristics**

•	0 0				Ta= -40	)°C~85°C
Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
		f <sub>SYS</sub> =f <sub>HIRC</sub> =2MHz	2.2	—	5.5	
	f <sub>SYS</sub> =f <sub>HIRC</sub> =4MHz	2.2	—	5.5	V	
	f <sub>sys</sub> =f <sub>HIRC</sub> =8MHz	2.2	_	5.5		
	Operating Voltage (LIRC)	f <sub>sys</sub> =f <sub>LIRC</sub> =32kHz	2.2	—	5.5	V

Ta=25°C

### **Operating Current Characteristics**

Symbol	Normal Operation		Test Conditions	Min.	Turn	Max.	Unit
Symbol		VDD	Conditions	IVIII.	Тур.	IVIAX.	Unit
		2.2V		—	8	16	μA
	SLOW Mode (LIRC)	3V	f <sub>sys</sub> =32kHz	—	10	20	
		5V		—	30	50	
		2.2V		_	0.15	0.20	
		3V	f <sub>sys</sub> =2MHz	—	0.20	0.30	mA
		5V		—	0.40	0.60	
I <sub>DD</sub>		2.2V			0.3	0.5	
	FAST Mode (HIRC)	3V	f <sub>sys</sub> =4MHz	—	0.4	0.6	mA
		5V		—	0.8	1.2	
		2.7V		—	0.6	1.0	
		3V	f <sub>sys</sub> =8MHz		0.8	1.2	mA
		5V			1.6	2.4	

Note: When using the characteristic table data, the following notes should be taken into consideration:

1. Any digital inputs are setup in a non-floating condition.

2. All measurements are taken under conditions of no load and with all peripherals in an off state.

3. There are no DC current paths.

4. All Operating Current values are measured using a continuous NOP instruction program loop.

### Standby Current Characteristics

Ta=25°C, unless otherwise specify

Querrah al	Cton dhu Mada		Test Conditions	Min	True	Max	Max.	11
Symbol	Standby Mode	VDD	Conditions	Min.	Тур.	Max.	85°C	Unit
		2.2V		—	1.2	2.4	2.9	
	SLEEP Mode	3V	WDT on	—	1.5	3	3.6	μA
		5V		_	3	5	6	
		2.2V		_	2.4	4.0	4.8	
	IDLE0 Mode ( LIRC)	3V	f <sub>SUB</sub> on	_	3	5	6	μA
		5V		_	5	10	12	
		2.2V	f <sub>SUB</sub> on, f <sub>SYS</sub> =2MHz	_	60	120	140	
ISTB		3V		_	70	140	160	μA
		5V		_	130	260	280	
		2.2V		_	90	200	220	
	IDLE1 Mode (HIRC)	3V	f <sub>SUB</sub> on, f <sub>SYS</sub> =4MHz	_	110	220	240	μA
	2	5V		_	210	420	460	
		2.7V		_	150	300	340	
		3V	f <sub>SUB</sub> on, f <sub>SYS</sub> =8MHz	_	180	360	400	μA
		5V	]		370	740	800	

Note: When using the characteristic table data, the following notes should be taken into consideration:

1. Any digital inputs are setup in a non-floating condition.

2. All measurements are taken under conditions of no load and with all peripherals in an off state.

3. There are no DC current paths.

4. All Standby Current values are taken after a HALT instruction execution thus stopping all instruction execution.



# A.C. Characteristics

For data in the following tables, note that factors such as oscillator type, operating voltage, operating frequency and temperature etc., can all exert an influence on the measured values.

### High Speed Internal Oscillator Frequency Accuracy

During the program writing operation the writer will trim the HIRC oscillator at a user selected HIRC frequency and user selected voltage of either 3V or 5V.

Sympol	Parameter	Test C	Conditions	Min.	Turn	Max.	Unit
Symbol	Parameter	V <sub>DD</sub>	Temp.	win.	Тур.	wax.	Unit
			25°C	-1%	2	+1%	
		3V/5V	-20°C~60°C	-2%	2	+2%	
	2MHz Writer Trimmed HIRC Frequency		-40°C~85°C	-3%	2	+3%	MHz
		2.2V~5.5V	25°C	-9%	2	+9%	
		2.20~5.50	-40°C~85°C	-10%	2	+10%	
		3V/5V	25°C	-1%	4	+1%	
f <sub>HIRC</sub>		30/30	-40°C~85°C	-2%	4	+2%	
HIRC	4MHz Writer Trimmed HIRC Frequency	2.2V~5.5V	25°C	-2.5%	4	+2.5%	MHz
		2.20~5.50	-40°C~85°C	-3%	4	+3%	
		2.4V~3.3V	-20°C~50°C	-2%	4	+2%	
		3V/5V	25°C	-1%	8	+1%	
		30/30	-40°C~85°C	-10%	8	+2%	
	8MHz Writer Trimmed HIRC Frequency	2.2V~5.5V	25°C	-10%	8	+3%	MHz
		2.20~0.00	-40°C~85°C	-15%	8	+5%	

Note: 1. The 3V/5V values for V<sub>DD</sub> are provided as these are the two selectable fixed voltages at which the HIRC frequency is trimmed by the writer.

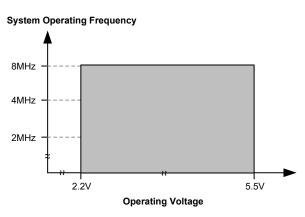
- 2. The row below the 3V/5V trim voltage row is provided to show the values for the full  $V_{DD}$  range operating voltage. It is recommended that the trim voltage is fixed at 3V for application voltage ranges from 2.2V to 3.6V and fixed at 5V for application voltage ranges from 3.3V to 5.5V.
- 3. The minimum and maximum tolerance values provided in the table are only for the frequency at which the writer trims the HIRC oscillator. After trimming at this chosen specific frequency any change in HIRC oscillator frequency using the oscillator register control bits by the application program will give a frequency tolerance to within  $\pm 20\%$ .

### Low Speed Internal Oscillator Characteristics – LIRC

Sumbol	Deremeter	Test	Min.	Turn	Max.	Unit	
Symbol	Parameter	V <sub>DD</sub>	Temp.	with.	Тур.	wax.	Unit
f		2.2V~5.5V	25°C	-5%	32	+5%	kHz
† <sub>LIRC</sub>	LIRC Frequency	2.20~5.50	-40°C ~ 85°C	-10%	32	+10%	KIIZ
<b>t</b> START	LIRC Start-up Time	—	—		_	100	μs



### **Operating Frequency Characteristic Curves**



### System Start Up Time Characteristics

Ta=-40°C~85°C

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
	System Start-up Time	f <sub>SYS</sub> =f <sub>H</sub> ~ f <sub>H</sub> /64, f <sub>H</sub> =f <sub>HIRC</sub>	_	16	_	t <sub>HIRC</sub>
	Wake-up from condition where $f_{\mbox{\scriptsize SYS}}$ is off	f <sub>SYS</sub> =f <sub>SUB</sub> =f <sub>LIRC</sub>	_	2	—	t <sub>LIRC</sub>
	System Start-up Time	$f_{SYS}=f_H \sim f_H/64, f_H=f_{HIRC}$	_	2	—	tн
tsst	Wake-up from condition where $f_{\text{SYS}}$ is on	f <sub>SYS</sub> =f <sub>SUB</sub> =f <sub>LIRC</sub>	_	2	_	t <sub>suв</sub>
System Speed Switch Tim FAST to SLOW Mode or SLOW to FAST Mode		$f_{\text{HIRC}}  \text{switches from off} \to \text{on}$	_	16	_	t <sub>HIRC</sub>
	System Reset Delay Time Reset source from Power-on reset or LVR hardware reset	RR <sub>POR</sub> =5V/ms	42 48		54	
t <sub>RSTD</sub>	System Reset Delay Time WDTC register software reset	—				ms
	System Reset Delay Time WDT overflow reset	_	14	16	18	
t <sub>SRESET</sub>	Minimum Software Reset Width to Reset	_	45	90	120	μs

Note: For the System Start-up time values, whether  $f_{SYS}$  is on or off depends upon the mode type and the chosen  $f_{SYS}$  system oscillator. Details are provided in the System Operating Modes section.

- 1. The time units, shown by the symbols  $t_{HIRC}$  etc. are the inverse of the corresponding frequency values as provided in the frequency tables. For example  $t_{HIRC}=1/f_{HIRC}$ ,  $t_{SYS}=1/f_{SYS}$  etc.
- 2. If the LIRC is used as the system clock and if it is off when in the SLEEP Mode, then an additional LIRC start up time, t<sub>START</sub>, as provided in the LIRC frequency table, must be added to the t<sub>SST</sub> time in the table above.
- 3. The System Speed Switch Time is effectively the time taken for the newly activated oscillator to start up.



# **Input/Output Characteristics**

Symbol	Parameter		Test Conditions	Min.	Turn	Max.	Unit
Symbol	Parameter	VDD	Conditions		Тур.	wax.	Unit
Vil	Input Low Voltage for I/O	5V		0	_	1.5	v
VIL	Ports	_		0	_	0.2V <sub>DD</sub>	
VIH	Input High Voltage for I/O	5V		3.5	—	5.0	v
VIH	Ports	—		$0.8V_{DD}$	_	V <sub>DD</sub>	v
IOL	Sink Current for I/O Ports	3V	Voi=0.1VDD	16	32		mA
IOL		5V		32	65		
		3V	V <sub>OH</sub> =0.9V <sub>DD</sub> , SLEDC[m+1, m]=00B	-0.7	-1.5		
		5V	(m=0, 2, 4, 6)	-1.5	-2.9		
		3V	V <sub>OH</sub> =0.9V <sub>DD</sub> , SLEDC[m+1, m]=01B	-1.3	-2.5		
Юн	Source Current for I/O Ports	5V	(m=0, 2, 4, 6)	-2.5	-5.1		mA
1011		3V	V <sub>OH</sub> =0.9V <sub>DD</sub> , SLEDC[m+1, m]=10B	-1.8	-3.6		
		5V         (m=0, 2, 4, 6)         -3.6         -7.3            3V         V <sub>OH</sub> =0.9V <sub>DD</sub> , SLEDC[m+1, m]=11B         -4         -8            5V         (m=0, 2, 4, 6)         -8         -16	(m=0, 2, 4, 6)	-3.6	-7.3		
				-			
		5V	(m=0, 2, 4, 6)	-8	-16		
Rph	Pull-High Resistance for I/O	3V		20	60	100	kΩ
	Ports (Note)	5V		10	30	50	
ILEAK	Input Leakage Current	5V	V <sub>IN</sub> =V <sub>DD</sub> or V <sub>IN</sub> =V <sub>SS</sub>	—	_	±1	μA
I <sub>TOL</sub>	Total Sink Current for all I/O Ports	5V	_	80	—	_	mA
I <sub>тон</sub>	Total Source Current for all I/O Ports	5V	_	-80	—	-	mA
	STM STPI Input Pin Minimum Pulse Width			0.3	_	_	
t <sub>TPI</sub>	PTM PTPI Input Pin Minimum Pulse Width	_	_	0.1		_	μs
t <sub>тск</sub>	STM/PTM TCK Input Pin Minimum Pulse Width	_	_	0.3		_	μs
t <sub>INT</sub>	Interrupt Pin Minimum Pulse Width	_	_	10	_	_	μs
f <sub>TMCLK</sub>	PTM Maximum Timer Clock Source Frequency	5V	_	_	_	1	f <sub>sys</sub>
t <sub>CPW</sub>	PTM Minimum Capture Pulse Width	_	_	2	_	_	t <sub>TMCL</sub>

Note: 1. The R<sub>PH</sub> internal pull high resistance value is calculated by connecting to ground and enabling input pin with pull-high resistor and then measuring the pin current at the specified supply voltage level. Dividing the voltage by this measured current provides the R<sub>PH</sub> value.

 $2. \ t_{\text{TMCLK}}{=}1/f_{\text{TMCLK}}$ 



# **Memory Characteristics**

	Ta=-40°C~85°C, unless otherwise specified									
Symbol	Parameter		Test Conditions	Min.	Тур.	Max.	Unit			
Symbol	Falailletei	$V_{\text{DD}}$	Conditions	IVIIII.	тур.	Wax.	Unit			
V <sub>RW</sub>	V <sub>DD</sub> for Read/Write	—	—	$V_{\text{DDmin}}$	_	V <sub>DDmax</sub>	V			
Flash Pr	ogram / Data EEPROM Memory									
t <sub>DEW</sub>	Erase/Write Cycle Time – Flash Program Memory	_	_	_	2	3	ms			
	Write Cycle Time – Data EEPROM Memory	—	—	—	4	6	ms			
IDDPGM	Programming/Erase Current on VDD	—	—	—	—	5.0	mA			
-	Cell Endurance – Flash Program Memory	—	_	10K	_	—	E/W			
Ep	Cell Endurance – Data EEPROM Memory	—	_	100K	_	_	E/W			
t <sub>RETD</sub>	ROM Data Retention Time		Ta=25°C	_	40	—	Year			
RAM Da	ta Memory									
V <sub>DR</sub>	RAM Data Retention Voltage	_	Device in SLEEP Mode	1.0	_	_	V			

# LVD & LVR Electrical Characteristics

Symbol	Parameter		Test Conditions	Min.	Turn	Max.	Unit
Symbol	Parameter	VDD	Conditions		Тур.	wax.	Unit
V <sub>LVR</sub>	Low Voltage Reset Voltage	_	LVR enable	-5%	2.1	+5%	V
			LVD enable, voltage select 2.2V		2.2		
			LVD enable, voltage select 2.4V	]	2.4	]	
			LVD enable, voltage select 2.7V	1	2.7	1	
VLVD	Low Voltage Detection Voltage	_	LVD enable, voltage select 3.0V	-5%	3.0	+5%	V
			LVD enable, voltage select 3.3V		3.3	1	
			LVD enable, voltage select 3.6V		3.6		
			LVD enable, voltage select 4.0V		4.0		
		3V	LVD enable, LVR enable,	_	15 20		
	Operating Current	5V	VBGEN=0	_	20	25	μA
LVRLVDBG	Operating Current	3V	LVD enable, LVR enable,	_	20	25	
		5V	VBGEN=1	_	25	30	μA
	LVDO Stable Time		For LVR enable, VBGEN=0, LVD off $\rightarrow$ on		_	15	
t <sub>LVDS</sub>		_	For LVR disable, VBGEN=0, LVD off $\rightarrow$ on		_	150	μs
t <sub>LVR</sub>	Minimum Low Voltage Width to Reset			120	240	480	μs
t <sub>LVD</sub>	Minimum Low Voltage Width to Interrupt			60	120	240	μs
I <sub>LVR</sub>	Additional Current for LVR Enable	_	LVD disable, VBGEN=0	_	20	25	μA

Ta=25°C



# A/D Converter Electrical Characteristics

						Т	ā=25°C
Symbol	Parameter		Test Conditions	Min.	Turn	Max.	Unit
Symbol	Falameter	VDD	Conditions	IVIIII.	Тур.	Wax.	Unit
V <sub>DD</sub>	Operating Voltage	—	_	2.2	—	5.5	V
Vadi	Input Voltage		_	0	—	VREF	V
V <sub>REF</sub>	Reference Voltage	_		2	_	V <sub>DD</sub>	V
		3V					
<b>D</b>		5V	Vref=Vdd, tadck=0.5µs				
DNL	Differential Non-linearity	3V			_	±3	LSB
		5V	V <sub>REF</sub> =V <sub>DD</sub> , t <sub>ADCK</sub> =8µs				
		3V					
		5V	VREF=VDD, tadck=0.5µs				
INL	Integral Non-linearity	3V		-	_	±4	LSB
		5V	VREF=VDD, tadck=8µs				
	Additional Current for A/D	3V	N 1 1 1 0 5	_	1	2	
ADC	Converter Enable	5V	No load, t <sub>ΑDCK</sub> =0.5μs	_	1.5	3	mA
t <sub>ADCK</sub>	Clock Period	_	_	0.5	_	10	μs
t <sub>ON2ST</sub>	A/D Converter On-to-Start Time	_	_	4	_	_	μs
t <sub>ADS</sub>	Sampling time	_	_	_	4	_	<b>t</b> ADCK
t <sub>ADC</sub>	Conversion Time (Include A/D Sample and Hold Time)	_	_		16	_	t <sub>adck</sub>

# Internal Reference Voltage Characteristics

Ta=25°C, unless otherwise specify

Symbol	Parameter	Te	est Conditions	Min.	Tun	Max.	Unit
Symbol	Falameter	V <sub>DD</sub>	Conditions		Тур.	Wax.	Unit
Vdd	Operating Voltage	_	—	2.2	_	5.5	V
V <sub>BGREF</sub>	Bandgap Reference Voltage	—	Ta=-40°C~85°C	-1.0%	1.2	+1.0%	V
IBGREF	Operating Current	5.5V	Ta=-40°C~85°C	—	25	40	μA
PSRR	Power Supply Rejection Ratio	_	V <sub>RIPPLE</sub> =1V <sub>P-P</sub> , f <sub>RIPPLE</sub> =100Hz	75		_	dB
En	Output Noise	_	No load current, f=0.1Hz~10Hz	_	300	_	μV <sub>RMS</sub>
I <sub>DRV</sub>	Buffer Driving Capability	_	$\Delta V_{BGREF}$ =-1%	1	_	_	mA
Isd	Shutdown Current	_	VBGREN=0	_	_	0.1	μA
<b>t</b> start	Startup Time	2.2V~5.5V	_	_	_	400	μs

Note: 1. All the above parameters are measured under conditions of no load condition unless otherwise described.

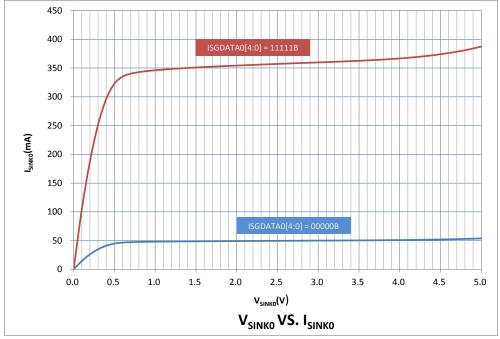
2. A 0.1µF ceramic capacitor should be connected between VDD and GND.

3. The  $V_{\text{BGREF}}$  voltage can be used as the A/D converter input.



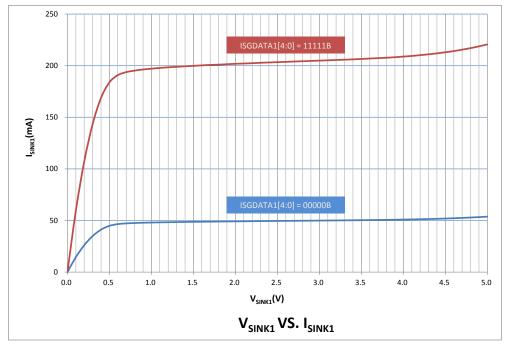
Sympol	Parameter		Test Conditions	Min.	Turn	Max.	Unit
Symbol	Parameter	VDD	Conditions	wiin.	Тур.	wax.	Unit
		5V	Ta=25°C, V <sub>ISINK0</sub> =3.0V, ISGDATA0[4:0]=00000B	47.5	50.0	52.5	
		—	Ta=-40°C~85°C, V <sub>ISINK0</sub> =1.0V~4.5V, ISGDATA0[4:0]=00000B	41	50	59	
1	Sink Current for ISINK0 Pin	_	Ta=-40°C~85°C, V <sub>ISINK0</sub> =0.7V~1.0V, ISGDATA0[4:0]=00000B	37.5	50.0	50.0	mA
Isinko		5V	Ta=25°C, V <sub>ISINK0</sub> =3.0V, ISGDATA0[4:0]=11111B	330	360	390	mA
		_	Ta=-40°C~85°C, V <sub>ISINK0</sub> =1.0V~4.5V, ISGDATA0[4:0]=11111B	295	360	425	
		—	Ta=-40°C~85°C, V <sub>ISINK0</sub> =0.7V~1.0V, ISGDATA0[4:0]=11111B	270	360	360	
		5V	Ta=25°C, V <sub>ISINK1</sub> =3.0V, ISGDATA1[4:0]=00000B	45	50	55	
		_	Ta=-40°C~85°C, V <sub>ISINK1</sub> =1.0V~4.5V, ISGDATA1[4:0]=00000B	41	50	59	
I <sub>SINK1</sub>	Sink Current for ISINK1 Pin	_	Ta=-40°C~85°C, V <sub>ISINK1</sub> =0.7V~1.0V, ISGDATA1[4:0]=00000B	37.5	50.0	50.0	mA
ISINK1		5V	Ta=25°C, V <sub>ISINK1</sub> =3.0V, ISGDATA1[4:0]=11111B	178	205	231	IIIA
		_	Ta=-40°C~85°C, V <sub>ISINK1</sub> =1.0V~4.5V, ISGDATA1[4:0]=11111B	168	205	242	
		_	Ta=-40°C~85°C, V <sub>ISINK1</sub> =0.7V~1.0V, ISGDATA1[4:0]=11111B	154	205	205	

# **Sink Current Generator Electrical Characteristics**



ISINKO Characteristic Curve





ISINK1 Characteristic Curve

# **Operational Amplifier Electrical Characteristics**

### For Smoke Detector AFE

VDD-3V, 1a40 C~03 C, typical 1a- 23 C, unless otherwise spec										
Symbol	Parameter		Test Conditions	Min.	Тур.	Max.	Unit			
Symbol	raidilietei	VDD	VDD Conditions		тур.	IVIAA.	Unit			
V <sub>DD</sub>	Operating Voltage	_		2.2	—	5.5	V			
			SDAmBW[1:0]=00B (m=0,1), no load	_	3.0	5.0				
	Operating Current	5V	SDAmBW[1:0]=01B (m=0,1), no load	_	10	16				
IOPA	Operating Current	50	SDAmBW[1:0]=10B (m=0,1), no load	_	80	128	μA			
			SDAmBW[1:0]=11B (m=0,1), no load	_	200	320				
Vos	Input Offset Voltage	5V	Without calibration (SDAmOF[5:0]=100000B (m=0,1))	-15	_	15	mV			
			With calibration	-2	—	+2				
los	Input Offset Current	5V	V <sub>IN</sub> =1/2V <sub>CM</sub>	_	1	10	nA			
V <sub>CM</sub>	Common Mode Voltage Range	5V	SDAmBW[1:0]=00,01,10,11 (m=0,1)	Vss	_	V <sub>DD</sub> -1.4	V			
PSRR	Power Supply Rejection Ratio	5V	SDAmBW[1:0]=00,01,10,11 (m=0,1)	50	70		dB			
CMRR	Common Mode Rejection Ratio	5V	SDAmBW[1:0]=00,01,10,11 (m=0,1)	50	80	_	dB			
Aol	Open Loop Gain	5V	SDAmBW[1:0]=00,01,10,11 (m=0,1)	60	80		dB			

V<sub>DD</sub>=5V, Ta=-40°C~85°C, typical Ta= 25°C, unless otherwise specify



Cumb al	Deveneeter	Test Conditions			-		11	
Symbol	Parameter	VDD	Conditions	Min.	Тур.	Max.	Unit	
SR Slew Rate			R <sub>LOAD</sub> =1MΩ, C <sub>LOAD</sub> =60pF, SDAmBW[1:0]=00 (m=0,1)	0.5	1.5	_		
	Slew Rate	5V	R <sub>LOAD</sub> =1MΩ, C <sub>LOAD</sub> =60pF, SDAmBW[1:0]=01 (m=0,1)	5	15			
SK	Siew Rale	57	R <sub>LOAD</sub> =1MΩ, C <sub>LOAD</sub> =60pF, SDAmBW[1:0]=10 (m=0,1)	180	500		v/1115	
		R <sub>LOAD</sub> =1MΩ, C <sub>LOAD</sub> =60pF, SDAmBW[1:0]=11 (m=0,1)	600	1800				
		5V	R <sub>LOAD</sub> =1MΩ, C <sub>LOAD</sub> =60pF, SDAmBW[1:0]=00 (m=0,1)	2.5	5.0			
GBW	Gain Bandwidth		R <sub>LOAD</sub> =1MΩ, C <sub>LOAD</sub> =60pF, SDAmBW[1:0]=01 (m=0,1)	20	40		kHz	
GBW			R <sub>LOAD</sub> =1MΩ, C <sub>LOAD</sub> =60pF, SDAmBW[1:0]=10 (m=0,1)	400	600		KIIZ	
			R <sub>LOAD</sub> =1MΩ, C <sub>LOAD</sub> =60pF, SDAmBW[1:0]=11 (m=0,1)	1300	2000			
Vor	Maximum Qutput Valtage Pange		SDAmBW[1:0]=00,01 (m=0,1) R <sub>LOAD</sub> =5k $\Omega$ to V <sub>DD</sub> /2	V <sub>ss</sub> +140	—	V <sub>DD</sub> -160	mV	
VOR	Vor Maximum Output Voltage Range	5V	SDAmBW[1:0]=10,11 (m=0,1) R <sub>LOAD</sub> =5kΩ to V <sub>DD</sub> /2	V <sub>SS</sub> +120	_	V <sub>DD</sub> -140	IIIV	
lsc	Output Short Circuit Current	5V	R <sub>LOAD</sub> =5.1Ω, SDAmBW[1:0]=00,01 (m=0,1)	±6	±12		mA	
			R <sub>LOAD</sub> =5.1Ω, SDAmBW[1:0]=10,11 (m=0,1)	±10	±20			

Note: These parameters are characterized but not tested.

Symbol	Parameter		Test Conditions		Тур.	Max.	Unit	
Symbol Parameter	Falailetei	VDD	Conditions	Min.	Typ.	IVIAA.	Unit	
			SDAmBW[1:0]=00B (m=0,1), no load		2.5	4.0	μA	
IOPA	Operating Current		SDAmBW[1:0]=01B (m=0,1), no load		10	16		
IOPA		_	SDAmBW[1:0]=10B (m=0,1), no load	—	80	128	μΑ	
			SDAmBW[1:0]=11B (m=0,1), no load	_	200	320		
Vos	Vos Input Offset Voltage		Without calibration (SDAmOF[5:0]=100000B (m=0,1))	-15	_	15	mV	
			With calibration	-2 — +2				
los	Input Offset Current	—	V <sub>IN</sub> =1/2V <sub>CM</sub>	—	1	10	nA	
Vсм	Common Mode Voltage Range	_	SDAmBW[1:0]=00, 01,10, 11(m=0,1)	Vss	_	V <sub>DD</sub> -1.4	V	
PSRR	Power Supply Rejection Ratio	—	SDAmBW[1:0]=00, 01,10, 11(m=0,1)	50	70		dB	
CMRR	Common Mode Rejection Ratio	—	SDAmBW[1:0]=00, 01,10, 11(m=0,1)	50	80	_	dB	
Aol	Open Loop Gain	—	SDAmBW[1:0]=00, 01,10, 11(m=0,1)	60	80		dB	
			R <sub>LOAD</sub> =1MΩ, C <sub>LOAD</sub> =60pF, SDAmBW[1:0]=00 (m=0,1)	0.5	1.5	_		
SR Slew Rate	Class Data		R <sub>LOAD</sub> =1MΩ, C <sub>LOAD</sub> =60pF, SDAmBW[1:0]=01 (m=0,1)	5	15	_	\//maa	
	Siew Rale	_	R <sub>LOAD</sub> =1MΩ, C <sub>LOAD</sub> =60pF, SDAmBW[1:0]=10 (m=0,1)	180	500	_	V/ms	
			R <sub>LOAD</sub> =1MΩ, C <sub>LOAD</sub> =60pF, SDAmBW[1:0]=11 (m=0,1)	600	1800			



Symbol	Parameter	Test Conditions			Tun	Max.	Unit	
Symbol Parame	Fardilleler	$V_{\text{DD}}$	Conditions	Min.	Тур.	IVIAX.	Unit	
		R <sub>LOAD</sub> =1MΩ, C <sub>LOAD</sub> =60pF, SDAmBW[1:0]=00 (m=0,1)	2	5	_			
CDW	Cain Bandwidth		R <sub>LOAD</sub> =1MΩ, C <sub>LOAD</sub> =60pF, SDAmBW[1:0]=01 (m=0,1)	15	40	_		
GBW Gain Bandwidth		R <sub>LOAD</sub> =1MΩ, C <sub>LOAD</sub> =60pF, SDAmBW[1:0]=10 (m=0,1)	250	600	_	kHz		
		R <sub>LOAD</sub> =1MΩ, C <sub>LOAD</sub> =60pF, SDAmBW[1:0]=11 (m=0,1)	800	2000	_			
			SDAmBW[1:0]=00,01 (m=0,1) R <sub>LOAD</sub> =5kΩ to V <sub>DD</sub> /2	V <sub>SS</sub> +140	_	V <sub>DD</sub> -160	mV	
Vor Maximum Output Voltage Range	Maximum Output voltage Range		SDAmBW[1:0]=10,11 (m=0,1) R <sub>LOAD</sub> =5kΩ to V <sub>DD</sub> /2	V <sub>SS</sub> +120	_	V <sub>DD</sub> -140	mv	
	Output Short Circuit Current	R <sub>LOAD</sub> =5.1Ω, SDAmBW[1:0]=00,01 (m=0,1)	±1.2	±12	_	mA		
Isc Output Short Cir	Output Short Circuit Current		R <sub>LOAD</sub> =5.1Ω, SDAmBW[1:0]=10,11 (m=0,1)	±2	±20	_	ША	

Note: These parameters are characterized but not tested.

# **Calendar Electrical Characteristics**

### D.C. Characteristics

							Ta=25°C
Symbol	<b>D</b>		Test Conditions		_	Max.	
Symbol	Parameter	VDD	Conditions	Min.	Тур.	wax.	Unit
I <sub>STB</sub> Standby Current	Standby Current	3V			—	100	nA
	Standby Current	5V		_	—	100	nA
IDD Operating Current 3V 5V	No load	_	0.50	0.70			
	5V	NO IOAU	_	0.85	1.00	μA	
	Source Current	3V	V <sub>он</sub> =2.7V	-0.35	-0.70	_	
Іон	Source Current	5V	V <sub>OH</sub> =4.5V	-0.50	-1.00	_	mA
	Sink Oursent	3V	V <sub>OL</sub> =0.3V	1.20	2.50	—	A
IOL	Sink Current	5V	V <sub>OL</sub> =0.5V	2.00	4.00	_	mA
V		3V		2.00	_	_	V
V <sub>IH</sub> "H" Input Voltage		5V	1 —	2.00	_	_	V
V	"I " Input \/oltogo	3V		_	_	0.60	V
VIL	"L" Input Voltage	5V	1 —	_	_	0.80	V

Note:  $I_{STB}$  is measured when SCLK, DIO,  $\overline{REST}$  are fixed at a low level, and the Clock Halt bit must be set to logic 1 (oscillator disabled).

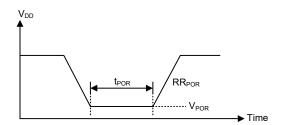


### A.C. Characteristics

							Ta=25°C
Symphol	Parameter	Tes	st Conditions	Min.	<b></b>		Unit
Symbol		V <sub>DD</sub>	Conditions		Тур.	Max.	Unit
+	Data ta Clask Satur	3V	_	100	_	_	
t <sub>DC</sub>	Data to Clock Setup	5V		50	_	_	ns
tсрн	Clock to Data Hold	3V	—	140	—	—	nc
LCDH		5V	—	70	_	—	ns
t <sub>CDD</sub> Clock to Data Delay	Clock to Data Dalay	3V	—	—	—	400	<b>n</b> 0
	5V	_	_	_	200	ns	
+	Clock Low Time	3V	—	500	—	—	ns
t <sub>CL</sub>	Clock Low Time	5V	—	250	_	_	115
t <sub>CH</sub> Clock High Time	Cleak Lligh Time	3V		500	_	_	
	Clock High Time	5V		250	_	_	ns
fscik	Cleak Fraguenay	3V	—	—	_	1.0	MHz
ISCLK	Clock Frequency	5V	—	—	_	2.0	IVITIZ
tr/tr	Clock Rise and Fall Time	3V	—	—	_	1000	20
Lr/ Lf	Clock Rise and Fall Time	5V	—	—	—	500	ns
t <sub>cc</sub>	Reset to Clock Setup	3V	—	2	_	—	
LCC	Reset to Clock Setup	5V	—	1	—	—	μs
t <sub>ссн</sub>	Clock to Reset Hold	3V		120	_	_	<b>n</b> 0
CCH	Clock to Reset Hold	5V	—	60	_	—	ns
tсwн	Reset Inactive Time	3V		2	—	_	
CWH		5V	—	1	—		μs
+	Popot to DIO High Impodores	3V		—	_	140	
t <sub>CDZ</sub>	Reset to DIO High Impedance	5V		_		70	ns

# **Power-on Reset Characteristics**

						Т	a=25°C
Symbol	Parameter	Те	st Conditions	Min.	Turn	Max.	Unit
Symbol	Parameter	VDD	Conditions		Тур.	wax.	Unit
VPOR	V <sub>DD</sub> Start Voltage to Ensure Power-on Reset	_	—	_	—	100	mV
RRPOR	V <sub>DD</sub> Rising Rate to Ensure Power-on Reset	_	—	0.035	_	—	V/ms
t <sub>POR</sub>	Minimum Time for V <sub>DD</sub> Stays at V <sub>POR</sub> to Ensure Power-on Reset	_		1			ms



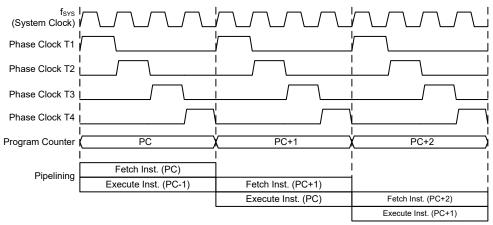


### System Architecture

A key factor in the high-performance features of the Holtek range of microcontrollers is attributed to their internal system architecture. The range of the devices take advantage of the usual features found within RISC microcontrollers providing increased speed of operation and enhanced performance. The pipelining scheme is implemented in such a way that instruction fetching and instruction execution are overlapped, hence instructions are effectively executed in one or two cycles for most of the standard or extended instructions respectively. The exceptions to these are branch or call instructions which need one more cycle. An 8-bit wide ALU is used in practically all instruction set operations, which carries out arithmetic operations, logic operations, rotation, increment, decrement, branch decisions, etc. The internal data path is simplified by moving data through the Accumulator and the ALU. Certain internal registers are implemented in the Data Memory and can be directly or indirectly addressed. The simple addressing methods of these registers along with additional architectural features ensure that a minimum of external components is required to provide a functional I/O and A/D control system with maximum reliability and flexibility. This makes the devices suitable for low-cost, high-volume production for controller applications.

### **Clocking and Pipelining**

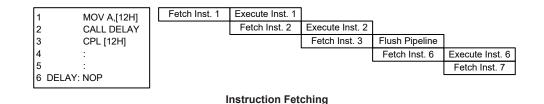
The main system clock, derived from either a HIRC or LIRC oscillator is subdivided into four internally generated non-overlapping clocks, T1~T4. The Program Counter is incremented at the beginning of the T1 clock during which time a new instruction is fetched. The remaining T2~T4 clocks carry out the decoding and execution functions. In this way, one T1~T4 clock cycle forms one instruction cycle. Although the fetching and execution of instructions takes place in consecutive instruction cycles, the pipelining structure of the microcontroller ensures that instructions are effectively executed in one instruction cycle. The exception to this are instructions where the contents of the Program Counter are changed, such as subroutine calls or jumps, in which case the instruction will take one more instruction cycle to execute.



#### System Clocking and Pipelining

For instructions involving branches, such as jump or call instructions, two machine cycles are required to complete instruction execution. An extra cycle is required as the program takes one cycle to first obtain the actual jump or call address and then another cycle to actually execute the branch. The requirement for this extra cycle should be taken into account by programmers in timing sensitive applications.





#### Program Counter

During program execution, the Program Counter is used to keep track of the address of the next instruction to be executed. It is automatically incremented by one each time an instruction is executed except for instructions, such as "JMP" or "CALL" that demands a jump to a non-consecutive Program Memory address. Only the lower 8 bits, known as the Program Counter Low Register, are directly addressable by the application program.

When executing instructions requiring jumps to non-consecutive addresses such as a jump instruction, a subroutine call, interrupt or reset, etc., the microcontroller manages program control by loading the required address into the Program Counter. For conditional skip instructions, once the condition has been met, the next instruction, which has already been fetched during the present instruction execution, is discarded and a dummy cycle takes its place while the correct instruction is obtained.

Program Counter						
High Byte Low Byte (PCL)						
PC11~PC8	PCL7~PCL0					
_						

#### **Program Counter**

The lower byte of the Program Counter, known as the Program Counter Low register, PCL, is available for program control and is a readable and writeable register. By transferring data directly into this register, a short program jump can be executed directly; however, as only this low byte is available for manipulation, the jumps are limited to the present page of memory that is 256 locations. When such program jumps are executed it should also be noted that a dummy cycle will be inserted. Manipulating the PCL register may cause program branching, so an extra cycle is needed to pre-fetch.

#### Stack

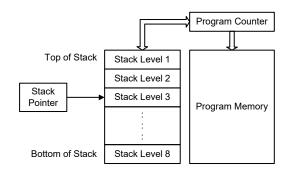
This is a special part of the memory which is used to save the contents of the Program Counter only. The stack has multiple levels and neither part of the data nor part of the program space, and is neither readable nor writeable. The activated level is indexed by the Stack Pointer, and is neither readable nor writeable. At a subroutine call or interrupt acknowledge signal, the contents of the Program Counter are pushed onto the stack. At the end of a subroutine or an interrupt routine, signaled by a return instruction, RET or RETI, the Program Counter is restored to its previous value from the stack. After a device reset, the Stack Pointer will point to the top of the stack.

If the stack is full and an enabled interrupt takes place, the interrupt request flag will be recorded but the acknowledge signal will be inhibited. When the Stack Pointer is decremented, by RET or RETI, the interrupt will be serviced. This feature prevents stack overflow allowing the programmer to use the structure more easily. However, when the stack is full, a CALL subroutine instruction can still be executed which will result in a stack overflow. Precautions should be taken to avoid such cases which might cause unpredictable program branching.

If the stack is overflow, the first Program Counter save in the stack will be lost.

27





### Arithmetic and Logic Unit – ALU

The arithmetic-logic unit or ALU is a critical area of the microcontroller that carries out arithmetic and logic operations of the instruction set. Connected to the main microcontroller data bus, the ALU receives related instruction codes and performs the required arithmetic or logical operations after which the result will be placed in the specified register. As these ALU calculation or operations may result in carry, borrow or other status changes, the status register will be correspondingly updated to reflect these changes. The ALU supports the following functions:

- Arithmetic operations: ADD, ADDM, ADC, ADCM, SUB, SUBM, SBC, SBCM, DAA, LADD, LADDM, LADC, LADCM, LSUB, LSUBM, LSBC, LSBCM, LDAA
- Logic operations: AND, OR, XOR, ANDM, ORM, XORM, CPL, CPLA, LAND, LOR, LXOR, LANDM, LORM, LXORM, LCPL, LCPLA
- Rotation: RRA, RR, RRCA, RRC, RLA, RL, RLCA, RLC, LRR, LRRA, LRRCA, LRRC, LRLA, LRL, LRLCA, LRLC
- Increment and Decrement: INCA, INC, DECA, DEC, LINCA, LINC, LDECA, LDEC
- Branch decision: JMP, SZ, SZA, SNZ, SIZ, SDZ, SIZA, SDZA, CALL, RET, RETI, LSZ, LSZA, LSNZ, LSIZ, LSDZ, LSIZA, LSDZA

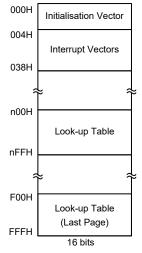


### Flash Program Memory

The Program Memory is the location where the user code or program is stored. For the devices the Program Memory is Flash type, which means it can be programmed and re-programmed a large number of times, allowing the user the convenience of code modification on the same device. By using the appropriate programming tools, the Flash devices offer users the flexibility to conveniently debug and develop their applications while also offering a means of field programming and updating.

### Structure

The Program Memory has a capacity of  $4K \times 16$  bits. The Program Memory is addressed by the Program Counter and also contains data, table information and interrupt entries. Table data, which can be setup in any location within the Program Memory, is addressed by a separate table pointer register.



**Program Memory Structure** 

### **Special Vectors**

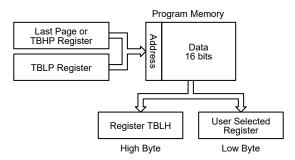
Within the Program Memory, certain locations are reserved for the reset and interrupts. The location 000H is reserved for use by the device reset for program initialisation. After a device reset is initiated, the program will jump to this location and begin execution.

#### Look-up Table

Any location within the Program Memory can be defined as a look-up table where programmers can store fixed data. To use the look-up table, the table pointer must first be setup by placing the address of the look up data to be retrieved in the table pointer registers, TBLP and TBHP. These registers define the total address of the look-up table.

After setting up the table pointer, the table data can be retrieved from the Program Memory using the corresponding table read instruction such as "TABRD [m]" or "TABRDL [m]" respectively when the memory [m] is located in sector 0. If the memory [m] is located in other sectors, the data can be retrieved from the program memory using the corresponding extended table read instruction such as "LTABRD [m]" or "LTABRDL [m]" respectively. When the instruction is executed, the lower order table byte from the Program Memory will be transferred to the user defined Data Memory register [m] as specified in the instruction. The higher order table data byte from the Program Memory will be transferred to the TBLH special register.

The accompanying diagram illustrates the addressing data flow of the look-up table.



### **Table Program Example**

The following example shows how the table pointer and table data is defined and retrieved from the microcontroller. This example uses raw table data located in the Program Memory which is stored there using the ORG statement. The value at this ORG statement is "0F00H" which refers to the start address of the last page within the 4K Program Memory of the devices. The table pointer low byte register is setup here to have an initial value of "06H". This will ensure that the first data read from the data table will be at the Program Memory address "0F06H" or 6 locations after the start of the last page. Note that the value for the table pointer is referenced to the first address specified by TBLP and TBHP if the "TABRD [m]" or "LTABRD [m]" instruction is being used. The high byte of the table data which in this case is equal to zero will be transferred to the TBLH register automatically when the "TABRD [m]" or "LTABRD [m]" instruction is executed.

Because the TBLH register is a read/write register and can be restored, care should be taken to ensure its protection if both the main routine and Interrupt Service Routine use table read instructions. If using the table read instructions, the Interrupt Service Routines may change the value of the TBLH and subsequently cause errors if used again by the main routine. As a rule it is recommended that simultaneous use of the table read instructions should be avoided. However, in situations where simultaneous use cannot be avoided, the interrupts should be disabled prior to the execution of any main routine table-read instructions. Note that all table related instructions require two instruction cycles to complete their operation.

### Table Read Program Example

```
tempreg1 db ? ; temporary register #1
tempreq2 db ? ; temporary register #2
:
mov a,06h
               ; initialise low table pointer - note that this address is referenced
mov tblp,a
               ; to the last page or the page that the pointed
mov a,OFh
               ; initialise high table pointer
mov tbhp,a
tabrd tempreq1 ; transfers value in table referenced by table pointer data at program
               ; memory address "OFO6H" transferred to tempreq1 and TBLH
dec tblp
               ; reduce value of table pointer by one
tabrd tempreq2 ; transfers value in table referenced by table pointer data at program
               ; memory address "OFO5H" transferred to tempreg2 and TBLH in this example
               ; the data "1AH" is transferred to tempreg1 and data "OFH" to register
               ; tempreq2
:
org OFOOh
               ; sets initial address of program memory
dc 00Ah, 00Bh, 00Ch, 00Dh, 00Eh, 00Fh, 01Ah, 01Bh
:
```



### In Circuit Programming – ICP

The provision of Flash type Program Memory provides the user with a means of convenient and easy upgrades and modifications to their programs on the same device. As an additional convenience, Holtek has provided a means of programming the microcontroller in-circuit using a 4-pin interface. This provides manufacturers with the possibility of manufacturing their circuit boards complete with a programmed or un-programmed microcontroller, and then programming or upgrading the program at a later stage. This enables product manufacturers to easily keep their manufactured products supplied with the latest program releases without removal and re-insertion of the devices.

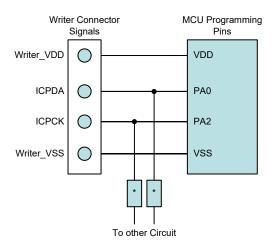
Holtek Writer Pins	MCU Programming Pins	Pin Description
ICPDA	PA0	Programming Serial Data/Address
ICPCK	PA2	Programming Clock
VDD	VDD	Power Supply
VSS	VSS	Ground

The Holtek Flash MCU to Writer Programming Pin correspondence table is as follows:

Note: All the VDD pins in the MCU must be connected to the Writer VDD pins.

The Program Memory can be programmed serially in-circuit using this 4-wire interface. Data is downloaded and uploaded serially on a single pin with an additional line for the clock. Two additional lines are required for the power supply. The technical details regarding the in-circuit programming of the devices is beyond the scope of this document and will be supplied in supplementary literature.

During the programming process, the user must take care of the PA0 and PA2 pins for data and clock programming purposes to ensure that no other outputs are connected to these two pins.



Note: \* may be resistor or capacitor. The resistance of \* must be greater than  $1k\Omega$  or the capacitance of \* must be less than 1nF.

### On Chip Debug Support – OCDS

There is an EV chip which is used to emulate the real MCU device. The EV chip device also provides an "On-Chip Debug" function to debug the real MCU device during the development process. The EV chip and the real MCU device are almost functionally compatible except for "On-Chip Debug" function. Users can use the EV chip device to emulate the real chip device behavior by connecting the OCDSDA and OCDSCK pins to the Holtek HT-IDE development tools. The OCDSDA pin is the OCDS Data/Address input/output pin while the OCDSCK pin is the OCDS clock input pin. When users use the EV chip for debugging, other functions which are shared with

the OCDSDA and OCDSCK pins in the device will have no effect in the EV chip. However, the two OCDS pins which are pin-shared with the ICP programming pins are still used as the Flash Memory programming pins for ICP. For more detailed OCDS information, refer to the corresponding document named "Holtek e-Link for 8-bit MCU OCDS User's Guide".

Holtek e-Link Pins	EV Chip Pins	Pin Description
OCDSDA	OCDSDA	On-Chip Debug Support Data/Address input/output
OCDSCK	OCDSCK	On-Chip Debug Support Clock input
VDD	VDD	Power Supply
VSS	VSS	Ground

Note: All the VDD pins in the EV chip must be connected to the e-Link VDD pins.

### **Data Memory**

The Data Memory is a volatile area of 8-bit wide RAM internal memory and is the location where temporary information is stored.

Categorized into two types, the first of these is an area of RAM where special function registers are located. These registers have fixed locations and are necessary for correct operation of the devices. Many of these registers can be read from and written to directly under program control, however, some remain protected from user manipulation. The second area of Data Memory is reserved for general purpose use. All locations within this area are read and write accessible under program control.

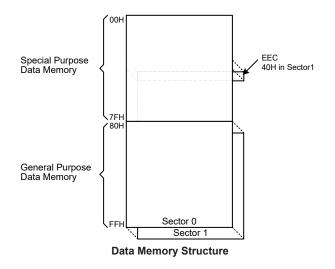
Switching between the different Data Memory sectors is achieved by properly setting the Memory Pointers to correct value when using the indirect addressing method.

### Structure

The Data Memory is subdivided into several sectors, all of which are implemented in 8-bit wide RAM. Each of the Data Memory Sector is categorized into two types, the special Purpose Data Memory and the General Purpose Data Memory. The address range of the Special Purpose Data Memory for the devices is from 00H to 7FH while the General Purpose Data Memory address range is from 80H to FFH.

Special Purpose Data Memory	General Purpose Data Memory		
Located Sectors	Capacity	Sector: Address	
0: 00H~7FH 1: 40H (EEC only)	256×8	0: 80H~FFH 1: 80H~FFH	





### **Data Memory Addressing**

For the devices that support the extended instructions, there is no Bank Pointer for Data Memory addressing. The desired Sector is pointed by the MP1H or MP2H register and the certain Data Memory address in the selected sector is specified by the MP1L or MP2L register when using indirect addressing access.

Direct Addressing can be used in all sectors using the corresponding instruction which can address all available data memory space. For the accessed data memory which is located in any data memory sectors except sector 0, the extended instructions can be used to access the data memory instead of using the indirect addressing access. The main difference between standard instructions and extended instructions is that the data memory address "m" in the extended instructions has 9 valid bits, the high byte indicates a sector and the low byte indicates a specific address within the sector.

#### **General Purpose Data Memory**

All microcontroller programs require an area of read/write memory where temporary data can be stored and retrieved for use later. It is this area of RAM memory that is known as General Purpose Data Memory. This area of Data Memory is fully accessible by the user programing for both reading and writing operations. By using the bit operation instructions individual bits can be set or reset under program control giving the user a large range of flexibility for bit manipulation in the Data Memory.

### Special Purpose Data Memory

This area of Data Memory is where registers, necessary for the correct operation of the microcontroller, are stored. Most of the registers are both readable and writeable but some are protected and are readable only, the details of which are located under the relevant Special Function Register section. Note that for locations that are unused, any read instruction to these addresses will return the value "00H".



	Sector 0	Sector 1		Sector 0	Sector 1
00H	IAR0		40H		EEC
01H	MP0		41H	SIMC0	
02H	IAR1		42H	SIMC1/UUCR1	
03H	MP1L		43H	SIMC2/SIMA/UUCR2	
	MP1H		43H		
04H				SIMD/UTXR_RXR	
05H	ACC		45H	SIMTOC/UBRG	
06H	PCL		46H	UUSR	
07H	TBLP		47H	INTEG	
08H	TBLH		48H	INTC0	
09H	TBHP		49H	INTC1	
0AH	STATUS		4AH	INTC2	
0BH	VBGRC		4BH	INTC3	
0CH	IAR2		4CH	PAS0	
	MP2L		40H		
0DH				PAS1	
0EH	MP2H		4EH	PBS0	
0FH	RSTFC		4FH	PBS1	
10H	TB0C		50H	IFS0	
11H	TB1C		51H		
12H	SCC		52H	PTMC0	
13H	HIRCC		53H	PTMC1	
14H	PA		54H	PTMC2	
	PAC		55H		
15H				PTMDL	
16H	PAPU		56H	PTMDH	
17H	PAWU		57H	PTMAL	
18H	PB		58H	PTMAH	
19H	PBC		59H	PTMBL	
1AH	PBPU		5AH	PTMBH	
1BH	SLEDC		5BH	PTMRPL	
1CH			5CH	PTMRPH	
1DH	PSCR		5DH	ISGENC	
1EH	LVDC		5EH	ISGDATA0	
1FH			5FH	ISGDATA1	
20H	SDSW		60H		
21H	SDPGAC0		61H		
22H	SDPGAC1		62H		
23H	SDA0C		63H		
24H	SDA0VOS		64H		
25H	SDA1C		65H		
26H	SDA1VOS		66H		
27H	STMC0		67H		
28H	STMC1		68H		
29H	STMDL		69H		
2AH	STMDH		6AH		
2BH	STMAL		6BH		
2CH	STMAH		6CH		
2DH	SADOL		6DH		
2EH	SADOH		6EH		
2FH	SADC0		6FH		
30H	SADC1		70H		
31H			70H		
	200000000000000000000000000000000000000				
32H			72H		
33H	000000000000000000000000000000000000000		73H		
34H			74H		
35H	8888888888888		75H		
36H	200000000000000000000000000000000000000		76H		
37H			77H		
38H	6222222222222		78H		
39H	2000000000000		79H		
3AH			7AH		
	62222222222222				
	800000000000		7BH		
3BH			7CH		
3CH					
3CH 3DH	WDTC		7DH		
3CH	WDTC EEA				

: Unused, read as 00H : Reserved, cannot be changed

Special Purpose Data Memory Structure



# **Special Function Register Description**

Most of the Special Function Register details will be described in the relevant functional sections however several registers require a separate description in this section.

### Indirect Addressing Registers – IAR0, IAR1, IAR2

The Indirect Addressing Registers, IAR0, IAR1 and IAR2, although having their locations in normal RAM register space, do not actually physically exist as normal registers. The method of indirect addressing for RAM data manipulation uses these Indirect Addressing Registers and Memory Pointers, in contrast to direct memory addressing, where the actual memory address is specified. Actions on the IAR0, IAR1 and IAR2 registers will result in no actual read or write operation to these registers but rather to the memory location specified by their corresponding Memory Pointers, MP0, MP1L/MP1H or MP2L/MP2H. Acting as a pair, IAR0 and MP0 can together access data only from Sector 0 while the IAR1 register together with the MP1L/MP1H register pair and IAR2 register together with the MP2L/MP2H register pair can access data from any Data Memory Sector. As the Indirect Addressing Registers are not physically implemented, reading the Indirect Addressing Registers will result of "00H" and writing to the registers will result in no operation.

### Memory Pointers – MP0, MP1L, MP1H, MP2L, MP2H

Five Memory Pointers, known as MP0, MP1L, MP1H, MP2L, MP2H, are provided. These Memory Pointers are physically implemented in the Data Memory and can be manipulated in the same way as normal registers providing a convenient way with which to address and track data. When any operation to the relevant Indirect Addressing Registers is carried out, the actual address that the microcontroller is directed to is the address specified by the related Memory Pointer. MP0, together with Indirect Addressing Register, IAR0, are used to access data from Sector 0, while MP1L/MP1H together with IAR1 and MP2L/MP2H together with IAR2 are used to access data from all sectors according to the corresponding MP1H or MP2H register. Direct Addressing can be used in all sectors using the corresponding instruction which can address all available data memory space.

The following example shows how to clear a section of four Data Memory locations already defined as locations adres1 to adres4.

#### Indirect Addressing Program Example 1

```
data .section 'data
adres1 db ?
        db ?
adres2
adres3
        db ?
adres4 db ?
        db ?
block
code .section at 0 'code'
org OOh
start:
     mov a, 04h
                             ; setup size of block
    mov block, a
    mov a, offset adres1
                             ; Accumulator loaded with first RAM address
    mov mp0, a
                             ; setup memory pointer with first RAM address
loop:
     clr IAR0
                             ; clear the data at address defined by MPO
     inc mp0
                             ; increment memory pointer
     sdz block
                              ; check if last memory location has been cleared
     jmp loop
continue:
```



#### Indirect Addressing Program Example 2

```
data .section 'data'
adres1 db ?
adres2 db ?
adres3 db ?
adres4 db ?
block db ?
code .section at 0 'code'
org OOh
start:
    mov a, 04h
                           ; setup size of block
    mov block, a
    mov a, 01h
                            ; setup the memory sector
    mov mplh, a
    mov a, offset adres1
                           ; Accumulator loaded with first RAM address
    mov mp11, a
                            ; setup memory pointer with first RAM address
loop:
    clr IAR1
                            ; clear the data at address defined by MP1L
    inc mpll
                            ; increment memory pointer MP1L
    sdz block
                            ; check if last memory location has been cleared
     jmp loop
continue:
```

The important point to note here is that in the example shown above, no reference is made to specific Data Memory addresses.

#### **Direct Addressing Program Example using extended instructions**

```
data .section 'data'
temp db ?
code .section at 0 'code'
org 00h
start:
                       ; move [m] data to acc
    lmov a, [m]
    lsub a, [m+1]
                       ; compare [m] and [m+1] data
    snz c
                        ; [m]>[m+1]?
     jmp continue
                        ; no
                        ; yes, exchange [m] and [m+1] data
    lmov a, [m]
    mov temp, a
    lmov a, [m+1]
    lmov [m], a
    mov a, temp
    lmov [m+1], a
continue:
```

Note: here "m" is a data memory address located in any data memory sectors. For example, m=1F0H, it indicates address 0F0H in Sector 1.

#### Accumulator – ACC

The Accumulator is central to the operation of any microcontroller and is closely related with operations carried out by the ALU. The Accumulator is the place where all intermediate results from the ALU are stored. Without the Accumulator it would be necessary to write the result of each calculation or logical operation such as addition, subtraction, shift, etc., to the Data Memory resulting in higher programming and timing overheads. Data transfer operations usually involve the temporary storage function of the Accumulator; for example, when transferring data between one user-defined register and another, it is necessary to do this by passing the data through the Accumulator as no direct transfer between two registers is permitted.



### Program Counter Low Register – PCL

To provide additional program control functions, the low byte of the Program Counter is made accessible to programmers by locating it within the Special Purpose area of the Data Memory. By manipulating this register, direct jumps to other program locations are easily implemented. Loading a value directly into this PCL register will cause a jump to the specified Program Memory location, however, as the register is only 8-bit wide, only jumps within the current Program Memory page are permitted. When such operations are used, note that a dummy cycle will be inserted.

# Look-up Table Registers – TBLP, TBHP, TBLH

These three special function registers are used to control operation of the look-up table which is stored in the Program Memory. TBLP and TBHP are the table pointers and indicate the location where the table data is located. Their value must be setup before any table read commands are executed. Their value can be changed, for example using the "INC" or "DEC" instructions, allowing for easy table data pointing and reading. TBLH is the location where the high order byte of the table data is stored after a table read data instruction has been executed. Note that the lower order table data byte is transferred to a user defined location.

## Status Register – STATUS

This 8-bit register contains the SC flag, CZ flag, zero flag (Z), carry flag (C), auxiliary carry flag (AC), overflow flag (OV), power down flag (PDF), and watchdog time-out flag (TO). These arithmetic/ logical operation and system management flags are used to record the status and operation of the microcontroller.

With the exception of the TO and PDF flags, bits in the status register can be altered by instructions like most other registers. Any data written into the status register will not change the TO or PDF flag. In addition, operations related to the status register may give different results due to the different instruction operations. The TO flag can be affected only by a system power-up, a WDT time-out or by executing the "CLR WDT" or "HALT" instruction. The PDF flag is affected only by executing the "HALT" or "CLR WDT" instruction or during a system power-up.

The Z, OV, AC, C, SC and CZ flags generally reflect the status of the latest operations.

- C is set if an operation results in a carry during an addition operation or if a borrow does not take place during a subtraction operation; otherwise C is cleared. C is also affected by a rotate through carry instruction.
- AC is set if an operation results in a carry out of the low nibbles in addition, or no borrow from the high nibble into the low nibble in subtraction; otherwise AC is cleared.
- Z is set if the result of an arithmetic or logical operation is zero; otherwise Z is cleared.
- OV is set if an operation results in a carry into the highest-order bit but not a carry out of the highest-order bit, or vice versa; otherwise OV is cleared.
- PDF is cleared by a system power-up or executing the "CLR WDT" instruction. PDF is set by executing the "HALT" instruction.
- TO is cleared by a system power-up or executing the "CLR WDT" or "HALT" instruction. TO is set by a WDT time-out.
- CZ is the operational result of different flags for different instructions. Refer to register definitions for more details.
- SC is the result of the "XOR" operation which is performed by the OV flag and the MSB of the current instruction operation result.



In addition, on entering an interrupt sequence or executing a subroutine call, the status register will not be pushed onto the stack automatically. If the contents of the status registers are important and if the subroutine can corrupt the status register, precautions must be taken to correctly save it.

### STATUS Register

Bit	7	6	5	4	3	2	1	0
Name	SC	CZ	TO	PDF	OV	Z	AC	С
R/W	R/W	R/W	R	R	R/W	R/W	R/W	R/W
POR	х	х	0	0	х	х	х	х
							">	k": unknowr
Bit 7		result of th the instruct			nich is perf	ormed by th	ne OV flag	and the
Bit 6	For SUB For SBC result wh flag.	S/SUBM/LS C/SBCM/L	SUB/LSUB SBC/LSBC ormed by th	M instructi CM instruction ne previous	ons, the CZ tions, the ( operation (	ent instruct flag is equ CZ flag is CZ flag and	al to the Z : the "AND'	operation
Bit 5	<b>TO</b> : Wat 0: Afte	tchdog Tim	e-out flag or executir	ng the "CLI		"HALT" ir	struction	
Bit 4	0: Afte	ower down i er power up executing th	or executin			struction		
Bit 3	0: No 0 1: An 0	erflow flag overflow operation re lest-order b			e highest-or	rder bit but	not a carry	out of the
Bit 2		flag result of ar result of ar						
Bit 1	0: No a 1: An o	kiliary flag auxiliary ca operation re 1 the high n	sults in a c			bles in addi	tion, or no	borrow
Bit 0	C: Carry 0: No o 1: An o not t	r flag carry-out	sults in a c	arry during straction op	an additior eration	n operation		ow does



# **EEPROM Data Memory**

The devices contain an area of internal EEPROM Data Memory. EEPROM is by its nature a nonvolatile form of re-programmable memory, with data retention even when its power supply is removed. By incorporating this kind of data memory, a whole new host of application possibilities are made available to the designer. The availability of EEPROM storage allows information such as product identification numbers, calibration values, specific user data, system setup data or other product information to be stored directly within the product microcontroller. The process of reading and writing data to the EEPROM memory has been reduced to a very trivial affair.

## **EEPROM Data Memory Structure**

The device EEPROM Data Memory capacity is  $64 \times 8$  bits. Unlike the Program Memory and RAM Data Memory, the EEPROM Data Memory is not directly mapped into memory space and is therefore not directly addressable in the same way as the other types of memory. Read and Write operations to the EEPROM are carried out in single byte operations using an address and a data register in Sector 0 and a single control register in Sector 1.

# **EEPROM Registers**

Three registers control the overall operation of the internal EEPROM Data Memory. These are the address register, EEA, the data register, EED and a single control register, EEC. As both the EEA and EED registers are located in Sector 0, they can be directly accessed in the same way as any other Special Function Register. The EEC register however, being located in Sector 1, can only be read from or written to indirectly using the MP1L/MP1H or MP2L/MP2H Memory Pointer and Indirect Addressing Register, IAR1/IAR2. Because the EEC control register is located at address 40H in Sector 1, the MP1L or MP2L Memory Pointer must first be set to the value 40H and the MP1H or MP2H Memory Pointer high byte set to the value, 01H, before any operations on the EEC register are executed.

Register		Bit											
Name	7	6	5	4	3	2	1	0					
EEA	—	—	EEA5	EEA4	EEA3	EEA2	EEA1	EEA0					
EED	D7	D6	D5	D4	D3	D2	D1	D0					
EEC		—	—		WREN	WR	RDEN	RD					

### **EEPROM Register List**

### • EEA Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	EEA5	EEA4	EEA3	EEA2	EEA1	EEA0
R/W	—	_	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	—	0	0	0	0	0	0

Bit 7~6 Unimplemented, read as "0"

Bit 5~0 EEA5~EEA0: Data EEPROM address

Data EEPROM address bit 5 ~ bit 0



### • EED Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: Data EEPROM data

Data EEPROM data bit 7 ~ bit 0

#### • EEC Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	WREN	WR	RDEN	RD
R/W	—	—	—	—	R/W	R/W	R/W	R/W
POR	—	—	—	—	0	0	0	0

Bit 7~4 Unimplemented, read as "0"

Bit 3 WREN: Data EEPROM Write Enable

0: Disable

1: Enable

This is the Data EEPROM Write Enable Bit which must be set high before Data EEPROM write operations are carried out. Clearing this bit to zero will inhibit Data EEPROM write operations.

### Bit 2 WR: EEPROM Write Control

0: Write cycle has finished

1: Activate a write cycle

This is the Data EEPROM Write Control Bit and when set high by the application program will activate a write cycle. This bit will be automatically reset to zero by the hardware after the write cycle has finished. Setting this bit high will have no effect if the WREN has not first been set high.

#### Bit 1 **RDEN**: Data EEPROM Read Enable

0: Disable

1: Enable

This is the Data EEPROM Read Enable Bit which must be set high before Data EEPROM read operations are carried out. Clearing this bit to zero will inhibit Data EEPROM read operations.

### Bit 0 RD: EEPROM Read Control

0: Read cycle has finished

1: Activate a read cycle

This is the Data EEPROM Read Control Bit and when set high by the application program will activate a read cycle. This bit will be automatically reset to zero by the hardware after the read cycle has finished. Setting this bit high will have no effect if the RDEN has not first been set high.

- Note: 1. The WREN, WR, RDEN and RD cannot be set high at the same time in one instruction. The WR and RD cannot be set high at the same time.
  - 2. Ensure that the  $f_{SUB}$  clock is stable before executing the write operation.
  - 3. Ensure that the write operation is totally complete before changing the contents of the EEPROM related registers.



### Reading Data from the EEPROM

To read data from the EEPROM, the EEPROM address of the data to be read must first be placed in the EEA register. Then the read enable bit, RDEN, in the EEC register must then be set high to enable the read function. If the RD bit in the EEC register is now set high, a read cycle will be initiated. Setting the RD bit high will not initiate a read operation if the RDEN bit has not been set. When the read cycle terminates, the RD bit will be automatically cleared to zero, after which the data can be read from the EED register. The data will remain in the EED register until another read or write operation is executed. The application program can poll the RD bit to determine when the data is valid for reading.

### Writing Data to the EEPROM

To write data to the EEPROM, the EEPROM address of the data to be written must first be placed in the EEA register and the data placed in the EED register. To initiate a write cycle, the write enable bit, WREN, in the EEC register must first be set high to enable the write function. After this, the WR bit in the EEC register must be immediately set high to initiate a write cycle. These two instructions must be executed in two consecutive instruction cycles. The global interrupt bit EMI should also first be cleared before implementing any write operations, and then set again after the write cycle has started. Note that setting the WR bit high will not initiate a write cycle if the WREN bit has not been set. As the EEPROM write cycle is controlled using an internal timer whose operation is asynchronous to microcontroller system clock, a certain time will elapse before the data will have been written into the EEPROM. Detecting when the write cycle has finished can be implemented either by polling the WR bit in the EEC register or by using the EEPROM interrupt. When the write cycle terminates, the WR bit will be automatically cleared to zero by the microcontroller, informing the user that the data has been written to the EEPROM. The application program can therefore poll the WR bit to determine when the write cycle has ended.

### Write Protection

Protection against inadvertent write operation is provided in several ways. After the devices are powered-on the Write Enable bit in the control register will be cleared preventing any write operations. Also at power-on the Memory Pointer high byte register, MP1H or MP2H, will be reset to zero, which means that Data Memory Sector 0 will be selected. As the EEPROM control register is located in Sector 1, this adds a further measure of protection against spurious write operations. During normal program operation, ensuring that the Write Enable bit in the control register is cleared will safeguard against incorrect write operations.

### **EEPROM** Interrupt

The EEPROM write interrupt is generated when an EEPROM write cycle has ended. The EEPROM interrupt must first be enabled by setting the DEE bit in the relevant interrupt register. When an EEPROM write cycle ends, the DEF request flag will be set. If the global and EEPROM interrupts are enabled and the stack is not full, a jump to the Data EEPROM Interrupt vector will take place. When the interrupt is serviced the EEPROM interrupt flag will be automatically reset. More details can be obtained in the Interrupt section.

### Programming Considerations

Care must be taken that data is not inadvertently written to the EEPROM. Protection can be enhanced by ensuring that the Write Enable bit is normally cleared to zero when not writing. Also the Memory Pointer high byte register, MP1H or MP2H, could be normally cleared to zero as this would inhibit access to Sector 1 where the EEPROM control register exists. Although certainly not



necessary, consideration might be given in the application program to the checking of the validity of new write data by a simple read back process.

When writing data the WR bit must be set high immediately after the WREN bit has been set high, to ensure the write cycle executes correctly. The global interrupt bit EMI should also be cleared before a write cycle is executed and then re-enabled after the write cycle starts. Note that the devices should not enter the IDLE or SLEEP mode until the EEPROM read or write operation is totally complete. Otherwise, the EEPROM read or write operation will fail.

#### **Programming Examples**

#### Reading data from the EEPROM - polling method

	A, EEPROM_ADRES EEA, A	;	user defined address
	A, 040H	;	setup memory pointer MP1L
MOV	MP1L, A	;	MP1L points to EEC register
MOV	A, 01H	;	setup memory pointer MP1H
MOV	MP1H, A		
SET	IAR1.1	;	set RDEN bit, enable read operations
SET	IAR1.0	;	start Read Cycle - set RD bit
BACK	:		
SZ	IAR1.0	;	check for read cycle end
JMP	BACK		
CLR	IAR1	;	disable EEPROM read if no more read operations are required
CLR	MP1H		
MOV	A, EED	;	move read data to register
MOV	READ_DATA, A		

Note: For each read operation, the address register should be re-specified followed by setting the RD bit high to activate a read cycle even if the target address is consecutive.

#### Writing Data to the EEPROM – polling method

	A, EEPROM_ADRES	;	user defined address
MOV	EEA, A		
MOV	A, EEPROM_DATA	;	user defined data
MOV	EED, A		
MOV	A, 040H	;	setup memory pointer MP1L
MOV	MP1L, A	;	MP1L points to EEC register
MOV	A, OlH	;	setup memory pointer MP1H
MOV	MP1H, A		
CLR	EMI		
SET	IAR1.3	;	set WREN bit, enable write operations
SET	IAR1.2	;	start Write Cycle - set WR bit - executed immediately
		;	after set WREN bit
SET	EMI		
BACK	:		
SZ	IAR1.2	;	check for write cycle end
JMP	BACK		
CLR	MP1H		



# Oscillators

Various oscillator options offer the user a wide range of functions according to their various application requirements. The flexible features of the oscillator functions ensure that the best optimisation can be achieved in terms of speed and power saving. Oscillator selections and operation are selected only through the application program by using some control registers.

# **Oscillator Overview**

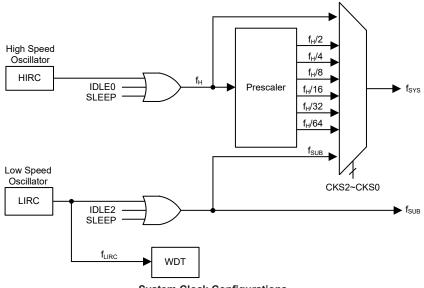
In addition to being the source of the main system clock the oscillators also provide clock sources for the Watchdog Timer and Time Base Interrupts. Two fully integrated internal oscillators, requiring no external components, are provided to form a wide range of both fast and slow system oscillators. All oscillator options all can be selected through register programming. The higher frequency oscillators provide higher performance but carry with it the disadvantage of higher power requirements, while the opposite is of course true for the lower frequency oscillators. With the capability of dynamically switching between fast and slow system clock, the devices have the flexibility to optimize the performance/power ratio, a feature especially important in power sensitive portable applications.

Туре	Name	Frequency
Internal High Speed RC	HIRC	2/4/8MHz
Internal Low Speed RC	LIRC	32kHz

Oscillator T	/pes
--------------	------

## System Clock Configurations

There are two oscillator sources, one high speed oscillator and one low speed oscillator. The high speed system clocks is sourced from the internal 2/4/8MHz RC oscillator, HIRC. The low speed oscillator is the internal 32kHz RC oscillator, LIRC. Selecting whether the low or high speed oscillator is used as the system oscillator is implemented using the CKS2~CKS0 bits in the SCC register and as the system clock can be dynamically selected.



System Clock Configurations



### Internal High Speed RC Oscillator - HIRC

The internal RC oscillator is a fully integrated system oscillator requiring no external components. The internal RC oscillator has three selectable frequencies of 2MHz, 4MHz and 8MHz. Device trimming during the manufacturing process and the inclusion of internal frequency compensation circuits are used to ensure that the influence of the power supply voltage, temperature and process variations on the oscillation frequency are minimised.

### Internal 32kHz Oscillator – LIRC

The Internal 32kHz System Oscillator is also a fully integrated RC oscillator with a typical frequency of 32kHz at at a full voltage range, requiring no external components for its implementation. Device trimming during the manufacturing process and the inclusion of internal frequency compensation circuits are used to ensure that the influence of the power supply voltage, temperature and process variations on the oscillation frequency are minimised.

# **Operating Modes and System Clocks**

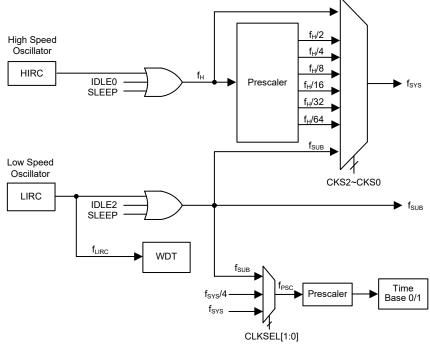
Present day applications require that their microcontrollers have high performance but often still demand that they consume as little power as possible, conflicting requirements that are especially true in battery powered portable applications. The fast clocks required for high performance will by their nature increase current consumption and of course vice versa, lower speed clocks reduce current consumption. As Holtek has provided the devices with both high and low speed clock sources and the means to switch between them dynamically, the user can optimise the operation of their microcontroller to achieve the best performance/power ratio.

### System Clocks

The devices have many different clock sources for both the CPU and peripheral function operation. By providing the user with a wide range of clock options using register programming, a clock system can be configured to obtain maximum application performance.

The main system clock, can come from a high frequency,  $f_{H}$ , or low frequency,  $f_{SUB}$ , source, and is selected using the CKS2~CKS0 bits in the SCC register. The high speed system clock is sourced from the HIRC oscillator, while the low speed system clock source is sourced from the internal clock  $f_{SUB}$  which is sourced by the LIRC oscillator. The other choice, which is a divided version of the high speed system oscillator has a range of  $f_{H}/2~f_{H}/64$ .





**Device Clock Configurations** 

Note: When the system clock source  $f_{SYS}$  is switched to  $f_{SUB}$  from  $f_H$ , the high speed oscillator can be stop to conserve the power by clearing the corresponding high speed oscillator enable control bit. However there is no  $f_H \sim f_H/64$  for peripheral circuit to use.

### **System Operation Modes**

There are six different modes of operation for the microcontroller, each one with its own special characteristics and which can be chosen according to the specific performance and power requirements of the application. There are two modes allowing normal operation of the microcontroller, the FAST Mode and SLOW Mode. The remaining four modes, the SLEEP, IDLE0, IDLE1 and IDLE2 Mode are used when the microcontroller CPU is switched off to conserve power.

Operation	CPU	Re	egister Se	tting	fsys	f⊦	£	f <sub>LIRC</sub>	
Mode	CFU	FHIDEN	FSIDEN	CKS2~CKS0	ISYS	IH	fsuв	LIKC	
FAST	On	х	х	000~110	f <sub>H</sub> ~f <sub>H</sub> /64	On	On	On	
SLOW	On	х	х	111	fsuв	On/Off <sup>(1)</sup>	On	On	
IDLE0	Off	0	1	000~110	Off	Off	On	On	
IDLEU	Oli	0		111	On	OII	OII	On	
IDLE1	Off	1	1	XXX	On	On	On	On	
IDLE2	Off	1	0	000~110	On	0.5	0#	0.5	
IDLEZ	Oli		0	111	Off	On	Off	On	
SLEEP	Off	0	0	XXX	Off	Off	Off	On (2)	
							(1.1) D		

"x": Don't care

- Note: 1. The  $f_{\rm H}$  clock will be switched on or off by configuring the corresponding oscillator enable bit in the SLOW mode.
  - 2. The  $f_{LIRC}$  clock will be switched on since the WDT function is always enabled even in the SLEEP mode.



#### **FAST Mode**

This is one of the main operating modes where the microcontroller has all of its functions operational and where the system clock is provided the high speed oscillator. This mode operates allowing the microcontroller to operate normally with a clock source from the HIRC high speed oscillator. The high speed oscillator will however first be divided by a ratio ranging from 1 to 64, the actual ratio being selected by the CKS2~CKS0 bits in the SCC register. Although a high speed oscillator is used, running the microcontroller at a divided clock ratio reduces the operating current.

#### SLOW Mode

This is also a mode where the microcontroller operates normally although now with a slower speed clock source. The clock source used will be from  $f_{SUB}$ . The  $f_{SUB}$  clock is derived from the LIRC oscillator.

#### **SLEEP Mode**

The SLEEP Mode is entered when an HALT instruction is executed and when the FHIDEN and FSIDEN bit both are low. In the SLEEP mode the CPU will be stopped. The  $f_{SUB}$  clock provided to the peripheral function will also be stopped. However the  $f_{LIRC}$  clock still continues to operate since the WDT function is enabled.

#### **IDLE0 Mode**

The IDLE0 Mode is entered when an HALT instruction is executed and when the FHIDEN bit in the SCC register is low and the FSIDEN bit in the SCC register is high. In the IDLE0 Mode the CPU will be switched off but the low speed oscillator will be turned on to drive some peripheral functions.

#### **IDLE1 Mode**

The IDLE1 Mode is entered when an HALT instruction is executed and when the FHIDEN bit in the SCC register is high and the FSIDEN bit in the SCC register is high. In the IDLE1 Mode the CPU will be switched off but both the high and low speed oscillators will be turned on to provide a clock source to keep some peripheral functions operational.

#### **IDLE2 Mode**

The IDLE2 Mode is entered when an HALT instruction is executed and when the FHIDEN bit in the SCC register is high and the FSIDEN bit in the SCC register is low. In the IDLE2 Mode the CPU will be switched off but the high speed oscillator will be turned on to provide a clock source to keep some peripheral functions operational.

### **Control Registers**

The registers, SCC and HIRCC, are used to control the system clock and the HIRC oscillator configurations.

Register		Bit											
Name	7	6	5	4	3	2	1	0					
SCC	CKS2	CKS1	CKS0	_	—	—	FHIDEN	FSIDEN					
HIRCC				_	HIRC1	HIRC0	HIRCF	HIRCEN					

System Operating Mode Control Register List



#### SCC Register

		v	Э	4	3	2	1	U
Name C	CKS2	CKS1	CKS0	—	—	_	FHIDEN	FSIDEN
R/W I	R/W	R/W	R/W	—	_	_	R/W	R/W
POR	0	0	0	—		_	0	0

Bit 7~5 CKS2~CKS0: System clock selection

	000: $f_{\rm H}$
	$001: f_{\rm H}/2$
	010: f <sub>H</sub> /4
	$011: f_{\rm H}/8$
	100: f <sub>H</sub> /16
	101: f <sub>H</sub> /32
	110: f <sub>H</sub> /64
	111: f <sub>SUB</sub>
	These three bits are used to select which clock is used as the system clock source. In addition to the system clock source directly derived from $f_H$ or $f_{SUB}$ , a divided version of the high speed system oscillator can also be chosen as the system clock source.
Bit 4~2	Unimplemented, read as 0.
Bit 1	FHIDEN: High Frequency oscillator control when CPU is switched off 0: Disable 1: Enable
	This bit is used to control whether the high speed oscillator is activated or stopped when the CPU is switched off by executing an "HALT" instruction.
D:+ 0	ESIDEN, Low Frequency escillator control when CDL is gwitched off

### Bit 0 FSIDEN: Low Frequency oscillator control when CPU is switched off

0: Disable

1: Enable

This bit is used to control whether the low speed oscillator is activated or stopped when the CPU is switched off by executing an "HALT" instruction. The LIRC oscillator is controlled by this bit together with the WDT function enable control. If this bit is cleared to 0 but the WDT function is enabled, the  $f_{LIRC}$  clock will also be enabled.

### HIRCC Register

Bit	7	6	5	4	3	2	1	0
Name	_	_	—	—	HIRC1	HIRC0	HIRCF	HIRCEN
R/W	_	_	—	—	R/W	R/W	R	R/W
POR	—	—	_	—	0	0	0	1

Bit 7~4 Unimplemented, read as 0.

00: 2MHz

- 01: 4MHz
- 10: 8MHz
- 11: 2MHz

When the HIRC oscillator is enabled, the HIRC frequency is changed by changing these two bits, the clock frequency will automatically be changed after the HIRCF flag is set to 1.

### Bit 1 HIRCF: HIRC oscillator stable flag

- 0: HIRC unstable
- 1: HIRC stable

This bit is used to indicate whether the HIRC oscillator is stable or not. When the HIRCEN bit is set to 1 to enable the HIRC oscillator or the HIRC frequency selection is changed by application program, the HIRCF bit will first be cleared to 0 and then set to 1 after the HIRC oscillator is stable.

### Bit 0 HIRCEN: HIRC oscillator enable control

- 0: Disable
- 1: Enable

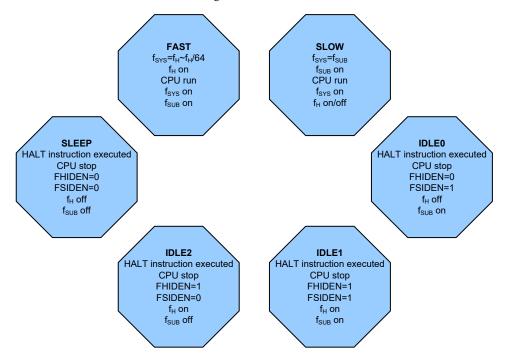
Bit 3~2 HIRC1~HIRC0: HIRC frequency selection



### **Operating Mode Switching**

The devices can switch between operating modes dynamically allowing the user to select the best performance/power ratio for the present task in hand. In this way microcontroller operations that do not require high performance can be executed using slower clocks thus requiring less operating current and prolonging battery life in portable applications.

In simple terms, Mode Switching between the FAST Mode and SLOW Mode is executed using the CKS2~CKS0 bits in the SCC register while Mode Switching from the FAST/SLOW Modes to the SLEEP/IDLE Modes is executed via the HALT instruction. When an HALT instruction is executed, whether the devices enter the IDLE Mode or the SLEEP Mode is determined by the condition of the FHIDEN and FSIDEN bits in the SCC register.

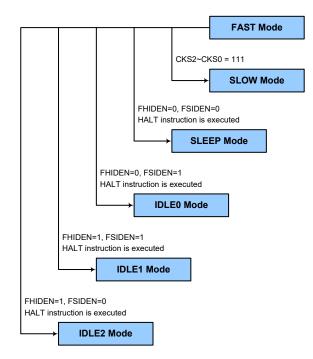


### FAST Mode to SLOW Mode Switching

When running in the FAST Mode, which uses the high speed system oscillator, and therefore consumes more power, the system clock can switch to run in the SLOW Mode by set the CKS2~CKS0 bits to "111" in the SCC register. This will then use the low speed system oscillator which will consume less power. Users may decide to do this for certain operations which do not require high performance and can subsequently reduce power consumption.

The SLOW Mode is sourced from the LIRC oscillator and therefore requires this oscillator to be stable before full mode switching occurs.

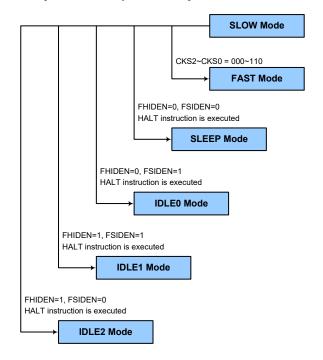




### SLOW Mode to FAST Mode Switching

In SLOW mode the system clock is derived from  $f_{SUB}$ . When system clock is switched back to the FAST mode from  $f_{SUB}$ , the CKS2~CKS0 bits should be set to "000"~"110" and then the system clock will respectively be switched to  $f_{\rm H} \sim f_{\rm H}/64$ .

However, if  $f_H$  is not used in SLOW mode and thus switched off, it will take some time to reoscillate and stabilise when switching to the FAST mode from the SLOW Mode. This is monitored using the HIRCF bit in the HIRCC register. The time duration required for the high speed system oscillator stabilization is specified in the System Start Up Time Characteristics.





### Entering the SLEEP Mode

There is only one way for the devices to enter the SLEEP Mode and that is to execute the "HALT" instruction in the application program with both the FHIDEN and FSIDEN bits in the SCC register equal to "0". In this mode all the clocks and functions will be switched off except the WDT function. When this instruction is executed under the conditions described above, the following will occur:

- The system clock will be stopped and the application program will stop at the "HALT" instruction.
- · The Data Memory contents and registers will maintain their present condition.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag, PDF, will be set and the Watchdog time-out flag, TO, will be cleared.
- The WDT will be cleared and resume counting as the WDT function is always enabled.

#### Entering the IDLE0 Mode

There is only one way for the devices to enter the IDLE0 Mode and that is to execute the "HALT" instruction in the application program with the FHIDEN bit in the SCC register equal to "0" and the FSIDEN bit in the SCC register equal to "1". When this instruction is executed under the conditions described above, the following will occur:

- The  $f_{\rm H}$  clock will be stopped and the application program will stop at the "HALT" instruction, but the  $f_{SUB}$  clock will be on.
- The Data Memory contents and registers will maintain their present condition.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag, PDF, will be set and the Watchdog time-out flag, TO, will be cleared.
- The WDT will be cleared and resume counting as the WDT function is always enabled.

### Entering the IDLE1 Mode

There is only one way for the devices to enter the IDLE1 Mode and that is to execute the "HALT" instruction in the application program with both the FHIDEN and FSIDEN bits in the SCC register equal to "1". When this instruction is executed under the conditions described above, the following will occur:

- The  $f_H$  and  $f_{SUB}$  clocks will be on but the application program will stop at the "HALT" instruction.
- The Data Memory contents and registers will maintain their present condition.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag, PDF, will be set and the Watchdog time-out flag, TO, will be cleared.
- The WDT will be cleared and resume counting as the WDT function is always enabled.

### Entering the IDLE2 Mode

There is only one way for the devices to enter the IDLE2 Mode and that is to execute the "HALT" instruction in the application program with the FHIDEN bit in the SCC register equal to "1" and the FSIDEN bit in the SCC register equal to "0". When this instruction is executed under the conditions described above, the following will occur:

- The  $f_{\rm H}$  clock will be on but the  $f_{SUB}$  clock will be off and the application program will stop at the "HALT" instruction.
- The Data Memory contents and registers will maintain their present condition.



- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag, PDF, will be set and the Watchdog time-out flag, TO, will be cleared.
- The WDT will be cleared and resume counting as the WDT function is always enabled.

### **Standby Current Considerations**

As the main reason for entering the SLEEP or IDLE Mode is to keep the current consumption of the devices to as low a value as possible, perhaps only in the order of several micro-amps in the IDLE0 and SLEEP Mode, there are other considerations which must also be taken into account by the circuit designer if the power consumption is to be minimised. Special attention must be made to the I/O pins on the devices. All high-impedance input pins must be connected to either a fixed high or low level as any floating input pins could create internal oscillations and result in increased current consumption. This also applies to the devices which has different package types, as there may be unbonded pins. These must either be setup as outputs or if setup as inputs must have pull-high resistors connected.

Care must also be taken with the loads, which are connected to I/O pins, which are setup as outputs. These should be placed in a condition in which minimum current is drawn or connected only to external circuits that do not draw current, such as other CMOS inputs.

In the IDLE1 and IDLE2 Mode the high speed oscillator is on, if the peripheral function clock source is derived from the high speed oscillator, the additional standby current will also be perhaps in the order of several hundred micro-amps.

### Wake-up

To minimise power consumption the devices can enter the SLEEP or any IDLE Mode, where the CPU will be switched off. However, when the devices are woken up again, they will take a considerable time for the original system oscillator to restart, stabilise and allow normal operation to resume.

After the system enters the SLEEP or IDLE Mode, it can be woken up from one of various sources listed as follows:

- An external falling edge on Port A
- · A system interrupt
- A WDT overflow

When the devices execute the "HALT" instruction, it will enter the Power down mode and the PDF flag will be set high. The PDF flag is cleared to 0 if the devices experience a system power-up or execut the clear Watchdog Timer instruction.

If the system is woken up by a WDT overflow, a Watchdog Timer Time-out reset will be initiated and the TO flag will be set to 1. The TO flag is set high if a WDT time-out occurs, and causes a wake-up that only resets the Program Counter and Stack Pointer, the other flags remain in their original status.

Each pin on Port A can be setup using the PAWU register to permit a negative transition on the pin to wake-up the system. When a pin wake-up occurs, the program will resume execution at the instruction following the "HALT" instruction.

If the system is woken up by an interrupt, then two possible situations may occur. The first is where the related interrupt is disabled or the interrupt is enabled but the stack is full, in which case the program will resume execution at the instruction following the "HALT" instruction. In this situation, the interrupt which woke-up the devices will not be immediately serviced, but will rather be serviced



later when the related interrupt is finally enabled or when a stack level becomes free. The other situation is where the related interrupt is enabled and the stack is not full, in which case the regular interrupt response takes place. If an interrupt request flag is set high before entering the SLEEP or IDLE Mode, the wake-up function of the related interrupt will be disabled.

# Watchdog Timer

The Watchdog Timer is provided to prevent program malfunctions or sequences from jumping to unknown locations, due to certain uncontrollable external events such as electrical noise.

### Watchdog Timer Clock Source

The Watchdog Timer clock source is provided by the internal clock,  $f_{LIRC}$  which is sourced from the LIRC oscillator. The LIRC internal oscillator has an approximate frequency of 32kHz and this specified internal clock period can vary with  $V_{DD}$ , temperature and process variations. The Watchdog Timer source clock is then subdivided by a ratio of  $2^8$  to  $2^{18}$  to give longer timeouts, the actual value being chosen using the WS2~WS0 bits in the WDTC register.

### Watchdog Timer Control Register

A single register, WDTC, controls the required timeout period as well as the enable and reset MCU operation.

### WDTC Register

Bit	7	6	5	4	3	2	1	0
Name	WE4	WE3	WE2	WE1	WE0	WS2	WS1	WS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	1	0	1	0	0	1	1

#### Bit 7~3 WE4~WE0: WDT function software control

01010/10101: Enable

Others: Reset MCU

When these bits are changed by the environmental noise or software setting to reset the microcontroller, the reset operation will be activated after a delay time,  $t_{SRESET}$  and the WRF bit in the RSTFC register will be set high.

Bit 2~0 WS2~WS0: WDT time-out period selection

 $\begin{array}{l} 000:\ 2^8/f_{LIRC}\\ 001:\ 2^{10}/f_{LIRC}\\ 010:\ 2^{12}/f_{LIRC}\\ 011:\ 2^{14}/f_{LIRC}\\ 100:\ 2^{15}/f_{LIRC}\\ 101:\ 2^{16}/f_{LIRC}\\ 110:\ 2^{17}/f_{LIRC}\\ 111:\ 2^{18}/f_{LIRC} \end{array}$ 

These three bits determine the division ratio of the watchdog timer source clock, which in turn determines the time-out period.



### RSTFC Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	LVRF	—	WRF
R/W	—	—	—	—	—	R/W	—	R/W
POR	—	—	—	—	_	х	—	0
"x": unknown								
Bit 7~3	it 7~3 Unimplemented, read as "0"							

Bit 2 LVRF: LVR function reset flag Described elsewhere

- Bit 1 Unimplemented, read as "0"

WRF: WDTC register software reset flag Bit 0 0: Not occurred

1: Occurred

This bit is set to 1 by the WDTC register software reset and cleared by the application program. Note that this bit can be cleared to 0 only by the application program.

### Watchdog Timer Operation

The Watchdog Timer operates by providing a device reset when its timer overflows. This means that in the application program and during normal operation the user has to strategically clear the Watchdog Timer before it overflows to prevent the Watchdog Timer from executing a reset. This is done using the clear watchdog instruction. If the program malfunctions for whatever reason, jumps to an unknown location, or enters an endless loop, the clear instruction will not be executed in the correct manner, in which case the Watchdog Timer will overflow and reset the devices. There are five bits, WE4~WE0, in the WDTC register to offer the enable control and reset control of the Watchdog Timer. The WDT function will be enabled if the WE4~WE0 bits are equal to 01010B or 10101B. If the WE4~WE0 bits are set to any other values, other than 01010B and 10101B, it will reset the devices after a delay time, t<sub>SRESET</sub>. After power on these bits will have a value of 01010B.

WE4~WE0 Bits	WDT Function
01010B or 10101B	Enable
Any other values	Reset MCU

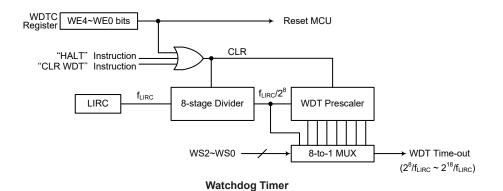
Watchdog Timer Enable/Reset Control

Under normal program operation, a Watchdog Timer time-out will initialise a device reset and set the status bit TO high. However, if the system is in the SLEEP or IDLE Mode, when a Watchdog Timer time-out occurs, the TO bit in the status register will be set high and only the Program Counter and Stack Pointer will be reset. Three methods can be adopted to clear the contents of the Watchdog Timer. The first is a WDTC software reset, which means a certain value except 01010B and 10101B written into the WE4~WE0 bits, the second is using the Watchdog Timer software clear instruction and the third is via a HALT instruction.

There is only one method of using software instruction to clear the Watchdog Timer. That is to use the single "CLR WDT" instruction to clear the WDT.

The maximum time out period is when the  $2^{18}$  division ratio is selected. As an example, with a 32kHz LIRC oscillator as its source clock, this will give a maximum watchdog period of around 8 seconds for the 2<sup>18</sup> division ratio, and a minimum timeout of 8ms for the 2<sup>8</sup> division ration.





# **Reset and Initialisation**

A reset function is a fundamental part of any microcontroller ensuring that the devices can be set to some predetermined condition irrespective of outside parameters. The most important reset condition is after power is first applied to the microcontroller. In this case, internal circuitry will ensure that the microcontroller, after a short delay, will be in a well-defined state and ready to execute the first program instruction. After this power-on reset, certain important internal registers will be set to defined states before the program commences. One of these registers is the Program Counter, which will be reset to zero forcing the microcontroller to begin program execution from the lowest Program Memory address.

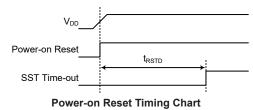
In addition to the power-on reset, another reset exists in the form of a Low Voltage Reset, LVR, where a full reset is implemented in situations where the power supply voltage falls below a certain threshold. Another type of reset is when the Watchdog Timer overflows and resets the microcontroller. All types of reset operations result in different register conditions being setup.

### **Reset Functions**

There are several ways in which a microcontroller reset can occur, through events occurring internally:

### **Power-on Reset**

The most fundamental and unavoidable reset is the one that occurs after power is first applied to the microcontroller. As well as ensuring that the Program Memory begins execution from the first memory address, a power-on reset also ensures that certain other registers are preset to known conditions. All the I/O port and port control registers will power up in a high condition ensuring that all pins will be first set to inputs.



### Low Voltage Reset – LVR

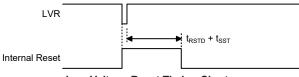
The microcontrollers contain a low voltage reset circuit in order to monitor the supply voltage of the devices and provide an MCU reset when the value falls below a certain predefined level.

The LVR function is always enabled in normal operation with a specific LVR voltage  $V_{LVR}$ . For the devices the  $V_{LVR}$  value is fixed at 2.1V. If the supply voltage of the devices drop to within a range of



 $0.9V \sim V_{LVR}$  such as might occur when changing the battery in battery powered applications, the LVR will automatically reset the devices internally and the LVRF bit in the RSTFC register will also be set to 1. For a valid LVR signal, a low supply voltage, i.e., a voltage in the range between  $0.9V \sim V_{LVR}$  must exist for a time greater than that specified by  $t_{LVR}$  in the LVD & LVR Electrical Characteristics. If the low supply voltage state does not exceed this value, the LVR will ignore the low supply voltage and will not perform a reset function.

Note that the LVR function will be automatically disabled when the devices enter the SLEEP or IDLE mode.



Low Voltage Reset Timing Chart

### RSTFC Register

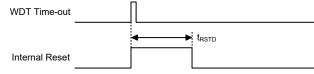
Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	LVRF	—	WRF
R/W	_	_	—	—	—	R/W	—	R/W
POR	—	—	—	—	_	х	_	0

"x": unknown

Bit 7~3	Unimplemented, read as "0"
Bit 2	LVRF: LVR function reset flag
	0: Not occur
	1: Occurred
	This bit is set to 1 when an actual Low Voltage Reset situation condition occurs. This
	bit can be cleared to 0 only by the application program.
Bit 1	Unimplemented, read as "0"
Bit 0	WRF: WDTC register software reset flag
	0: Not occurred
	1: Occurred
	This bit is set to 1 by the WDTC register software reset and cleared by the application
	program. Note that this bit can be cleared to 0 only by the application program.

### Watchdog Time-out Reset during Normal Operation

The Watchdog time-out Reset during normal operation in the FAST or SLOW mode is the same as a LVR reset except that the Watchdog time-out flag TO will be set to "1" and the LVRF flag bit will be unchanged.

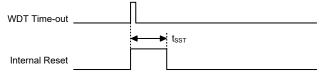


WDT Time-out Reset during Normal Operation Timing Chart

### Watchdog Time-out Reset during SLEEP or IDLE Mode

The Watchdog time-out Reset during SLEEP or IDLE Mode is a little different from other kinds of reset. Most of the conditions remain unchanged except that the Program Counter and the Stack Pointer will be cleared to "0" and the TO flag will be set to "1". Refer to the System Start Up Time Characteristics for  $t_{SST}$  details.





WDT Time-out Reset during SLEEP or IDLE Mode Timing Chart

### **Reset Initial Conditions**

The different types of reset described affect the reset flags in different ways. These flags, known as PDF and TO are located in the status register and are controlled by various microcontroller operations, such as the SLEEP or IDLE Mode function or Watchdog Timer. The reset flags are shown in the table:

то	PDF	RESET Conditions		
0	0	Power-on reset		
u	u	LVR reset during Normal operation		
1	u	WDT time-out reset during Normal operation		
1	1	WDT time-out reset during IDLE or SLEEP Mode		

"u" stands for unchanged

The following table indicates the way in which the various components of the microcontroller are affected after a power-on reset occurs.

Item	Condition After RESET
Program Counter	Reset to zero
Interrupts	All interrupts will be disabled
WDT, Time Bases	Clear after reset, WDT begins counting
Timer Modules	Timer Modules will be turned off
Input/Output Ports	I/O ports will be setup as inputs
Stack Pointer	Stack Pointer will point to the top of the stack

The different kinds of resets all affect the internal registers of the microcontroller in different ways. To ensure reliable continuation of normal program execution after a reset occurs, it is important to know what condition the microcontroller is in after a particular reset occurs. The following table describes how each type of reset affects each of the microcontroller internal registers. Note that where more than one package type exists the table will reflect the situation for the larger package type.

Register	Power On Reset	LVR Reset	WDT Time-out (Normal Operation)	WDT Time-out (IDLE/SLEEP)
IAR0	0000 0000	0000 0000	0000 0000	uuuu uuuu
MP0	0000 0000	0000 0000	0000 0000	uuuu uuuu
IAR1	0000 0000	0000 0000	0000 0000	uuuu uuuu
MP1L	0000 0000	0000 0000	0000 0000	uuuu uuuu
MP1H	0000 0000	0000 0000	0000 0000	uuuu uuuu
ACC	XXXX XXXX	uuuu uuuu	uuuu uuuu	uuuu uuuu
PCL	0000 0000	0000 0000	0000 0000	0000 0000
TBLP	XXXX XXXX	uuuu uuuu	uuuu uuuu	uuuu uuuu
TBLH	XXXX XXXX	uuuu uuuu	uuuu uuuu	uuuu uuuu
ТВНР	x x x x	uuuu	uuuu	uuuu
STATUS	xx00 xxxx	xxuu uuuu	xx1u uuuu	uu11 uuuu
VBGRC	0	0	0	u
IAR2	0000 0000	0000 0000	0000 0000	uuuu uuuu



Register	Power On Reset	LVR Reset	WDT Time-out (Normal Operation)	WDT Time-out (IDLE/SLEEP)
MP2L	0000 0000	0000 0000	0000 0000	uuuu uuuu
MP2H	0000 0000	0000 0000	0000 0000	uuuu uuuu
RSTFC	x - 0	1 - u	u - u	u - u
TB0C	0000	0000	0000	uuuu
TB1C	0000	0000	0000	uuuu
SCC	00000	00000	00000	uuuuu
HIRCC	0001	0001	0001	uuuu
PA	1111 1111	1111 1111	1111 1111	uuuu uuuu
PAC	1111 1111	1111 1111	1111 1111	uuuu uuuu
PAPU	0000 0000	0000 0000	0000 0000	uuuu uuuu
PAWU	0000 0000	0000 0000	0000 0000	uuuu uuuu
PB	1 1111	1 1111	1 1111	u uuuu
PBC	1 1111	1 1111	1 1111	u uuuu
PBPU	0 0000	0 0000	0 0000	u uuuu
SLEDC	0000 0000	0000 0000	0000 0000	<u>uuuu uuuu</u>
PSCR	0 0	0 0	00	u u
LVDC	00 0000	00 0000	00 0000	
SDSW	-000 0000	-000 0000	-000 0000	-uuu uuuu
SDPGAC0	00 0000	00 0000	00 0000	
SDPGAC1	0000 0000	0000 0000	0000 0000	
SDA0C	-0000	-0000	-0000	- uu uu
SDA0VOS	0010 0000	0010 0000	0010 0000	
SDA1C	-0000	-0000	-0000	- uu uu
SDA1VOS	0010 0000	0010 0000	0010 0000	
STMC0	0000 0000	0000 0000	0000 0000	
STMC1	0000 0000	0000 0000	0000 0000	
STMDL	0000 0000	0000 0000	0000 0000	
STMDE	0 0	0 0	0 0	u u
STMAL	0000 0000	0000 0000	0000 0000	
STMAL	0 0	0 0	00	u u
SADOL				uuuu (ADRFS=0)
SADOL	X X X X	XXXX	X X X X	uuuu uuuu (ADRFS=1)
SADOH	xxxx xxxx	xxxx xxxx	xxxx xxxx	uuuu uuuu (ADRFS=0)
				(ADRFS=1)
SADC0	0000 0000	0000 0000	0000 0000	<u>uuuu uuuu</u>
SADC1	0000 0000	0000 0000	0000 0000	<u>uuuu uuuu</u>
WDTC	0101 0011	0101 0011	0101 0011	uuuu uuuu
EEA	00 0000	00 0000	00 0000	uu uuuu
EED	0000 0000	0000 0000	0000 0000	uuuu uuuu
EEC	0000	0000	0000	uuuu
SIMC0	1110 0000	1110 0000	1110 0000	uuuu uuuu
SIMC1 (UMD=0)	1000 0001	1000 0001	1000 0001	uuuu uuuu
UUCR1* (UMD=1)	0000 00x0	0000 00x0	0000 00x0	uuuu uuuu



Register	Power On Reset	LVR Reset	WDT Time-out (Normal Operation)	WDT Time-out (IDLE/SLEEP)
SIMD/UTXR_RXR	XXXX XXXX	XXXX XXXX	x x x x x x x x x x	uuuu uuuu
SIMA/SIMC2/UUCR2	0000 0000	0000 0000	0000 0000	uuuu uuuu
SIMTOC (UMD=0)	0000 0000	0000 0000	0000 0000	uuuu uuuu
UBRG* (UMD=1)	XXXX XXXX	XXXX XXXX	XXXX XXXX	uuuu uuuu
UUSR	0000 1011	0000 1011	0000 1011	uuuu uuuu
INTEG	0000	0000	0000	uuuu
INTC0	-000 0000	-000 0000	-000 0000	-uuu uuuu
INTC1	0000 0000	0000 0000	0000 0000	uuuu uuuu
INTC2	0000 0000	0000 0000	0000 0000	uuuu uuuu
INTC3	-000 -000	-000 -000	-000 -000	-uuu -uuu
PAS0	0000 0000	0000 0000	0000 0000	uuuu uuuu
PAS1	0000 0000	0000 0000	0000 0000	uuuu uuuu
PBS0	0000 0000	0000 0000	0000 0000	uuuu uuuu
PBS1	00	00	00	u u
IFS0	0000 0000	0000 0000	0000 0000	uuuu uuuu
PTMC0	0000 0	0000 0	0000 0	uuuu u
PTMC1	0000 0000	0000 0000	0000 0000	uuuu uuuu
PTMC2	000	000	000	u u u
PTMDL	0000 0000	0000 0000	0000 0000	uuuu uuuu
PTMDH	00	00	00	u u
PTMAL	0000 0000	0000 0000	0000 0000	uuuu uuuu
PTMAH	00	00	00	u u
PTMBL	0000 0000	0000 0000	0000 0000	uuuu uuuu
РТМВН	00	00	00	u u
PTMRPL	0000 0000	0000 0000	0000 0000	uuuu uuuu
PTMRPH	00	00	00	u u
ISGENC	0 0 0	0 0 0	0 0 0	uuu
ISGDATA0	0 0000	0 0000	0 0000	u uuuu
ISGDATA1	0 0000	0 0000	0 0000	u uuuu

Note: "u" stands for unchanged

"x" stands for unknown

"-" stands for unimplemented

"\*": The UUCR1 and SIMC1 registers share the same memory address while the UBRG and SIMTOC registers share the same memory address. The default value of the UUCR1 or UBRG register can be obtained when the UMD bit is set high by application program after a reset.



# Input/Output Ports

Holtek microcontrollers offer considerable flexibility on their I/O ports. With the input or output designation of every pin fully under user program control, pull-high selections for all ports and wake-up selections on certain pins, the user is provided with an I/O structure to meet the needs of a wide range of application possibilities.

The devices provide bidirectional input/output lines labeled with port names PA~PB. These I/O ports are mapped to the RAM Data Memory with specific addresses as shown in the Special Purpose Data Memory table. All of these I/O ports can be used for input and output operations. For input operation, these ports are non-latching, which means the inputs must be ready at the T2 rising edge of instruction "MOV A, [m]", where m denotes the port address. For output operation, all the data is latched and remains unchanged until the output latch is rewritten.

Register		Bit											
Name	7	6	5	4	3	2	1	0					
PA	PA7	PA6	PA5	PA4	PA3	PA2	PA1	PA0					
PAC	PAC7	PAC6	PAC5	PAC4	PAC3	PAC2	PAC1	PAC0					
PAPU	PAPU7	PAPU6	PAPU5	PAPU4	PAPU3	PAPU2	PAPU1	PAPU0					
PAWU	PAWU7	PAWU6	PAWU5	PAWU4	PAWU3	PAWU2	PAWU1	PAWU0					
PB	_	—	_	PB4	PB3	PB2	PB1	PB0					
PBC	_	—		PBC4	PBC3	PBC2	PBC1	PBC0					
PBPU	_	—		PBPU4	PBPU3	PBPU2	PBPU1	PBPU0					

I/O Logic Function Register List

Note: The unbonded lines, PA7, PB0~PB4 in the BA45F5740-2 device, PA5, PA7, PB0~PB4 in the BA45F5740 16-pin NSOP package and PA5, PB0~PB2 in the BA45F5740 20-pin SOP package, are also controlled by the corresponding bits in the registers. They should be properly configured to avoid the current consumption resulting from an input floating condition. Refer to the "Standby Current Considerations" section.

### Pull-high Resistors

Many product applications require pull-high resistors for their switch inputs usually requiring the use of an external resistor. To eliminate the need for these external resistors, all I/O pins, when configured as an input have the capability of being connected to an internal pull-high resistor. These pull-high resistors are selected using registers PAPU~PBPU, and are implemented using weak PMOS transistors.

Note that the pull-high resistor can be controlled by the relevant pull-high control register only when the pin-shared functional pin is selected as a digital input or NMOS output. Otherwise, the pull-high resistors cannot be enabled.

### PxPU Register

Bit	7	6	5	4	3	2	1	0
Name	PxPU7	PxPU6	PxPU5	PxPU4	PxPU3	PxPU2	PxPU1	PxPU0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

**PxPUn**: I/O Port x Pin pull-high function control

- 0: Disable
- 1: Enable

The PxPUn bit is used to control the pin pull-high function. Here the "x" can be A or B. However, the actual available bits for each I/O Port may be different.



### Port A Wake-up

The HALT instruction forces the microcontroller into the SLEEP or IDLE Mode which preserves power, a feature that is important for battery and other low-power applications. Various methods exist to wake-up the microcontroller, one of which is to change the logic condition on one of the Port A pins from high to low. This function is especially suitable for applications that can be woken up via external switches. Each pin on Port A can be selected individually to have this wake-up feature using the PAWU register.

Note that the wake-up function can be controlled by the wake-up control registers only when the pin is selected as a general purpose input and the MCU enters the IDLE or SLEEP mode.

PAWU Register

Bit	7	6	5	4	3	2	1	0
Name	PAWU7	PAWU6	PAWU5	PAWU4	PAWU3	PAWU2	PAWU1	PAWU0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **PAWU7~PAWU0**: PA7~PA0 wake-up function control 0: Disable

1: Enable

### I/O Port Control Registers

Each I/O port has its own control register known as PAC~PBC, to control the input/output configuration. With this control register, each CMOS output or input can be reconfigured dynamically under software control. Each pin of the I/O ports is directly mapped to a bit in its associated port control register. For the I/O pin to function as an input, the corresponding bit of the control register must be written as a "1". This will then allow the logic state of the input pin to be directly read by instructions. When the corresponding bit of the control register is written as a "0", the I/O pin will be setup as a CMOS output. If the pin is currently setup as an output, instructions can still be used to read the output register. However, it should be noted that the program will in fact only read the status of the output data latch and not the actual logic status of the output pin.

#### PxC Register

Bit	7	6	5	4	3	2	1	0
Name	PxC7	PxC6	PxC5	PxC4	PxC3	PxC2	PxC1	PxC0
R/W								
POR	1	1	1	1	1	1	1	1

**PxCn**: I/O Port x Pin type selection

0: Output

1: Input

The PxCn bit is used to control the pin type selection. Here the "x" can be A or B. However, the actual available bits for each I/O Port may be different.

Note that for part of package types, the REST and DIO pins are pin-shared with I/O pins respectively which should be properly configured by using the corresponding I/O port control registers and pin-shared function selection registers to implement calendar control, refer to the Pin-shared Functions section for more details.

#### Source Current Selection

Each pin in these devices can be configured with different output source current which is selected by the corresponding source current selection bits. These source current selection bits are available when the corresponding pin is configured as a CMOS output. Otherwise, these select bits have no



effect. Users should refer to the Input/Output Characteristics section to obtain the exact value for different applications.

### SLEDC Register

	cgister											
Bit	7	6	5	4	3	2	1	0				
Name	SLEDC7	SLEDC6	SLEDC5	SLEDC4	SLEDC3	SLEDC2	SLEDC1	SLEDC0				
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W				
POR	0	0	0	0	0	0	0	0				
Bit 7~6	SLEDC	7~SLEDC	6: PB4 sour	ce current	selection							
210, 0		urce curren										
		urce curren										
	10: So	urce curren	t=Level 2									
	11: Source current=Level 3 (Max.) For the BA45F5740-2 device, these two bits are reserved and should be fixed at "00".											
	For the I	BA45F5740	-2 device,	these two b	its are reser	ved and sh	ould be fixe	ed at "00".				
Bit 5~4 SLEDC5~SLEDC4: PB3~PB0 source current selection												
	00: Source current=Level 0 (Min.)											
	01: Source current=Level 1											
	10: Source current=Level 2 11: Source current=Level 3 (Max.)											
				,				1				
	For the I	3A45F574(	-2 device,	these two b	its are reser	rved and sh	ould be fixe	ed at "00".				
Bit 3~2					rrent select	ion						
		urce curren	(	Min.)								
	011.00	urce curren										
		urce curren										
		urce curren					6					
	For the I	3A45F5740	)-2 device,	these two b	its are avail	lable for PA	.6~PA4.					
Bit 1~0					rrent selecti	ion						
		urce curren	(	Min.)								
		urce curren										
	10: So	urce curren	t=Level 2									

- 11: Source current=Level 3 (Max.)
- Note: Users should refer to the Input/Output Characteristics section to obtain the exact value for different applications

### Sink Current Generator

The sink current source generator could provide constant current no matter what  $V_{ISINK}$  voltage is from 1.0V~4.5V. The constant current value is controlled by the ISGDATA0/ISGDATA1 register, and the sink current range is 50mA ~ 360mA.

#### ISGENC Register

Bit	7	6	5	4	3	2	1	0
Name	ISGEN	—	—	—		—	ISGS1	ISGS0
R/W	R/W	—	—	—	—	—	R/W	R/W
POR	0	—	—	—	—	—	0	0

Bit 7 ISGEN: Sink current generator enable control

- 0: Disable
- 1: Enable

When the ISGEN bit is cleared to zero to disable the sink current generator, the ISINK0&ISINK1 pin status is:  $V_{ISINK0}$ &V<sub>ISINK1</sub>=floating, I<sub>SINK0</sub>&I<sub>SINK1</sub>=0.

Bit 6~2 Unimplemented, read as "0"



Bit 1	ISGS1: Enable/Disable ISINK1 pin sink current control
	0: Disable
	1: Enable

Bit 0 ISGS0: Enable/Disable ISINK0 pin sink current control 0: Disable

1: Enable

#### ISGDATA0 Register

Bit	7	6	5	4	3	2	1	0
Name	_	_		D4	D3	D2	D1	D0
R/W	—	_	—	R/W	R/W	R/W	R/W	R/W
POR	—	_	—	0	0	0	0	0

Bit 7~5 Unimplemented, read as "0"

Bit 4~0 **D4~D0**: Sink current generator control for ISINK0 pin

Current value (mA) =  $50 + 10 \times (ISGDATA0[4:0])$ 

Refer to "Sink Current Generator Electrical Characteristics" table for more details.

### ISGDATA1 Register

Bit	7	6	5	4	3	2	1	0
Name	_	_	_	D4	D3	D2	D1	D0
R/W	_	—	—	R/W	R/W	R/W	R/W	R/W
POR	_	_	_	0	0	0	0	0

Bit 7~5 Unimplemented, read as "0"

Bit 4~0 **D4~D0**: Sink current generator control for ISINK1 pin

Current value (mA) =  $50 + 5 \times (ISGDATA1[4:0])$ 

Refer to "Sink Current Generator Electrical Characteristics" table for more details.

### **Pin-shared Functions**

The flexibility of the microcontroller range is greatly enhanced by the use of pins that have more than one function. Limited numbers of pins can force serious design constraints on designers but by supplying pins with multi-functions, many of these difficulties can be overcome. For these pins, the desired function of the multi-function I/O pins is selected by a series of registers via the application program control.

### **Pin-shared Function Selection Registers**

The limited number of supplied pins in a package can impose restrictions on the amount of functions a certain device can contain. However by allowing the same pins to share several different functions and providing a means of function selection, a wide range of different functions can be incorporated into even relatively small package sizes. The devices include Port "x" Output Function Selection register "n", labeled as PxSn, and Input Function Selection register, labeled as IFS0, which can select the desired functions of the multi-function pin-shared pins.

The most important point to note is to make sure that the desired pin-shared function is properly selected and also deselected. For most pin-shared functions, to select the desired pin-shared function, the pin-shared function should first be correctly selected using the corresponding pin-shared control register. After that the corresponding peripheral functional setting should be configured and then the peripheral function can be enabled.

However, special point must be noted for some digital input pins, such as INTn, xTCK, xTPI, etc, which share the same pin-shared control configuration with their corresponding general purpose I/O functions when setting the revelant pin-shared control bit fields. To select these pin functions,



in addition to the necessary pin-shared control and peripheral functional setup forementioned, they must also be setup as input by setting the corresponding bit in the I/O port control register.

To correctly deselect the pin-shared function, the peripheral function should first be disabled and then the corresponding pin-shared function control register can be modified to select other pinshared functions.

Register		Bit											
Name	7	6	5	4	3	2	1	0					
PAS0	PAS07	PAS06	PAS05	PAS04	PAS03	PAS02	PAS01	PAS00					
PAS1	PAS17	PAS16	PAS15	PAS14	PAS13	PAS12	PAS11	PAS10					
PBS0	PBS07	PBS06	PBS05	PBS04	PBS03	PBS02	PBS01	PBS00					
PBS1	—	—	—	—	—	—	PBS11	PBS10					
IFS0	IFS07	IFS06	IFS05	IFS04	IFS03	IFS02	IFS01	IFS00					

**Pin-shared Function Selection Register List** 

#### PAS0 Register

Bit	7	6	5	4	3	2	1	0
Name	PAS07	PAS06	PAS05	PAS04	PAS03	PAS02	PAS01	PAS00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

00:	PA3
0.1	CDO/TT

- 01: SDO/TX 10: PTPB
- 11: AN3

Bit 5~4 PAS05~PAS04: PA2 Pin-Shared function selection

For BA45F5740/BA45F5740-2 16-pin NSOP:

- 00: PA2
- 01: Reserved
- 10~11: PA2
- For other packages:
  - 00: PA2
  - 01: SDI/SDA/RX
  - 10: PA2
  - 11: PA2
- Note: For the BA45F5740/BA45F5740-2 16-pin NSOP package type, the DIO pin is pin-shared with PA2 and should be controlled by PA2 for calendar normal operation, therefore the PAS05~PAS04 bits must be fixed at "00" or "10~11". In addition the PAC2 bit in the PAC register should be properly configured according to the Timing Diagrams in the Calendar section.

Bit 3~2 PAS03~PAS02: PA1 Pin-Shared function selection

- For BA45F5740/BA45F5740-2 16-pin NSOP:
  - 00: PA1
  - 01~10: Reserved
  - 11: A1PI
- For other packages:
- 00: PA1/INT1
- 01:  $\overline{\text{SCS}}$
- 10: AN2
- 11: A1PI
- Note: 1. For the BA45F5740 16-pin NSOP package type, the REST pin is pin-shared with PA1 and should be controlled by PA1 for calendar normal operation,

therefore the PAS03~PAS02 bits are recommended to be fixed at "00". In addition the PA1 should be setup as an output by clearing the PAC1 bit in the PAC register to zero to send control commands according to the Timing Diagrams in the Calendar section. If the A1PI pin function is selected, it may affect the calendar function, only for OPA waveform observation during development.

2. For the BA45F5740-2 16-pin NSOP package type in which the PA1 and PA5 share one external pin, it is important to note that PA5 should be configured as an input with the pull-high function disabled when using these bits to select the pin function on PA1 pin. In this case, the REST pin is pin-shared with PA1 and should be controlled by PA1 for calendar normal operation, therefore the PAS03~PAS02 bits are recommended to be fixed at "00". In addition the PA1 should be setup as an output by clearing the PAC1 bit in the PAC register to zero to send control commands according to the Timing Diagrams in the Calendar section. If the A1PI pin function is selected, it may affect the calendar function, only for OPA waveform observation during development.

PAS01~PAS00: PA0 Pin-Shared function selection

- 00: PA0 01: SCL/SCK 10: PA0
- 11: PA0

### PAS1 Register

Bit 1~0

Bit	7	6	5	4	3	2	1	0
Name	PAS17	PAS16	PAS15	PAS14	PAS13	PAS12	PAS11	PAS10
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0
Dit 7 6 DAS17 DAS16, DA7 Din Shared function coloring								

Bit 7~6 PAS17~PAS16: PA7 Pin-Shared function selection ΡI

- 01: SCK/SCL
- 10: AN1
- 11: PA7/STPI/PTPI
- Note: For the BA45F5740-2 device, these two bits are reserved and should be fixed at "00"
- Bit 5~4 PAS15~PAS14: PA6 Pin-Shared function selection
  - 00: PA6
  - 01: PTP
  - 10: SDI/SDA/RX
  - 11: VREF

Bit 3~2 PAS13~PAS12: PA5 Pin-Shared function selection

- For BA45F5740-2 16-pin NSOP:
  - 00: PA5
  - 01: A1O
  - 10~11: PA5
- For BA45F5740 24-pin SOP:
  - 00: PA5/STCK
  - 01: A1O
  - 10: PA5/STCK
  - 11: PA5/STCK
- Note: For the BA45F5740-2 16-pin NSOP package type in which the PA1 and PA5 share one external pin, it is important to note that PA1 should be configured as an input with the pull-high function disabled when using these bits to select the pin function on PA5 pin. In this case, the REST pin is pin-shared with PA5 and should be controlled by PA5 for calendar normal operation, therefore the PAS13~PAS12 bits are recommended to be fixed at "00" or "10~11". In addition



the PA5 should be setup as an output by clearing the PAC5 bit in the PAC register to zero to send control commands according to the Timing Diagrams in the Calendar section. If the A1O pin function is selected, it may affect the calendar function, only for OPA waveform observation during development.

Bit 1~0 PAS11~PAS10: PA4 Pin-Shared function selection

00: PA4/PTCK 01: STPB 10: AN0 11: A00

#### PBS0 Register

Bit	7	6	5	4	3	2	1	0
Name	PBS07	PBS06	PBS05	PBS04	PBS03	PBS02	PBS01	PBS00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PBS07~PBS06**: PB3 Pin-Shared function selection

For BA45F5740 20-pin SOP:

- 00: PB3
- 01~10: Reserved
- 11: PB3
- For BA45F5740 24-pin SOP:
- 00: PB3
- 01: Reserved
- 10: SDI/SDA/RX
- 11: PB3
- Note: 1. For the BA45F5740 20-pin SOP package type, the REST pin is pin-shared with PB3 and should be controlled by PB3 for calendar normal operation, therefore the PBS07~PBS06 bits must be fixed at "00" or "11". In addition the PB3 should be setup as an output by clearing the PBC3 bit in the PBC register to zero to send control commands according to the Timing Diagrams in the Calendar section.
  - 2. For the BA45F5740-2 device, these two bits are reserved and should be fixed at "00".

#### Bit 5~4 PBS05~PBS04: PB2 Pin-Shared function selection

- 00: PB2
- 01: Reserved
- 10: SCK/SCL
- 11: PB2

Note: For the BA45F5740-2 device, these two bits are reserved and should be fixed at "00".

Bit 3~2

### -2 **PBS03~PBS02**: PB1 Pin-Shared function selection

- 00: PB1
- 01: Reserved
- 10: Reserved
- 11: PB1
- Note: 1. For the BA45F5740 24-pin SOP package type, the REST pin is pin-shared with PB1 and should be controlled by PB1 for calendar normal operation, therefore the PBS03~PBS02 bits must be fixed at "00" or "11". In addition the PB1 should be setup as an output by clearing the PBC1 bit in the PBC register to zero to send control commands according to the Timing Diagrams in the Calendar section.
  - 2. For the BA45F5740-2 device, these two bits are reserved and should be fixed at "00".



Bit 1~0 **PBS01~PBS00**: PB0 Pin-Shared function selection

00:	PB0/INT0

- $01: \overline{\text{SCS}}$
- 10: STP
- 11: A0PB

Note: For the BA45F5740-2 device, these two bits are reserved and should be fixed at "00".

#### PBS1 Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	PBS11	PBS10
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	_	—	—	_	—	0	0

Bit 7~2 Unimplemented, read as "0"

Bit 1~0 PBS11~PBS10: PB4 Pin-Shared function selection

00:	PB4
01:	SCS
10	<b>DD</b> 4

- 10: PB4 11: PB4
- 11. I D4

Note: For the BA45F5740-2 device, these two bits are reserved and should be fixed at "00".

### IFS0 Register

Bit	7	6	5	4	3	2	1	0
Name	IFS07	IFS06	IFS05	IFS04	IFS03	IFS02	IFS01	IFS00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 IFS07~IFS06: PTPI input source selection

- 00: Reserved
- 01: PTPI on PA7
- 10: Reserved
- 11: Reserved

Note: 1. If the PTPI input signal is used, these bits should be fixed at "01".

2. For the BA45F5740-2 device, these two bits are reserved and should be fixed at "00".

Bit 5~4 IFS05~IFS04: SCS input source pin selection

- 00:  $\overline{\text{SCS}}$  on PB0
- 01:  $\overline{\text{SCS}}$  on PB4
- 10:  $\overline{\text{SCS}}$  on PA1
- 11: SCS on PB0

Note: For the BA45F5740-2 device, these two bits are reserved and should be fixed at "00".

Bit 3~2 IFS03~IFS02: SCK/SCL input source pin selection

- 00: SCK/SCL on PA0
- 01: SCK/SCL on PB2
- 10: SCK/SCL on PA7
- 11: SCK/SCL on PB2
- Note: 1. If the SPI Master mode is selected, when the SIMEN bit is set high, the PA0 and PB2 pins both can be used as the SCK pin function ignoring the IFS0[3:2] bit settings.
  - 2. For the BA45F5740-2 device, the PA7 and PB2 pins are unavailable, so these bits should be fixed at "00".

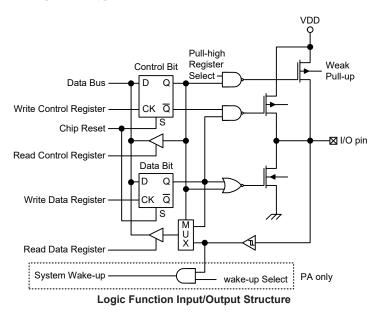


Bit 1~0 IFS01~IFS00: SDI/SDA/RX input source pin selection

- 00: SDI/SDA/RX on PA2 01: SDI/SDA/RX on PA6 10: SDI/SDA/RX on PB3
  - 11: SDI/SDA/RX on PA2
- Note: 1. For part of BA45F5740 package types, the PA2 or PB3 pin is not pin-shared with SDI/SDA/RX, so these bits should be properly configured.
  - 2. For the BA45F5740-2 device, if the USIM function is used, these bits should be fixed at "01".

### I/O Pin Structures

The accompanying diagram illustrates the internal structures of the I/O logic function. As the exact logical construction of the I/O pin will differ from this diagram, it is supplied as a guide only to assist with the functional understanding of the logic function I/O pins. The wide range of pin-shared structures does not permit all types to be shown.



### **Programming Considerations**

Within the user program, one of the first things to consider is port initialisation. After a reset, all of the I/O data and port control registers will be set high. This means that all I/O pins will default to an input state, the level of which depends on the other connected circuitry and whether pull-high selections have been chosen. If the port control registers are then programmed to setup some pins as outputs, these output pins will have an initial high output value unless the associated port data registers are first programmed. Selecting which pins are inputs and which are outputs can be achieved byte-wide by loading the correct values into the appropriate port control register or by programming individual bits in the port control register using the "SET [m].i" and "CLR [m].i" instructions. Note that when using these bit control instructions, a read-modify-write operation takes place. The microcontroller must first read in the data on the entire port, modify it to the required new bit values and then rewrite this data back to the output ports.

Port A has the additional capability of providing wake-up functions. When the devices are in the SLEEP or IDLE Mode, various methods are available to wake the devices up. One of these is a high to low transition of any of the Port A pins. Single or multiple pins on Port A can be setup to have this function.



# Timer Modules – TM

One of the most fundamental functions in any microcontroller the devices are the ability to control and measure time. To implement time related functions the devices include several Timer Modules, abbreviated to the name TM. The TMs are multi-purpose timing units and serve to provide operations such as Timer/Counter, Input Capture, Compare Match Output and Single Pulse Output as well as being the functional unit for the generation of PWM signals. Each of the TMs has two individual interrupts. The addition of input and output pins for each TM ensures that users are provided with timing units with a wide and flexible range of features.

The common features of the different TM types are described here with more detailed information provided in the individual Standard and Periodic TM sections.

#### Introduction

The devices contain two TMs and each individual TM can be categorised as a certain type, namely Standard Type TM or Periodic Type TM. Although similar in nature, the different TM types vary in their feature complexity. The common features to all of the Standard and Periodic TMs will be described in this section. The detailed operation regarding each of the TM types will be described in separate sections. The main features and differences between the two types of TMs are summarised in the accompanying table.

Function	STM	PTM
Timer/Counter	$\checkmark$	$\checkmark$
Input Capture	$\checkmark$	$\checkmark$
Compare Match Output	$\checkmark$	$\checkmark$
PWM Channels	1	1
Single Pulse Output	1	1
PWM Alignment	Edge	Edge
PWM Adjustment Period & Duty	Duty or Period	Duty or Period

**TM Function Summary** 

### **TM Operation**

The different types of TM offer a diverse range of functions, from simple timing operations to PWM signal generation. The key to understanding how the TM operates is to see it in terms of a free running counter whose value is then compared with the value of pre-programmed internal comparators. When the free running counter has the same value as the pre-programmed comparator, known as a compare match situation, a TM interrupt signal will be generated which can clear the counter and perhaps also change the condition of the TM output pin. The internal TM counter is driven by a user selectable clock source, which can be an internal clock or an external pin.

### **TM Clock Source**

The clock source which drives the main counter in each TM can originate from various sources. The selection of the required clock source is implemented using the xTCK2-xTCK0 bits in the xTM control registers, where "x" stands for S or P type TM. The clock source can be a ratio of the system clock f<sub>SYS</sub> or the internal high clock f<sub>H</sub>, the f<sub>SUB</sub> clock source or the external xTCK pin. The xTCK pin clock source is used to allow an external signal to drive the TM as an external clock source or for event counting.

#### **TM Interrupts**

The Standard and Periodic type TMs each have two internal interrupts, one for each of the internal comparator A or comparator P, which generate a TM interrupt when a compare match condition occurs. When a TM interrupt is generated it can be used to clear the counter and also to change the state of the TM output pin.



### **TM External Pins**

Each of the TMs, irrespective of what type, has two TM input pins, with the label xTCK and xTPI respectively. The xTM input pin, xTCK, is essentially a clock source for the xTM and is selected using the xTCK2~xTCK0 bits in the xTMC0 register. This external xTCK input pin allows an external clock source to drive the internal TM. The xTCK input pin can be chosen to have either a rising or falling active edge. The xTCK pin is also used as the external trigger input pin in single pulse output mode.

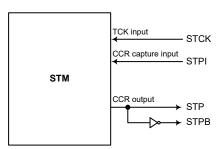
The other xTM input pin, xTPI, is the capture input whose active edge can be a rising edge, a falling edge or both rising and falling edges and the active edge transition type is selected using the xTIO1~xTIO0 bits in the xTMC1 register. There is another capture input, PTCK, for PTM capture input mode, which can be used as the external trigger input source.

The TMs each have two output pins with the label xTP and xTPB. The xTPB pin outputs the inverted signal of the xTP. When the TM is in the Compare Match Output Mode, these pins can be controlled by the TM to switch to a high or low level or to toggle when a compare match situation occurs. The external output pin is also the pin where the TM generates the PWM output waveform.

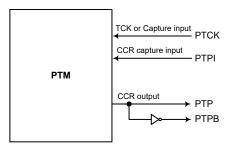
As the TM input and output pins are pin-shared with other functions, the TM input or output function must first be setup using relevant pin-shared function selection register. The details of the pin-shared function selection are described in the pin-shared function section.

SI	ſM	PTM			
Input Output		Input	Output		
STCK, STPI	STP, STPB	PTCK, PTPI	PTP, PTPB		

Note: For the BA45F5740-2 device, the STCK, STPI, STP and PTPI pins are unavailable. **TM External Pins** 







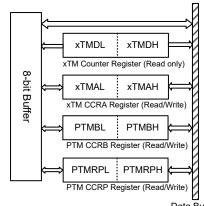
Note: For the BA45F5740-2 device, the PTPI pin is unavailable. **PTM Function Pin Block Diagram** 



### **Programming Considerations**

The TM Counter Registers and the Capture/Compare CCRA and CCRP registers as well as the PTM CCRB register, all have a low and high byte structure. The high bytes can be directly accessed, but as the low bytes can only be accessed via an internal 8-bit buffer, reading or writing to these register pairs must be carried out in a specific way. The important point to note is that data transfer to and from the 8-bit buffer and its related low byte only takes place when a write or read operation to its corresponding high byte is executed.

As the CCRA, CCRP and CCRB registers are implemented in the way shown in the following diagram and accessing these register pairs is carried out in a specific way as described above, it is recommended to use the "MOV" instruction to access the CCRA, CCRP and CCRB low byte registers, named xTMAL, PTMRPL, PTMBL, using the following access procedures. Accessing the CCRA, CCRB or CCRP low byte register without following these access procedures will result in unpredictable values.



Data Bus

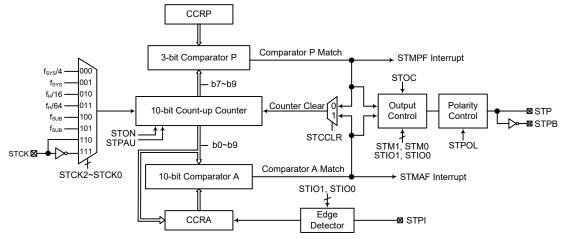
The following steps show the read and write procedures:

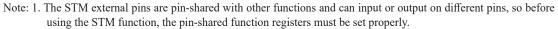
- Writing Data to CCRA, CCRB or CCRP
  - Step 1. Write data to Low Byte xTMAL, PTMBL or PTMRPL
     Note that here data is only written to the 8-bit buffer.
  - Step 2. Write data to High Byte xTMAH, PTMBH or PTMRPH
    - Here data is written directly to the high byte registers and simultaneously data is latched from the 8-bit buffer to the Low Byte registers.
- Reading Data from the Counter Registers, CCRA, CCRB or CCRP
  - ♦ Step 1. Read data from the High Byte xTMDH, xTMAH, PTMBH or PTMRPH
  - Here data is read directly from the High Byte registers and simultaneously data is latched from the Low Byte register into the 8-bit buffer.
  - ♦ Step 2. Read data from the Low Byte xTMDL, xTMAL, PTMBL or PTMRPL
    - This step reads data from the 8-bit buffer.



# Standard Type TM – STM

The Standard Type TM contains five operating modes, which are Compare Match Output, Timer/ Event Counter, Capture Input, Single Pulse Output and PWM Output modes.





2. For the BA45F5740-2 device, the STCK, STPI and STP pins are unavailable.

Standard Type TM Block Diagram

## Standard TM Operation

The size of Standard TM is 10-bit wide and its core is a 10-bit count-up counter which is driven by a user selectable internal or external clock source. There are also two internal comparators with the names, Comparator A and Comparator P. These comparators will compare the value in the counter with CCRP and CCRA registers. The CCRP comparator is 3-bit wide whose value is compared with the highest 3 bits in the counter while the CCRA is 10 bits and therefore compares all counter bits.

The only way of changing the value of the 10-bit counter using the application program, is to clear the counter by changing the STON bit from low to high. The counter will also be cleared automatically by a counter overflow or a compare match with one of its associated comparators. When these conditions occur, a STM interrupt signal will also usually be generated. The Standard Type TM can operate in a number of different operational modes, can be driven by different clock sources including an input pin and can also control two output pins. All operating setup conditions are selected using relevant internal registers.

# Standard Type TM Register Description

Overall operation of the Standard TM is controlled using a series of registers. A read only register pair exists to store the internal counter 10-bit value, while a read/write register pair exists to store the internal 10-bit CCRA value. The remaining two registers are control registers which setup the different operating and control modes as well as three CCRP bits.

Register	Bit									
Name	7	6	5	4	3	2	1	0		
STMC0	STPAU	STCK2	STCK1	STCK0	STON	STRP2	STRP1	STRP0		
STMC1	STM1	STM0	STIO1	STIO0	STOC	STPOL	STDPX	STCCLR		
STMDL	D7	D6	D5	D4	D3	D2	D1	D0		
STMDH	—	—	—	—	—	_	D9	D8		
STMAL	D7	D6	D5	D4	D3	D2	D1	D0		
STMAH	—	—		—			D9	D8		

#### 10-bit Standard TM Register List

#### STMC0 Register

Bit	7	6	5	4	3	2	1	0
Name	STPAU	STCK2	STCK1	STCK0	STON	STRP2	STRP1	STRP0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 STPAU: STM Counter Pause Control

0: Run

1: Pause

The counter can be paused by setting this bit high. Clearing the bit to zero restores normal counter operation. When in a Pause condition the STM will remain powered up and continue to consume power. The counter will retain its residual value when this bit changes from low to high and resume counting from this value when the bit changes to a low value again.

#### Bit 6~4 STCK2~STCK0: Select STM Counter Clock

- 000:  $f_{\text{SYS}}/4$
- 001: f<sub>sys</sub>

010: f<sub>H</sub>/16

- 011: f<sub>H</sub>/64
- 100: f<sub>sub</sub>
- 101: f<sub>SUB</sub>
- 110: STCK rising edge clock (unavailable for BA45F5740-2)

111: STCK falling edge clock (unavailable for BA45F5740-2)

These three bits are used to select the clock source for the STM. The external pin clock source can be chosen to be active on the rising or falling edge. The clock source  $f_{SYS}$  is the system clock, while  $f_H$  and  $f_{SUB}$  are other internal clocks, the details of which can be found in the oscillator section.

Bit 3

- STON: STM Counter On/Off Control 0: Off
  - 1: On

This bit controls the overall on/off function of the STM. Setting the bit high enables the counter to run while clearing the bit disables the STM. Clearing this bit to zero will stop the counter from counting and turn off the STM which will reduce its power consumption. When the bit changes state from low to high the internal counter value will be reset to zero, however when the bit changes from high to low, the internal counter will retain its residual value until the bit returns high again. If the STM is in the Compare Match Output Mode or the PWM output Mode or Single Pulse Output Mode then the STM output pin will be reset to its initial condition, as specified by the STOC bit, when the STON bit changes from low to high.

Bit 2~0 STRP2~STRP0: STM CCRP 3-bit register, compared with the STM counter bit 9~bit 7 Comparator P Match Period =

000: 1024 STM clocks 001: 128 STM clocks 010: 256 STM clocks



011: 384 STM clocks 100: 512 STM clocks 101: 640 STM clocks 110: 768 STM clocks 111: 896 STM clocks

These three bits are used to setup the value on the internal CCRP 3-bit register, which are then compared with the internal counter's highest three bits. The result of this comparison can be selected to clear the internal counter if the STCCLR bit is set to zero. Setting the STCCLR bit to zero ensures that a compare match with the CCRP values will reset the internal counter. As the CCRP bits are only compared with the highest three bits, the compare values exist in 128 clock cycle multiples. Clearing all three bits to zero is in effect allowing the counter to overflow at its maximum value.

#### STMC1 Register

Bit	7	6	5	4	3	2	1	0
Name	STM1	STM0	STIO1	STIO0	STOC	STPOL	STDPX	STCCLR
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 STM1~STM0: Select STM Operating Mode

00: Compare Match Output Mode

01: Capture Input Mode

10: PWM Output Mode or Single Pulse Output Mode

11: Timer/Counter Mode

These bits setup the required operating mode for the STM. To ensure reliable operation the STM should be switched off before any changes are made to the STM1 and STM0 bits. In the Timer/Counter Mode, the STM output pin state is undefined.

## Bit 5~4 STIO1~STIO0: Select STM External Pins Function

Compare Match Output Mode

- 00: No change
- 01: Output low
- 10: Output high
- 11: Toggle output
- PWM Output Mode/Single Pulse Output Mode
  - 00: PWM output inactive state
  - 01: PWM output active state
  - 10: PWM output
  - 11: Single Pulse Output
- Capture Input Mode
  - 00: Input capture at rising edge of STPI
  - 01: Input capture at falling edge of STPI
  - 10: Input capture at both rising and falling edges of STPI
  - 11: Input capture disabled
- Timer/Counter Mode

Unused

These two bits are used to determine how the STM functions when a certain condition is reached. The function that these bits select depends upon in which mode the STM is running.

In the Compare Match Output Mode, the STIO1 and STIO0 bits determine how the STM output pin changes state when a compare match occurs from the Comparator A. The STM output pin can be setup to switch high, switch low or to toggle its present state when a compare match occurs from the Comparator A. When the bits are both zero, then no change will take place on the output. The initial value of the STM output pin should be setup using the STOC bit in the STMC1 register. Note that the output level requested by the STIO1 and STIO0 bits must be different from the initial value

setup using the STOC bit otherwise no change will occur on the STM output pin when a compare match occurs. After the STM output pin changes state, it can be reset to its initial level by changing the level of the STON bit from low to high.

In the PWM Output Mode, the STIO1 and STIO0 bits determine how the STM output pin changes state when a certain compare match condition occurs. The PWM output function is modified by changing these two bits. It is necessary to only change the values of the STIO1 and STIO0 bits only after the STM has been switched off. Unpredictable PWM outputs will occur if the STIO1 and STIO0 bits are changed when the STM is running.

Bit 3

#### STOC: STP Output Control

Compare Match Output Mode

0: Initial low

1: Initial high

PWM Output Mode/Single Pulse Output Mode

0: Active low

1: Active high

This is the output control bit for the STM output. Its operation depends upon whether STM is being used in the Compare Match Output Mode or in the PWM Output Mode/ Single Pulse Output Mode. It has no effect if the STM is in the Timer/Counter Mode. In the Compare Match Output Mode it determines the logic level of the STM output pin before a compare match occurs. In the PWM Output Mode/Single Pulse Output Mode it determines if the PWM signal is active high or active low.

#### Bit 2 STPOL: STP Output Polarity Control

0: Non-invert

1: Invert

This bit controls the polarity of the STP output. When the bit is set high the STM output will be inverted and not inverted when the bit is zero. It has no effect if the STM is in the Timer/Counter Mode.

#### Bit 1 STDPX: STM PWM Duty/Period Control

0: CCRP - period; CCRA - duty

1: CCRP - duty; CCRA - period

This bit determines which of the CCRA and CCRP registers are used for period and duty control of the PWM waveform.

#### Bit 0 STCCLR: STM Counter Clear Condition Selection

0: Comparator P match

1: Comparator A match

This bit is used to select the method which clears the counter. Remember that the Standard TM contains two comparators, Comparator A and Comparator P, either of which can be selected to clear the internal counter. With the STCCLR bit set high, the counter will be cleared when a compare match occurs from the Comparator A. When the bit is low, the counter will be cleared when a compare match occurs from the Comparator P or with a counter overflow. A counter overflow clearing method can only be implemented if the CCRP bits are all cleared to zero. The STCCLR bit is not used in the PWM Output, Single Pulse Output or Capture Input Mode.

## STMDL Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R	R	R	R	R	R	R	R
POR	0	0	0	0	0	0	0	0

Bit 7~0

D7~D0: STM Counter Low Byte Register bit 7 ~ bit 0 STM 10-bit Counter bit 7 ~ bit 0



## STMDH Register

Bit	7	6	5	4	3	2	1	0
Name	_	—	—	—	—	—	D9	D8
R/W	_	_	—	—	_	_	R	R
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as "0"

Bit 1~0 **D9~D8**: STM Counter High Byte Register bit 1 ~ bit 0 STM 10-bit Counter bit 9 ~ bit 8

## STMAL Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: STM CCRA Low Byte Register bit 7 ~ bit 0

STM 10-bit CCRA bit 7 ~ bit 0

## STMAH Register

Bit	7	6	5	4	3	2	1	0
Name	—	_	—	—	—	—	D9	D8
R/W	—	_	—	—	—	—	R/W	R/W
POR	—	—	_	_	_	—	0	0

Bit 7~2 Unimplemented, read as "0"

Bit 1~0 **D9~D8**: STM CCRA High Byte Register bit 1 ~ bit 0 STM 10-bit CCRA bit 9 ~ bit 8

## Standard Type TM Operation Modes

The Standard Type TM can operate in one of five operating modes, Compare Match Output Mode, PWM Output Mode, Single Pulse Output Mode, Capture Input Mode or Timer/Counter Mode. The operating mode is selected using the STM1 and STM0 bits in the STMC1 register.

## Compare Match Output Mode

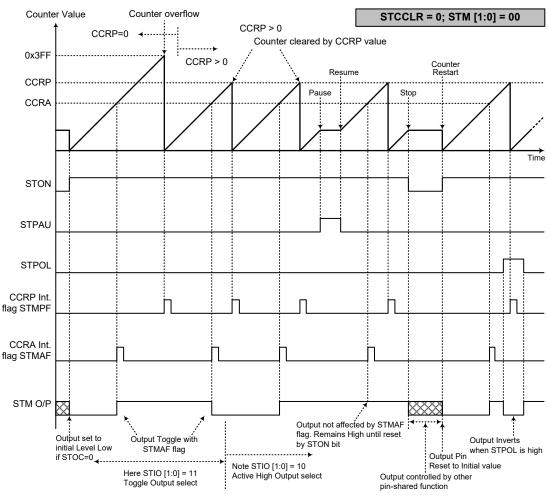
To select this mode, bits STM1 and STM0 in the STMC1 register, should be set to 00 respectively. In this mode once the counter is enabled and running it can be cleared by three methods. These are a counter overflow, a compare match from Comparator A and a compare match from Comparator P. When the STCCLR bit is low, there are two ways in which the counter can be cleared. One is when a compare match from Comparator P, the other is when the CCRP bits are all zero which allows the counter to overflow. Here both STMAF and STMPF interrupt request flags for Comparator A and Comparator P respectively, will both be generated.

If the STCCLR bit in the STMC1 register is high then the counter will be cleared when a compare match occurs from Comparator A. However, here only the STMAF interrupt request flag will be generated even if the value of the CCRP bits is less than that of the CCRA registers. Therefore when STCCLR is high no STMPF interrupt request flag will be generated. In the Compare Match Output Mode, the CCRA can not be set to "0".

If the CCRA bits are all zero, the counter will overflow when it reaches its maximum 10-bit, 3FF Hex, value, however here the STMAF interrupt request flag will not be generated.

As the name of the mode suggests, after a comparison is made, the STM output pin, will change state. The STM output pin condition however only changes state when a STMAF interrupt request

flag is generated after a compare match occurs from Comparator A. The STMPF interrupt request flag, generated from a compare match occurs from Comparator P, will have no effect on the STM output pin. The way in which the STM output pin changes state are determined by the condition of the STIO1 and STIO0 bits in the STMC1 register. The STM output pin can be selected using the STIO1 and STIO0 bits to go high, to go low or to toggle from its present condition when a compare match occurs from Comparator A. The initial condition of the STM output, which is setup after the STON bit changes from low to high, is setup using the STOC bit. Note that if the STIO1 and STIO0 bits are zero then no pin change will take place.



Compare Match Output Mode – STCCLR=0

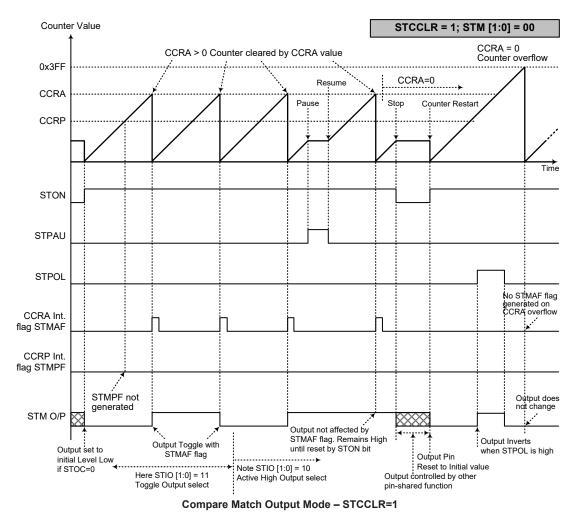
Note: 1. With STCCLR=0 a Comparator P match will clear the counter

2. The STM output is controlled only by the STMAF flag

3. The STM output is reset to itsinitial state by a STON bit rising edge

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Note: 1. With STCCLR=1 a Comparator A match will clear the counter

2. The STM output is controlled only by the STMAF flag

3. The STM output is reset to its initial state by a STON bit rising edge

4. A STMPF flag is not generated when STCCLR=1



#### **Timer/Counter Mode**

To select this mode, bits STM1 and STM0 in the STMC1 register should be set to 11 respectively. The Timer/Counter Mode operates in an identical way to the Compare Match Output Mode generating the same interrupt flags. The exception is that in the Timer/Counter Mode the STM output pin is not used. Therefore the above description and Timing Diagrams for the Compare Match Output Mode can be used to understand its function. As the STM output pin is not used in this mode, the pin can be used as a normal I/O pin or other pin-shared function.

#### **PWM Output Mode**

To select this mode, bits STM1 and STM0 in the STMC1 register should be set to 10 respectively and also the STIO1 and STIO0 bits should be set to 10 respectively. The PWM function within the STM is useful for applications which require functions such as motor control, heating control, illumination control etc. By providing a signal of fixed frequency but of varying duty cycle on the STM output pin, a square wave AC waveform can be generated with varying equivalent DC RMS values.

As both the period and duty cycle of the PWM waveform can be controlled, the choice of generated waveform is extremely flexible. In the PWM Output mode, the STCCLR bit has no effect as the PWM period. Both of the CCRA and CCRP registers are used to generate the PWM waveform, one register is used to clear the internal counter and thus control the PWM waveform frequency, while the other one is used to control the duty cycle. Which register is used to control either frequency or duty cycle is determined using the STDPX bit in the STMC1 register. The PWM waveform frequency and duty cycle can therefore be controlled by the values in the CCRA and CCRP registers.

An interrupt flag, one for each of the CCRA and CCRP, will be generated when a compare match occurs from either Comparator A or Comparator P. The STOC bit in the STMC1 register is used to select the required polarity of the PWM waveform while the two STIO1 and STIO0 bits are used to enable the PWM output or to force the STM output pin to a fixed high or low level. The STPOL bit is used to reverse the polarity of the PWM output waveform.

#### • 10-bit STM, PWM Output Mode, Edge-aligned Mode, STDPX=0

CCRP	1~7	0			
Period	CCRP×128	1024			
Duty	CCRA				

If  $f_{SYS}$ =4MHz, STM clock source is  $f_{SYS}$ /4, CCRP=4 and CCRA =128,

The STM PWM output frequency= $(f_{SYS}/4)/(4 \times 128) = f_{SYS}/2048 = 2kHz$ , duty= $128/(4 \times 128) = 25\%$ .

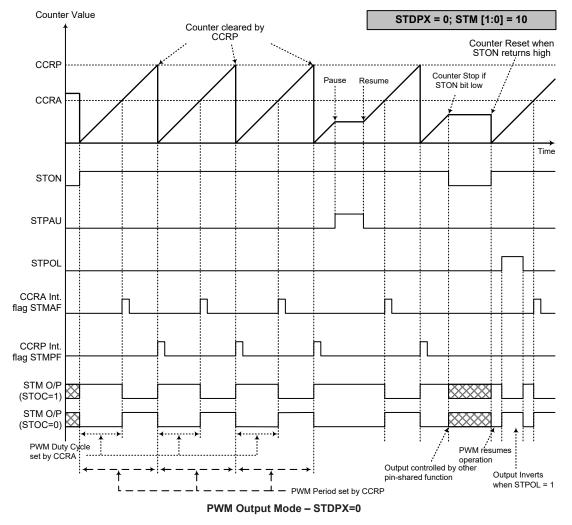
If the Duty value defined by the CCRA register is equal to or greater than the Period value, then the PWM output duty is 100%.

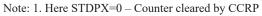
#### • 10-bit STM, PWM Output Mode, Edge-aligned Mode, STDPX=1

CCRP	1~7 0				
Period	CCRA				
Duty	CCRP×128	1024			

The PWM output period is determined by the CCRA register value together with the STM clock while the PWM duty cycle is defined by the CCRP register value.



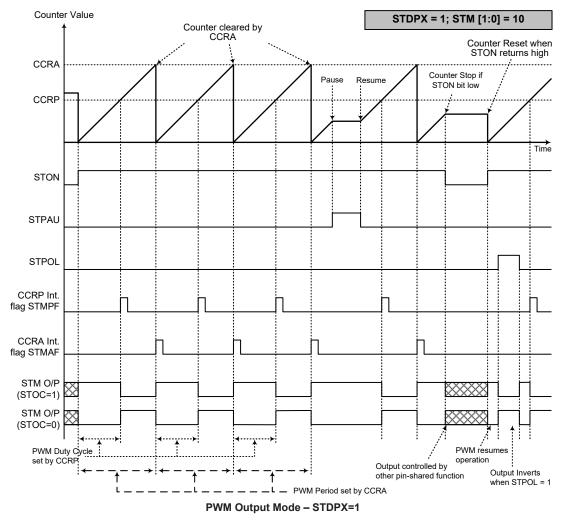


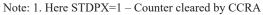


2. A counter clear sets the PWM Period

- 3. The internal PWM function continues running even when STIO [1:0]=00 or 01
- 4. The STCCLR bit has no influence on PWM operation







2. A counter clear sets the PWM Period

- 3. The internal PWM function continues even when STIO [1:0]=00 or 01
- 4. The STCCLR bit has no influence on PWM operation

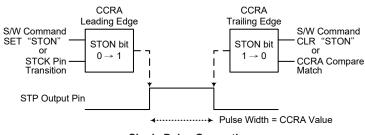


## Single Pulse Output Mode

To select this mode, bits STM1 and STM0 in the STMC1 register should be set to 10 respectively and also the STIO1 and STIO0 bits should be set to 11 respectively. The Single Pulse Output Mode, as the name suggests, will generate a single shot pulse on the STM output pin.

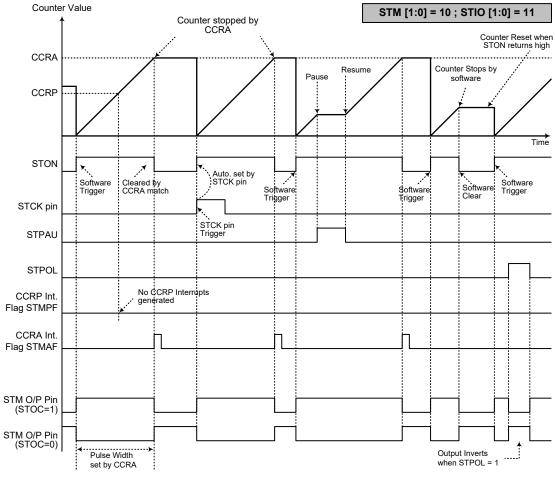
The trigger for the pulse output leading edge is a low to high transition of the STON bit, which can be implemented using the application program. However in the Single Pulse Mode, the STON bit can also be made to automatically change from low to high using the external STCK pin, which will in turn initiate the Single Pulse output. When the STON bit transitions to a high level, the counter will start running and the pulse leading edge will be generated. The STON bit should remain high when the pulse is in its active state. The generated pulse trailing edge will be generated when the STON bit is cleared to zero, which can be implemented using the application program or when a compare match occurs from Comparator A.

However a compare match from Comparator A will also automatically clear the STON bit and thus generate the Single Pulse output trailing edge. In this way the CCRA value can be used to control the pulse width. A compare match from Comparator A will also generate a STM interrupt. The counter can only be reset back to zero when the STON bit changes from low to high when the counter restarts. In the Single Pulse Mode CCRP is not used. The STCCLR and STDPX bits are not used in this Mode.



Single Pulse Generation





Single Pulse Mode

Note: 1. Counter stopped by CCRA

2. CCRP is not used

- 3. The pulse triggered by the STCK pin or by setting the STON bit high
- 4. A STCK pin active edge will automatically set the STON bit high.
- 5. In the Single Pulse Mode, STIO [1:0] must be set to "11" and can not be changed

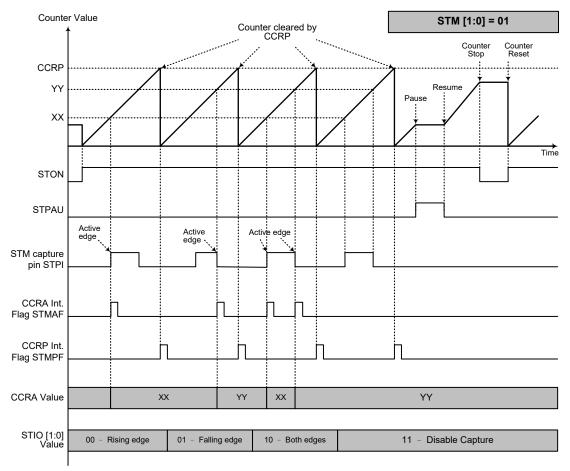


## Capture Input Mode

To select this mode bits STM1 and STM0 in the STMC1 register should be set to 01 respectively. This mode enables external signals to capture and store the present value of the internal counter and can therefore be used for applications such as pulse width measurements. The external signal is supplied on the STPI pin, whose active edge can be a rising edge, a falling edge or both rising and falling edges; the active edge transition type is selected using the STIO1 and STIO0 bits in the STMC1 register. The counter is started when the STON bit changes from low to high which is initiated using the application program.

When the required edge transition appears on the STPI pin the present value in the counter will be latched into the CCRA registers and a STM interrupt generated. Irrespective of what events occur on the STPI pin the counter will continue to free run until the STON bit changes from high to low. When a CCRP compare match occurs the counter will reset back to zero; in this way the CCRP value can be used to control the maximum counter value. When a CCRP compare match occurs from Comparator P, a STM interrupt will also be generated. Counting the number of overflow interrupt signals from the CCRP can be a useful method in measuring long pulse widths. The STIO1 and STIO0 bits can select the active trigger edge on the STPI pin to be a rising edge, falling edge or both edge types. If the STIO1 and STIO0 bits are both set high, then no capture operation will take place irrespective of what happens on the STPI pin, however it must be noted that the counter will continue to run. The STCCLR and STDPX bits are not used in this Mode.





## **Capture Input Mode**

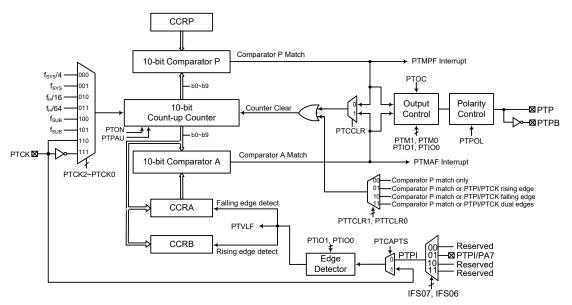
Note: 1. STM [1:0]=01 and active edge set by the STIO [1:0] bits

- 2. A STM Capture input pin active edge transfers the counter value to CCRA
- 3. STCCLR bit not used
- 4. No output function STOC and STPOL bits are not used
- 5. CCRP determines the counter value and the counter has a maximum count value when CCRP is equal to zero



# Periodic Type TM – PTM

The Periodic Type TM contains five operating modes, which are Compare Match Output, Timer/ Event Counter, Capture Input, Single Pulse Output and PWM Output modes. The Periodic TM can be controlled with external input pins and can drive some external output pins.



- Note: 1. The PTM capture input can be sourced from the PTPI signal or from the external PTCK pin, if the PTPI signal is used, the IFS0[7:6] bits must be set to 01 to select the external PTPI pin input.
  - 2. For the BA45F5740-2 device, the PTPI pin is unavailable, so the capture input can only be sourced from the PTCK pin.
  - 3. If the PTM external pins will be used and as these pins are pin-shared with other functions, before using the PTM function, the pin-shared function registers should be set properly.

#### Periodic Type TM Block Diagram

## Periodic TM Operation

The Periodic Type TM core is a 10-bit count-up counter which is driven by a user selectable internal or external clock source. There are also two internal comparators with the names, Comparator A and Comparator P. These comparators will compare the value in the counter with CCRP and CCRA registers. The CCRP comparator is 10-bit wide.

The only way of changing the value of the 10-bit counter using the application program, is to clear the counter by changing the PTON bit from low to high. The counter will also be cleared automatically by a counter overflow or a compare match with one of its associated comparators or the active trigger edge in Capture input mode by PTTCLR[1:0]. When these conditions occur, a PTM interrupt signal will also usually be generated. The Periodic Type TM can operate in a number of different operational modes and can be driven by different clock sources including two input pins and also control more than one output pins. All operating setup conditions are selected using relevant internal registers.



## Periodic Type TM Register Description

Overall operation of the Periodic Type TM is controlled using a series of registers. A read only register pair exists to store the internal 10-bit counter value, while three read/write register pairs exist to store the internal 10-bit CCRA value, CCRP value and CCRB value. The remaining three registers are control registers which setup the different operating and control modes.

Register				В	it			
Name	7	6	5	4	3	2	1	0
PTMC0	PTPAU	PTCK2	PTCK1	PTCK0	PTON	_	—	_
PTMC1	PTM1	PTM0	PTIO1	PTIO0	PTOC	PTPOL	PTCAPTS	PTCCLR
PTMC2	_	—	—	—	—	PTTCLR1	PTTCLR0	PTVLF
PTMDL	D7	D6	D5	D4	D3	D2	D1	D0
PTMDH	—	—	—	—	—	—	D9	D8
PTMAL	D7	D6	D5	D4	D3	D2	D1	D0
PTMAH	_	—	—	—	—	—	D9	D8
PTMBL	D7	D6	D5	D4	D3	D2	D1	D0
PTMBH	_	—	—	_	_	—	D9	D8
PTMRPL	D7	D6	D5	D4	D3	D2	D1	D0
PTMRPH	_	_	_	_		_	D9	D8

**10-bit Periodic TM Register List** 

## PTMC0 Register

Bit	7	6	5	4	3	2	1	0
Name	PTPAU	PTCK2	PTCK1	PTCK0	PTON	_	—	—
R/W	R/W	R/W	R/W	R/W	R/W	_	—	—
POR	0	0	0	0	0	—	—	—

Bit 7 PTPAU: PTM Counter Pause Control

0: Run

1: Pause

The counter can be paused by setting this bit high. Clearing the bit to zero restores normal counter operation. When in a Pause condition the PTM will remain powered up and continue to consume power. The counter will retain its residual value when this bit changes from low to high and resume counting from this value when the bit changes to a low value again.

## Bit 6~4 PTCK2~PTCK0: Select PTM Counter clock

- 000:  $f_{SYS}/4$
- 001: f<sub>sys</sub>
- 010: f<sub>H</sub>/16
- 011: f<sub>H</sub>/64 100: f<sub>SUB</sub>
- 100. ISUB $101: f_{SUB}$
- 110: PTCK rising edge clock
- 111: PTCK falling edge clock

These three bits are used to select the clock source for the PTM. The external pin clock source can be chosen to be active on the rising or falling edge. The clock source  $f_{SYS}$  is the system clock, while  $f_H$  and  $f_{SUB}$  are other internal clocks, the details of which can be found in the oscillator section.

Bit 3 PTON: PTM Counter On/Off Control

- 0: Off
- 1: On

This bit controls the overall on/off function of the PTM. Setting the bit high enables the counter to run, clearing the bit disables the PTM. Clearing this bit to zero will



stop the counter from counting and turn off the PTM which will reduce its power consumption. When the bit changes state from low to high the internal counter value will be reset to zero, however when the bit changes from high to low, the internal counter will retain its residual value until the bit returns high again.

If the PTM is in the Compare Match Output Mode, PWM output Mode or Single Pulse Output Mode then the PTM output pin will be reset to its initial condition, as specified by the PTOC bit, when the PTON bit changes from low to high.

Bit 2~0 Unimplemented, read as "0"

#### PTMC1 Register

Bit	7	6	5	4	3	2	1	0
Name	PTM1	PTM0	PTIO1	PTIO0	PTOC	PTPOL	PTCAPTS	PTCCLR
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6

PTM1~PTM0: Select PTM Operating Mode 00: Compare Match Output Mode

01: Capture Input Mode

10: PWM Output Mode or Single Pulse Output Mode

11: Timer/Counter Mode

These bits setup the required operating mode for the PTM. To ensure reliable operation the PTM should be switched off before any changes are made to the PTM1 and PTM0 bits. In the Timer/Counter Mode, the PTM output pin state is undefined.

## Bit 5~4 PTIO1~PTIO0: Select PTM External Pins Function

Compare Match Output Mode

00: No change

- 01: Output low
- 10: Output high
- 11: Toggle output

PWM Output Mode/Single Pulse Output Mode

- 00: PWM output inactive state
- 01: PWM output active state
- 10: PWM output
- 11: Single pulse output

Capture Input Mode

PTTCLR[1:0]=00B:

- 00: Input capture at rising edge of PTPI or PTCK, and the counter value will be latched into CCRA
- 01: Input capture at falling edge of PTPI or PTCK, and the counter value will be latched into CCRA
- 10: Input capture at both falling and rising edges of PTPI or PTCK, and the counter value will be latched into CCRA

11: Input capture disabled

- PTTCLR[1:0]=01B,10B or 11B:
  - 00: Input capture at rising edge of PTPI or PTCK, and the counter value will be latched into CCRB
  - 01: Input capture at falling edge of PTPI or PTCK, and the counter value will be latched into CCRA
  - 10: Input capture at both falling and rising edges of PTPI or PTCK, and the counter value will be latched into CCRA at falling edge or CCRB at rising edge
  - 11: Input capture disabled
- Timer/Counter Mode
- Unused

These two bits are used to determine how the PTM functions when a certain condition is reached. The function that these bits select depends upon in which mode the PTM is running.

In the Compare Match Output Mode, the PTIO1 and PTIO0 bits determine how the PTM output changes state when a compare match occurs from the Comparator A. The PTM output can be setup to switch high, switch low or to toggle its present state when a compare match occurs from the Comparator A. When the bits are both zero, then no change will take place on the output. The initial value of the PTM output should be setup using the PTOC bit in the PTMC1 register. Note that the output level requested by the PTIO1 and PTIO0 bits must be different from the initial value setup using the PTOC bit otherwise no change will occur on the PTM output when a compare match occurs. After the PTM output changes state, it can be reset to its initial level by changing the level of the PTON bit from low to high.

In the PWM Output Mode, the PTIO1 and PTIO0 bits determine how the PTM output changes state when a certain compare match condition occurs. The PWM output function is modified by changing these two bits. It is necessary to only change the values of the PTIO1 and PTIO0 bits only after the TM has been switched off. Unpredictable PWM outputs will occur if the PTIO1 and PTIO0 bits are changed when the PTM is running.

	when the PTM is running.
Bit 3	PTOC: PTM PTP Output control bit
	Compare Match Output Mode 0: Initial low 1: Initial high
	PWM Output Mode/Single Pulse Output Mode 0: Active low 1: Active high
	This is the output control bit for the PTM output. Its operation depends upon whether PTM is being used in the Compare Match Output Mode or in the PWM Output Mode/ Single Pulse Output Mode. It has no effect if the PTM is in the Timer/Counter Mode. In the Compare Match Output Mode it determines the logic level of the PTM output before a compare match occurs. In the PWM Output Mode it determines if the PWM signal is active high or active low. In the Single Pulse Output Mode it determines the logic level of the PTM output when the PTON bit changes from low to high.
Bit 2	<b>PTPOL</b> : PTM PTP Output polarity Control 0: Non-invert 1: Invert
	This bit controls the polarity of the PTP output. When the bit is set high the PTM output will be inverted and not inverted when the bit is zero. It has no effect if the PTM is in the Timer/Counter Mode.
Bit 1	<ul> <li>PTCAPTS: PTM Capture Trigger Source Selection</li> <li>0: From PTPI input signal</li> <li>1: From PTCK input</li> <li>Note: For the BA45F5740-2 device, the PTPI input source is unavailable, so this bit should be fixed at "1".</li> </ul>
Bit 0	PTCCLR: Select PTM Counter clear condition 0: PTM Comparator P match 1: PTM Comparator A match
	This bit is used to select the method which clears the counter. Remember that the Periodic TM contains two comparators, Comparator A and Comparator P, either of which can be selected to clear the internal counter. With the PTCCLR bit set high, the counter will be cleared when a compare match occurs from the Comparator A. When the bit is low, the counter will be cleared when a compare match occurs from the Comparator P or with a counter overflow. A counter overflow clearing method can only be implemented if the CCRP bits are all cleared to zero. The PTCCLR bit is not

used in the PWM Output Mode, Single Pulse Output Mode or Capture Input Mode.



## PTMC2 Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	PTTCLR1	PTTCLR0	PTVLF
R/W	_	_	—	—	—	R/W	R/W	R
POR	—	—	—	—	—	0	0	0

Bit 7~3 Unimplemented, read as "0"

Bit 2~1 **PTTCLR1~PTTCLR0**: Select PTM Counter clear condition in capture input mode only 00: Comparator P match

01: Comparator P match or PTCK/PTPI rising edge

10: Comparator P match or PTCK/PTPI falling edge

11: Comparator P match or PTCK/PTPI dual edges

Note that these bits selection can be available only when the PTM operates in the Capture Input Mode.

Bit 0 **PTVLF**: PTM counter value latch edge flag

0: Falling edge trigger the counter value latch

1: Rising edge trigger the counter value latch

Note: When the PTTCLR1~PTTCLR0 bits equal to 00B, ignore this flag status.

## PTMDL Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R	R	R	R	R	R	R	R
POR	0	0	0	0	0	0	0	0

Bit 7~0D7~D0: PTM Counter Low Byte Register bit 7 ~ bit 0PTM 10-bit Counter bit 7 ~ bit 0

## PTMDH Register

Bit	7	6	5	4	3	2	1	0
Name	—	_	—	—	—	—	D9	D8
R/W	_	_	_	_	_	_	R	R
POR		_	_	_	_		0	0

Bit 7~2 Unimplemented, read as "0"

Bit 1~0 **D9~D8**: PTM Counter High Byte Register bit 1 ~ bit 0 PTM1 10-bit Counter bit 9 ~ bit 8

## PTMAL Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

 Bit 7~0
 D7~D0: PTM CCRA Low Byte Register bit 7 ~ bit 0

 PTM 10-bit CCRA bit 7 ~ bit 0



## PTMAH Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	_	—	—	—	D9	D8
R/W	_	_	—	—	—	_	R/W	R/W
POR	—	—	—	—	_	—	0	0

Bit 7~2 Unimplemented, read as "0"

Bit 1~0 **D9~D8**: PTM CCRA High Byte Register bit 1 ~ bit 0 PTM 10-bit CCRA bit 9 ~ bit 8

## PTMBL Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: PTM CCRB Low Byte Register bit 7 ~ bit 0

PTM 10-bit CCRB bit 7 ~ bit 0

## PTMBH Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	D9	D8
R/W	—	_	—	—	—	—	R/W	R/W
POR	—	—	_	—	_	—	0	0

Bit 7~2 Unimplemented, read as "0"

Bit 1~0 **D9~D8**: PTM CCRB High Byte Register bit 1 ~ bit 0 PTM 10-bit CCRB bit 9 ~ bit 8

## PTMRPL Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 D7~D0: PTM CCRP Low Byte Register bit 7 ~ bit 0 PTM 10-bit CCRP bit 7 ~ bit 0

## PTMRPH Register

Bit	7	6	5	4	3	2	1	0
Name	_	_	—	—	—	—	D9	D8
R/W	_	—	—	—	—	—	R/W	R/W
POR	_	—	_	—			0	0

Bit 7~2 Unimplemented, read as "0"

Bit 1~0 D9~D8: PTM CCRP High Byte Register bit 1 ~ bit 0 PTM 10-bit CCRP bit 9 ~ bit 8



## Periodic Type TM Operating Modes

The Periodic Type TM can operate in one of five operating modes, Compare Match Output Mode, PWM Output Mode, Single Pulse Output Mode, Capture Input Mode or Timer/Counter Mode. The operating mode is selected using the PTM1 and PTM0 bits in the PTMC1 register.

## **Compare Match Output Mode**

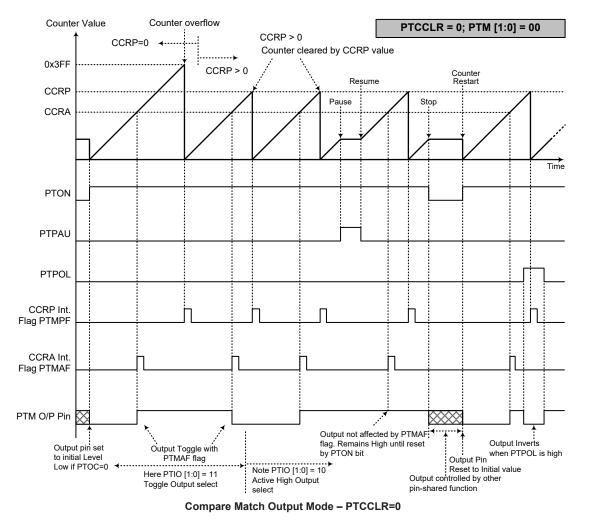
To select this mode, bits PTM1 and PTM0 in the PTMC1 register, should be set to 00 respectively. In this mode once the counter is enabled and running it can be cleared by three methods. These are a counter overflow, a compare match from Comparator A and a compare match from Comparator P. When the PTCCLR bit is low, there are two ways in which the counter can be cleared. One is when a compare match from Comparator P, the other is when the CCRP bits are all zero which allows the counter to overflow. Here both PTMAF and PTMPF interrupt request flags for Comparator A and Comparator P respectively, will both be generated.

If the PTCCLR bit in the PTMC1 register is high then the counter will be cleared when a compare match occurs from Comparator A. However, here only the PTMAF interrupt request flag will be generated even if the value of the CCRP bits is less than that of the CCRA registers. Therefore when PTCCLR is high no PTMPF interrupt request flag will be generated. In the Compare Match Output Mode, the CCRA can not be cleared to zero.

If the CCRA bits are all zero, the counter will overflow when its reaches its maximum 3FF Hex value, however here the PTMAF interrupt request flag will not be generated.

As the name of the mode suggests, after a comparison is made, the PTM output will change state. The PTM output condition however only changes state when a PTMAF interrupt request flag is generated after a compare match occurs from Comparator A. The PTMPF interrupt request flag, generated from a compare match occurs from Comparator P, will have no effect on the PTM output. The way in which the PTM output changes state are determined by the condition of the PTIO1 and PTIO0 bits in the PTMC1 register. The PTM output can be selected using the PTIO1 and PTIO0 bits to go high, to go low or to toggle from its present condition when a compare match occurs from Comparator A. The initial condition of the PTM output, which is setup after the PTON bit changes from low to high, is setup using the PTOC bit. Note that if the PTIO1 and PTIO0 bits are zero then no output change will take place.



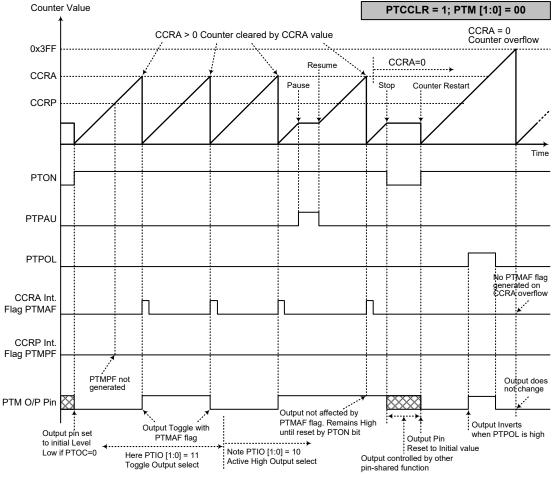




2. The PTM output is controlled only by the PTMAF flag

3. The output is reset to its initial state by a PTON bit rising edge





Compare Match Output Mode – PTCCLR=1

Note: 1. With PTCCLR=1 a Comparator A match will clear the counter

2. The PTM output is controlled only by the PTMAF flag

3. The output is reset to its initial state by a PTON bit rising edge

4. A PTMPF flag is not generated when PTCCLR=1



#### **Timer/Counter Mode**

To select this mode, bits PTM1 and PTM0 in the PTMC1 register should be set to 11 respectively. The Timer/Counter Mode operates in an identical way to the Compare Match Output Mode generating the same interrupt flags. The exception is that in the Timer/Counter Mode the PTM output pins are not used. Therefore the above description and Timing Diagrams for the Compare Match Output Mode can be used to understand its function. As the PTM output pins are not used in this mode, the pins can be used as normal I/O pins or other pin-shared function.

#### **PWM Output Mode**

To select this mode, bits PTM1 and PTM0 in the PTMC1 register should be set to 10 respectively. The PWM function within the PTM is useful for applications which require functions such as motor control, heating control, illumination control etc. By providing a signal of fixed frequency but of varying duty cycle on the PTM output pin, a square wave AC waveform can be generated with varying equivalent DC RMS values.

As both the period and duty cycle of the PWM waveform can be controlled, the choice of generated waveform is extremely flexible. In the PWM Output Mode, the PTCCLR bit has no effect on the PWM operation. Both of the CCRA and CCRP registers are used to generate the PWM waveform, one register is used to clear the internal counter and thus control the PWM waveform frequency, while the other one is used to control the duty cycle. The PWM waveform frequency and duty cycle can therefore be controlled by the values in the CCRA and CCRP registers.

An interrupt flag, one for each of the CCRA and CCRP, will be generated when a compare match occurs from either Comparator A or Comparator P. The PTOC bit in the PTMC1 register is used to select the required polarity of the PWM waveform while the two PTIO1 and PTIO0 bits are used to enable the PWM output or to force the PTM output to a fixed high or low level. The PTPOL bit is used to reverse the polarity of the PWM output waveform.

#### • 10-bit PTM, PWM Output Mode, Edge-aligned Mode

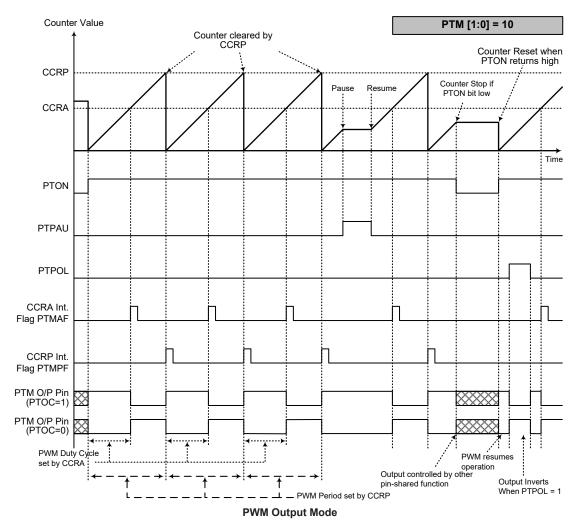
CCRP	1~1023	0
Period	1~1023	1024
Duty	CC	RA

If f<sub>SYS</sub>=8MHz, PTM clock source select f<sub>SYS</sub>/4, CCRP=512 and CCRA=128,

The PTM PWM output frequency=  $(f_{SYS}/4)/512 = f_{SYS}/2048 = 4kHz$ , duty=128/512=25%.

If the Duty value defined by the CCRA register is equal to or greater than the Period value, then the PWM output duty is 100%.





Note: 1. Counter cleared by CCRP

- 2. A counter clear sets the PWM Period
- 3. The internal PWM function continues running even when PTIO[1:0]=00 or 01
- 4. The PTCCLR bit has no influence on PWM operation

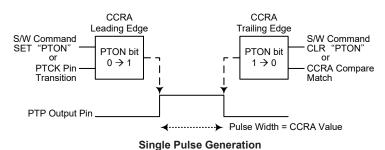


#### Single Pulse Output Mode

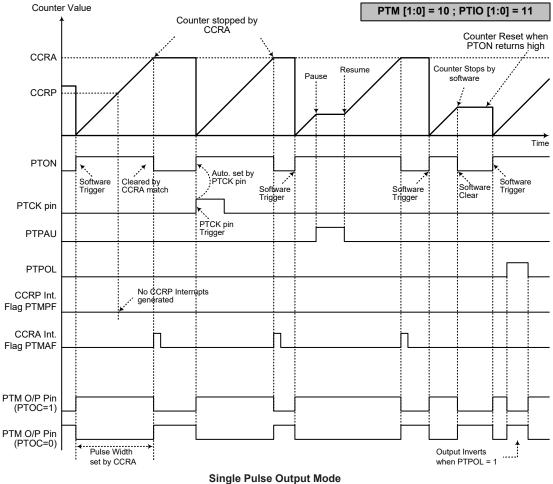
To select this mode, bits PTM1 and PTM0 in the PTMC1 register should be set to 10 respectively and also the PTIO1 and PTIO0 bits should be set to 11 respectively. The Single Pulse Output Mode, as the name suggests, will generate a single shot pulse on the PTM output pin.

The trigger for the pulse output leading edge is a low to high transition of the PTON bit, which can be implemented using the application program. However in the Single Pulse Mode, the PTON bit can also be made to automatically change from low to high using the external PTCK pin, which will in turn initiate the Single Pulse output. When the PTON bit transitions to a high level, the counter will start running and the pulse leading edge will be generated. The PTON bit should remain high when the pulse is in its active state. The generated pulse trailing edge will be generated when the PTON bit is cleared to zero, which can be implemented using the application program or when a compare match occurs from Comparator A.

However a compare match from Comparator A will also automatically clear the PTON bit and thus generate the Single Pulse output trailing edge. In this way the CCRA value can be used to control the pulse width. A compare match from Comparator A will also generate a PTM interrupt. The counter can only be reset back to zero when the PTON bit changes from low to high when the counter restarts. In the Single Pulse Output Mode CCRP is not used. The PTCCLR bit is not used in this Mode.







Note: 1. Counter stopped by CCRA

2. CCRP is not used

- 3. The pulse is triggered by the PTCK pin or by setting the PTON bit high
- 4. A PTCK pin active edge will automatically set the PTON bit high
- 5. In the Single Pulse Mode, PTIO[1:0] must be set to "11" and cannot be changed



### **Capture Input Mode**

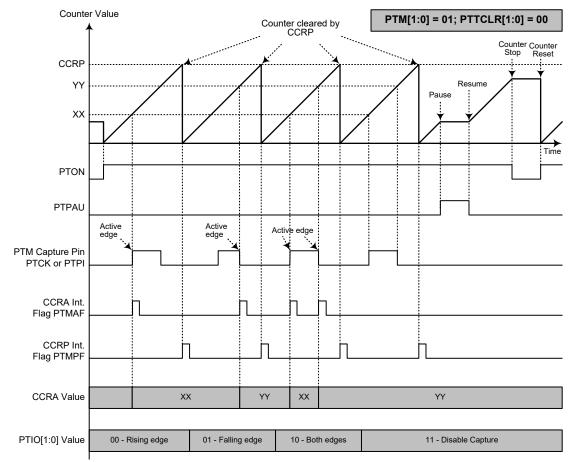
To select this mode bits PTM1 and PTM0 in the PTMC1 register should be set to 01 respectively. This mode enables external or internal signals to capture and store the present value of the internal counter and can therefore be used for applications such as pulse width measurements. The external or internal signal is supplied on the PTPI or PTCK pin which is selected using the IFS07~IFS06 bits in the IFS0 register and the PTCAPTS bit in the PTMC1 register. The input signal active edge can be either a rising edge, a falling edge or both rising and falling edges; the active edge transition type is selected using the PTIO1 and PTIO0 bits in the PTMC1 register. The counter is started when the PTON bit changes from low to high which is initiated using the application program.

The PTIO1 and PTIO0 bits decide which active edge transition type to be latched and interrupted. The PTTCLR1 and PTTCLR0 bits decide the condition that the counter reset back to zero. The present counter value latched into CCRA or CCRB is decided by both PTIO1~PTIO0 and PTTCLR1~PTTCLR0 setting. The PTIO1~PTIO0 and PTTCLR1~PTTCLR0 are independent on and uninfluenced each other.

When the required edge transition appears on the input signal, the present value in the counter will be latched into the CCRA registers or CCRB registers and a PTM interrupt generated. Irrespective of what events occur on the input signal, the counter will continue to free run until the PTON bit changes from high to low. When a CCRP compare match occurs the counter will reset back to zero; in this way the CCRP value can be used to control the maximum counter value. When a CCRP compare match occurs from Comparator P, a PTM interrupt will also be generated. Counting the number of overflow interrupt signals from the CCRP can be a useful method in measuring long pulse widths. The PTIO1 and PTIO0 bits can select the active trigger edge on the input signal to be a rising edge, falling edge or both edge types. If the PTIO1 and PTIO0 bits are both set high, then no capture operation will take place irrespective of what happens on the input signal, however it must be noted that the counter will continue to run.

If the capture pulse width is less than two timer clock cycles, it may be ignored by hardware. The timer clock source must be equal to or less than 50MHz, otherwise the counter may fail to count. The PTCCLR, PTOC and PTPOL bits are not used in this Mode.





Capture Input Mode – PTTCLR[1:0]=00

Note: 1. PTM[1:0]=01, PTTCLR[1:0]=00 and active edge set by the PTIO[1:0] bits

2. A PTM Capture input active edge transfers the counter value to CCRA

3. Comparator P match will clear the counter

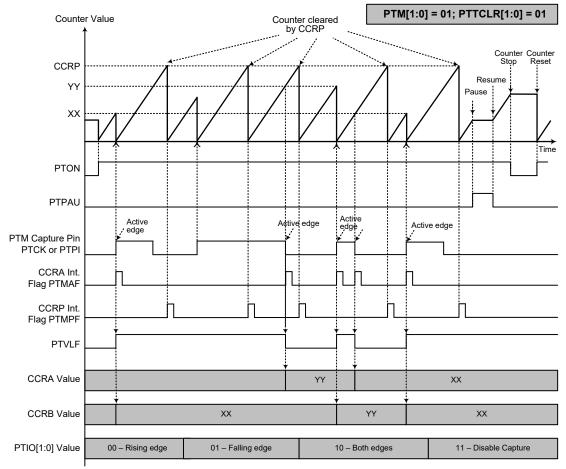
4. PTCCLR bit is not used

5. No output function – PTOC and PTPOL bits are not used

6. CCRP determines the counter value and the counter has a maximum count value when CCRP is equal to zero

7. Ignore the PTVLF bit status when PTTCLR[1:0]=00



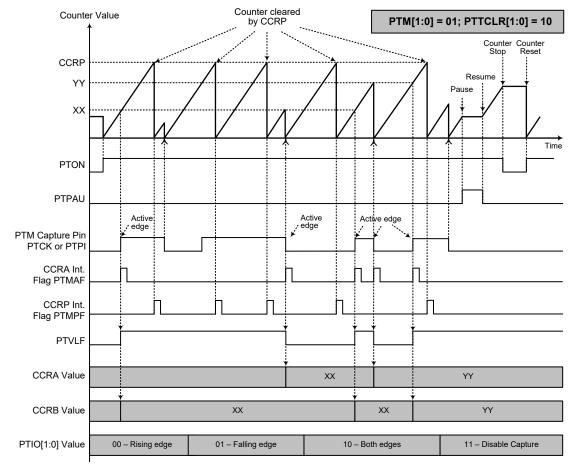


## Capture Input Mode – PTTCLR[1:0]=01

Note: 1. PTM[1:0]=01, PTTCLR[1:0]=01 and active edge set by the PTIO[1:0] bits

- 2. A PTM Capture input active edge transfers the counter value to CCRA or CCRB
- 3. Comparator P match or PTM capture input rising edge will clear the counter
- 4. PTCCLR bit is not used
- 5. No output function PTOC and PTPOL bits are not used
- 6. CCRP determines the counter value and the counter has a maximum count value when CCRP is equal to zero





Capture Input Mode – PTTCLR[1:0]=10

Note: 1. PTM[1:0]=01, PTTCLR[1:0]=10 and active edge set by the PTIO[1:0] bits

2. A PTM Capture input active edge transfers the counter value to CCRA or CCRB

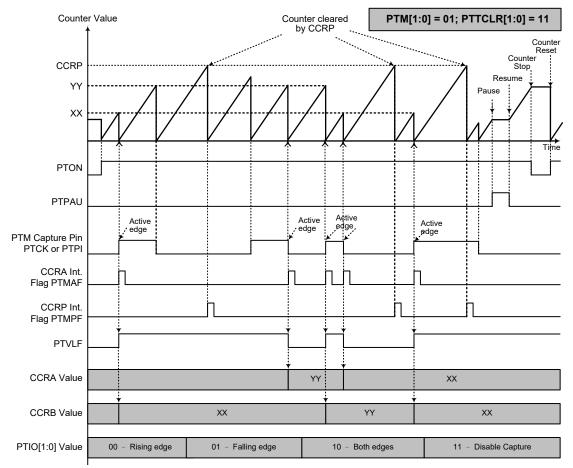
3. Comparator P match or PTM capture input falling edge will clear the counter

4. PTCCLR bit is not used

5. No output function - PTOC and PTPOL bits are not used

6. CCRP determines the counter value and the counter has a maximum count value when CCRP is equal to zero





Capture Input Mode – PTTCLR[1:0]=11

Note: 1. PTM[1:0]=01, PTTCLR[1:0]=11 and active edge set by the PTIO[1:0] bits

2. A PTM Capture input active edge transfers the counter value to CCRA or CCRB

3. Comparator P match or PTM capture input rising or falling edge will clear the counter

4. PTCCLR bit is not used

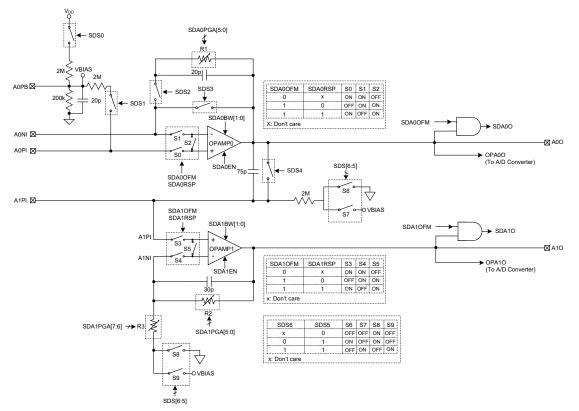
5. No output function - PTOC and PTPOL bits are not used

6. CCRP determines the counter value and the counter has a maximum count value when CCRP is equal to zero



# **Smoke Detector AFE**

The devices provide a Smoke Detector AFE circuit which can be used for optical signal detection in Smoke Detector applications. The circuit consists of two fully integrated Operational Amplifiers. The optical signal can be detected and processed by the operational amplifiers.



Note: For the BA45F5740-2 device, the A0PB external input pin is unavailable. Smoke Detector AFE Block Diagram

## Smoke Detector AFE Registers

Overall operation of the Smoke Detector AFE circuit is controlled using a series of registers. The SDSW register is used to control the switches on or off thus controlling the OPAMPs operating mode. The SDPGAC0 and SDPGAC1 registers are used to select the R1, R2 and R3 resistance. The SDAnC register where  $n=0\sim1$ , is used to control the SD OPAMPn enable/disable and bandwidth functions as well as stores the output status. The SDAnVOS register is used to select and control the SD OPAMPn input offset voltage calibration function.

Register				E	Bit			
Name	7	6	5	4	3	2	1	0
SDSW	—	SDS6	SDS5	SDS4	SDS3	SDS2	SDS1	SDS0
SDPGAC0	—	_	SDA0PGA5	SDA0PGA4	SDA0PGA3	SDA0PGA2	SDA0PGA1	SDA0PGA0
SDPGAC1	SDA1PGA7	SDA1PGA6	SDA1PGA5	SDA1PGA4	SDA1PGA3	SDA1PGA2	SDA1PGA1	SDA1PGA0
SDA0C	_	SDA0EN	SDA0O	_	_	_	SDA0BW1	SDA0BW0
SDA1C	_	SDA1EN	SDA10	_	_	_	SDA1BW1	SDA1BW0
SDA0VOS	SDA00FM	SDA0RSP	SDA0OF5	SDA00F4	SDA0OF3	SDA00F2	SDA00F1	SDA0OF0
SDA1VOS	SDA10FM	SDA1RSP	SDA10F5	SDA10F4	SDA10F3	SDA10F2	SDA10F1	SDA1OF0

Smoke Detector AFE Register List

Rev. 1.00

### SDSW Register

Bit 6~5

Bit	7	6	5	4	3	2	1	0
Name	—	SDS6	SDS5	SDS4	SDS3	SDS2	SDS1	SDS0
R/W	—	R/W						
POR		0	0	0	0	0	0	0

Bit 7 Unimplemented, read as "0"

SDS6-	SDS5-	Mode	control
2020~	SD35.	woue	control

00: External mode

01: AC coupling mode

10: External mode

- 11: DC coupling mode
- Note: 1. When the AC coupling mode is selected, the Switches of S6 and S8 will be on. As the S8 has on-resistance  $R_{ON_S8}$ , the OPAMP1 gain will be decreased. OPAMP1 Gain=1+R2/(R3+R\_{ON\_S8}), where  $R_{ON_S8}\approx 150\Omega$  @  $V_{DD}=5V$ ,  $R_{ON_S8}\approx 170\Omega$  @  $V_{DD}=3V$ .
  - 2. When the DC coupling mode is selected, the Switches of S7 and S9 will be on and connected to the VBIAS. Therefore the original settings of the resistance amplification value cannot be for reference which should be noted.
- Bit 4 **SDS4**: SDS4 switch on/off control
  - 0: Off

- Bit 3 SDS3: SDS3 switch on/off control
  - 0: Off
- 1: On Bit 2 **SDS2**: SDS2 switch on/off control 0: Off 1: On
- Bit 1 SDS1: SDS1 switch on/off control 0: Off 1: On Bit 0 SDS0: SDS0 switch on/off control
  - 0: Off 1: On

## SDPGAC0 Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	SDA0PGA5	SDA0PGA4	SDA0PGA3	SDA0PGA2	SDA0PGA1	SDA0PGA0
R/W	—	—	R/W	R/W	R/W	R/W	R/W	R/W
POR	_		0	0	0	0	0	0

Bit 7~6 Unimplemented, read as "0"

**SDA0PGA5~ SDA0PGA0**: R1 resistance control R1= SDA0PGA[5:0] × 100kΩ

## SDPGAC1 Register

Bit 5~0

Bit	7	6	5	4	3	2	1	0
Name	SDA1PGA7	SDA1PGA6	SDA1PGA5	SDA1PGA4	SDA1PGA3	SDA1PGA2	SDA1PGA1	SDA1PGA0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **SDA1PGA7 ~ SDA1PGA6**: R3 resistance control

<sup>00: 10</sup>kΩ



10: 30kΩ 11: 40kΩ

## Bit 5~0 **SDA1PGA5~ SDA1PGA0**: R2 resistance control R2= SDA1PGA[5:0] × $100k\Omega$

## SDA0C Register

Bit	7	6	5	4	3	2	1	0
Name	_	SDA0EN	SDA0O	_	—	_	SDA0BW1	SDA0BW0
R/W	—	R/W	R	—	—	_	R/W	R/W
POR	—	0	0	—	—		0	0
Bit 7	Unimple	emented, rea	ad as "0"					
Bit 6	<b>SDA0E</b> 0: Disa 1: Ena		MP0 enable	e or disabl	e control			
Bit 5	SDA0O	: SD OPAM	IP0 output	status (pos	itive logic	)		
	This bit	is read only						
	When th	e SDA0OF	M bit is set	t to 1, SDA	A0O is def	ined as SI	O OPAMP0	output status,
	refer to t	he "Operati	onal Ampl	ifier Input	Offset Cal	ibration" s	section for de	etails.
	When th	e SDA0OF	M bit is cle	ared to 0,	this bit wil	ll be fixed	at a low leve	1.
Bit 4~2	Unimple	emented, rea	ad as "0"					
Bit 1~0	SDA0B <sup>v</sup>	W1~SDA0I	BW0: SD C	PAMP0 b	andwidth	control		
	00: 5k	Hz						
	01:40	kHz						
	10:60							
	11: 2N	1Hz						
	Refer to	"Operation	al Amplifie	r Electrica	1 Characte	ristics" for	r details.	

Refer to "Operational Amplifier Electrical Characteristics" for details.

## SDA1C Register

		-								
B	Bit	7	6	5	4	3	2	1	0	
Na	ime	—	SDA1EN	SDA10	—	—	—	SDA1BW1	SDA1BW0	
R/	/W	_	R/W	R	_	_		R/W	R/W	
PC	OR	—	0	0	—		—	0	0	
Bit 7		Unimple	emented, rea	ad as "0"						
Bit 6 SDA1EN: SD OPAMP1 enable or disable control										
		1	N: SD OPA		e or disabl	e control				

	0: Disable
	1: Enable
Bit 5	SDA10: SD OPAMP1 output status (positive logic)
	This bit is read only.
	<ul><li>When the SDA1OFM bit is set to 1, SDA1O is defined as SD OPAMP1 output status, refer to the "Operational Amplifier Input Offset Calibration" section for details.</li><li>When the SDA1OFM bit is cleared to 0, this bit will be fixed at a low level.</li></ul>
Bit 4~2	Unimplemented, read as "0"
Bit 1~0	SDA1BW1~SDA1BW0: SD OPAMP1 bandwidth control
	00: 5kHz
	01: 40kHz
	10: 600kHz
	11: 2MHz

Refer to "Operational Amplifier Electrical Characteristics" for details.

## SDA0VOS Register

Bit	7	6	5	4	3	2	1	0
Name	SDA0OFM	SDA0RSP	SDA0OF5	SDA0OF4	SDA0OF3	SDA00F2	SDA0OF1	SDA0OF0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	1	0	0	0	0	0

Bit 7 **SDA00FM**: SD OPAMP0 normal operation or input offset voltage calibration mode selection

0: Normal operation

1: Offset calibration mode

## Bit 6 **SDA0RSP**: SD OPAMP0 input offset voltage calibration reference selection 0: Input reference voltage comes from A0NI

1: Input reference voltage comes from A0PI

#### Bit 5~0 SDA0OF5~SDA0OF0: SD OPAMP0 input offset voltage calibration control

This 6-bit field is used to perform the operational amplifier input offset calibration operation and the value for the SD OPAMP0 input offset Calibration can be restored into this bit field. More detailed information is described in the "Operational Amplifier Input Offset Calibration" section.

## SDA1VOS Register

Bit	7	6	5	4	3	2	1	0
Name	SDA10FM	SDA1RSP	SDA10F5	SDA10F4	SDA10F3	SDA10F2	SDA10F1	SDA1OF0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	1	0	0	0	0	0

Bit 7 **SDA10FM**: SD OPAMP1 normal operation or input offset voltage calibration mode selection

0: Normal operation

1: Offset calibration mode

- Bit 6 **SDA1RSP**: SD OPAMP1 input offset voltage calibration reference selection 0: Input reference voltage comes from A1NI
  - 1: Input reference voltage comes from A1PI
- Bit 5~0 **SDA1OF5~SDA1OF0**: SD OPAMP1 input offset voltage calibration control This 6-bit field is used to perform the operational amplifier input offset calibration operation and the value for the SD OPAMP1 input offset Calibration can be restored into this bit field. More detailed information is described in the "Operational Amplifier Input Offset Calibration" section.

## **Operational Amplifier Operation**

There are two fully integrated Operational Amplifiers in the devices, OPAMP1 and OPAMP0. These OPAMPs can be used for signal amplification according to specific user requirements. The OPAMPs can be disabled or enabled entirely under software control using the internal registers. With specific control registers, some OPAMP related applications can be more flexible and easier to be implemented, such as Unit Gain Buffer, Non-Inverting Amplifier, Inverting Amplifier and various kinds of filters, etc.

## **Operational Amplifier Input Offset Calibration**

Note that if the SD Operational Amplifier inputs are pin-shared with I/O pins, they should be configured as the SD Operational Amplifier input function before the Input Offset Calibration.

Step 1. Set SDAnOFM=1 and SDAnRSP=1, the SD Operational Amplifier n is now under the input offset Calibration mode. To make sure the V<sub>AnOS</sub> as minimal as possible after calibration, the input reference voltage in calibration should be the same as input DC operating voltage in normal operation.



- Step 2. Set SDAnOF[5:0]=000000 and then read the SDAnO bit.
- Step 3. Increase the SDAnOF[5:0] value by 1 and then read the SDAnO bit.

If the SDAnO bit state has not changed, then repeat Step 3 until the SDAnO bit state has changed.

If the SDAnO bit state has changed, record the SDAnOF[5:0] value as  $V_{\text{AnOS1}}$  and then go to Step 4.

- Step 4. Set SDAnOF[5:0]=111111 and read the SDAnO bit.
- Step 5. Decrease the SDAnOF[5:0] value by 1 and then read the SDAnO bit.

If the SDAnO bit state has not changed, then repeat Step 5 until the SDAnO bit state has changed.

If the SDAnO bit state has changed, record the SDAnOF[5:0] value as  $V_{\text{AnOS2}}$  and then go to Step 6.

 $\begin{array}{l} \mbox{Step 6. Restore the SD Operational Amplifier n input offset calibration value $V_{AnOS}$ into the $SDAnOF[5:0]$ bit field. The offset Calibration procedure is now finished. $V_{AnOS}=(V_{AnOS1}+V_{AnOS2})/2.$ \end{array}$ 

If  $(V_{AnOS1}+V_{AnOS2})/2$  is not integral, discard the decimal.

# Analog to Digital Converter

The need to interface to real world analog signals is a common requirement for many electronic systems. However, to properly process these signals by a microcontroller, they must first be converted into digital signals by A/D converters. By integrating the A/D conversion electronic circuitry into the microcontroller, the need for external components is reduced significantly with the corresponding follow-on benefits of lower costs and reduced component space requirements.

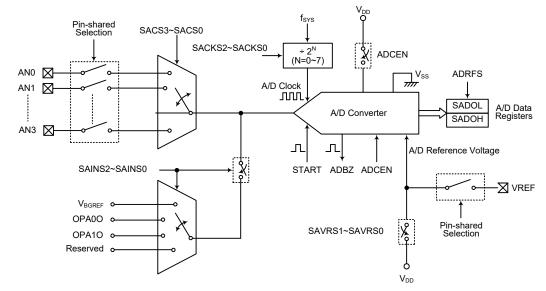
## A/D Converter Overview

The devices contain a multi-channel analog to digital converter which can directly interface to external analog signals, such as that from sensors or other control signals and convert these signals directly into a 12-bit digital value. It also can convert the internal signals, the high performance bandgap reference voltage,  $V_{BGREF}$ , the SD operational amplifier 0 output signal OPA0O and SD operational amplifier 1 output signal OPA1O, into a 12-bit digital value. The external or internal analog signal to be converted is determined by the SAINS2~SAINS0 bits together with the SACS3~SACS0 bits. More detailed information about the A/D input signal is described in the "A/D Converter Control Registers" and "A/D Converter Input Signals" sections respectively.

Device	External Input Channels	Internal Signals	Channel Select Bits		
BA45F5740	AN0~AN3	V <sub>BGREF</sub> , OPA0O, OPA1O	SAINS2~SAINS0,		
BA45F5740-2	AN0, AN3	VBGREF, OFAUO, OFATO	SACS3~SACS0		

The accompanying block diagram shows the overall internal structure of the A/D converter, together with its associated registers.





Note: For the BA45F5740-2 device, the AN1~AN2 external input pins are unavailable. A/D Converter Structure

## A/D Converter Register Description

Overall operation of the A/D converter is controlled using several registers. A read only register pair exists to store the A/D converter data 12-bit value. The remaining two registers are control registers which setup the operating and control function of the A/D converter.

Deviator Nome		Bit											
Register Name	7	6	5	4	3	2	1	0					
SADOH (ADRFS=0)	D11	D10	D9	D8	D7	D6	D5	D4					
SADOH (ADRFS=1)	—	_	—	_	D11	D10	D9	D8					
SADOL (ADRFS=0)	D3	D2	D1	D0	_	_	_	_					
SADOL (ADRFS=1)	D7	D6	D5	D4	D3	D2	D1	D0					
SADC0	START	ADBZ	ADCEN	ADRFS	SACS3	SACS2	SACS1	SACS0					
SADC1	SAINS2	SAINS1	SAINS0	SAVRS1	SAVRS0	SACKS2	SACKS1	SACKS0					

### A/D Converter Register List

#### A/D Converter Data Registers – SADOL, SADOH

As the devices contain an internal 12-bit A/D converter, it requires two data registers to store the converted value. These are a high byte register, known as SADOH, and a low byte register, known as SADOL. After the conversion process takes place, these registers can be directly read by the microcontroller to obtain the digitised conversion value. As only 12 bits of the 16-bit register space is utilised, the format in which the data is stored is controlled by the ADRFS bit in the SADC0 register as shown in the accompanying table. D0~D11 are the A/D conversion result data bits. Any unused bits will be read as zero. Note that A/D Converter data register contents will be unchanged if the A/D converter is disabled.

ADRFS		SADOH								SADOL						
	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
0	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	0	0	0	0
1	0	0	0	0	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0

A/D Data Registers



# A/D Converter Control Registers – SADC0, SADC1

To control the function and operation of the A/D converter, two control registers known as SADC0 and SADC1 are provided. These 8-bit registers define functions such as the selection of which analog channel is connected to the internal A/D converter, the digitised data format, the A/D clock source as well as controlling the start function and monitoring the A/D converter busy status. As the devices contain only one actual analog to digital converter hardware circuit, each of the external or internal analog signal inputs must be routed to the converter. The SACS3~SACS0 bits in the SADC0 register are used to determine which external channel input is selected to be converted. The SAINS2~SAINS0 bits in the SADC1 register are used to determine that the analog signal to be converted comes from the internal analog signal or external analog channel input.

The relevant pin-shared function selection bits determine which pins on I/O Ports are used as analog inputs for the A/D converter input and which pins are not to be used as the A/D converter input. When the pin is selected to be an A/D input, its original function whether it is an I/O or other pin-shared function will be removed. In addition, any internal pull-high resistor connected to the pin will be automatically removed if the pin is selected to be an A/D converter input.

Bit	7	6	5	4	3	2	1	0				
Name	START	ADBZ	ADCEN	ADRFS	SACS3	SACS2	SACS1	SACS0				
R/W	R/W	R	R/W	R/W	R/W	R/W	R/W	R/W				
POR	0	0	0	0	0	0	0	0				
Bit 7	0→1-	Start the A →0: Start is used to ir			on process.	The bit is	normally lo	w but if s				
	high and then cleared low again, the A/D converter will initiate a conversion process.											
Bit 6	ADBZ: A/D converter busy flag 0: No A/D conversion is in progress 1: A/D conversion is in progress											
	This read only flag is used to indicate whether the A/D conversion is in progress or not. When the START bit is set from low to high and then to low again, the ADBZ flag will be set to 1 to indicate that the A/D conversion is initiated. The ADBZ flag will be cleared to 0 after the A/D conversion is complete.											
Bit 5	ADCEN 0: Disa 1: Ena		erter functi	on enable c	control							
	This bit controls the A/D internal function. This bit should be set to one to enabl the A/D converter. If the bit is set low, then the A/D converter will be switche off reducing the devices power consumption. When the A/D converter function is disabled, the contents of the A/D data register pair known as SADOH and SADO will be unchanged.											
Bit 4	0: A/D 1: A/D This bit	converter	data format data format e format o	$\rightarrow$ SADOI $\rightarrow$ SADOI f the 12-bi	H=D[11:4]; H=D[11:8]; t converted	SADOL=I SADOL=I A/D value	D[7:0]	o A/D da				
3it 3~0	SACS3~ 0000: 2 0001: 2 0010: 2 0010: 2 0100~ Note: F	- <b>SACS0</b> : A AN0 AN1 AN2 AN3 1111: Non	/D converte	er external a	analog chan put will be	nel input so		ut pins a:				

#### SADC0 Register

## SADC1 Register

Bit	7	6	5	4	3	2	1	0
Name	SAINS2	SAINS1	SAINS0	SAVRS1	SAVRS0	SACKS2	SACKS1	SACKS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~5 SAINS2~SAINS0: A/D converter input signal select

000: External input – External analog channel input

001: Internal input – Internal high performance bandgap reference voltage,  $V_{\text{BGREF}}$ 

010: Internal input – Internal SD operational amplifier 0 output signal, OPA00

011: Internal input - Internal SD operational amplifier 1 output signal, OPA10

100: Reserved

101~111: External input - External analog channel input

Care must be taken if the SAINS2~SAINS0 bits are set from "001" to "011" to select the internal analog signal to be converted. When the internal analog signal is selected to be converted, the external input pin must never be selected as the A/D input signal by properly setting the SACS3~SACS0 bits. Otherwise, the external channel input will be connected together with the internal analog signal. This will result in unpredictable situations such as an irreversible damage.

Bit 4~3 SAVRS1~SAVRS0: A/D converter reference voltage select

00: From external VREF pin

01: Internal A/D converter power,  $V_{\text{DD}}$ 

1x: From external VREF pin

These bits are used to select the A/D converter reference voltage. Care must be taken if the SAVRS1~SAVRS0 bits are set to "01" to select the internal A/D converter power as the reference voltage source. When the internal A/D converter power is selected as the reference voltage, the VREF pin cannot be configured as the reference voltage input by properly configuring the corresponding pin-shared function control bits. Otherwise, the external input voltage on VREF pin will be connected to the internal A/D converter power.

Bit 2~0 SACKS2~SACKS0: A/D conversion clock source select

 $\begin{array}{l} 000: \, f_{SYS} \\ 001: \, f_{SYS}/2 \\ 010: \, f_{SYS}/4 \\ 011: \, f_{SYS}/8 \\ 100: \, f_{SYS}/16 \\ 101: \, f_{SYS}/32 \\ 110: \, f_{SYS}/64 \\ 111: \, f_{SYS}/128 \end{array}$ 

These three bits are used to select the clock source for the A/D converter.

# **A/D Converter Operation**

The START bit in the SADC0 register is used to start the AD conversion. When the microcontroller sets this bit from low to high and then low again, an analog to digital conversion cycle will be initiated.

The ADBZ bit in the SADC0 register is used to indicate whether the analog to digital conversion process is in progress or not. This bit will be automatically set to 1 by the microcontroller after an A/D conversion is successfully initiated. When the A/D conversion is complete, the ADBZ will be cleared to 0. In addition, the corresponding A/D interrupt request flag will be set in the interrupt control register, and if the interrupts are enabled, an appropriate internal interrupt signal will be generated. This A/D internal interrupt signal will direct the program flow to the associated A/D internal interrupt address for processing. If the A/D internal interrupt is disabled, the microcontroller can poll the ADBZ bit in the SADC0 register to check whether it has been cleared as an alternative



method of detecting the end of an A/D conversion cycle.

The clock source for the A/D converter, which originates from the system clock  $f_{SYS}$ , can be chosen to be either  $f_{SYS}$  or a subdivided version of  $f_{SYS}$ . The division ratio value is determined by the SACKS2~SACKS0 bits in the SADC1 register. Although the A/D clock source is determined by the system clock  $f_{SYS}$  and by bits SACKS2~SACKS0, there are some limitations on the maximum A/D clock source speed that can be selected. As the recommended range of permissible A/D clock period,  $t_{ADCK}$ , is from 0.5µs to 10µs, care must be taken for system clock frequencies. For example, as the system clock operates at a frequency of 8MHz, the SACKS2~SACKS0 bits should not be set to 000, 001 or 111. Doing so will give A/D clock periods that are less or larger than the minimum or maximum A/D clock period which may result in inaccurate A/D conversion values. Refer to the following table for examples, where values marked with an asterisk \* show where, special care must be taken, as the values may be less or larger than the specified A/D Clock Period range.

		A/D Clock Period (t <sub>ADCK</sub> )											
f <sub>sys</sub>	SACKS [2:0]=000 (f <sub>SYS</sub> )	SACKS [2:0]=001 (f <sub>SYS</sub> /2)	SACKS [2:0]=010 (f <sub>SYS</sub> /4)	SACKS [2:0]=011 (f <sub>SYS</sub> /8)	SACKS [2:0]=100 (f <sub>SYS</sub> /16)	SACKS [2:0]=101 (f <sub>SYS</sub> /32)	SACKS [2:0]=110 (f <sub>SYS</sub> /64)	SACKS [2:0]=111 (f <sub>SYS</sub> /128)					
1MHz	1µs	2µs	4µs	8µs	16µs *	32µs *	64µs *	128µs *					
2MHz	500ns	1µs	2µs	4µs	8µs	16µs *	32µs *	64µs *					
4MHz	250ns *	500ns	1µs	2µs	4µs	8µs	16µs *	32µs *					
8MHz	125ns *	250ns *	500ns	1µs	2µs	4µs	8µs	16µs *					

A/D Clock Period Examples

Controlling the power on/off function of the A/D converter circuitry is implemented using the ADCEN bit in the SADC0 register. This bit must be set high to power on the A/D converter. When the ADCEN bit is set high to power on the A/D converter internal circuitry a certain delay, as indicated in the timing diagram, must be allowed before an A/D conversion is initiated. Even if no pins are selected for use as A/D inputs, if the ADCEN bit is high, then some power will still be consumed. In power conscious applications it is therefore recommended that the ADCEN is set low to reduce power consumption when the A/D converter function is not being used.

# A/D Converter Reference Voltage

The reference voltage supply to the A/D converter can be supplied from the power supply  $V_{DD}$ , or from an external reference source supplied on pin VREF. The desired selection is made using the SAVRS1 and SAVRS0 bits. When the SAVRS bit field is set to "01", the A/D converter reference voltage will come from the V<sub>DD</sub>. Otherwise, if the SAVRS bit field is set to any other value except "01", the A/D converter reference voltage will come from the VREF pin. As the VREF pin is pin-shared with other functions, when the VREF pin is selected as the reference voltage supply pin, the VREF pin should be configured as the reference voltage input function by properly configuring the corresponding pin-shared function selection bits PAS15~PAS14. However, if the internal A/D converter power V<sub>DD</sub> is selected as the reference voltage, the VREF pin must not be configured as the reference voltage.

# A/D Converter Input Signals

All the external A/D analog channel input pins are pin-shared with the I/O pins as well as other functions. The corresponding control bits for each A/D external input pin in the PxS0 and PxS1 registers determine whether the input pins are setup as A/D converter analog input channel or whether they have other functions. If the pin is setup to be as an A/D analog channel input, the

original pin functions will be disabled. In this way, pins can be changed under program control to change their function between A/D inputs and other functions. All pull high resistors, which are setup through register programming, will be automatically disconnected if the pins are setup as A/D inputs. Note that it is not necessary to first setup the A/D pin as an input in the port control register to enable the A/D input as when the pin-shared function control bits enable an A/D input, the status of the port control register will be overridden.

There are some internal analog signals derived from the high performance bandgap reference voltage  $V_{BGREF}$ , the SD operational amplifier 0 output signal OPA0O and the SD operational amplifier 1 output signal OPA1O, which can be connected to the A/D converter as the analog input signal by configuring the SAINS2~SAINS0 bits. If the external channel input is selected to be converted, the SAINS2~SAINS0 bits should be set to "000" or "101~111" and the SACS3~SACS0 bits can determine which external channel is selected. If the internal analog signal is selected to be converted, the SACS3~SACS0 bits must be configured with an appropriate value to switch off the external analog channel input. Otherwise, the internal analog signal will be connected together with the external channel input. This will result in unpredictable situations.

This  $V_{BGREF}$  is the internal high performance bandgap voltage reference with driver capability. It has accurate voltage reference output when input supply voltage  $V_{DD}$  change or temperature variation. And, this bandgap will startup at a low supply voltage. Therefore, this voltage reference has high power supply rejection ratio (PSRR) for low dropout regulator (LDO) is presented.

SAINS[2:0]	SACS[3:0]	Input Signals	Description
000, 101~111	0000~0011	AN0~AN3 (BA45F5740) AN0, AN3 (BA45F5740-2)	External pin analog input
101~111	0100~1111	_	Non-existed channel, input is floating.
001	0100~1111	V <sub>BGREF</sub>	Internal high performance Bandgap reference voltage
010	0100~1111	OPA0O	Internal SD operational amplifier 0 output signal
011	0100~1111	OPA10	Internal SD operational amplifier 1 output signal
100	0100~1111		Reserved

A/D Converter Input Signal Selection

## VBGRC Register

Bit	7	6	5	4	3	2	1	0
Name	—	_	—	—	—	_	—	VBGREN
R/W	—	—	—	—	_	—	_	R/W
POR	_	_	—	—	_	_	—	0

Bit 7~1 Unimplemented, read as "0"

Bit 0 VBGREN: Bandgap enable or disable control

0: Disable

1: Enable

When the VBGREN bit is cleared to zero, the Bandgap voltage output is in a high impedance state.

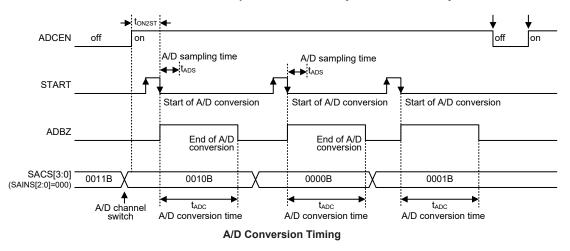
# **Conversion Rate and Timing Diagram**

A complete A/D conversion contains two parts, data sampling and data conversion. The data sampling which is defined as  $t_{ADS}$  takes 4 A/D clock cycles and the data conversion takes 12 A/D clock cycles. Therefore a total of 16 A/D clock cycles for an external input A/D conversion which is defined as  $t_{ADC}$  are necessary.

Maximum single A/D conversion rate=A/D clock period/16



The accompanying diagram shows graphically the various stages involved in an analog to digital conversion process and its associated timing. After an A/D conversion process has been initiated by the application program, the microcontroller internal hardware will begin to carry out the conversion, during which time the program can continue with other functions. The time taken for the A/D conversion is 16  $t_{ADCK}$  clock cycles where  $t_{ADCK}$  is equal to the A/D clock period.



# Summary of A/D Conversion Steps

The following summarises the individual steps that should be executed in order to implement an A/D conversion process.

• Step 1

Select the required A/D conversion clock by correctly programming bits SACKS2~SACKS0 in the SADC1 register.

• Step 2

Enable the A/D by setting the ADCEN bit in the SADC0 register to 1.

• Step 3

Select which signal is to be connected to the internal A/D converter by correctly configuring the SAINS2~SAINS0 bits

Select the external channel input to be converted, go to Step 4. Select the internal analog signal to be converted, go to Step 5.

• Step 4

If the A/D input signal comes from the external channel input selected by configuring the SAINS2~SAINS0 bits field, the corresponding pins should be configured as A/D input function by configuring the relevant pin-shared function control bits. The desired analog channel then should be selected by configuring the SACS3~SACS0 bits field. After this step, go to Step 6.

• Step 5

Before the A/D input signal is selected to come from the internal analog signal by configuring the SAINS2~SAINS0 bits field, the corresponding external input pin must be switched to a non-existed channel input by properly configured the SACS3~SACS0 bits. The desired internal analog signal then can be selected by configuring the SAINS2~SAINS0 bits field. After this step, go to Step 6.

• Step 6

Select the reference voltage source by configuring the SAVRS1~SAVRS0 bits in the SADC1 register.



• Step 7

Select A/D converter output data format by setting the ADRFS bit in the SADC0 register.

• Step 8

If A/D conversion interrupt is used, the interrupt control registers must be correctly configured to ensure the A/D interrupt function is active. The master interrupt control bit, EMI, and the A/D conversion interrupt control bit, ADE, must both be set high in advance.

• Step 9

The A/D conversion procedure can now be initialized by setting the START bit from low to high and then low again.

• Step 10

If A/D conversion is in progress, the ADBZ flag will be set high. After the A/D conversion process is complete, the ADBZ flag will go low and then the output data can be read from SADOH and SADOL registers.

Note: When checking for the end of the conversion process, if the method of polling the ADBZ bit in the SADC0 register is used, the interrupt enable step above can be omitted.

# **Programming Considerations**

During microcontroller operations where the A/D converter is not being used, the A/D internal circuitry can be switched off to reduce power consumption, by clearing bit ADCEN to 0 in the SADC0 register. When this happens, the internal A/D converter circuits will not consume power irrespective of what analog voltage is applied to their input lines. If the A/D converter input lines are used as normal I/O pins, then care must be taken as if the input voltage is not at a valid logic level, then this may lead to some increase in power consumption.

# **A/D Conversion Function**

As the devices contain a 12-bit A/D converter, its full-scale converted digitised value is equal to FFFH. Since the full-scale analog input value is equal to the actual A/D converter reference voltage,  $V_{REF}$ , this gives a single bit analog input value of  $V_{REF}$  divided by 4096.

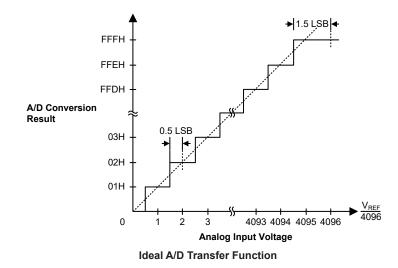
$$1 \text{ LSB}{=}V_{\text{REF}} \div 4096$$

The A/D Converter input voltage value can be calculated using the following equation:

A/D input voltage=A/D output digital value  $\times$   $V_{\text{REF}} \div 4096$ 

The diagram shows the ideal transfer function between the analog input value and the digitised output value for the A/D converter. Except for the digitised zero value, the subsequent digitised values will change at a point 0.5 LSB below where they would change without the offset, and the last full scale digitised value will change at a point 1.5 LSB below the  $V_{REF}$  level. Note that here the  $V_{REF}$  voltage is the actual A/D converter reference voltage determined by the SAVRS field.





# A/D Conversion Programming Examples

The following two programming examples illustrate how to setup and implement an A/D conversion. In the first example, the method of polling the ADBZ bit in the SADC0 register is used to detect when the conversion cycle is complete, whereas in the second example, the A/D interrupt is used to determine when the conversion is complete.

### Example: using an ADBZ polling method to detect the end of conversion

LAU		ADE		disable ADC interrupt
	mov	a,03H	,	
	mov	SADC1,a	;	select f <sub>sys</sub> /8 as A/D clock
	set	ADCEN		
	mov	a,02h	;	setup PAS1 register to configure pin ANO
	mov	PAS1,a		
	mov	a,20h		
	mov	SADCO, a	;	enable and connect ANO channel to A/D converter
	:			
	star	t conversion:		
	clr	START	;	high pulse on start bit to initiate conversion
	set	START	;	reset A/D
	clr	START	;	start A/D
	poll	ing EOC:		
	SZ	_	;	poll the SADCO register ADBZ bit to detect end of A/D conversion
	jmp	polling_EOC	;	continue polling
	mov	a,SADOL	;	read low byte conversion result value
	mov	SADOL_buffer,a	;	save result to user defined register
				read high byte conversion result value
	mov	SADOH_buffer,a	;	save result to user defined register
	:			
	:			
	jmp	start_conversion	;	start next A/D conversion
_				
Exa		•		method to detect the end of conversion
	clr	ADE	;	disable ADC interrupt

	0		
clr	ADE	;	disable ADC interrupt
mov	a,03H		
mov	SADC1,a	;	select f <sub>sys</sub> /8 as A/D clock
set	ADCEN		
mov	a, 02h	;	setup PAS1 register to configure pin ANO
mov	PAS1,a		



mov a,20h

```
mov SADCO,a
                       ; enable and connect ANO channel to A/D converter
Start conversion:
clr START
                       ; high pulse on START bit to initiate conversion
set START
                       ; reset A/D
clr START
                       ; start A/D
clr ADF
                       ; clear ADC interrupt request flag
set ADE
                       ; enable ADC interrupt
set EMI
                       ; enable global interrupt
:
:
; ADC interrupt service routine
ADC ISR:
mov acc stack, a
                  ; save ACC to user defined memory
mov a, STATUS
mov status stack, a ; save STATUS to user defined memory
:
:
mov a,SADOL
                       ; read low byte conversion result value
mov SADOL_buffer,a
                     ; save result to user defined register
mov a,SADOH
                     ; read high byte conversion result value
mov SADOH buffer,a
                     ; save result to user defined register
:
:
EXIT INT ISR:
mov a, status stack
mov STATUS,a
                     ; restore STATUS from user defined memory
mov a,acc stack
                      ; restore ACC from user defined memory
reti
```

# Universal Serial Interface Module – USIM

The devices contain a Universal Serial Interface Module, which includes the four-line SPI interface, the two-line I<sup>2</sup>C interface and the two-line UART interface types, to allow an easy method of communication with external peripheral hardware. Having relatively simple communication protocols, these serial interface types allow the microcontroller to interface to external SPI, I<sup>2</sup>C or UART based hardware such as sensors, Flash or EEPROM memory, etc. The USIM interface pins are pin-shared with other I/O pins therefore the USIM interface functional pins must first be selected using the corresponding pin-shared function selection bits. As all the interface types share the same pins and registers, the choice of whether the UART, SPI or I<sup>2</sup>C type is used is made using the UART mode selection bit, named UMD, and the SPI/I<sup>2</sup>C operating mode control bits, named SIM2~SIM0, in the SIMC0 register. These pull-high resistors of the USIM pin-shared I/O are selected using pull-high control registers when the USIM function is enabled and the corresponding pins are used as USIM input pins.

# **SPI Interface**

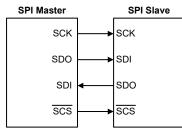
The SPI interface is often used to communicate with external peripheral devices such as sensors, Flash or EEPROM memory devices etc. Originally developed by Motorola, the four line SPI interface is a synchronous serial data interface that has a relatively simple communication protocol simplifying the programming requirements when communicating with external hardware devices.

The communication is full duplex and operates as a slave/master type, where the device can be either master or slave. Although the SPI interface specification can control multiple slave devices from a single master, but the device provides only one  $\overline{\text{SCS}}$  pin. If the master needs to control multiple slave devices from a single master, the master can use I/O pin to select the slave devices.



## **SPI Interface Operation**

The SPI interface is a full duplex synchronous serial data link. It is a four line interface with pin names SDI, SDO, SCK and  $\overline{SCS}$ . Pins SDI and SDO are the Serial Data Input and Serial Data Output lines, the SCK pin is the Serial Clock line and  $\overline{SCS}$  is the Slave Select line. As the SPI interface pins are pin-shared with normal I/O pins and with the I<sup>2</sup>C/UART function pins, the SPI interface pins must first be selected by configuring the pin-shared function selection bits and setting the correct bits in the SIMC0 and SIMC2 registers. Communication between devices connected to the SPI interface is carried out in a slave/master mode with all data transfer initiations being implemented by the master. The Master also controls the clock signal. As the device only contains a single  $\overline{SCS}$  pin only one slave device can be utilized. The  $\overline{SCS}$  pin is controlled by software, set CSEN bit to 1 to enable  $\overline{SCS}$  pin function, set CSEN bit to 0 the  $\overline{SCS}$  pin will be floating state.

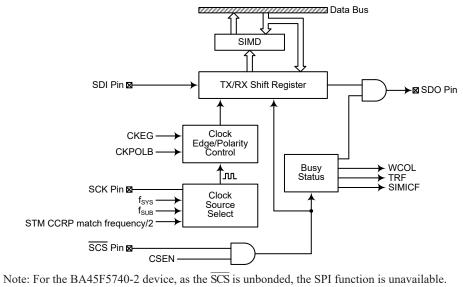


SPI Master/Slave Connection

The SPI function offers the following features:

- Full duplex synchronous data transfer
- Both Master and Slave modes
- · LSB first or MSB first data transmission modes
- Transmission complete flag
- · Rising or falling active clock edge

The status of the SPI interface pins is determined by a number of factors such as whether the device is in the master or slave mode and upon the condition of certain control bits such as CSEN and SIMEN.



SPI Block Diagram



## **SPI Registers**

There are three internal registers which control the overall operation of the SPI interface. These are the SIMD data register and two control registers, SIMC0 and SIMC2. Note that the SIMC2 and SIMD registers and their POR values are only available when the SPI mode is selected by properly configuring the UMD and SIM2~SIM0 bits in the SIMC0 register.

Register		Bit											
Name	7	6	5	4	3	2	1	0					
SIMC0	SIM2	SIM1	SIM0	UMD	SIMDEB1	SIMDEB0	SIMEN	SIMICF					
SIMC2	D7	D6	CKPOLB	CKEG	MLS	CSEN	WCOL	TRF					
SIMD	D7	D6	D5	D4	D3	D2	D1	D0					

### **SPI Data Register**

The SIMD register is used to store the data being transmitted and received. The same register is used by both the SPI and I<sup>2</sup>C functions. Before the device writes data to the SPI bus, the actual data to be transmitted must be placed in the SIMD register. After the data is received from the SPI bus, the device can read it from the SIMD register. Any transmission or reception of data from the SPI bus must be made via the SIMD register.

### SIMD Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	х	х	х	х	х	х	х	х

"x": unknown

Bit 7~0 **D7~D0**: USIM SPI/I<sup>2</sup>C data register bit 7 ~ bit 0

## **SPI Control Registers**

There are also two control registers for the SPI interface, SIMC0 and SIMC2. The SIMC0 register is used to control the enable/disable function and to set the data transmission clock frequency. The SIMC2 register is used for other control functions such as LSB/MSB selection, write collision flag etc.

### SIMC0 Register

Bit	7	6	5	4	3	2	1	0
Name	SIM2	SIM1	SIM0	UMD	SIMDEB1	SIMDEB0	SIMEN	SIMICF
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	1	1	1	0	0	0	0	0

Bit 7~5 SIM2~SIM0: USIM SPI/I<sup>2</sup>C Operating Mode Control (SPI unavailable for

BA45F5740-2)

000: SPI master mode; SPI clock is fsys/4

001: SPI master mode; SPI clock is  $f_{SYS}/16$ 

010: SPI master mode; SPI clock is  $f_{SYS}/64$ 

011: SPI master mode; SPI clock is  $f_{\text{SUB}}$ 

100: SPI master mode; SPI clock is STM CCRP match frequency/2

101: SPI slave mode

110: I<sup>2</sup>C slave mode

111: Unused mode

When the UMD bit is cleared to zero, these bits setup the SPI or I<sup>2</sup>C operating mode of the USIM function. As well as selecting if the I<sup>2</sup>C or SPI function, they are used



to control the SPI Master/Slave selection and the SPI Master clock frequency. The SPI clock is a function of the system clock but can also be chosen to be sourced from STM and  $f_{SUB}$ . If the SPI Slave Mode is selected then the clock will be supplied by an external Master device.

Bit 4 UMD: UART mode selection bit

0: SPI or I<sup>2</sup>C mode

1: UART mode

This bit is used to select the UART mode. When this bit is cleared to zero, the actual SPI or  $I^2C$  mode can be selected using the SIM2~SIM0 bits. Note that the UMD bit must be cleared to zero for SPI or  $I^2C$  mode.

# Bit 3~2 SIMDEB1~SIMDEB0: I<sup>2</sup>C Debounce Time Selection

These bits are only available when the USIM is configured to operate in the I<sup>2</sup>C mode. Refer to the I<sup>2</sup>C register section.

# Bit 1 SIMEN: USIM SPI/I<sup>2</sup>C Enable Control

0: Disable 1: Enable

The bit is the overall on/off control for the USIM SPI/I<sup>2</sup>C interface. When the SIMEN bit is cleared to zero to disable the USIM SPI/I<sup>2</sup>C interface, the SDI, SDO, SCK and SCS, or SDA and SCL lines will lose their SPI or I<sup>2</sup>C function and the USIM operating current will be reduced to a minimum value. When the bit is high the USIM SPI/I<sup>2</sup>C interface is enabled. If the USIM is configured to operate as an SPI interface via the UMD and SIM2~SIM0 bits, the contents of the SPI control registers will remain at the previous settings when the SIMEN bit changes from low to high and should therefore be first initialised by the application program. If the USIM is configured to operate as an I<sup>2</sup>C interface via the UMD and SIM2~SIM0 bits such as HTX and TXAK will remain at the previous settings and should therefore be first initialised by the application program while the relevant I<sup>2</sup>C flags such as HCF, HAAS, HBB, SRW and RXAK will be set to their default states.

Bit 0

## SIMICF: USIM SPI Incomplete Flag

0: USIM SPI incomplete condition is not occurred

1: USIM SPI incomplete condition is occurred

This bit is only available when the USIM is configured to operate in an SPI slave mode. If the SPI operates in the slave mode with the SIMEN and CSEN bits both being set to 1 but the  $\overline{SCS}$  line is pulled high by the external master device before the SPI data transfer is completely finished, the SIMICF bit will be set to 1 together with the TRF bit. When this condition occurs, the corresponding interrupt will occur if the interrupt function is enabled. However, the TRF bit will not be set to 1 if the SIMICF bit is set to 1 by software application program.

## SIMC2 Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	CKPOLB	CKEG	MLS	CSEN	WCOL	TRF
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **D7~D6**: Undefined bits

These bits can be read or written by the application program.

Bit 5

CKPOLB: SPI clock line base condition selection

0: The SCK line will be high when the clock is inactive

1: The SCK line will be low when the clock is inactive

The CKPOLB bit determines the base condition of the clock line, if the bit is high, then the SCK line will be low when the clock is inactive. When the CKPOLB bit is low, then the SCK line will be high when the clock is inactive.

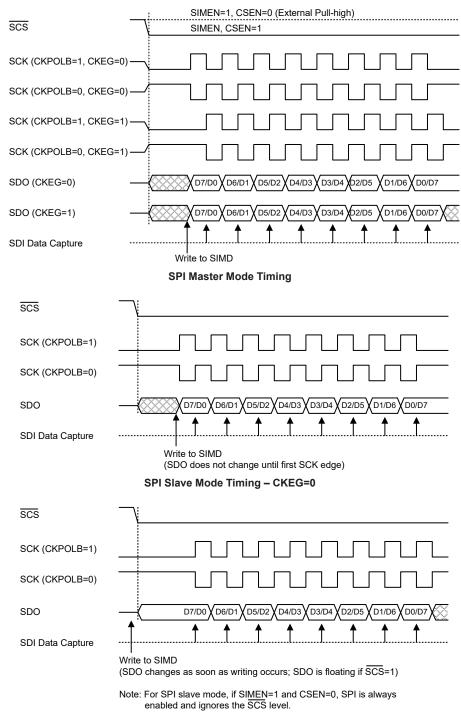
Bit 4	CKEG: SPI SCK clock active edge type selection
	CKPOLB=0 0: SCK is high base level and data capture at SCK rising edge
	1: SCK is high base level and data capture at SCK falling edge
	CKPOLB=1 0: SCK is low base level and data capture at SCK falling edge
	1: SCK is low base level and data capture at SCK rising edge
	The CKEG and CKPOLB bits are used to setup the way that the clock signal outputs and inputs data on the SPI bus. These two bits must be configured before data transfer is executed otherwise an erroneous clock edge may be generated. The CKPOLB bit determines the base condition of the clock line, if the bit is high, then the SCK line will be low when the clock is inactive. When the CKPOLB bit is low, then the SCK line will be high when the clock is inactive. The CKEG bit determines active clock edge type which depends upon the condition of CKPOLB bit.
Bit 3	MLS: SPI data shift order
	0: LSB first
	1: MSB first This is the data shift select bit and is used to select how the data is transferred, either
	MSB or LSB first. Setting the bit high will select MSB first and low for LSB first.
Bit 2	CSEN: SPI $\overline{SCS}$ pin control
	0: Disable 1: Enable
	The CSEN bit is used as an enable/disable for the $\overline{\text{SCS}}$ pin. If this bit is low, then the $\overline{\text{SCS}}$ pin will be disabled and placed into a floating condition. If the bit is high the $\overline{\text{SCS}}$ pin will be enabled and used as a select pin.
Bit 1	WCOL: SPI write collision flag
	0: No collision 1: Collision
	The WCOL flag is used to detect if a data collision has occurred. If this bit is high it means that data has been attempted to be written to the SIMD register during a data transfer operation. This writing operation will be ignored if data is being transferred. The bit can be cleared by the application program.
Bit 0	<b>TRF</b> : SPI Transmit/Receive complete flag 0: SPI data is being transferred 1: SPI data transmission is completed
	The TRF bit is the Transmit/Receive Complete flag and is set "1" automatically when an SPI data transmission is completed, but must set to "0" by the application program. It can be used to generate an interrupt.

# **SPI** Communication

After the SPI interface is enabled by setting the SIMEN bit high, then in the Master Mode, when data is written to the SIMD register, transmission/reception will begin simultaneously. When the data transfer is completed, the TRF flag will be set high automatically, but must be cleared using the application program. In the Slave Mode, when the clock signal from the master has been received, any data in the SIMD register will be transmitted and any data on the SDI pin will be shifted into the SIMD register. The master should output an SCS signal to enable the slave devices before a clock signal is provided. The slave data to be transferred should be well prepared at the appropriate moment relative to the SCK signal depending upon the configurations of the CKPOLB bit and CKEG bit. The accompanying timing diagram shows the relationship between the slave data and SCK signal for various configurations of the CKPOLB and CKEG bits.

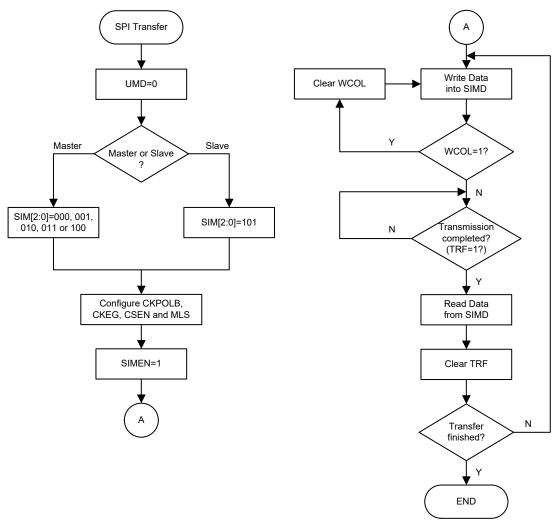
The SPI will continue to function in certain IDLE Modes if the clock source used by the SPI interface is still active.





SPI Slave Mode Timing – CKEG=1





**SPI Transfer Control Flowchart** 

# SPI Bus Enable/Disable

To enable the SPI bus, set CSEN=1 and  $\overline{SCS}$ =0, then wait for data to be written into the SIMD (TXRX buffer) register. For the Master Mode, after data has been written to the SIMD (TXRX buffer) register, then transmission or reception will start automatically. When all the data has been transferred, the TRF bit should be set. For the Slave Mode, when clock pulses are received on SCK, data in the TXRX buffer will be shifted out or data on SDI will be shifted in.

When the SPI bus is disabled, SCK, SDI, SDO and  $\overline{SCS}$  can become I/O pins or other pin-shared functions using the corresponding pin-shared control bits.

# SPI Operation Steps

All communication is carried out using the 4-line interface for either Master or Slave Mode.

The CSEN bit in the SIMC2 register controls the overall function of the SPI interface. Setting this bit high will enable the SPI interface by allowing the  $\overline{SCS}$  line to be active, which can then be used to control the SPI interface. If the CSEN bit is low, the SPI interface will be disabled and the  $\overline{SCS}$  line will be in a floating condition and can therefore not be used for control of the SPI interface. If the CSEN bit in the SIMC0 are set high, this will place the SDI line in a



floating condition and the SDO line high. If in Master Mode the SCK line will be either high or low depending upon the clock polarity selection bit CKPOLB in the SIMC2 register. If in Slave Mode the SCK line will be in a floating condition. If the SIMEN bit is low, then the bus will be disabled and  $\overline{SCS}$ , SDI, SDO and SCK will all become I/O pins or the other functions using the corresponding pin-shared control bits. In the Master Mode the Master will always generate the clock signal. The clock and data transmission will be initiated after data has been written into the SIMD register. In the Slave Mode, the clock signal will be received from an external master device for both data transmission and reception. The following sequences show the order to be followed for data transfer in both Master and Slave Mode.

## Master Mode:

• Step 1

Select the SPI Master mode and clock source using the UMD and SIM2~SIM0 bits in the SIMC0 control register.

• Step 2

Setup the CSEN bit and setup the MLS bit to choose if the data is MSB or LSB first, this setting must be the same with the Slave devices.

• Step 3

Setup the SIMEN bit in the SIMC0 control register to enable the SPI interface.

• Step 4

For write operations: write the data to the SIMD register, which will actually place the data into the TXRX buffer. Then use the SCK and  $\overline{\text{SCS}}$  lines to output the data. After this, go to step 5. For read operations: the data transferred in on the SDI line will be stored in the TXRX buffer until all the data has been received at which point it will be latched into the SIMD register.

• Step 5

Check the WCOL bit if set high then a collision error has occurred so return to step 4. If equal to zero then go to the following step.

• Step 6

Check the TRF bit or wait for a USIM SPI serial bus interrupt.

• Step 7

Read data from the SIMD register.

- Step 8 Clear TRF.
- Step 9 Go to step 4.

## Slave Mode:

• Step 1

Select the SPI Slave mode using the UMD and SIM2~SIM0 bits in the SIMC0 control register

• Step 2

Setup the CSEN bit and setup the MLS bit to choose if the data is MSB or LSB first, this setting must be the same with the Master devices.

• Step 3

Setup the SIMEN bit in the SIMC0 control register to enable the SPI interface.

• Step 4

For write operations: write the data to the SIMD register, which will actually place the data into the TXRX buffer. Then wait for the master clock SCK and  $\overline{SCS}$  signal. After this, go to step 5.



For read operations: the data transferred in on the SDI line will be stored in the TXRX buffer until all the data has been received at which point it will be latched into the SIMD register.

• Step 5

Check the WCOL bit if set high then a collision error has occurred so return to step 4. If equal to zero then go to the following step.

• Step 6

Check the TRF bit or wait for a USIM SPI serial bus interrupt.

• Step 7

Read data from the SIMD register.

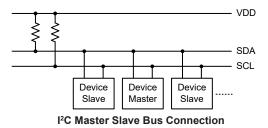
- Step 8
- Clear TRF.
- Step 9 Go to step 4.

# **Error Detection**

The WCOL bit in the SIMC2 register is provided to indicate errors during data transfer. The bit is set by the SPI serial Interface but must be cleared by the application program. This bit indicates that a data collision has occurred which happens if a write to the SIMD register takes place during a data transfer operation and will prevent the write operation from continuing.

# I<sup>2</sup>C Interface

The I<sup>2</sup>C interface is used to communicate with external peripheral devices such as sensors, EEPROM memory etc. Originally developed by Philips, it is a two line low speed serial interface for synchronous serial data transfer. The advantage of only two lines for communication, relatively simple communication protocol and the ability to accommodate multiple devices on the same bus has made it an extremely popular interface type for many applications.

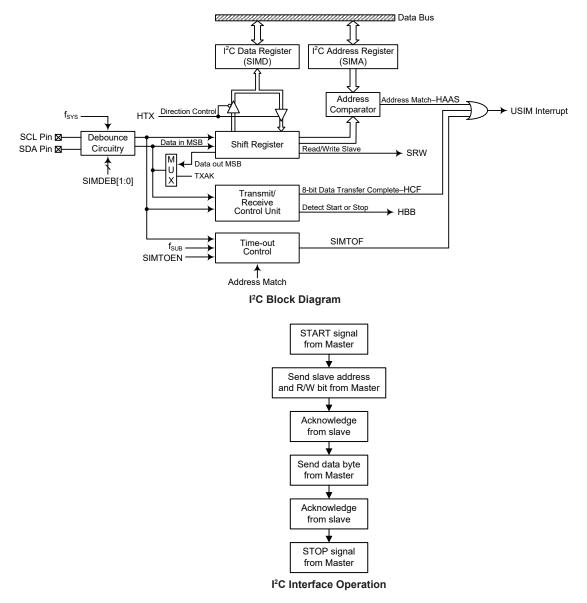


### I<sup>2</sup>C Interface Operation

The I<sup>2</sup>C serial interface is a two line interface, a serial data line, SDA, and serial clock line, SCL. As many devices may be connected together on the same bus, their outputs are both open drain types. For this reason it is necessary that external pull-high resistors are connected to these outputs. Note that no chip select line exists, as each device on the I<sup>2</sup>C bus is identified by a unique address which will be transmitted and received on the I<sup>2</sup>C bus.

When two devices communicate with each other on the bidirectional I<sup>2</sup>C bus, one is known as the master device and one as the slave device. Both master and slave can transmit and receive data, however, it is the master device that has overall control of the bus. For the device, which only operates in slave mode, there are two methods of transferring data on the I<sup>2</sup>C bus, the slave transmit mode and the slave receive mode. The pull-high control function pin-shared with SCL/SDA pin is still applicable even if I<sup>2</sup>C device is activated and the related internal pull-high register could be controlled by its corresponding pull-high control register.





The SIMDEB1 and SIMDEB0 bits determine the debounce time of the I<sup>2</sup>C interface. This uses the internal clock to in effect add a debounce time to the external clock to reduce the possibility of glitches on the clock line causing erroneous operation. The debounce time, if selected, can be chosen to be either 2 or 4 system clocks. To achieve the required I<sup>2</sup>C data transfer speed, there exists a relationship between the system clock,  $f_{SYS}$ , and the I<sup>2</sup>C debounce time. For either the I<sup>2</sup>C Standard or Fast mode operation, users must take care of the selected system clock frequency and the configured debounce time to match the criterion shown in the following table.

I <sup>2</sup> C Debounce Time Selection	I <sup>2</sup> C Standard Mode (100kHz)	I <sup>2</sup> C Fast Mode (400kHz)
No Debounce	f <sub>sys</sub> > 2 MHz	f <sub>sys</sub> > 5 MHz
2 system clock debounce	f <sub>SYS</sub> > 4 MHz	f <sub>sys</sub> > 10 MHz
4 system clock debounce	f <sub>sys</sub> > 8 MHz	f <sub>sys</sub> > 20 MHz

I<sup>2</sup>C Minimum f<sub>SYS</sub> Frequency Requirements



## I<sup>2</sup>C Registers

There are three control registers associated with the I<sup>2</sup>C bus, SIMC0, SIMC1 and SIMTOC, one address register SIMA and one data register, SIMD. Note that the SIMC1, SIMD, SIMA and SIMTOC registers and their POR values are only available when the I<sup>2</sup>C mode is selected by properly configuring the UMD and SIM2~SIM0 bits in the SIMC0 register.

Register	Bit										
Name	7	6	5	4	3	2	1	0			
SIMC0	SIM2	SIM1	SIM0	UMD	SIMDEB1	SIMDEB0	SIMEN	SIMICF			
SIMC1	HCF	HAAS	HBB	HTX	TXAK	SRW	IAMWU	RXAK			
SIMD	D7	D6	D5	D4	D3	D2	D1	D0			
SIMA	SIMA6	SIMA5	SIMA4	SIMA3	SIMA2	SIMA1	SIMA0	D0			
SIMTOC	SIMTOEN	SIMTOF	SIMTOS5	SIMTOS4	SIMTOS3	SIMTOS2	SIMTOS1	SIMTOS0			

### I<sup>2</sup>C Register List

## I<sup>2</sup>C Data Register

The SIMD register is used to store the data being transmitted and received. The same register is used by both the SPI and I<sup>2</sup>C functions. Before the device writes data to the I<sup>2</sup>C bus, the actual data to be transmitted must be placed in the SIMD register. After the data is received from the I<sup>2</sup>C bus, the device can read it from the SIMD register. Any transmission or reception of data from the I<sup>2</sup>C bus must be made via the SIMD register.

### SIMD Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	х	х	х	х	х	х	х	х

"x": unknown

Bit 7~0 **D7~D0**: USIM SPI/I<sup>2</sup>C data register bit 7 ~ bit 0

### I<sup>2</sup>C Address Register

The SIMA register is also used by the SPI interface but has the name SIMC2. The SIMA register is the location where the 7-bit slave address of the slave device is stored. Bits  $7\sim1$  of the SIMA register define the device slave address. Bit 0 is not defined. When a master device, which is connected to the I<sup>2</sup>C bus, sends out an address, which matches the slave address in the SIMA register, the slave device will be selected.

### SIMA Register

Bit	7	6	5	4	3	2	1	0
Name	SIMA6	SIMA5	SIMA4	SIMA3	SIMA2	SIMA1	SIMA0	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~1 SIMA6~SIMA0: I<sup>2</sup>C slave address

SIMA6~SIMA0 is the I<sup>2</sup>C slave address bit 6~bit 0.

Bit 0 **D0**: Reserved bit, can be read or written



# I<sup>2</sup>C Control Registers

There are three control registers for the I<sup>2</sup>C interface, SIMC0, SIMC1 and SIMTOC. The SIMC0 register is used to control the enable/disable function and to set the data transmission clock frequency. The SIMC1 register contains the relevant flags which are used to indicate the I<sup>2</sup>C communication status. Another register, SIMTOC, is used to control the I<sup>2</sup>C time-out function and is described in the corresponding section.

### SIMC0 Register

Bit 4

Bit 1

Bit	7	6	5	4	3	2	1	0
Name	SIM2	SIM1	SIM0	UMD	SIMDEB1	SIMDEB0	SIMEN	SIMICF
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	1	1	1	0	0	0	0	0

Bit 7~5 SIM2~SIM0: USIM SPI/I<sup>2</sup>C Operating Mode Control (SPI unavailable for BA45F5740-2)

000: SPI master mode; SPI clock is f<sub>SYS</sub>/4

001: SPI master mode; SPI clock is f<sub>SYS</sub>/16

010: SPI master mode; SPI clock is f<sub>SYS</sub>/64

011: SPI master mode; SPI clock is f<sub>SUB</sub>

100: SPI master mode; SPI clock is STM CCRP match frequency/2

101: SPI slave mode

110: I<sup>2</sup>C slave mode

111: Unused mode

When the UMD bit is cleared to zero, these bits setup the SPI or  $I^2C$  operating mode of the USIM function. As well as selecting if the  $I^2C$  or SPI function, they are used to control the SPI Master/Slave selection and the SPI Master clock frequency. The SPI clock is a function of the system clock but can also be chosen to be sourced from STM and  $f_{SUB}$ . If the SPI Slave Mode is selected then the clock will be supplied by an external Master device.

UMD: UART mode selection bit

0: SPI or I<sup>2</sup>C mode

1: UART mode

This bit is used to select the UART mode. When this bit is cleared to zero, the actual SPI or I<sup>2</sup>C mode can be selected using the SIM2~SIM0 bits. Note that the UMD bit must be set low for SPI or I<sup>2</sup>C mode.

# Bit 3~2 SIMDEB1~SIMDEB0: I<sup>2</sup>C Debounce Time Selection

### 00: No debounce

01: 2 system clock debounce

1x: 4 system clock debounce

These bits are used to select the  $I^2C$  debounce time when the USIM is configured as the  $I^2C$  interface function by setting the UMD bit to "0" and the SIM2~SIM0 bits to "110".

# SIMEN: USIM SPI/I<sup>2</sup>C Enable Control

0: Disable

1: Enable

The bit is the overall on/off control for the USIM SPI/I<sup>2</sup>C interface. When the SIMEN bit is cleared to zero to disable the USIM SPI/I<sup>2</sup>C interface, the SDI, SDO, SCK and  $\overline{\text{SCS}}$ , or SDA and SCL lines will lose their SPI or I<sup>2</sup>C function and the USIM operating current will be reduced to a minimum value. When the bit is high the USIM SPI/I<sup>2</sup>C interface is enabled. If the USIM is configured to operate as an SPI interface via the UMD and SIM2~SIM0 bits, the contents of the SPI control registers will remain at the previous settings when the SIMEN bit changes from low to high and should therefore be first initialised by the application program. If the USIM is configured to operate as an I<sup>2</sup>C interface via the UMD and SIM2~SIM0 bits and the SIMEN bit changes from

low to high, the contents of the I<sup>2</sup>C control bits such as HTX and TXAK will remain at the previous settings and should therefore be first initialised by the application program while the relevant I<sup>2</sup>C flags such as HCF, HAAS, HBB, SRW and RXAK will be set to their default states.

Bit 0 SIMICF: USIM SPI Incomplete Flag This bit is only available when the USIM is configured to operate in an SPI slave mode. Refer to the SPI register section.

### SIMC1 Register

Bit	7	6	5	4	3	2	1	0	
Name	HCF	HAAS	HBB	HTX	TXAK	SRW	IAMWU	RXAK	
R/W	R	R	R	R/W	R/W	R	R/W	R	
POR	1	0	0	0	0	0	0	1	
Bit 7	<ul> <li>HCF: I<sup>2</sup>C Bus data transfer completion flag</li> <li>0: Data is being transferred</li> <li>1: Completion of an 8-bit data transfer</li> <li>The HCF flag is the data transfer flag. This flag will be zero when data is being transferred. Upon completion of an 8-bit data transfer the flag will go high and an interrupt will be generated.</li> </ul>								
Bit 6	0: Not 1: Add The HA device a	ddress is th	ntch the address he same as t	match flag		dress. If the	e addresses	if the slave match then	
Bit 5	0: I <sup>2</sup> C 1: I <sup>2</sup> C The HB which w	vill occur w	busy he I <sup>2</sup> C busy hen a STA	RT signal i		The flag w	vill be set to	ous is busy o "0" when	
Bit 4	HTX: I <sup>2</sup> 0: Slav	C slave dev ve device is	vice is trans the receive	mitter or re er	ceiver selec				
Bit 3	<ol> <li>Slave device is the transmitter</li> <li>TXAK: I<sup>2</sup>C Bus transmit acknowledge flag</li> <li>O: Slave send acknowledge flag</li> <li>I: Slave do not send acknowledge flag</li> <li>The TXAK bit is the transmit acknowledge flag. After the slave device receipt of 8 bits of data, this bit will be transmitted to the bus on the 9th clock from the slave device. The slave device must always set TXAK bit to "0" before further data is received.</li> </ol>								
Bit 2	0: Slav 1: Slav The SR the mass transmitt the slave mode or from the is zero,	ter device ted address e device wil receive me e bus, so th	nould be in nould be in the I <sup>2</sup> C SI wishes to and slave a ll check the pode. If the te slave de will write o	receive mo transmit mo ave Read/ transmit o ddress is n SRW flag SRW flag i vice should	ode Write flag. r receive d hatch, that is to determin s high, the l be in tran	ata from the s when the e whether i master is r smit mode	he I <sup>2</sup> C bus. HAAS flag t should be requesting t . When the	es whether When the is set high, in transmit o read data e SRW flag hould be in	



#### Bit 1 IAMWU: I<sup>2</sup>C Address Match Wake-up control

0: Disable

1: Enable

This bit should be set to 1 to enable the I<sup>2</sup>C address match wake up from the SLEEP or IDLE Mode. If the IAMWU bit has been set before entering either the SLEEP or IDLE mode to enable the I<sup>2</sup>C address match wake up, then this bit must be cleared by the application program after wake-up to ensure correction device operation.

Bit 0 RXAK: I<sup>2</sup>C Bus Receive acknowledge flag

0: Slave receive acknowledge flag

1: Slave does not receive acknowledge flag

The RXAK flag is the receiver acknowledge flag. When the RXAK flag is "0", it means that a acknowledge signal has been received at the 9th clock, after 8 bits of data have been transmitted. When the slave device in the transmit mode, the slave device checks the RXAK flag to determine if the master receiver wishes to receive the next byte. The slave transmitter will therefore continue sending out data until the RXAK flag is "1". When this occurs, the slave transmitter will release the SDA line to allow the master to send a STOP signal to release the I<sup>2</sup>C Bus.

## I<sup>2</sup>C Bus Communication

Communication on the I<sup>2</sup>C bus requires four separate steps, a START signal, a slave device address transmission, a data transmission and finally a STOP signal. When a START signal is placed on the I<sup>2</sup>C bus, all devices on the bus will receive this signal and be notified of the imminent arrival of data on the bus. The first seven bits of the data will be the slave address with the first bit being the MSB. If the address of the slave device matches that of the transmitted address, the HAAS bit in the SIMC1 register will be set and an USIM interrupt will be generated. After entering the interrupt service routine, the slave device must first check the condition of the HAAS and SIMTOF bits to determine whether the interrupt source originates from an address match or from the completion of an 8-bit data transfer completion or from the I<sup>2</sup>C bus time-out occurrence. During a data transfer, note that after the 7-bit slave address has been transmitted, the following bit, which is the 8th bit, is the read/write bit whose value will be placed in the SRW bit. This bit will be checked by the slave device to determine whether to go into transmit or receive mode. Before any transfer of data to or from the I<sup>2</sup>C bus, the microcontroller must initialise the bus, the following are steps to achieve this:

• Step 1

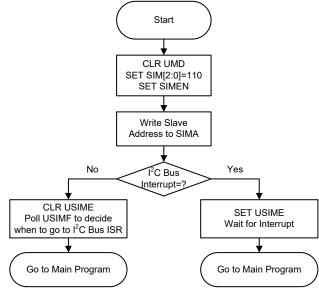
Set the UMD, SIM2~SIM0 and SIMEN bits in the SIMC0 register to "0", "110" and "1" respectively to enable the  $I^2C$  bus.

• Step 2

Write the slave address of the device to the I<sup>2</sup>C bus address register SIMA.

• Step 3

Set the USIME interrupt enable bit of the interrupt control register to enable the USIM interrupt.



I<sup>2</sup>C Bus Initialisation Flow Chart

## I<sup>2</sup>C Bus Start Signal

The START signal can only be generated by the master device connected to the I<sup>2</sup>C bus and not by the slave device. This START signal will be detected by all devices connected to the I<sup>2</sup>C bus. When detected, this indicates that the I<sup>2</sup>C bus is busy and therefore the HBB bit will be set. A START condition occurs when a high to low transition on the SDA line takes place when the SCL line remains high.

### I<sup>2</sup>C Slave Address

The transmission of a START signal by the master will be detected by all devices on the I<sup>2</sup>C bus. To determine which slave device the master wishes to communicate with, the address of the slave device will be sent out immediately following the START signal. All slave devices, after receiving this 7-bit address data, will compare it with their own 7-bit slave address. If the address sent out by the master matches the internal address of the microcontroller slave device, then an internal USIM I<sup>2</sup>C bus interrupt signal will be generated. The next bit following the address, which is the 8th bit, defines the read/write status and will be saved to the SRW bit of the SIMC1 register. The slave device will then transmit an acknowledge bit, which is a low level, as the 9th bit. The slave device will also set the status flag HAAS when the addresses match.

As an USIM I<sup>2</sup>C bus interrupt can come from three sources, when the program enters the interrupt subroutine, the HAAS and SIMTOF bits should be examined to see whether the interrupt source has come from a matching slave address or from the completion of a data byte transfer or from the I<sup>2</sup>C bus time-out occurrence. When a slave address is matched, the device must be placed in either the transmit mode and then write data to the SIMD register, or in the receive mode where it must implement a dummy read from the SIMD register to release the SCL line.

## I<sup>2</sup>C Bus Read/Write Signal

The SRW bit in the SIMC1 register defines whether the master device wishes to read data from the I<sup>2</sup>C bus or write data to the I<sup>2</sup>C bus. The slave device should examine this bit to determine if it is to be a transmitter or a receiver. If the SRW flag is "1" then this indicates that the master device wishes to read data from the I<sup>2</sup>C bus, therefore the slave device must be setup to send data to the I<sup>2</sup>C bus as a transmitter. If the SRW flag is "0" then this indicates that the master wishes to send data to the I<sup>2</sup>C bus, therefore the slave device that the master wishes to send data to the I<sup>2</sup>C bus, therefore the slave device that the master wishes to send data to the I<sup>2</sup>C bus, therefore the slave device must be setup to read data from the I<sup>2</sup>C bus as a receiver.



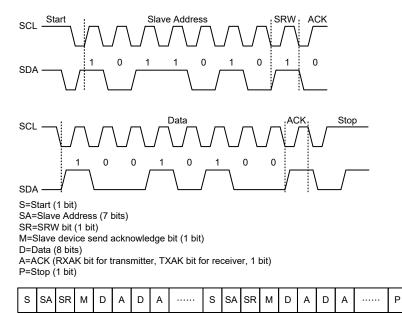
## I<sup>2</sup>C Bus Slave Address Acknowledge Signal

After the master has transmitted a calling address, any slave device on the I<sup>2</sup>C bus, whose own internal address matches the calling address, must generate an acknowledge signal. The acknowledge signal will inform the master that a slave device has accepted its calling address. If no acknowledge signal is received by the master then a STOP signal must be transmitted by the master to end the communication. When the HAAS flag is high, the addresses have matched and the slave device must check the SRW flag to determine if it is to be a transmitter or a receiver. If the SRW flag is high, the slave device should be setup to be a transmitter so the HTX bit in the SIMC1 register should be set to "1". If the SRW flag is low, then the microcontroller slave device should be setup as a receiver and the HTX bit in the SIMC1 register should be set to "0".

### I<sup>2</sup>C Bus Data and Acknowledge Signal

The transmitted data is 8-bit wide and is transmitted after the slave device has acknowledged receipt of its slave address. The order of serial bit transmission is the MSB first and the LSB last. After receipt of 8 bits of data, the receiver must transmit an acknowledge signal, level "0", before it can receive the next data byte. If the slave transmitter does not receive an acknowledge bit signal from the master receiver, then the slave transmitter will release the SDA line to allow the master to send a STOP signal to release the I<sup>2</sup>C Bus. The corresponding data will be stored in the SIMD register. If setup as a transmitter, the slave device must first write the data to be transmitted into the SIMD register. If setup as a receiver, the slave device must read the transmitted data from the SIMD register.

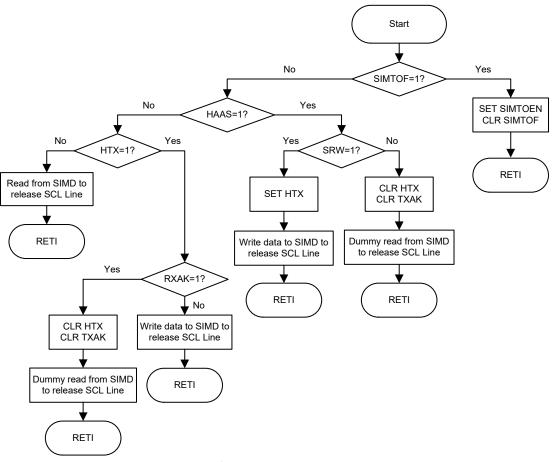
When the slave receiver receives the data byte, it must generate an acknowledge bit, known as TXAK, on the 9th clock. The slave device, which is setup as a transmitter will check the RXAK bit in the SIMC1 register to determine if it is to send another data byte, if not then it will release the SDA line and await the receipt of a STOP signal from the master.



#### I<sup>2</sup>C Communication Timing Diagram

Note: When a slave address is matched, the device must be placed in either the transmit mode and then write data to the SIMD register, or in the receive mode where it must implement a dummy read from the SIMD register to release the SCL line.



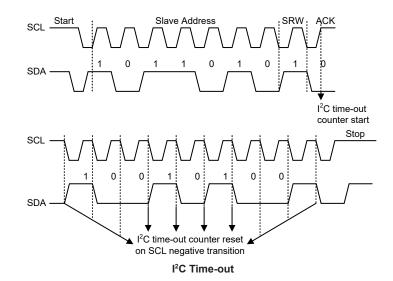


I<sup>2</sup>C Bus ISR Flow Chart

## I<sup>2</sup>C Time-out Control

In order to reduce the problem of  $I^2C$  lockup due to reception of erroneous clock sources, a time-out function is provided. If the clock source to the  $I^2C$  is not received for a while, then the  $I^2C$  circuitry and registers will be reset after a certain time-out period. The time-out counter starts counting on an  $I^2C$  bus "START" & "address match" condition, and is cleared by an SCL falling edge. Before the next SCL falling edge arrives, if the time elapsed is greater than the time-out setup by the SIMTOC register, then a time-out condition will occur. The time-out function will stop when an  $I^2C$  "STOP" condition occurs.





When an I<sup>2</sup>C time-out counter overflow occurs, the counter will stop and the SIMTOEN bit will be cleared to zero and the SIMTOF bit will be set high to indicate that a time-out condition has occurred. The time-out condition will also generate an interrupt which uses the USIM interrupt vector. When an I<sup>2</sup>C time-out occurs, the I<sup>2</sup>C internal circuitry will be reset and the registers will be reset into the following condition:

Registers	After I <sup>2</sup> C Time-out
SIMD, SIMA, SIMC0	No change
SIMC1	Reset to POR condition

I<sup>2</sup>C Registers after Time-out

The SIMTOF flag can be cleared by the application program. There are 64 time-out periods which can be selected using SIMTOS bit field in the SIMTOC register. The time-out time is given by the formula:  $((1\sim64)\times32)/f_{SUB}$ . This gives a time-out period which ranges from about 1ms to 64ms.

# SIMTOC Register

Bit	7	6	5	4	3	2	1	0
Name	SIMTOEN	SIMTOF	SIMTOS5	SIMTOS4	SIMTOS3	SIMTOS2	SIMTOS1	SIMTOS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 SIMTOEN: USIM I<sup>2</sup>C Time-out control

0: Disable

1: Enable

Bit 6 SIMTOF: USIM I<sup>2</sup>C Time-out flag

0: No time-out occurred

1: Time-out occurred

This bit is set high when time-out occurs and can only be cleared by application

program.

Bit 5~0 SIMTOS5~SIMTOS0: USIM I<sup>2</sup>C Time-out period selection

 $\rm I^2C$  time-out clock source is  $f_{SUB}/32.$ 

I<sup>2</sup>C time-out time is equal to (SIMTOS[5:0]+1) × (32/f<sub>SUB</sub>).

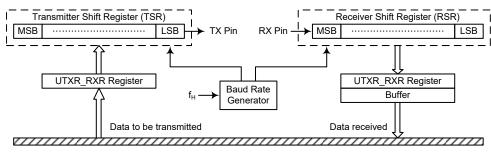


# **UART Interface**

The devices contain an integrated full-duplex asynchronous serial communications UART interface that enables communication with external devices that contain a serial interface. The UART function has many features and can transmit and receive data serially by transferring a frame of data with eight or nine data bits per transmission as well as being able to detect errors when the data is overwritten or incorrectly framed. The UART function shares the same internal interrupt vector with the SPI and I<sup>2</sup>C interfaces which can be used to indicate when a reception occurs or when a transmission terminates.

The integrated UART function contains the following features:

- Full-duplex, asynchronous communication
- 8 or 9 bits character length
- Even, odd or no parity options
- One or two stop bits
- Baud rate generator with 8-bit prescaler
- · Parity, framing, noise and overrun error detection
- Support for interrupt on address detect (last character bit=1)
- · Separately enabled transmitter and receiver
- 2-byte Deep FIFO Receive Data Buffer
- RX pin wake-up function
- Transmit and receive interrupts
- Interrupts can be triggered by the following conditions:
  - Transmitter Empty
  - Transmitter Idle
  - Receiver Full
  - Receiver Overrun
  - ♦ Address Mode Detect



MCU Data Bus

UART Data Transfer Block Diagram

## **UART External Pins**

To communicate with an external serial interface, the internal UART has two external pins known as TX and RX. The TX and RX pins are the UART transmitter and receiver pins respectively. The TX and RX pin function should first be selected by the corresponding pin-shared function selection register before the UART function is used. Along with the UMD bit, the UREN bit, the UTXEN and URXEN bits, if set, will setup these pins to their respective TX output and RX input conditions and disable any pull-high resistor option which may exist on the TX and RX pins. When the TX or RX



pin function is disabled by clearing the UMD, UREN, UTXEN or URXEN bit, the TX or RX pin will be set to a floating state. At this time whether the internal pull-high resistor is connected to the TX or RX pin or not is determined by the corresponding I/O pull-high function control bit.

## UART Data Transfer Scheme

The above block diagram shows the overall data transfer structure arrangement for the UART. The actual data to be transmitted from the MCU is first transferred to the UTXR\_RXR register by the application program. The data will then be transferred to the Transmit Shift Register from where it will be shifted out, LSB first, onto the TX pin at a rate controlled by the Baud Rate Generator. Only the UTXR\_RXR register is mapped onto the MCU Data Memory, the Transmit Shift Register is not mapped and is therefore inaccessible to the application program.

Data to be received by the UART is accepted on the external RX pin, from where it is shifted in, LSB first, to the Receiver Shift Register at a rate controlled by the Baud Rate Generator. When the shift register is full, the data will then be transferred from the shift register to the internal UTXR\_RXR register, where it is buffered and can be manipulated by the application program. Only the UTXR\_RXR register is mapped onto the MCU Data Memory, the Receiver Shift Register is not mapped and is therefore inaccessible to the application program.

It should be noted that the actual register for data transmission and reception only exists as a single shared register in the Data Memory. This shared register known as the UTXR\_RXR register is used for both data transmission and data reception.

### UART Status and Control Registers

There are six control registers associated with the UART function. The UMD bit in the SIMC0 register can be used to select the UART mode. The UUSR, UUCR1 and UUCR2 registers control the overall function of the UART, while the UBRG register controls the Baud rate. The actual data to be transmitted and received on the serial interface is managed through the UTXR\_RXR data register. Note that UART related registers and their POR values are only available when the UART mode is selected by setting the UMD bit in the SIMC0 register to "1".

Register	Bit										
Name	7	6	5	4	3	2	1	0			
SIMC0	SIM2	SIM1	SIM0	UMD	SIMDEB1	SIMDEB0	SIMEN	SIMICF			
UUSR	UPERR	UNF	UFERR	UOERR	URIDLE	URXIF	UTIDLE	UTXIF			
UUCR1	UREN	UBNO	UPREN	UPRT	USTOPS	UTXBRK	URX8	UTX8			
UUCR2	UTXEN	URXEN	UBRGH	UADDEN	UWAKE	URIE	UTIIE	UTEIE			
UTXR_ RXR	UTXRX7	UTXRX6	UTXRX5	UTXRX4	UTXRX3	UTXRX2	UTXRX1	UTXRX0			
UBRG	UBRG7	UBRG6	UBRG5	UBRG4	UBRG3	UBRG2	UBRG1	UBRG0			

**UART Register List** 

## SIMC0 Register

Bit	7	6	5	4	3	2	1	0
Name	SIM2	SIM1	SIM0	UMD	SIMDEB1	SIMDEB0	SIMEN	SIMICF
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	1	1	1	0	0	0	0	0

Bit 7~5 SIM2~SIM0: USIM SPI/I<sup>2</sup>C Operating Mode Control (SPI unavailable for BA45F5740-2)

When the UMD bit is cleared to zero, these bits setup the SPI or  $I^2C$  operating mode of the USIM function. Refer to the SPI or  $I^2C$  register section for more details.



Bit 4	UMD: UART mode selection bit
	0: SPI or I <sup>2</sup> C mode
	1: UART mode
	This bit is used to select the UART mode. When this bit is cleared to zero, the actual
	SPI or I <sup>2</sup> C mode can be selected using the SIM2~SIM0 bits. Note that the UMD bit must be set low for SPI or I <sup>2</sup> C mode.
Bit 3~2	SIMDEB1~SIMDEB0: I <sup>2</sup> C Debounce Time Selection
	Refer to the I <sup>2</sup> C register section.
Bit 1	SIMEN: USIM SPI/I <sup>2</sup> C Enable Control
	This bit is only available when the USIM is configured to operate in an SPI or $I^2C$ mode with the UMD bit set low. Refer to the SPI or $I^2C$ register section for more details.
Bit 0	SIMICF: USIM SPI Incomplete Flag
	Refer to the SPI register section.

### • UUSR Register

The UUSR register is the status register for the UART, which can be read by the program to determine the present status of the UART. All flags within the UUSR register are read only. Further explanation on each of the flags is given below:

Bit	7	6	5	4	3	2	1	0
Name	UPERR	UNF	UFERR	UOERR	URIDLE	URXIF	UTIDLE	UTXIF
R/W	R	R	R	R	R	R	R	R
POR	0	0	0	0	1	0	1	1

Bit 7 UPERR: Parity error flag

0: No parity error is detected

1: Parity error is detected

The UPERR flag is the parity error flag. When this read only flag is "0", it indicates a parity error has not been detected. When the flag is "1", it indicates that the parity of the received word is incorrect. This error flag is applicable only if Parity mode (odd or even) is selected. The flag can also be cleared by a software sequence which involves a read to the status register UUSR followed by an access to the UTXR\_RXR data register.

#### Bit 6 UNF: Noise flag 0: No noise is detected 1: Noise is detected The UNF flag is the noise flag. When this read only flag is "0", it indicates no noise condition. When the flag is "1", it indicates that the UART has detected noise on the receiver input. The UNF flag is set during the same cycle as the URXIF flag but will not be set in the case of as overrun. The UNF flag can be cleared by a software sequence which will involve a read to the status register UUSR followed by an access to the UTXR RXR data register. Bit 5 UFERR: Framing error flag 0: No framing error is detected 1: Framing error is detected The UFERR flag is the framing error flag. When this read only flag is "0", it indicates that there is no framing error. When the flag is "1", it indicates that a framing error has been detected for the current character. The flag can also be cleared by a software sequence which will involve a read to the status register UUSR followed by an access to the UTXR RXR data register.



Bit 4	UOERR: Overrun error flag 0: No overrun error is detected 1: Overrun error is detected
	The UOERR flag is the overrun error flag which indicates when the receiver buffer has overflowed. When this read only flag is "0", it indicates that there is no overrun error. When the flag is "1", it indicates that an overrun error occurs which will inhibit further transfers to the UTXR_RXR receive data register. The flag is cleared by a software sequence, which is a read to the status register UUSR followed by an access to the UTXR_RXR data register.
Bit 3	<ul><li>URIDLE: Receiver status</li><li>0: Data reception is in progress (Data being received)</li><li>1: No data reception is in progress (Receiver is idle)</li></ul>
	The URIDLE flag is the receiver status flag. When this read only flag is "0", it indicates that the receiver is between the initial detection of the start bit and the completion of the stop bit. When the flag is "1", it indicates that the receiver is idle. Between the completion of the stop bit and the detection of the next start bit, the URIDLE bit is "1" indicating that the UART receiver is idle and the RX pin stays in logic high condition.
Bit 2	URXIF: Receive UTXR_RXR data register status 0: UTXR_RXR data register is empty 1: UTXR_RXR data register has available data
	The URXIF flag is the receive data register status flag. When this read only flag is "0", it indicates that the UTXR_RXR read data register is empty. When the flag is "1", it indicates that the UTXR_RXR read data register contains new data. When the contents of the shift register are transferred to the UTXR_RXR register, an interrupt is generated if URIE=1 in the UUCR2 register. If one or more errors are detected in the received word, the appropriate receive-related flags UNF, UFERR, and/or UPERR are set within the same clock cycle. The URXIF flag will eventually be cleared when the UUSR register is read with URXIF set, followed by a read from the UTXR_RXR register, and if the UTXR_RXR register has no more new data available.
Bit 1	UTIDLE: Transmission idle 0: Data transmission is in progress (Data being transmitted)
	1: No data transmission is in progress (Transmitter is idle) The UTIDLE flag is known as the transmission complete flag. When this read only flag is "0", it indicates that a transmission is in progress. This flag will be set high when the UTXIF flag is "1" and when there is no transmit data or break character being transmitted. When UTIDLE is equal to "1", the TX pin becomes idle with the pin state in logic high condition. The UTIDLE flag is cleared by reading the UUSR register with UTIDLE set and then writing to the UTXR_RXR register. The flag is not generated when a data character or a break is queued and ready to be sent.
Bit 0	<ul> <li>UTXIF: Transmit UTXR_RXR data register status</li> <li>0: Character is not transferred to the transmit shift register</li> <li>1: Character has transferred to the transmit shift register (UTXR_RXR data register is empty)</li> </ul>
	The UTXIF flag is the transmit data register empty flag. When this read only flag is "0", it indicates that the character is not transferred to the transmitter shift register. When the flag is "1", it indicates that the transmitter shift register has received a character from the UTXR_RXR data register. The UTXIF flag is cleared by reading the UART status register (UUSR) with UTXIF set and then writing to the UTXR_RXR data register. Note that when the UTXEN bit is set, the UTXIF flag bit will also be set since

the transmit data register is not yet full.



# UUCR1 Register

The UUCR1 register together with the UUCR2 register are the two UART control registers that are used to set the various options for the UART function, such as overall on/off control, parity control, data transfer bit length etc. Further explanation on each of the bits is given below:

Bit	7	6	5	4	3	2	1	0
Name	UREN	UBNO	UPREN	UPRT	USTOPS	UTXBRK	URX8	UTX8
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R	W
POR	0	0	0	0	0	0	х	0

"x": unknown

Bit 7

# UREN: UART function enable control

0: Disable UART. TX and RX pins are in a floating state

	1: Enable UART. TX and RX pins function as UART pins
	The UREN bit is the UART enable bit. When this bit is equal to "0", the UART will be disabled and the RX pin as well as the TX pin will be set in a floating state. When the bit is equal to "1", the UART will be enabled if the UMD bit is set and the TX and RX pins will function as defined by the UTXEN and URXEN enable control bits.
	When the UART is disabled, it will empty the buffer so any character remaining in the buffer will be discarded. In addition, the value of the baud rate counter will be reset. If the UART is disabled, all error and status flags will be reset. Also the UTXEN, URXEN, UTXBRK, URXIF, UOERR, UFERR, UPERR and UNF bits will be cleared, while the UTIDLE, UTXIF and URIDLE bits will be set. Other control bits in UUCR1, UUCR2 and UBRG registers will remain unaffected. If the UART is active and the UREN bit is cleared, all pending transmissions and receptions will be terminated and the module will be reset as defined above. When the UART is re- enabled, it will restart in the same configuration.
Bit 6	<b>UBNO</b> : Number of data transfer bits selection 0: 8-bit data transfer 1: 9-bit data transfer
	This bit is used to select the data length format, which can have a choice of either 8-bit or 9-bit format. When this bit is equal to "1", a 9-bit data length format will be selected. If the bit is equal to "0", then an 8-bit data length format will be selected. If 9-bit data length format is selected, then bits URX8 and UTX8 will be used to store the 9th bit of the received and transmitted data respectively.
Bit 5	<b>UPREN</b> : Parity function enable control 0: Parity function is disabled 1: Parity function is enabled
	This is the parity enable bit. When this bit is equal to "1", the parity function will be enabled. If the bit is equal to "0", then the parity function will be disabled.
Bit 4	UPRT: Parity type selection bit 0: Even parity for parity generator 1: Odd parity for parity generator
	This bit is the parity type selection bit. When this bit is equal to "1", odd parity type will be selected. If the bit is equal to "0", then even parity type will be selected.
Bit 3	USTOPS: Number of Stop bits selection 0: One stop bit format is used 1: Two stop bits format is used
	This bit determines if one or two stop bits are to be used. When this bit is equal to "1", two stop bits are used. If this bit is equal to "0", then only one stop bit is used.
Bit 2	UTXBRK: Transmit break character 0: No break character is transmitted 1: Break characters transmit
	The UTYPPK hit is the Transmit Break Character bit. When this bit is "0" there are

The UTXBRK bit is the Transmit Break Character bit. When this bit is "0", there are no break characters and the TX pin operates normally. When the bit is "1", there are



transmit break characters and the transmitter will send logic zeros. When this bit is equal to "1", after the buffered data has been transmitted, the transmitter output is held low for a minimum of a 13-bit length and until the UTXBRK bit is reset.

- Bit 1 URX8: Receive data bit 8 for 9-bit data transfer format (read only) This bit is only used if 9-bit data transfers are used, in which case this bit location will store the 9th bit of the received data known as URX8. The UBNO bit is used to determine whether data transfers are in 8-bit or 9-bit format.
- Bit 0 UTX8: Transmit data bit 8 for 9-bit data transfer format (write only) This bit is only used if 9-bit data transfers are used, in which case this bit location will store the 9th bit of the transmitted data known as UTX8. The UBNO bit is used to determine whether data transfers are in 8-bit or 9-bit format.

# UUCR2 Register

The UUCR2 register is the second of the two UART control registers and serves several purposes. One of its main functions is to control the basic enable/disable operation of the UART Transmitter and Receiver as well as enabling the various USIM UART mode interrupt sources. The register also serves to control the baud rate speed, receiver wake-up enable and the address detect enable. Further explanation on each of the bits is given below:

Bit	7	6	5	4	3	2	1	0
Name	UTXEN	URXEN	UBRGH	UADDEN	UWAKE	URIE	UTIIE	UTEIE
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

# Bit 7 UTXEN: UART Transmitter enabled control

0: UART transmitter is disabled

1: UART transmitter is enabled

The bit named UTXEN is the Transmitter Enable Bit. When this bit is equal to "0", the transmitter will be disabled with any pending data transmissions being aborted. In addition the buffers will be reset. In this situation the TX pin will be set in a floating state.

If the UTXEN bit is equal to "1" and the UMD and UREN bit are also equal to "1", the transmitter will be enabled and the TX pin will be controlled by the UART. Clearing the UTXEN bit during a transmission will cause the data transmission to be aborted and will reset the transmitter. If this situation occurs, the TX pin will be set in a floating state.

Bit 6

# URXEN: UART Receiver enabled control

0: UART receiver is disabled

1: UART receiver is enabled

The bit named URXEN is the Receiver Enable Bit. When this bit is equal to "0", the receiver will be disabled with any pending data receptions being aborted. In addition the receive buffers will be reset. In this situation the RX pin will be set in a floating state. If the URXEN bit is equal to "1" and the UMD and UREN bit are also equal to "1", the receiver will be enabled and the RX pin will be controlled by the UART. Clearing the URXEN bit during a reception will cause the data reception to be aborted and will reset the receiver. If this situation occurs, the RX pin will be set in a floating state.

### Bit 5

- **UBRGH**: Baud Rate speed selection 0: Low speed baud rate
  - 1: High speed baud rate

The bit named UBRGH selects the high or low speed mode of the Baud Rate Generator. This bit, together with the value placed in the baud rate register UBRG, controls the Baud Rate of the UART. If this bit is equal to "1", the high speed mode is selected. If the bit is equal to "0", the low speed mode is selected.



Bit 4 UADDEN: Address detect function enable control

0: Address detect function is disabled

1: Address detect function is enabled

The bit named UADDEN is the address detect function enable control bit. When this bit is equal to "1", the address detect function is enabled. When it occurs, if the 8th bit, which corresponds to UTXRX7 if UBNO=0 or the 9th bit, which corresponds to URX8 if UBNO=1, has a value of "1", then the received word will be identified as an address, rather than data. If the corresponding interrupt is enabled, an interrupt request will be generated each time the received word has the address bit set, which is the 8th or 9th bit depending on the value of UBNO. If the address bit known as the 8th or 9th bit of the received word is "0" with the address detect function being enabled, an interrupt will not be generated and the received data will be discarded.

### Bit 3 UWAKE: RX pin wake-up UART function enable control

0: RX pin wake-up UART function is disabled

1: RX pin wake-up UART function is enabled

This bit is used to control the wake-up UART function when a falling edge on the RX pin occurs. Note that this bit is only available when the UART clock  $(f_H)$  is switched off. There will be no RX pin wake-up UART function if the UART clock  $(f_H)$  exists. If the UWAKE bit is set to 1 as the UART clock  $(f_H)$  is switched off, a UART wake-up request will be initiated when a falling edge on the RX pin occurs. When this request happens and the corresponding interrupt is enabled, an RX pin wake-up UART interrupt will be generated to inform the MCU to wake up the UART function by switching on the UART clock  $(f_H)$  via the application program. Otherwise, the UART function cannot resume even if there is a falling edge on the RX pin when the UWAKE bit is cleared to 0.

### Bit 2 URIE: Receiver interrupt enable control

0: Receiver related interrupt is disabled

1: Receiver related interrupt is enabled

This bit enables or disables the receiver interrupt. If this bit is equal to "1" and when the receiver overrun flag UOERR or receive data available flag URXIF is set, the USIM interrupt request flag USIMF will be set. If this bit is equal to "0", the USIM interrupt request flag USIMF will not be influenced by the condition of the UOERR or URXIF flags.

UTIIE: Transmitter Idle interrupt enable control

0: Transmitter idle interrupt is disabled

1: Transmitter idle interrupt is enabled

This bit enables or disables the transmitter idle interrupt. If this bit is equal to "1" and when the transmitter idle flag UTIDLE is set, due to a transmitter idle condition, the USIM interrupt request flag USIMF will be set. If this bit is equal to "0", the USIM interrupt request flag USIMF will not be influenced by the condition of the UTIDLE flag.

UTEIE: Transmitter Empty interrupt enable control

0: Transmitter empty interrupt is disabled

1: Transmitter empty interrupt is enabled

This bit enables or disables the transmitter empty interrupt. If this bit is equal to "1" and when the transmitter empty flag UTXIF is set, due to a transmitter empty condition, the USIM interrupt request flag USIMF will be set. If this bit is equal to "0", the USIM interrupt request flag USIMF will not be influenced by the condition of the UTXIF flag.

Bit 1

Bit 0



# • UTXR\_RXR Register

The UTXR\_RXR register is the data register which is used to store the data to be transmitted on the TX pin or being received from the RX pin.

Bit	7	6	5	4	3	2	1	0
Name	UTXRX7	UTXRX6	UTXRX5	UTXRX4	UTXRX3	UTXRX2	UTXRX1	UTXRX0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	х	х	х	х	х	х	х	х

"x": unknown

### Bit 7~0 UTXRX7~UTXRX0: UART Transmit/Receive Data bit 7 ~ bit 0

### UBRG Register

Bit	7	6	5	4	3	2	1	0
Name	UBRG7	UBRG6	UBRG5	UBRG4	UBRG3	UBRG2	UBRG1	UBRG0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	х	х	х	х	х	х	х	х

"x": unknown

Bit 7~0 UBRG7~UBRG0: Baud Rate values

By programming the UBRGH bit in UUCR2 Register which allows selection of the related formula described above and programming the required value in the UBRG register, the required baud rate can be setup.

Note: Baud rate= $f_H/[64\times(N+1)]$  if UBRGH=0.

Baud rate= $f_H/[16\times(N+1)]$  if UBRGH=1.

### **Baud Rate Generator**

To setup the speed of the serial data communication, the UART function contains its own dedicated baud rate generator. The baud rate is controlled by its own internal free running 8-bit timer, the period of which is determined by two factors. The first of these is the value placed in the baud rate register UBRG and the second is the value of the UBRGH bit with the control register UUCR2. The UBRGH bit decides if the baud rate generator is to be used in a high speed mode or low speed mode, which in turn determines the formula that is used to calculate the baud rate. The value N in the UBRG register which is used in the following baud rate calculation formula determines the division factor. Note that N is the decimal value placed in the UBRG register and has a range of between 0 and 255.

UUCR2 UBRGH Bit	0	1
Baud Rate (BR)	f <sub>H</sub> /[64 (N+1)]	f <sub>H</sub> /[16 (N+1)]

By programming the UBRGH bit which allows selection of the related formula and programming the required value in the UBRG register, the required baud rate can be setup. Note that because the actual baud rate is determined using a discrete value, N, placed in the UBRG register, there will be an error associated between the actual and requested value. The following example shows how the UBRG register value N and the error value can be calculated.

## Calculating the Baud Rate and Error Values

For a clock frequency of 4MHz, and with UBRGH cleared to zero determine the UBRG register value N, the actual baud rate and the error value for a desired baud rate of 4800.

From the above table the desired baud rate  $BR=f_H/[64 (N+1)]$ 

Re-arranging this equation gives  $N=[f_H/(BR\times 64)] - 1$ 

Giving a value for N=[4000000/(4800×64)]-1=12.0208



To obtain the closest value, a decimal value of 12 should be placed into the UBRG register. This gives an actual or calculated baud rate value of BR=4000000/[ $64 \times (12+1)$ ]=4808

Therefore the error is equal to (4808-4800)/4800=0.16%

## UART Setup and Control

For data transfer, the UART function utilizes a non-return-to-zero, more commonly known as NRZ, format. This is composed of one start bit, eight or nine data bits, and one or two stop bits. Parity is supported by the UART hardware, and can be setup to be even, odd or no parity. For the most common data format, 8 data bits along with no parity and one stop bit, denoted as 8, N, 1, is used as the default setting, which is the setting at power-on. The number of data bits and stop bits, along with the parity, are setup by programming the corresponding UBNO, UPRT, UPREN, and USTOPS bits in the UUCR1 register. The baud rate used to transmit and receive data is setup using the internal 8-bit baud rate generator, while the data is transmitted and received LSB first. Although the UART transmitter and receiver are functionally independent, they both use the same data format and baud rate. In all cases stop bits will be used for data transmission.

## Enabling/Disabling the UART Interface

The basic on/off function of the internal UART function is controlled using the UREN bit in the UUCR1 register. When the UART mode is selected by setting the UMD bit in the SIMC0 register to "1", if the UREN, UTXEN and URXEN bits are set, then these two UART pins will act as normal TX output pin and RX input pin respectively. If no data is being transmitted on the TX pin, then it will default to a logic high value.

Clearing the UREN bit will disable the TX and RX pins and allow these two pins to be used as normal I/O or other pin-shared functional pins by configuring the corresponding pin-shared control bits. When the UART function is disabled the buffer will be reset to an empty condition, at the same time discarding any remaining residual data. Disabling the UART will also reset the error and status flags with bits UTXEN, URXEN, UTXBRK, URXIF, UOERR, UFERR, UPERR and UNF being cleared while bits UTIDLE, UTXIF and URIDLE will be set. The remaining control bits in the UUCR1, UUCR2 and UBRG registers will remain unaffected. If the UREN bit in the UUCR1 register is cleared while the UART is active, then all pending transmissions and receptions will be immediately suspended and the UART will be reset to a condition as defined above. If the UART is then subsequently re-enabled, it will restart again in the same configuration.

## Data, Parity and Stop Bit Selection

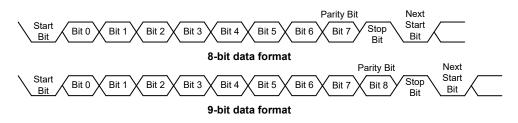
The format of the data to be transferred is composed of various factors such as data bit length, parity on/off, parity type, address bits and the number of stop bits. These factors are determined by the setup of various bits within the UUCR1 register. The UBNO bit controls the number of data bits which can be set to either 8 or 9, the UPRT bit controls the choice of odd or even parity, the UPREN bit controls the parity on/off function and the USTOPS bit decides whether one or two stop bits are to be used. The following table shows various formats for data transmission. The address bit, which is the MSB of the data byte, identifies the frame as an address character or data if the address detect function is enabled. The number of stop bits, which can be either one or two, is independent of the data length and is only used for the transmitter. There is only one stop bit for the receiver.



Start Bit	Data Bits	Address Bit	Parity Bit	Stop Bit					
Example of 8-bit Data Formats									
1	8	0	0	1					
1	7	0	1	1					
1	7	1	0	1					
Example of 9-	Example of 9-bit Data Formats								
1	9	0	0	1					
1	8	0	1	1					
1	8	1	0	1					

#### **Transmitter Receiver Data Format**

The following diagram shows the transmit and receive waveforms for both 8-bit and 9-bit data formats.



### **UART Transmitter**

Data word lengths of either 8 or 9 bits can be selected by programming the UBNO bit in the UUCR1 register. When UBNO bit is set, the word length will be set to 9 bits. In this case the 9th bit, which is the MSB, needs to be stored in the UTX8 bit in the UUCR1 register. At the transmitter core lies the Transmitter Shift Register, more commonly known as the TSR, whose data is obtained from the transmit data register, which is known as the UTXR RXR register. The data to be transmitted is loaded into this UTXR RXR register by the application program. The TSR register is not written to with new data until the stop bit from the previous transmission has been sent out. As soon as this stop bit has been transmitted, the TSR can then be loaded with new data from the UTXR RXR register, if it is available. It should be noted that the TSR register, unlike many other registers, is not directly mapped into the Data Memory area and as such is not available to the application program for direct read/write operations. An actual transmission of data will normally be enabled when the UTXEN bit is set, but the data will not be transmitted until the UTXR RXR register has been loaded with data and the baud rate generator has defined a shift clock source. However, the transmission can also be initiated by first loading data into the UTXR RXR register, after which the UTXEN bit can be set. When a transmission of data begins, the TSR is normally empty, in which case a transfer to the UTXR RXR register will result in an immediate transfer to the TSR. If during a transmission the UTXEN bit is cleared, the transmission will immediately cease and the transmitter will be reset. The TX output pin can then be configured as the I/O or other pin-shared function by configuring the corresponding pin-shared control bits.

## **Transmitting Data**

When the UART is transmitting data, the data is shifted on the TX pin from the shift register, with the least significant bit first. In the transmit mode, the UTXR\_RXR register forms a buffer between the internal bus and the transmitter shift register. It should be noted that if 9-bit data format has been selected, then the MSB will be taken from the UTX8 bit in the UUCR1 register. The steps to initiate a data transfer can be summarized as follows:



- Make the correct selection of the UBNO, UPRT, UPREN and USTOPS bits to define the required word length, parity type and number of stop bits.
- Setup the UBRG register to select the desired baud rate.
- Set the UTXEN bit to ensure that the TX pin is used as a UART transmitter pin.
- Access the UUSR register and write the data that is to be transmitted into the UTXR\_RXR register. Note that this step will clear the UTXIF bit.

This sequence of events can now be repeated to send additional data.

It should be noted that when UTXIF=0, data will be inhibited from being written to the UTXR\_RXR register. Clearing the UTXIF flag is always achieved using the following software sequence:

- 1. A UUSR register access
- 2. A UTXR\_RXR register write execution

The read-only UTXIF flag is set by the UART hardware and if set indicates that the UTXR\_RXR register is empty and that other data can now be written into the UTXR\_RXR register without overwriting the previous data. If the UTEIE bit is set then the UTXIF flag will generate an interrupt.

During a data transmission, a write instruction to the UTXR\_RXR register will place the data into the UTXR\_RXR register, which will be copied to the shift register at the end of the present transmission. When there is no data transmission in progress, a write instruction to the UTXR\_RXR register will place the data directly into the shift register, resulting in the commencement of data transmission, and the UTXIF bit being immediately set. When a frame transmission is complete, which happens after stop bits are sent or after the break frame, the UTIDLE bit will be set. To clear the UTIDLE bit the following software sequence is used:

- 1. A UUSR register access
- 2. A UTXR\_RXR register write execution

Note that both the UTXIF and UTIDLE bits are cleared by the same software sequence.

## **Transmit Break**

If the UTXBRK bit is set and the state keeps for a time greater than  $[(UBRG+1)\times t_H]$  then the break characters will be sent on the next transmission. Break character transmission consists of a start bit, followed by 13×N '0' bits and stop bits, where N=1, 2, etc. If a break character is to be transmitted then the UTXBRK bit must be first set by the application program, and then cleared to generate the stop bits. Transmitting a break character will not generate a transmit interrupt. Note that a break condition length is at least 13 bits long. If the UTXBRK bit is continually kept at a logic high level then the transmitter circuitry will transmit continuous break characters. After the application program has cleared the UTXBRK bit, the transmitter will finish transmitting the last break character and subsequently send out one or two stop bits. The automatic logic highs at the end of the last break character will ensure that the start bit of the next frame is recognized.

### **UART Receiver**

The UART is capable of receiving word lengths of either 8 or 9 bits. If the UBNO bit is set, the word length will be set to 9 bits with the MSB being stored in the URX8 bit of the UUCR1 register. At the receiver core lies the Receive Serial Shift Register, commonly known as the RSR. The data which is received on the RX external input pin is sent to the data recovery block. The data recovery block operating speed is 16 times that of the baud rate, while the main receive serial shifter operates at the baud rate. After the RX pin is sampled for the stop bit, the received data in RSR is transferred to the receive data register, if the register is empty. The data which is received on the external RX input pin is sampled three times by a majority detect circuit to determine the logic level that has been placed



onto the RX pin. It should be noted that the RSR register, unlike many other registers, is not directly mapped into the Data Memory area and as such is not available to the application program for direct read/write operations.

#### **Receiving Data**

When the UART receiver is receiving data, the data is serially shifted in on the external RX input pin, LSB first. In the read mode, the UTXR\_RXR register forms a buffer between the internal bus and the receiver shift register. The UTXR\_RXR register is a two byte deep FIFO data buffer, where two bytes can be held in the FIFO while a third byte can continue to be received. Note that the application program must ensure that the data is read from UTXR\_RXR before the third byte has been completely shifted in, otherwise this third byte will be discarded and an overrun error UOERR will be subsequently indicated. The steps to initiate a data transfer can be summarized as follows:

- Make the correct selection of UBNO, UPRT and UPREN bits to define the word length, parity type.
- Setup the UBRG register to select the desired baud rate.
- Set the URXEN bit to ensure that the RX pin is used as a UART receiver pin.

At this point the receiver will be enabled which will begin to look for a start bit.

When a character is received the following sequence of events will occur:

- The URXIF bit in the UUSR register will be set when the UTXR\_RXR register has data available. There will be at most one more character available before an overrun error occurs.
- When the contents of the shift register have been transferred to the UTXR\_RXR register, then if the URIE bit is set, an interrupt will be generated.
- If during reception, a frame error, noise error, parity error, or an overrun error has been detected, then the error flags can be set.

The URXIF bit can be cleared using the following software sequence:

- 1. A UUSR register access
- 2. A UTXR\_RXR register read execution

## Receive Break

Any break character received by the UART will be managed as a framing error. The receiver will count and expect a certain number of bit times as specified by the values programmed into the UBNO bit plus one stop bit. If the break is much longer than 13 bit times, the reception will be considered as complete after the number of bit times specified by UBNO plus one stop bit. The URXIF bit is set, UFERR is set, zeros are loaded into the receive data register, interrupts are generated if appropriate and the URIDLE bit is set. A break is regarded as a character that contains only zeros with the UFERR flag set. If a long break signal has been detected, the receiver will regard it as a data frame including a start bit, data bits and the invalid stop bit and the UFERR flag will be set. The receiver must wait for a valid stop bit before looking for the next start bit. The break character will not make the assumption that the break condition on the line is the next start bit. The break character will be loaded into the URIDLE read only flag will go high when the stop bits have not yet been received. The receiven of a break character on the UART registers will result in the following:

- The framing error flag, UFERR, will be set.
- The receive data register, UTXR\_RXR, will be cleared.
- The UOERR, UNF, UPERR, URIDLE or URXIF flags will possibly be set.



#### **Idle Status**

When the receiver is reading data, which means it will be in between the detection of a start bit and the reading of a stop bit, the receiver status flag in the UUSR register, otherwise known as the URIDLE flag, will have a zero value. In between the reception of a stop bit and the detection of the next start bit, the URIDLE flag will have a high value, which indicates the receiver is in an idle condition.

#### **Receiver Interrupt**

The read only receive interrupt flag URXIF in the UUSR register is set by an edge generated by the receiver. An interrupt is generated if URIE=1, when a word is transferred from the Receive Shift Register, RSR, to the Receive Data Register, UTXR\_RXR. An overrun error can also generate an interrupt if URIE=1.

#### Managing Receiver Errors

Several types of reception errors can occur within the UART module, the following section describes the various types and how they are managed by the UART.

#### Overrun Error – UOERR

The UTXR\_RXR register is composed of a two byte deep FIFO data buffer, where two bytes can be held in the FIFO register, while a third byte can continue to be received. Before this third byte has been entirely shifted in, the data should be read from the UTXR\_RXR register. If this is not done, the overrun error flag UOERR will be consequently indicated.

In the event of an overrun error occurring, the following will happen:

- The UOERR flag in the UUSR register will be set.
- The UTXR\_RXR contents will not be lost.
- The shift register will be overwritten.
- An interrupt will be generated if the URIE bit is set.

The UOERR flag can be cleared by an access to the UUSR register followed by a read to the UTXR\_RXR register.

#### Noise Error – UNF

Over-sampling is used for data recovery to identify valid incoming data and noise. If noise is detected within a frame the following will occur:

- The read only noise flag, UNF, in the UUSR register will be set on the rising edge of the URXIF bit.
- Data will be transferred from the Shift register to the UTXR\_RXR register.
- No interrupt will be generated. However this bit rises at the same time as the URXIF bit which itself generates an interrupt.

Note that the UNF flag is reset by a UUSR register read operation followed by a UTXR\_RXR register read operation.

#### Framing Error – UFERR

The read only framing error flag, UFERR, in the UUSR register, is set if a zero is detected instead of stop bits. If two stop bits are selected, both stop bits must be high; otherwise the UFERR flag will be set. The UFERR flag and the received data will be recorded in the UUSR and UTXR\_RXR registers respectively, and the flag is cleared in any reset.



### Parity Error – UPERR

The read only parity error flag, UPERR, in the UUSR register, is set if the parity of the received word is incorrect. This error flag is only applicable if the parity is enabled, UPREN = 1, and if the parity type, odd or even is selected. The read only UPERR flag and the received data will be recorded in the UUSR and UTXR\_RXR registers respectively. It is cleared on any reset, it should be noted that the flags, UFERR and UPERR, in the UUSR register should first be read by the application program before reading the data word.

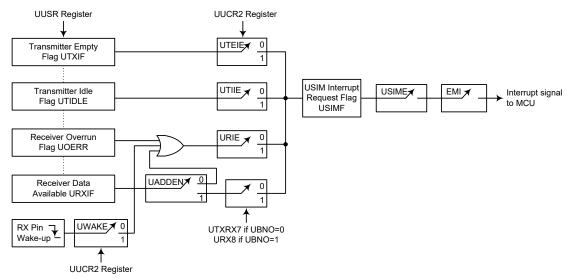
### UART Interrupt Structure

Several individual UART conditions can trigger an USIM interrupt. When these conditions exist, a low pulse will be generated to get the attention of the microcontroller. These conditions are a transmitter data register empty, transmitter idle, receiver data available, receiver overrun, address detect and an RX pin wake-up. When any of these conditions are created, if the global interrupt enable bit and the USIM interrupt control bit are enabled and the stack is not full, the program will jump to its corresponding interrupt vector where it can be serviced before returning to the main program. Four of these conditions have the corresponding UUSR register flags which will generate an USIM interrupt if its associated interrupt enable control bit in the UUCR2 register is set. The two transmitter interrupt conditions have their own corresponding enable control bits, while the two receiver interrupt conditions have a shared enable control bit. These enable bits can be used to mask out individual USIM UART mode interrupt sources.

The address detect condition, which is also an USIM UART mode interrupt source, does not have an associated flag, but will generate an USIM interrupt when an address detect condition occurs if its function is enabled by setting the UADDEN bit in the UUCR2 register. An RX pin wake-up, which is also an USIM UART mode interrupt source, does not have an associated flag, but will generate an USIM interrupt if the UART clock ( $f_H$ ) source is switched off and the UWAKE and URIE bits in the UUCR2 register are set when a falling edge on the RX pin occurs. Note that in the event of an RX wake-up interrupt occurring, there will be a certain period of delay, commonly known as the System Start-up Time, for the oscillator to restart and stabilize before the system resumes normal operation.

Note that the UUSR register flags are read only and cannot be cleared or set by the application program, neither will they be cleared when the program jumps to the corresponding interrupt servicing routine, as is the case for some of the other interrupts. The flags will be cleared automatically when certain actions are taken by the UART, the details of which are given in the UART register section. The overall UART interrupt can be disabled or enabled by the USIM interrupt enable control bit in the interrupt control register of the microcontroller to decide whether the interrupt requested by the UART module is masked out or allowed.





#### **UART Interrupt Structure**

#### Address Detect Mode

Setting the Address Detect Mode bit, UADDEN, in the UUCR2 register, enables this special mode. If this bit is enabled then an additional qualifier will be placed on the generation of a Receiver Data Available interrupt, which is requested by the URXIF flag. If the UADDEN bit is enabled, then when data is available, an interrupt will only be generated, if the highest received bit has a high value. Note that the USIME and EMI interrupt enable bits must also be enabled for correct interrupt generation. This highest address bit is the 9th bit if UBNO=1 or the 8th bit if UBNO=0. If this bit is high, then the received word will be defined as an address rather than data. A Data Available interrupt will be generated every time the last bit of the received word is set. If the UADDEN bit is not enabled, then a Receiver Data Available interrupt will be generated each time the URXIF flag is set, irrespective of the data last bit status. The address detect mode and parity enable are mutually exclusive functions. Therefore if the address detect mode is enabled, then to ensure correct operation, the parity function should be disabled by resetting the parity enable bit UPREN to zero.

UADDEN	9th bit if UBNO=1 8th bit if UBNO=0	USIM Interrupt Generated
0	0	$\checkmark$
0	1	$\checkmark$
4	0	×
	1	$\checkmark$

**UADDEN Bit Function** 

#### UART Power Down and Wake-up

When the UART clock  $(f_H)$  is off, the UART will cease to function, all clock sources to the module are shutdown. If the UART clock  $(f_H)$  is off while a transmission is still in progress, then the transmission will be paused until the UART clock source derived from the microcontroller is activated. In a similar way, if the MCU enters the IDLE or SLEEP Mode while receiving data, then the reception of data will likewise be paused. When the MCU enters the IDLE or SLEEP Mode, note that the UUSR, UUCR1, UUCR2, transmit and receive registers, as well as the UBRG register will not be affected. It is recommended to make sure first that the UART data transmission or reception has been finished before the microcontroller enters the IDLE or SLEEP mode.

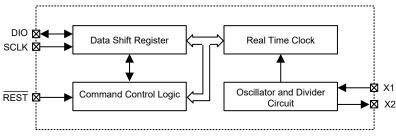


The UART function contains a receiver RX pin wake-up function, which is enabled or disabled by the UWAKE bit in the UUCR2 register. If this bit, along with the UART mode selection bit, UMD, the UART enable bit, UREN, the receiver enable bit, URXEN and the receiver interrupt bit, URIE, are all set when the UART clock ( $f_{\rm H}$ ) is off, then a falling edge on the RX pin will trigger an RX pin wake-up UART interrupt. Note that as it takes certain system clock cycles after a wake-up, before normal microcontroller operation resumes, any data received during this time on the RX pin will be ignored.

For a UART wake-up interrupt to occur, in addition to the bits for the wake-up being set, the global interrupt enable bit, EMI, and the USIM interrupt enable bit, USIME, must be set. If the EMI and USIME bits are not set then only a wake up event will occur and no interrupt will be generated. Note also that as it takes certain system clock cycles after a wake-up before normal microcontroller resumes, the USIM interrupt will not be generated until after this time has elapsed.

## Calendar

The Calendar has several registers to store the corresponding information with 8-bit data format. A 32768Hz crystal is required to provide the correct timing. In order to minimize the pin number, the Calendar uses a serial I/O transmission method to interface with the MCU. Only three wires are required: (1) REST, (2) SCLK and (3) DIO. Data can be delivered 1 byte at a time or in a burst of up to 8 bytes.



Calendar Block Diagram

The Calendar mainly contains the following internal elements: a data shift register array to store the clock/calendar data, command control logic, oscillator circuit and read timer clock. The clock is contained in eight read/write registers as shown below. Data contained in the clock register is in binary coded decimal format.

Two modes are available for transferring the data between the MCU and the Calendar. One is in single-byte mode and the other is in multiple-byte mode.

The Calendar also contains two additional bits, the Clock Halt bit (CH) and the Write Protect bit (WP). These bits control the operation of the oscillator and so data can be written to the register array. These two bits should first be specified in order to read from and write to the register array properly.

## **Command Byte**

For each data transfer, a Command Byte is initiated to specify which register is accessed. This is to determine whether a read, write, or test cycle is operated and whether a single byte or burst mode transfer is to occur. Refer to the table shown below and follow the steps to write the data to the Calendar. First give a Command Byte of the Calendar, and then write a data in the register.

This table illustrates the correlation between Command Byte and their bits:

Eurotian Departmention	Command Byte								
Function Description	C7	C6	C5	C4	C3	C2	C1	C0	
Select Read or Write Cycle	_	_	_	_	—	_	_	R/W	
Specify the Register to be Accessed	_	_	_	_	A2	A1	A0	—	
Clock Halt Flag	С		_	_	—	_	_	_	
For Calendar module Test Only	1	0	0	1	х	х	x	1	
Select Single Byte or Burst Mode	1	0	1	1	1	1	1	х	

"x": don't care

#### The following table shows the register address and its data format:

Register	Range			Reg	jister	Defini	tion			Address	Bit	Command
Name	Data	D7	D6	D5	D4	D3	D2	D1	D0	A2~A0	R/W	Byte
Seconds	00~59	СН		I0 SEC	2	SEC			000	W R	10000000 10000001	
Minutes	00~59	0		10 MIN	1	MIN			001	W R	10000010 10000011	
Hours	01~12 00~23	12\ 24	0 0	AP 10	HR HR	HOUR			010	W R	10000100 10000101	
Date	01~31	0	0	10 D	ATE		DATE			011	W R	10000110 10000111
Month	01~12	0	0	0	10M		MOI	NTH		100	W R	10001000 10001001
Day	01~07	0	0	0	0		D/	۹Y		101	W R	10001010 10001011
Year	00~99		10 Y	EAR	R YEAR			AR		110	W R	10001100 10001101
Write Protect	00~80	WP		ALWAYS ZERO				111	W R	10001110 10001111		

CH: Clock Halt bit

CH=0 oscillator enabled CH=1 oscillator disabled

WP: Write Protect bit

WP=0 register data can be written in

WP=1 register data can not be written in

Bit 7 of Reg2: 12/24 mode flag

bit 7=1, 12-hour mode bit 7=0, 24-hour mode

- Bit 5 of Reg2: AM/PM mode defined
  - AP=1 PM mode
  - AP=0 AM mode

#### **R/W Signal**

The LSB of the Command Byte determines whether the data in the register be read or be written to.

When it is set as "0" means that a write cycle is to take place otherwise this Calendar will be set into the read mode.

### A0~A2

A0 to A2 of the Command Byte is used to specify which registers are to be accessed. There are eight registers used to control the month data, etc., and each of these registers have to be set as a write cycle in the initial time.



## Burst Mode

When the Command Byte is 10111110 (or 10111111), the Calendar is configured in burst mode. In this mode the eight clock/calendar registers can be written (or read) in series, starting with bit 0 of register address 0 (see the following Burst Mode Transfer timing diagram).

#### **Test Mode**

When the Command Byte is set as 1001xxx1, the Calendar is configured in test mode. The test mode is used by Holtek only for testing purposes. If used generally, unpredictable conditions may occur.

#### Write Protect Register

This register is used to prevent a write operation to any other register. Data can be written into the designated register only if the Write Protect signal (WP) is set to logic 0. The Write Protect Register should be set first before restarting the system or before writing the new data to the system, and it should set as logic 1 in the read cycle. The Write Protect bit cannot be written to in the burst mode.

#### **Clock Halt Bit**

D7 of the Seconds Register is defined as the Clock Halt Flag (CH).

When this bit is set to logic 1, the clock oscillator is stopped and the Calendar goes into a low-power standby mode. When this bit is written to logic 0, the clock will start.

#### 12-hour/24-hour Mode

The D7 of the hour register is defined as the 12-hour or 24-hour mode select bit.

When this bit is in high level, the 12-hour mode is selected otherwise it's the 24-hour mode.

#### AM-PM Mode

These are two functions for the D5 of the hour register determined by the value D7 of the same register.

One is used in AM/PM selection on the 12-hour mode. When D5 is logic 1, it is PM, otherwise it's AM. The other is used to set the second 10-hour bit (20~23 hours) on the 24-hour mode.

## Reset and Serial Clock Control

The  $\overline{\text{REST}}$  pin is used to allow access data to the shift register like a toggle switch. When the  $\overline{\text{REST}}$  pin is taken high, the built-in control logic is turned on and the address/command sequence can access the corresponding shift register. The  $\overline{\text{REST}}$  pin is also used to terminate either single-byte or burst mode data format.

The input signal of SCLK is a sequence of a falling edge followed by a rising edge and it is used to synchronize the register data whether read or write. For data input, the data must be read after the rising edge of SCLK. The data on the DIO pin becomes output mode after the falling edge of the SCLK. All data transfer terminates if the REST pin is low and the DIO pin goes to a high impedance state.

## Data Input and Data Out

In writing a data byte with the Calendar, the read/write should first set as R/W=0 in the Command Byte and follow with the corresponding data register on the rising edge of the next eight SCLK cycles. Additional SCLK cycles are ignored. Data inputs are entered starting with bit 0.

In reading a data on the register of the Calendar, R/W=1 should first be entered as input. The data bit outputs on the falling edge of the next eight SCLK cycles. Note that the first data bit to be transmitted on the first falling edge after the last bit of the read command byte is written. Additional SCLK cycles re-transmits the data bytes as long as REST remains at high level. Data outputs are read starting with bit 0.

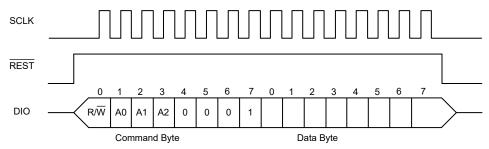


## **Crystal Selection**

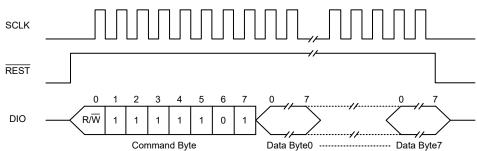
A 32768Hz crystal can be directly connected to the crystal X1 and X2 pins. In order to ensure that the desired frequency is achieved, it is recommended to use a crystal with a capacitance of 9.0pF. It is not recommended that additional load capacitors are connected to the X1 and X2 pins. Refer to the following table for the crystal specifications.

The following diagram shows the single and burst mode transfer:

#### Single Byte Transfer



#### **Burst Mode Transfer**



#### **Crystal Specifications**

Symbol	Parameter	Parameter Min. Typ.		Max.	Unit
f <sub>o</sub>	Nominal Frequency	_	32.768	—	kHz
ESR	Series Resistance	—	—	50	kΩ
CL	Load Capacitance	—	9.0	—	pF

Note: 1. It is strongly recommended to use a crystal with a load capacitance of 9.0pF. Never use a crystal with a load capacitance of 12.5pF.

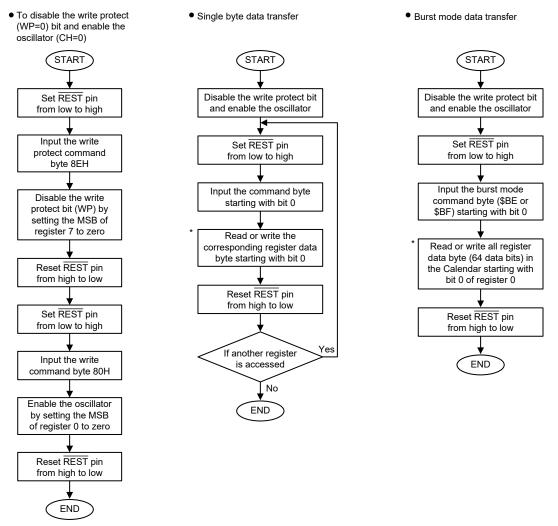
2. The oscillator selection can be optimized using a high quality resonator with a small ESR value. Refer to the crystal manufacturer for more details: <u>www.microcrystal.com</u>.

#### **Operating Flowchart**

To initiate any transfer of data, **REST** is taken high and an 8-bit command byte is first loaded into the control logic to provide the register address and command information. Following the command word, the clock/calendar data is serially transferred to or from the corresponding register. The **REST** pin must be taken low again after the transfer operation is completed. All data enter on the rising edge of SCLK and outputs on the falling edge of SCLK. In total, 16 clock pulses are needed for a single byte mode and 72 for burst mode. Both input and output data starts with bit 0.



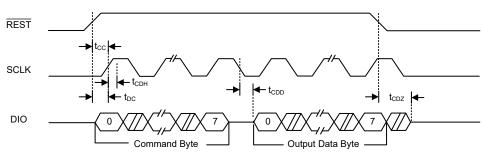
In using the Calendar, set first the WP and CH to 0 and wait for about 3 seconds, the oscillator will generate the clocks for internal use. Then, choose either single mode or burst mode to input the data. The read or write operating flowcharts are shown below.



Note: \* In reading data byte from the Calendar register, the first data bit to be transmitted at the first falling edge after the last bit of the command byte is written.

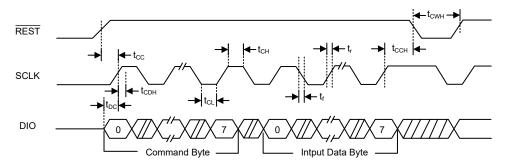
## **Timing Diagrams**

#### **Read Data Transfer**





## Write Data Transfer



## Low Voltage Detector – LVD

The devices have a Low Voltage Detector function, also known as LVD. This enabled the devices to monitor the power supply voltage,  $V_{DD}$ , and provide a warning signal should it fall below a certain level. This function may be especially useful in battery applications where the supply voltage will gradually reduce as the battery ages, as it allows an early warning battery low signal to be generated. The Low Voltage Detector also has the capability of generating an interrupt signal.

## LVD Register

The Low Voltage Detector function is controlled using a single register with the name LVDC. Three bits in this register, VLVD2~VLVD0, are used to select one of eight fixed voltages below which a low voltage condition will be determined. A low voltage condition is indicated when the LVDO bit is set. If the LVDO bit is low, this indicates that the V<sub>DD</sub> voltage is above the preset low voltage value. The LVDEN bit is used to control the overall on/off function of the low voltage detector. Setting the bit high will enable the low voltage detector. Clearing the bit to zero will switch off the internal low voltage detector circuits. As the low voltage detector will consume a certain amount of power, it may be desirable to switch off the circuit when not in use, an important consideration in power sensitive battery powered applications.

Bit	7	6	5	4	3	2	1	0	
Name	—	—	LVDO	LVDEN	VBGEN	VLVD2	VLVD1	VLVD0	
R/W	_	—	R	R/W	R/W	R/W	R/W	R/W	
POR	—	<u> </u>							
Bit 7~6	Unimplemented, read as "0"								
Bit 5	LVDO: LVD Output Flag 0: No Low Voltage Detected 1: Low Voltage Detected								
Bit 4	LVDEN 0: Disa 1: Ena	able	age Detecto	or Control					
Bit 3 VBGEN: Bandgap Buffer Control 0: Disable 1: Enable									
	Note that the Bandgap circuit is enabled when the LVD or the LVR function is enabled or when the VBGEN bit is set high.								

#### LVDC Register

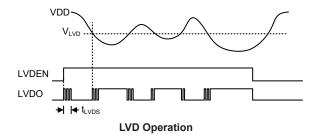
111: 4.0V



Bit 2~0 **VLVD2~VLVD0**: Select LVD Voltage 000: 2.0V 001: 2.2V 010: 2.4V 011: 2.7V 100: 3.0V 101: 3.3V 110: 3.6V

## **LVD Operation**

The Low Voltage Detector function operates by comparing the power supply voltage,  $V_{DD}$ , with a pre-specified voltage level stored in the LVDC register. This has a range of between 2.0V and 4.0V. When the power supply voltage,  $V_{DD}$ , falls below this pre-determined value, the LVDO bit will be set high indicating a low power supply voltage condition. The Low Voltage Detector function is supplied by a reference voltage which will be automatically enabled. When the devices are in the SLEEP mode, the low voltage detector will be disabled even if the LVDEN bit is high. After enabling the Low Voltage Detector, a time delay  $t_{LVDS}$  should be allowed for the circuitry to stabilise before reading the LVDO bit. Note also that as the  $V_{DD}$  voltage may rise and fall rather slowly, at the voltage nears that of  $V_{LVD}$ , there may be multiple bit LVDO transitions.



The Low Voltage Detector also has its own interrupt, providing an alternative means of low voltage detection, in addition to polling the LVDO bit. The interrupt will only be generated after a delay of  $t_{LVD}$  after the LVDO bit has been set high by a low voltage condition. In this case, the LVF interrupt request flag will be set, causing an interrupt to be generated if  $V_{DD}$  falls below the preset LVD voltage. This will cause the devices to wake-up from the IDLE Mode, however if the Low Voltage Detector wake up function is not required then the LVF flag should be first set high before the devices enter the IDLE Mode. The LVD function is always disabled in the SLEEP mode.



## Interrupts

Interrupts are an important part of any microcontroller system. When an external event or an internal function such as a Timer Module or an A/D converter requires microcontroller attention, their corresponding interrupt will enforce a temporary suspension of the main program allowing the microcontroller to direct attention to their respective needs. The devices contain several external interrupt and internal interrupts functions. The external interrupts are generated by the action of the external INT0~INT1 pins, while the internal interrupts are generated by various internal functions such as the TMs, Time Bases, LVD, EEPROM and the A/D converter.

## Interrupt Registers

Overall interrupt control, which basically means the setting of request flags when certain microcontroller conditions occur and the setting of interrupt enable bits by the application program, is controlled by a series of registers, located in the Special Purpose Data Memory. The first is the INTCO~INTC3 registers which setup the primary interrupts, the second is the INTEG register which setup the external interrupt trigger edge type.

Each register contains a number of enable bits to enable or disable individual registers as well as interrupt flags to indicate the presence of an interrupt request. The naming convention of these follows a specific pattern. First is listed an abbreviated interrupt type, then the (optional) number of that interrupt followed by either an "E" for enable/disable bit or "F" for request flag.

Function	Enable Bit	Request Flag	Notes
Global	EMI	—	—
INTn Pin	INTnE	INTnF	n=0~1
A/D Converter	ADE	ADF	—
Time Base	TBnE	TBnF	n=0~1
USIM	USIME	USIMF	_
LVD	LVE	LVF	
EEPROM	DEE	DEF	
STM	STMPE	STMPF	
	STMAE	STMAF	
PTM	PTMPE	PTMPF	
	PTMAE	PTMAF	

Note: For the BA45F5740-2 device, the INT0~INT1 external inputs are unavailable. Interrupt Register Bit Naming Conventions

Register		Bit								
Name	7	6	5	4	3	2	1	0		
INTEG	—	—	_	—	INT1S1	INT1S0	INT0S1	INT0S0		
INTC0	—	INT1F	<b>INT0F</b>	D4	INT1E	INT0E	D1	EMI		
INTC1	DEF	ADF	LVF	USIMF	DEE	ADE	LVE	USIME		
INTC2	STMAF	STMPF	PTMAF	PTMPF	STMAE	STMPE	PTMAE	PTMPE		
INTC3	_	D6	TB1F	TB0F	_	D2	TB1E	TB0E		

Interrupt Register List



#### INTEG Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	INT1S1	INT1S0	INT0S1	INT0S0
R/W	—	—	—	—	R/W	R/W	R/W	R/W
POR	—	—	_	—	0	0	0	0

Bit 7~4 Unimplemented, read as "0"

- Bit 3~2 INT1S1~INT1S0: Interrupt edge control for INT1 pin
  - 00: Disable
    - 01: Rising edge
    - 10: Falling edge
  - 11: Rising and falling edges
  - Note: For the BA45F5740-2 device, the INT1 external input pin is unavailable, so these bits should be fixed at "00".

#### Bit 1~0 INT0S1~INT0S0: Interrupt edge control for INT0 pin

- 00: Disable
- 01: Rising edge
- 10: Falling edge
- 11: Rising and falling edges
- Note: For the BA45F5740-2 device, the INT0 external input pin is unavailable, so these bits should be fixed at "00".

#### INTC0 Register

Bit	7	6	5	4	3	2	1	0			
Name	_	INT1F	INTOF	D4	INT1E	INT0E	D1	EMI			
R/W	—	R/W	R/W	R/W	R/W	R/W	R/W	R/W			
POR		0	0	0	0	0	0	0			
Bit 7	Unimple	emented, re	ad as "0"								
Bit 6	INT1F:	INT1F: INT1 interrupt request flag									
		0: No request									
		1: Interrupt request									
		or the BA4: t should be			NT1 extern	al input pin	ı is unavaila	able, so thi			
Bit 5	INT0F:	INT0F: INT0 interrupt request flag									
		request									
		1: Interrupt request									
	Note: For the BA45F5740-2 device, the INT0 external input pin is unavailable, so this bit should be fixed at "0".										
Bit 4	D4: Res	erved, shou	ld be fixed	at "0"							
Bit 3	INT1E:	INT1 inter	rupt control								
	0: Disa										
	1: Ena										
		or the BA4: t should be			NT1 extern	al input pin	i is unavaila	able, so thi			
Bit 2	INT0E:	INT0 inter	rupt control								
	0: Disa										
	1: Ena										
		or the BA4: t should be			NT0 extern	al input pin	ı is unavaila	able, so thi			
Bit 1	D1: Res	erved, shou	ld be fixed	at "0"							
Bit 0	EMI: G	lobal interr	upt control								
	0: Disa	able									
	1: Ena	ble									



## INTC1 Register

Bit	7	6	5	4	3	2	1	0
Name	DEF	ADF	LVF	USIMF	DEE	ADE	LVE	USIME
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0
Bit 7	<b>DEF</b> : Data EEPROM interrupt request flag 0: No request							
D'		rrupt reque		. 0				
Bit 6			er interrupt	request flag	5			
		request	_4					
Bit 5		rrupt reque: /D interrup		10				
DII J		request	i lequest lla	ig				
		rrupt request	at					
Bit 4		: USIM inte		est flao				
Dit i		request	inaprioque	ist mag				
		rrupt reque	st					
Bit 3		ata EEPRO		t control				
	0: Disa	able						
	1: Ena	ble						
Bit 2	ADE: A	/D Convert	er interrupt	control				
	0: Disa	able						
	1: Ena							
Bit 1		/D interrup	t control					
		0: Disable						
	1: Enable							
Bit 0	USIME: USIM interrupt control							
	0: Disa							
	1: Ena	ble						

## INTC2 Register

	<u> </u>								
Bit	7	6	5	4	3	2	1	0	
Name	STMAF	STMPF	PTMAF	PTMPF	STMAE	STMPE	PTMAE	PTMPE	
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	
POR	0	0 0 0 0 0 0 0 0							
Bit 7	<ul> <li>STMAF: STM Comparator A match interrupt request flag</li> <li>0: No request</li> <li>1: Interrupt request</li> </ul>								
Bit 6									
Bit 5	<b>PTMAF</b> : PTM Comparator A match interrupt request flag 0: No request 1: Interrupt request								
Bit 4	0: No 1	: PTM Con request rrupt reque		match inter	rupt request	t flag			
Bit 3	t 3 STMAE: STM Comparator A match interrupt control 0: Disable 1: Enable								
Bit 2	STMPE: STM Comparator P match interrupt control 0: Disable 1: Enable								



Bit 1	PTMAE: PTM Comparator A match interrupt control
	0: Disable
	1: Enable

- Bit 0 **PTMPE**: PTM Comparator P match interrupt control 0: Disable
  - 1: Enable

#### INTC3 Register

Bit	7	6	5	4	3	2	1	0
Name	—	D6	TB1F	TB0F	—	D2	TB1E	TB0E
R/W	—	R/W	R/W	R/W	—	R/W	R/W	R/W
POR	—	0	0	0		0	0	0
Bit 7	Unimplemented, read as "0"							
Bit 6	<b>D6</b> : Res	erved, shou	ld be fixed	at "0"				
Bit 5	<b>TB1F</b> : Time Base 1 interrupt request flag 0: No request 1: Interrupt request							
Bit 4	<b>TB0F</b> : Time Base 0 interrupt request flag 0: No request 1: Interrupt request							
Bit 3	Unimplemented, read as "0"							
Bit 2	<b>D2</b> : Rese	erved, shou	ld be fixed	at "0"				
Bit 1	<b>TB1E</b> : Time Base 1 interrupt control 0: Disable 1: Enable							
Bit 0	<b>TB0E</b> : Time Base 0 interrupt control 0: Disable 1: Enable							

## **Interrupt Operation**

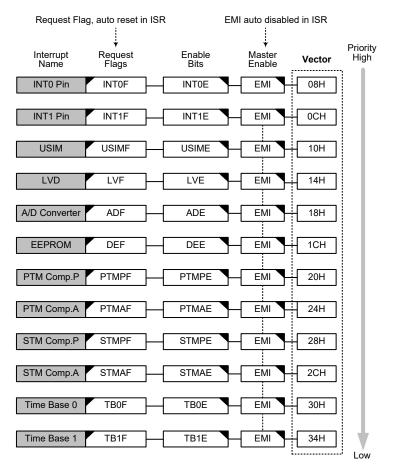
When the conditions for an interrupt event occur, such as a TM Comparator P, Comparator A match or A/D conversion completion etc., the relevant interrupt request flag will be set. Whether the request flag actually generates a program jump to the relevant interrupt vector is determined by the condition of the interrupt enable bit. If the enable bit is set high then the program will jump to its relevant vector; if the enable bit is zero then although the interrupt request flag is set an actual interrupt will not be generated and the program will not jump to the relevant interrupt vector. The global interrupt enable bit, if cleared to zero, will disable all interrupts.

When an interrupt is generated, the Program Counter, which stores the address of the next instruction to be executed, will be transferred onto the stack. The Program Counter will then be loaded with a new address which will be the value of the corresponding interrupt vector. The microcontroller will then fetch its next instruction from this interrupt vector. The instruction at this vector will usually be a "JMP" which will jump to another section of program which is known as the interrupt service routine. Here is located the code to control the appropriate interrupt. The interrupt service routine must be terminated with a "RETI", which retrieves the original Program Counter address from the stack and allows the microcontroller to continue with normal execution at the point where the interrupt occurred.

The various interrupt enable bits, together with their associated request flags, are shown in the accompanying diagrams with their order of priority. All interrupt sources have their own individual vector. Once an interrupt subroutine is serviced, all the other interrupts will be blocked, as the global interrupt enable bit, EMI bit will be cleared automatically. This will prevent any further interrupt nesting from occurring. However, if other interrupt requests occur during this interval, although the

interrupt will not be immediately serviced, the request flag will still be recorded.

If an interrupt requires immediate servicing while the program is already in another interrupt service routine, the EMI bit should be set after entering the routine, to allow interrupt nesting. If the stack is full, the interrupt request will not be acknowledged, even if the related interrupt is enabled, until the Stack Pointer is decremented. If immediate service is desired, the stack must be prevented from becoming full. In case of simultaneous requests, the accompanying diagram shows the priority that is applied. All of the interrupt request flags when set will wake-up the devices if it is in SLEEP or IDLE Mode, however to prevent a wake-up from occurring the corresponding flag should be set before the devices are in SLEEP or IDLE Mode.



Note: For the BA45F5740-2 device, the INT0~INT1 external input pins are unavailable. Interrupt Structure

## **External Interrupts**

The external interrupt is controlled by signal transitions on the INTn pin. An external interrupt request will take place when the external interrupt request flag, INTnF, is set, which will occur when a transition, whose type is chosen by the edge select bits, appears on the external interrupt pin. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, and respective external interrupt enable bit, INTnE, must first be set. Additionally the correct interrupt edge type must be selected using the INTEG register to enable the external interrupt function and to choose the trigger edge type. As the external interrupt pin is pin-shared with I/O pin, it can only be configured as external interrupt pin if its external interrupt enable bit

in the corresponding interrupt register has been set and the external interrupt pin is selected by the corresponding pin-shared function selection bits. The pin must also be setup as an input by setting the corresponding bit in the port control register. When the interrupt is enabled, the stack is not full and the correct transition type appears on the external interrupt pin, a subroutine call to the external interrupt vector, will take place. When the interrupt is serviced, the external interrupt request flag, INTnF, will be automatically reset and the EMI bit will be automatically cleared to disable other interrupts. Note that any pull-high resistor selections on the external interrupt pin will remain valid even if the pin is used as an external interrupt. The INTEG register is used to select the type of active edge that will trigger the external interrupt. Note that the INTEG register can also be used to disable the external interrupt function.

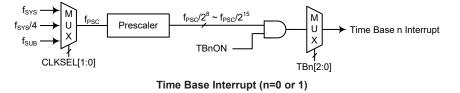
## A/D Converter Interrupt

The A/D Converter Interrupt is controlled by the termination of an A/D conversion process. An A/D Converter Interrupt request will take place when the A/D Converter Interrupt request flag, ADF, is set, which occurs when the A/D conversion process finishes. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, and A/D Interrupt enable bit, ADE, must first be set. When the interrupt is enabled, the stack is not full and the A/D conversion process has ended, a subroutine call to the A/D Converter Interrupt vector, will take place. When the interrupt is serviced, the A/D Converter Interrupt flag, ADF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

#### **Time Base Interrupts**

The function of the Time Base Interrupts is to provide regular time signal in the form of an internal interrupt. They are controlled by the overflow signals from their respective timer functions. When these happens their respective interrupt request flags, TBnF will be set. To allow the program to branch to their respective interrupt vector addresses, the global interrupt enable bit, EMI and Time Base enable bits, TBnE, must first be set. When the interrupt is enabled, the stack is not full and the Time Base overflows, a subroutine call to their respective vector locations will take place. When the interrupt is serviced, the respective interrupt request flag, TBnF, will be automatically reset and the EMI bit will be cleared to disable other interrupts.

The purpose of the Time Base Interrupt is to provide an interrupt signal at fixed time periods. Its clock source,  $f_{PSC}$ , originates from the internal clock source  $f_{SYS}$ ,  $f_{SYS}/4$  or  $f_{SUB}$  and then passes through a divider, the division ratio of which is selected by programming the appropriate bits in the TBnC register to obtain longer interrupt periods whose value ranges. The clock source which in turn controls the Time Base interrupt period is selected using the CLKSEL1~CLKSEL0 bits in the PSCR register.





#### PSCR Register

Bit	7	6	5	4	3	2	1	0
Name	_	_	—	_	—	—	CLKSEL1	CLKSEL0
R/W	_	_	_		_	_	R/W	R/W
POR	_	_	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as "0"

Bit 1~0 CLKSEL1~CLKSEL0: Prescaler clock source selection

#### • TBnC Register (n=0 or 1)

Bit	7	6	5	4	3	2	1	0
Name	TBnON	—	—	—	_	TBn2	TBn1	TBn0
R/W	R/W	_	_	_	_	R/W	R/W	R/W
POR	0	_			_	0	0	0

Bit 7	TBnON: Time Base n Control
	0: Disable
	1: Enable
Bit 6~3	Unimplemented, read as "0"
Bit 2~0	TBn2~TBn0: Select Time Base n Time-out Period
	$000: 2^8/f_{PSC}$
	$001: 2^{9}/f_{PSC}$
	010: $2^{10}/f_{PSC}$
	$011: 2^{11}/f_{PSC}$
	100: $2^{12}/f_{PSC}$
	$101: 2^{13}/f_{PSC}$
	110: $2^{14}/f_{PSC}$
	111: $2^{15}/f_{PSC}$
	$\begin{array}{c} 100:\ 2^{12}/f_{PSC} \\ 101:\ 2^{13}/f_{PSC} \\ 110:\ 2^{14}/f_{PSC} \end{array}$

## USIM Interrupt

The Universal Serial Interface Module Interrupt, also known as the USIM interrupt, will take place when the USIM Interrupt request flag, USIMF, is set. As the USIM interface can operate in three modes which are SPI mode, I<sup>2</sup>C mode and UART mode, the USIMF flag can be set by different conditions depending on the selected interface mode.

If the SPI or I<sup>2</sup>C mode is selected, the USIM interrupt can be triggered when a byte of data has been received or transmitted by the USIM SPI or I<sup>2</sup>C interface, or an I<sup>2</sup>C slave address match occurs, or an I<sup>2</sup>C bus time-out occurs. If the UART mode is selected, several individual UART conditions including a transmitter data register empty, transmitter idle, receiver data available, receiver overrun, address detect and an RX pin wake-up, can generate a USIM interrupt with the USIMF flag bit set high.

To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, and the Universal Serial Interface Module Interrupt enable bit, USIME, must first be set. When the interrupt is enabled, the stack is not full and any of the above described situations occurs, a subroutine call to the respective Interrupt vector, will take place. When the interrupt is serviced, the Universal Serial Interface Module Interrupt flag, USIMF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

Note that if the USIM interrupt is triggered by the UART interface, after the interrupt has been serviced, the UUSR register flags will only be cleared when certain actions are taken by the UART, the details of which are given in the UART section.

<sup>00:</sup> fsys

<sup>01:</sup>  $f_{SYS}/4$ 

 $<sup>1</sup>x: f_{SUB}$ 



## LVD Interrupt

An LVD Interrupt request will take place when the LVD Interrupt request flag, LVF, is set, which occurs when the Low Voltage Detector function detects a low power supply voltage. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, and Low Voltage Interrupt enable bit, LVE, must first be set. When the interrupt is enabled, the stack is not full and a low voltage condition occurs, a subroutine call to the LVD Interrupt vector, will take place. When the Low Voltage Interrupt is serviced, the LVD Interrupt flag, LVF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

## **EEPROM** Interrupt

An EEPROM Interrupt request will take place when the EEPROM Interrupt request flag, DEF, is set, which occurs when an EEPROM Write cycle ends. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, and EEPROM Interrupt enable bit, DEE, must first be set. When the interrupt is enabled, the stack is not full and an EEPROM Write cycle ends, a subroutine call to the EEPROM Interrupt vector will take place. When the EEPROM Interrupt is serviced, the EEPROM Interrupt flag, DEF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

## TM Interrupts

The Standard and Periodic Type TMs each has two interrupts, one comes from the comparator A match situation and the other comes from the comparator P match situation. For each of the Standard and Periodic Type TM there are two interrupt request flags xTMPF and xTMAF and two enable bits xTMPE and xTMAE. A TM interrupt request will take place when any of the TM request flags are set, a situation which occurs when a TM comparator P or A match situation happens.

To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, respective TM Interrupt enable bit must first be set. When the interrupt is enabled, the stack is not full and a TM comparator match situation occurs, a subroutine call to the TM comparator P or comparator A Interrupt vector locations, will take place. When the TM interrupt is serviced, the TM interrupt request flag will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

## Interrupt Wake-up Function

Each of the interrupt functions has the capability of waking up the microcontroller when in the SLEEP or IDLE Mode. A wake-up is generated when an interrupt request flag changes from low to high and is independent of whether the interrupt is enabled or not. Therefore, even though the devices are in the SLEEP or IDLE Mode and its system oscillator stopped, situations such as external edge transitions on the external interrupt pins or a low power supply voltage may cause their respective interrupt flag to be set high and consequently generate an interrupt. Care must therefore be taken if spurious wake-up situations are to be avoided. If an interrupt wake-up function is to be disabled then the corresponding interrupt request flag should be set high before the devices enter the SLEEP or IDLE Mode. The interrupt enable bits have no effect on the interrupt wake-up function.

## **Programming Considerations**

By disabling the relevant interrupt enable bits, a requested interrupt can be prevented from being serviced, however, once an interrupt request flag is set, it will remain in this condition in the interrupt register until the corresponding interrupt is serviced or until the request flag is cleared by the application program.

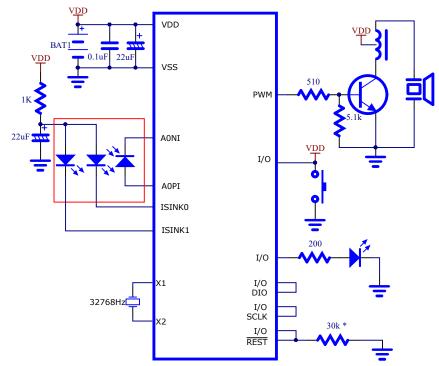
It is recommended that programs do not use the "CALL" instruction within the interrupt service subroutine. Interrupts often occur in an unpredictable manner or need to be serviced immediately. If only one stack is left and the interrupt is not well controlled, the original control sequence will be damaged once a CALL subroutine is executed in the interrupt subroutine.

Every interrupt has the capability of waking up the microcontroller when it is in SLEEP or IDLE Mode, the wake up being generated when the interrupt request flag changes from low to high. If it is required to prevent a certain interrupt from waking up the microcontroller then its respective request flag should be first set high before enter SLEEP or IDLE Mode.

As only the Program Counter is pushed onto the stack, then when the interrupt is serviced, if the contents of the accumulator, status register or other registers are altered by the interrupt service program, their contents should be saved to the memory at the beginning of the interrupt service routine. To return from an interrupt subroutine, either a RET or RETI instruction may be executed. The RETI instruction in addition to executing a return to the main program also automatically sets the EMI bit high to allow further interrupts. The RET instruction however only executes a return to the main program leaving the EMI bit in its present zero state and therefore disabling the execution of further interrupts.

# **Application Circuits**

## **Stand-alone Smoke Detector Applications**



Note: "\*" It is recommended to add a pull-down resistor.



## **Instruction Set**

## Introduction

Central to the successful operation of any microcontroller is its instruction set, which is a set of program instruction codes that directs the microcontroller to perform certain operations. In the case of Holtek microcontroller, a comprehensive and flexible set of over 60 instructions is provided to enable programmers to implement their application with the minimum of programming overheads.

For easier understanding of the various instruction codes, they have been subdivided into several functional groupings.

## **Instruction Timing**

Most instructions are implemented within one instruction cycle. The exceptions to this are branch, call, or table read instructions where two instruction cycles are required. One instruction cycle is equal to 4 system clock cycles, therefore in the case of an 8MHz system oscillator, most instructions would be implemented within 0.5µs and branch or call instructions would be implemented within 1µs. Although instructions which require one more cycle to implement are generally limited to the JMP, CALL, RET, RETI and table read instructions, it is important to realize that any other instructions which involve manipulation of the Program Counter Low register or PCL will also take one more cycle to implement. As instructions which change the contents of the PCL will imply a direct jump to that new address, one more cycle will be required. Examples of such instructions would be "CLR PCL" or "MOV PCL, A". For the case of skip instructions, it must be noted that if the result of the comparison involves a skip operation then this will also take one more cycle, if no skip is involved then only one cycle is required.

## Moving and Transferring Data

The transfer of data within the microcontroller program is one of the most frequently used operations. Making use of three kinds of MOV instructions, data can be transferred from registers to the Accumulator and vice-versa as well as being able to move specific immediate data directly into the Accumulator. One of the most important data transfer applications is to receive data from the input ports and transfer data to the output ports.

## **Arithmetic Operations**

The ability to perform certain arithmetic operations and data manipulation is a necessary feature of most microcontroller applications. Within the Holtek microcontroller instruction set are a range of add and subtract instruction mnemonics to enable the necessary arithmetic to be carried out. Care must be taken to ensure correct handling of carry and borrow data when results exceed 255 for addition and less than 0 for subtraction. The increment and decrement instructions INC, INCA, DEC and DECA provide a simple means of increasing or decreasing by a value of one of the values in the destination specified.



#### Logical and Rotate Operation

The standard logical operations such as AND, OR, XOR and CPL all have their own instruction within the Holtek microcontroller instruction set. As with the case of most instructions involving data manipulation, data must pass through the Accumulator which may involve additional programming steps. In all logical data operations, the zero flag may be set if the result of the operation is zero. Another form of logical data manipulation comes from the rotate instructions such as RR, RL, RRC and RLC which provide a simple means of rotating one Bit right or left. Different rotate instructions exist depending on program requirements. Rotate instructions are useful for serial port programming applications where data can be rotated from an internal register into the Carry Bit from where it can be examined and the necessary serial Bit set high or low. Another applications which rotate data operations are used is to implement multiplication and division calculations.

#### Branches and Control Transfer

Program branching takes the form of either jumps to specified locations using the JMP instruction or to a subroutine using the CALL instruction. They differ in the sense that in the case of a subroutine call, the program must return to the instruction immediately when the subroutine has been carried out. This is done by placing a return instruction "RET" in the subroutine which will cause the program to jump back to the address right after the CALL instruction. In the case of a JMP instruction, the program simply jumps to the desired location. There is no requirement to jump back to the original jumping off point as in the case of the CALL instruction. One special and extremely useful set of branch instructions are the conditional branches. Here a decision is first made regarding the condition of a certain data memory or individual Bits. Depending upon the conditions, the program will continue with the next instruction or skip over it and jump to the following instruction. These instructions are the key to decision making and branching within the program perhaps determined by the condition of certain input switches or by the condition of internal data Bits.

#### **Bit Operations**

The ability to provide single Bit operations on Data Memory is an extremely flexible feature of all Holtek microcontrollers. This feature is especially useful for output port Bit programming where individual Bits or port pins can be directly set high or low using either the "SET [m].i" or "CLR [m]. i" instructions respectively. The feature removes the need for programmers to first read the 8-Bit output port, manipulate the input data to ensure that other Bits are not changed and then output the port with the correct new data. This read-modify-write process is taken care of automatically when these Bit operation instructions are used.

#### Table Read Operations

Data storage is normally implemented by using registers. However, when working with large amounts of fixed data, the volume involved often makes it inconvenient to store the fixed data in the Data Memory. To overcome this problem, Holtek microcontrollers allow an area of Program Memory to be set as a table where data can be directly stored. A set of easy to use instructions provides the means by which this fixed data can be referenced and retrieved from the Program Memory.

#### **Other Operations**

In addition to the above functional instructions, a range of other instructions also exist such as the "HALT" instruction for Power-down operations and instructions to control the operation of the Watchdog Timer for reliable program operations under extreme electric or electromagnetic environments. For their relevant operations, refer to the functional related sections.



# **Instruction Set Summary**

The instructions related to the data memory access in the following table can be used when the desired data memory is located in Data Memory sector 0.

### Table Conventions

x: Bits immediate data

- m: Data Memory address
- A: Accumulator
- i: 0~7 number of bits
- addr: Program memory address

Mnemonic	Description	Cycles	Flag Affected
Arithmetic		1	1
ADD A,[m]	Add Data Memory to ACC	1	Z, C, AC, OV, SC
ADDM A,[m]	Add ACC to Data Memory	1 <sup>Note</sup>	Z, C, AC, OV, SC
ADD A,x	Add immediate data to ACC	1	Z, C, AC, OV, SC
ADC A,[m]	Add Data Memory to ACC with Carry	1	Z, C, AC, OV, SC
ADCM A,[m]	Add ACC to Data memory with Carry	1 <sup>Note</sup>	Z, C, AC, OV, SC
SUB A,x	Subtract immediate data from the ACC	1	Z, C, AC, OV, SC, CZ
SUB A,[m]	Subtract Data Memory from ACC	1	Z, C, AC, OV, SC, CZ
SUBM A,[m]	Subtract Data Memory from ACC with result in Data Memory	1 <sup>Note</sup>	Z, C, AC, OV, SC, CZ
SBC A,x	Subtract immediate data from ACC with Carry	1	Z, C, AC, OV, SC, CZ
SBC A,[m]	Subtract Data Memory from ACC with Carry	1	Z, C, AC, OV, SC, CZ
SBCM A,[m]	Subtract Data Memory from ACC with Carry, result in Data Memory	1 <sup>Note</sup>	Z, C, AC, OV, SC, CZ
DAA [m]	Decimal adjust ACC for Addition with result in Data Memory	1 <sup>Note</sup>	С
Logic Operation	on		·
AND A,[m]	Logical AND Data Memory to ACC	1	Z
OR A,[m]	Logical OR Data Memory to ACC	1	Z
XOR A,[m]	Logical XOR Data Memory to ACC	1	Z
ANDM A,[m]	Logical AND ACC to Data Memory	1 <sup>Note</sup>	Z
ORM A,[m]	Logical OR ACC to Data Memory	1 <sup>Note</sup>	Z
XORM A,[m]	Logical XOR ACC to Data Memory	1 <sup>Note</sup>	Z
AND A,x	Logical AND immediate Data to ACC	1	Z
OR A,x	Logical OR immediate Data to ACC	1	Z
XOR A,x	Logical XOR immediate Data to ACC	1	Z
CPL [m]	Complement Data Memory	1 <sup>Note</sup>	Z
CPLA [m]	Complement Data Memory with result in ACC	1	Z
Increment & D	ecrement		
INCA [m]	Increment Data Memory with result in ACC	1	Z
INC [m]	Increment Data Memory	1 <sup>Note</sup>	Z
DECA [m]	Decrement Data Memory with result in ACC	1	Z
DEC [m]	Decrement Data Memory	1 <sup>Note</sup>	Z
Rotate			
RRA [m]	Rotate Data Memory right with result in ACC	1	None
RR [m]	Rotate Data Memory right	1 <sup>Note</sup>	None
RRCA [m]	Rotate Data Memory right through Carry with result in ACC	1	С
RRC [m]	Rotate Data Memory right through Carry	1 <sup>Note</sup>	С
RLA [m]	Rotate Data Memory left with result in ACC	1	None
RL [m]	Rotate Data Memory left	1 <sup>Note</sup>	None
RLCA [m]	Rotate Data Memory left through Carry with result in ACC	1	С
RLC [m]	Rotate Data Memory left through Carry	1 <sup>Note</sup>	С



Mnemonic	Description	Cycles	Flag Affected
Data Move			
MOV A,[m]	Move Data Memory to ACC	1	None
MOV [m],A	Move ACC to Data Memory	1 <sup>Note</sup>	None
MOV A,x	Move immediate data to ACC	1	None
Bit Operatior	1		
CLR [m].i	Clear bit of Data Memory	1 <sup>Note</sup>	None
SET [m].i	Set bit of Data Memory	1 <sup>Note</sup>	None
Branch Oper	1		
JMP addr	Jump unconditionally	2	None
SZ [m]	Skip if Data Memory is zero	1 <sup>Note</sup>	None
SZA [m]	Skip if Data Memory is zero with data movement to ACC	1 <sup>Note</sup>	None
SZ [m].i	Skip if bit i of Data Memory is zero	1 <sup>Note</sup>	None
SNZ [m]	Skip if Data Memory is not zero	1 <sup>Note</sup>	None
SNZ [m].i	Skip if bit i of Data Memory is not zero	1 <sup>Note</sup>	None
SIZ [m]	Skip if increment Data Memory is zero	1 <sup>Note</sup>	None
SDZ [m]	Skip if decrement Data Memory is zero	1 <sup>Note</sup>	None
SIZA [m]	Skip if increment Data Memory is zero with result in ACC	1 <sup>Note</sup>	None
SDZA [m]	Skip if decrement Data Memory is zero with result in ACC	1 <sup>Note</sup>	None
CALL addr	Subroutine call	2	None
RET	Return from subroutine	2	None
RET A,x	Return from subroutine and load immediate data to ACC	2	None
RETI	Return from interrupt	2	None
Table Read C	Operation		
TABRD [m]	Read table (specific page) to TBLH and Data Memory	2 <sup>Note</sup>	None
TABRDL [m]	Read table (last page) to TBLH and Data Memory	2 <sup>Note</sup>	None
ITABRD [m]	Increment table pointer TBLP first and Read table to TBLH and Data Memory	2 <sup>Note</sup>	None
ITABRDL [m]	Increment table pointer TBLP first and Read table (last page) to TBLH and Data Memory	2 <sup>Note</sup>	None
Miscellaneou	IS		
NOP	No operation	1	None
CLR [m]	Clear Data Memory	1 <sup>Note</sup>	None
SET [m]	Set Data Memory	1 <sup>Note</sup>	None
CLR WDT	Clear Watchdog Timer	1	TO, PDF
SWAP [m]	Swap nibbles of Data Memory	1 <sup>Note</sup>	None
SWAPA [m]	Swap nibbles of Data Memory with result in ACC	1	None
HALT	Enter power down mode	1	TO, PDF

Note: 1. For skip instructions, if the result of the comparison involves a skip then up to three cycles are required, if no skip takes place only one cycle is required.

2. Any instruction which changes the contents of the PCL will also require 2 cycles for execution.

3. For the "CLR WDT" instruction the TO and PDF flags may be affected by the execution status. The TO and PDF flags are cleared after the "CLR WDT" instructions is executed. Otherwise the TO and PDF flags remain unchanged.



## Extended Instruction Set

The extended instructions are used to support the full range address access for the data memory. When the accessed data memory is located in any data memory sections except sector 0, the extended instruction can be used to access the data memory instead of using the indirect addressing access to improve the CPU firmware performance.

Mnemonic	Description	Description Cycles Flag Affect		
Arithmetic				
LADD A,[m]	Add Data Memory to ACC	2	Z, C, AC, OV, SC	
LADDM A,[m]	Add ACC to Data Memory	2 <sup>Note</sup>	Z, C, AC, OV, SC	
LADC A,[m]	Add Data Memory to ACC with Carry	2	Z, C, AC, OV, SC	
LADCM A,[m]	Add ACC to Data memory with Carry	2 <sup>Note</sup>	Z, C, AC, OV, SC	
LSUB A,[m]	Subtract Data Memory from ACC	2	Z, C, AC, OV, SC, CZ	
LSUBM A,[m]	Subtract Data Memory from ACC with result in Data Memory	2 <sup>Note</sup>	Z, C, AC, OV, SC, CZ	
LSBC A,[m]	Subtract Data Memory from ACC with Carry	2	Z, C, AC, OV, SC, CZ	
LSBCM A,[m]	Subtract Data Memory from ACC with Carry, result in Data Memory	2 <sup>Note</sup>	Z, C, AC, OV, SC, CZ	
LDAA [m]	Decimal adjust ACC for Addition with result in Data Memory	2 <sup>Note</sup>	С	
Logic Operation	n n			
LAND A,[m]	Logical AND Data Memory to ACC	2	Z	
LOR A,[m]	Logical OR Data Memory to ACC	2	Z	
LXOR A,[m]	Logical XOR Data Memory to ACC	2	Z	
LANDM A,[m]	Logical AND ACC to Data Memory	2 <sup>Note</sup>	Z	
LORM A,[m]	Logical OR ACC to Data Memory	2 <sup>Note</sup>	Z	
LXORM A,[m]	Logical XOR ACC to Data Memory	2 <sup>Note</sup>	Z	
LCPL [m]	Complement Data Memory	2 <sup>Note</sup>	Z	
LCPLA [m]	Complement Data Memory with result in ACC	2	Z	
Increment & D	ecrement			
LINCA [m]	Increment Data Memory with result in ACC	2	Z	
LINC [m]	Increment Data Memory	2 <sup>Note</sup>	Z	
LDECA [m]	Decrement Data Memory with result in ACC	2	Z	
LDEC [m]	Decrement Data Memory	2 <sup>Note</sup>	Z	
Rotate				
LRRA [m]	Rotate Data Memory right with result in ACC	2	None	
LRR [m]	Rotate Data Memory right	2 <sup>Note</sup>	None	
LRRCA [m]	Rotate Data Memory right through Carry with result in ACC	2	С	
LRRC [m]	Rotate Data Memory right through Carry	2 <sup>Note</sup>	C	
LRLA [m]	Rotate Data Memory left with result in ACC	2	None	
LRL [m]	Rotate Data Memory left	2 <sup>Note</sup>	None	
LRLCA [m]	Rotate Data Memory left through Carry with result in ACC	2	С	
LRLC [m]	Rotate Data Memory left through Carry	2 <sup>Note</sup>	С	
Data Move				
LMOV A,[m]	Move Data Memory to ACC	2	None	
LMOV [m],A	Move ACC to Data Memory	2 <sup>Note</sup> None		
Bit Operation				
LCLR [m].i	Clear bit of Data Memory	2 <sup>Note</sup>	None	
LSET [m].i	Set bit of Data Memory	2 <sup>Note</sup>	None	



Mnemonic	Description	Cycles	Flag Affected
Branch			
LSZ [m]	Skip if Data Memory is zero	2 <sup>Note</sup>	None
LSZA [m]	Skip if Data Memory is zero with data movement to ACC	2 <sup>Note</sup>	None
LSNZ [m]	Skip if Data Memory is not zero	2 <sup>Note</sup>	None
LSZ [m].i	Skip if bit i of Data Memory is zero	2 <sup>Note</sup>	None
LSNZ [m].i	Skip if bit i of Data Memory is not zero	2 <sup>Note</sup>	None
LSIZ [m]	Skip if increment Data Memory is zero	2 <sup>Note</sup>	None
LSDZ [m]	Skip if decrement Data Memory is zero	2 <sup>Note</sup>	None
LSIZA [m]	Skip if increment Data Memory is zero with result in ACC	2 <sup>Note</sup>	None
LSDZA [m]	Skip if decrement Data Memory is zero with result in ACC	2 <sup>Note</sup>	None
Table Read			
LTABRD [m]	Read table to TBLH and Data Memory	3 <sup>Note</sup>	None
LTABRDL [m]	Read table (last page) to TBLH and Data Memory	3 <sup>Note</sup>	None
LITABRD [m]	Increment table pointer TBLP first and Read table to TBLH and Data Memory	3 <sup>Note</sup>	None
LITABRDL [m]	Increment table pointer TBLP first and Read table (last page) to TBLH and Data Memory	3 <sup>Note</sup>	None
Miscellaneou	S		
LCLR [m]	Clear Data Memory	2 <sup>Note</sup>	None
LSET [m]	Set Data Memory	2 <sup>Note</sup>	None
LSWAP [m]	Swap nibbles of Data Memory	2 <sup>Note</sup>	None
LSWAPA [m]	Swap nibbles of Data Memory with result in ACC	2	None

Note: 1. For these extended skip instructions, if the result of the comparison involves a skip then up to four cycles are required, if no skip takes place two cycles is required.

2. Any extended instruction which changes the contents of the PCL register will also require three cycles for execution.



# Instruction Definition

ADC A,[m]	Add Data Memory to ACC with Carry
Description	The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC + [m] + C$
Affected flag(s)	OV, Z, AC, C, SC
ADCM A,[m]	Add ACC to Data Memory with Carry
Description	The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the specified Data Memory.
Operation	$[m] \leftarrow ACC + [m] + C$
Affected flag(s)	OV, Z, AC, C, SC
ADD A,[m]	Add Data Memory to ACC
Description	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC + [m]$
Affected flag(s)	OV, Z, AC, C, SC
ADD A,x	Add immediate data to ACC
Description	The contents of the Accumulator and the specified immediate data are added. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC + x$
Affected flag(s)	OV, Z, AC, C, SC
ADDM A,[m]	Add ACC to Data Memory
<b>ADDM A,[m]</b> Description	Add ACC to Data Memory The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory.
	The contents of the specified Data Memory and the Accumulator are added.
Description	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory.
Description Operation	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory. [m] ← ACC + [m]
Description Operation Affected flag(s)	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory. [m] ← ACC + [m] OV, Z, AC, C, SC
Description Operation Affected flag(s) AND A,[m] Description Operation	<ul> <li>The contents of the specified Data Memory and the Accumulator are added.</li> <li>The result is stored in the specified Data Memory.</li> <li>[m] ← ACC + [m]</li> <li>OV, Z, AC, C, SC</li> <li>Logical AND Data Memory to ACC</li> <li>Data in the Accumulator and the specified Data Memory perform a bitwise logical AND</li> </ul>
Description Operation Affected flag(s) AND A,[m] Description	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory. [m] ← ACC + [m] OV, Z, AC, C, SC Logical AND Data Memory to ACC Data in the Accumulator and the specified Data Memory perform a bitwise logical AND operation. The result is stored in the Accumulator.
Description Operation Affected flag(s) AND A,[m] Description Operation Affected flag(s)	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory. $[m] \leftarrow ACC + [m]$ OV, Z, AC, C, SC Logical AND Data Memory to ACC Data in the Accumulator and the specified Data Memory perform a bitwise logical AND operation. The result is stored in the Accumulator. $ACC \leftarrow ACC "AND" [m]$
Description Operation Affected flag(s) AND A,[m] Description Operation	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory. $[m] \leftarrow ACC + [m]$ OV, Z, AC, C, SC Logical AND Data Memory to ACC Data in the Accumulator and the specified Data Memory perform a bitwise logical AND operation. The result is stored in the Accumulator. $ACC \leftarrow ACC "AND" [m]$ Z
Description Operation Affected flag(s) AND A,[m] Description Operation Affected flag(s) AND A,x	<ul> <li>The contents of the specified Data Memory and the Accumulator are added.</li> <li>The result is stored in the specified Data Memory.</li> <li>[m] ← ACC + [m]</li> <li>OV, Z, AC, C, SC</li> <li>Logical AND Data Memory to ACC</li> <li>Data in the Accumulator and the specified Data Memory perform a bitwise logical AND operation. The result is stored in the Accumulator.</li> <li>ACC ← ACC "AND" [m]</li> <li>Z</li> <li>Logical AND immediate data to ACC</li> <li>Data in the Accumulator and the specified immediate data perform a bit wise logical AND</li> </ul>
Description Operation Affected flag(s) AND A,[m] Description Operation Affected flag(s) AND A,x Description	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory. $[m] \leftarrow ACC + [m]$ OV, Z, AC, C, SC Logical AND Data Memory to ACC Data in the Accumulator and the specified Data Memory perform a bitwise logical AND operation. The result is stored in the Accumulator. ACC $\leftarrow$ ACC "AND" [m] Z Logical AND immediate data to ACC Data in the Accumulator and the specified immediate data perform a bit wise logical AND operation. The result is stored in the Accumulator.
Description Operation Affected flag(s) AND A,[m] Description Operation Affected flag(s) AND A,x Description Operation	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory. $[m] \leftarrow ACC + [m]$ OV, Z, AC, C, SC Logical AND Data Memory to ACC Data in the Accumulator and the specified Data Memory perform a bitwise logical AND operation. The result is stored in the Accumulator. $ACC \leftarrow ACC "AND" [m]$ Z Logical AND immediate data to ACC Data in the Accumulator and the specified immediate data perform a bit wise logical AND operation. The result is stored in the Accumulator. $ACC \leftarrow ACC "AND" [m]$
Description Operation Affected flag(s) AND A,[m] Description Operation Affected flag(s) AND A,x Description Operation Affected flag(s)	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory. $[m] \leftarrow ACC + [m]$ OV, Z, AC, C, SC Logical AND Data Memory to ACC Data in the Accumulator and the specified Data Memory perform a bitwise logical AND operation. The result is stored in the Accumulator. $ACC \leftarrow ACC "AND" [m]$ Z Logical AND immediate data to ACC Data in the Accumulator and the specified immediate data perform a bit wise logical AND operation. The result is stored in the Accumulator. $ACC \leftarrow ACC "AND" [m]$ Z Logical AND immediate data to ACC Data in the Accumulator and the specified immediate data perform a bit wise logical AND operation. The result is stored in the Accumulator. $ACC \leftarrow ACC "AND" x$ Z
Description Operation Affected flag(s) AND A,[m] Description Operation Affected flag(s) AND A,x Description Operation Affected flag(s) AND A,x Description Affected flag(s) AMD A,[m]	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory. $[m] \leftarrow ACC + [m]$ OV, Z, AC, C, SC Logical AND Data Memory to ACC Data in the Accumulator and the specified Data Memory perform a bitwise logical AND operation. The result is stored in the Accumulator. $ACC \leftarrow ACC "AND" [m]$ Z Logical AND immediate data to ACC Data in the Accumulator and the specified immediate data perform a bit wise logical AND operation. The result is stored in the Accumulator. $ACC \leftarrow ACC "AND" [m]$ Z Logical AND immediate data to ACC Data in the Accumulator and the specified immediate data perform a bit wise logical AND operation. The result is stored in the Accumulator. $ACC \leftarrow ACC "AND" x$ Z Logical AND ACC to Data Memory Data in the specified Data Memory and the Accumulator perform a bitwise logical AND



	Subroutine call
Description	Unconditionally calls a subroutine at the specified address. The Program Counter then increments by 1 to obtain the address of the next instruction which is then pushed onto the stack. The specified address is then loaded and the program continues execution from this new address. As this instruction requires an additional operation, it is a two cycle instruction.
Operation	Stack ← Program Counter + 1 Program Counter ← addr
Affected flag(s)	None
CLR [m]	Clear Data Memory
Description	Each bit of the specified Data Memory is cleared to 0.
Operation	$[m] \leftarrow 00H$
Affected flag(s)	None
CLR [m].i	Clear bit of Data Memory
Description	Bit i of the specified Data Memory is cleared to 0.
Operation	$[m]$ .i $\leftarrow 0$
Affected flag(s)	None
CLR WDT	Clear Watchdog Timer
Description	The TO, PDF flags and the WDT are all cleared.
Operation	WDT cleared
	$TO \leftarrow 0$
	$PDF \leftarrow 0$
Affected flag(s)	TO, PDF
CPL [m]	Complement Data Memory
Description	Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa.
Operation	$[m] \leftarrow [m]$
Affected flag(s)	Z
CPLA [m]	Complement Data Memory with result in ACC
Description	Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa. The complemented result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
	Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa. The complemented result is stored in
Description	Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa. The complemented result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Description Operation	Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa. The complemented result is stored in the Accumulator and the contents of the Data Memory remain unchanged. ACC $\leftarrow [m]$
Description Operation Affected flag(s)	Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa. The complemented result is stored in the Accumulator and the contents of the Data Memory remain unchanged. ACC $\leftarrow [m]$ Z
Description Operation Affected flag(s) DAA [m]	Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa. The complemented result is stored in the Accumulator and the contents of the Data Memory remain unchanged. ACC $\leftarrow [m]$ Z Decimal-Adjust ACC for addition with result in Data Memory Convert the contents of the Accumulator value to a BCD (Binary Coded Decimal) value resulting from the previous addition of two BCD variables. If the low nibble is greater than 9 or if AC flag is set, then a value of 6 will be added to the low nibble. Otherwise the low nibble remains unchanged. If the high nibble is greater than 9 or if the C flag is set, then a value of 6 will be added to the high nibble. Essentially, the decimal conversion is performed by adding 00H, 06H, 60H or 66H depending on the Accumulator and flag conditions. Only the C flag may be affected by this instruction which indicates that if the original BCD sum is greater than



<b>DEC [m]</b> Description Operation Affected flag(s)	Decrement Data Memory Data in the specified Data Memory is decremented by 1. $[m] \leftarrow [m] - 1$ Z
<b>DECA [m]</b> Description	Decrement Data Memory with result in ACC Data in the specified Data Memory is decremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation Affected flag(s)	$ACC \leftarrow [m] - 1$ Z
HALT	Enter power down mode
Description	This instruction stops the program execution and turns off the system clock. The contents of the Data Memory and registers are retained. The WDT and prescaler are cleared. The power down flag PDF is set and the WDT time-out flag TO is cleared.
Operation	$TO \leftarrow 0$ $PDF \leftarrow 1$
Affected flag(s)	TO, PDF
INC [m]	Increment Data Memory
Description	Data in the specified Data Memory is incremented by 1.
Operation	$[m] \leftarrow [m] + 1$
Affected flag(s)	Z
INCA [m]	Increment Data Memory with result in ACC
Description	Data in the specified Data Memory is incremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	$ACC \leftarrow [m] + 1$
Affected flag(s)	Z
JMP addr	Jump unconditionally
Description	The contents of the Program Counter are replaced with the specified address. Program execution then continues from this new address. As this requires the insertion of a dummy instruction while the new address is loaded, it is a two cycle instruction.
Operation	Program Counter $\leftarrow$ addr
Affected flag(s)	None
MOV A,[m]	Move Data Memory to ACC
Description	The contents of the specified Data Memory are copied to the Accumulator.
Operation	$ACC \leftarrow [m]$
Affected flag(s)	None
MOV A,x	Move immediate data to ACC
Description	The immediate data specified is loaded into the Accumulator.
Operation	$ACC \leftarrow x$
Affected flag(s)	None
<b>MOV [m],A</b> Description Operation Affected flag(s)	Move ACC to Data Memory The contents of the Accumulator are copied to the specified Data Memory. [m] ← ACC None



NOP	No operation
Description	No operation is performed. Execution continues with the next instruction.
Operation	No operation
Affected flag(s)	None
2()	
OR A,[m]	Logical OR Data Memory to ACC
Description	Data in the Accumulator and the specified Data Memory perform a bitwise logical OR operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC "OR" [m]$
Affected flag(s)	Z
OR A,x	Logical OR immediate data to ACC
Description	Data in the Accumulator and the specified immediate data perform a bitwise logical OR operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC "OR" x$
Affected flag(s)	Z
ORM A,[m]	Logical OR ACC to Data Memory
Description	Data in the specified Data Memory and the Accumulator perform a bitwise logical OR
-	operation. The result is stored in the Data Memory.
Operation	[m] ← ACC "OR" [m]
Affected flag(s)	Z
RET	Return from subroutine
Description	The Program Counter is restored from the stack. Program execution continues at the restored address.
Operation	Program Counter ← Stack
Affected flag(s)	None
RET A,x	Return from subroutine and load immediate data to ACC
Description	The Program Counter is restored from the stack and the Accumulator loaded with the specified immediate data. Program execution continues at the restored address.
Operation	Program Counter $\leftarrow$ Stack ACC $\leftarrow$ x
Affected flag(s)	None
RETI	Return from interrupt
Description	The Program Counter is restored from the stack and the interrupts are re-enabled by setting the
Description	EMI bit. EMI is the master interrupt global enable bit. If an interrupt was pending when the RETI instruction is executed, the pending Interrupt routine will be processed before returning to the main program.
Operation	Program Counter ← Stack EMI ← 1
Affected flag(s)	None
DI faci	
RL [m]	Rotate Data Memory left The contents of the encoded Data Memory are rotated left by 1 bit with hit 7 rotated into bit 0
Description Operation	The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0. $[m].(i+1) \leftarrow [m].i; (i=0-6)$
Operation	$[m].(i+1) \leftarrow [m].i, (i-0 \sim 0)$ $[m].0 \leftarrow [m].7$
Affected flag(s)	None



RLA [m]	Rotate Data Memory left with result in ACC
Description	The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.(i+1) \leftarrow [m].i; (i=0\sim6)$ $ACC.0 \leftarrow [m].7$
Affected flag(s)	None
RLC [m]	Rotate Data Memory left through Carry
Description	The contents of the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into bit 0.
Operation	$[m].(i+1) \leftarrow [m].i; (i=0-6)$ $[m].0 \leftarrow C$ $C \leftarrow [m].7$
Affected flag(s)	C
RLCA [m]	Rotate Data Memory left through Carry with result in ACC
Description	Data in the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into the bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	ACC.(i+1) $\leftarrow$ [m].i; (i=0~6) ACC.0 $\leftarrow$ C C $\leftarrow$ [m].7
Affected flag(s)	C
RR [m]	Rotate Data Memory right
Description	The contents of the specified Data Memory are rotated right by 1 bit with bit 0 rotated into bit 7.
Operation	[m].i ← [m].(i+1); (i=0~6) [m].7 ← [m].0
Affected flag(s)	None
RRA [m]	Rotate Data Memory right with result in ACC
Description	Data in the specified Data Memory is rotated right by 1 bit with bit 0 rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.i \leftarrow [m].(i+1); (i=0\sim6)$ $ACC.7 \leftarrow [m].0$
Operation Affected flag(s)	
-	ACC.7 $\leftarrow$ [m].0
Affected flag(s)	ACC.7 $\leftarrow$ [m].0 None
Affected flag(s)	ACC.7 ← [m].0 None Rotate Data Memory right through Carry The contents of the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0



	Batata Data Mamany night through Computith nagult in ACC
RRCA [m]	Rotate Data Memory right through Carry with result in ACC
Description	Data in the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7. The rotated result is stored in the
	Accumulator and the contents of the Data Memory remain unchanged.
Operation	ACC.i $\leftarrow$ [m].(i+1); (i=0~6)
*	$ACC.7 \leftarrow C$
	$C \leftarrow [m].0$
Affected flag(s)	C
000 4 (1	
SBC A,[m]	Subtract Data Memory from ACC with Carry
Description	The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - [m] - \overline{C}$
Affected flag(s)	OV, Z, AC, C, SC, CZ
	- ·,_,,-,-,-,-
SBC A, x	Subtract immediate data from ACC with Carry
Description	The immediate data and the complement of the carry flag are subtracted from the
	Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is
	negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - [m] - \overline{C}$
Affected flag(s)	OV, Z, AC, C, SC, CZ
Ameeted hug(s)	01, 2, 110, 0, 50, 02
SBCM A,[m]	Subtract Data Memory from ACC with Carry and result in Data Memory
Description	The contents of the specified Data Memory and the complement of the carry flag are
	subtracted from the Accumulator. The result is stored in the Data Memory. Note that if the
	result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$[m] \leftarrow ACC - [m] - \overline{C}$
Affected flag(s)	OV, Z, AC, C, SC, CZ
Affected flag(s)	01, 2, AC, C, 5C, C2
SDZ [m]	Skip if decrement Data Memory is 0
Description	The contents of the specified Data Memory are first decremented by 1. If the result is 0 the
*	following instruction is skipped. As this requires the insertion of a dummy instruction while
	the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$[m] \leftarrow [m] - 1$
Operation	Skip if $[m]=0$
Affected flag(s)	None
<b>C</b> ()	
SDZA [m]	Skip if decrement Data Memory is zero with result in ACC
Description	The contents of the specified Data Memory are first decremented by 1. If the result is 0, the
	following instruction is skipped. The result is stored in the Accumulator but the specified
	Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0,
	the program proceeds with the following instruction.
Operation	$ACC \leftarrow [m] - 1$
	Skip if ACC=0
Affected flag(s)	None

# BA45F5740/BA45F5740-2 Smoke Detector Flash MCU with Calendar



<b>SET [m]</b> Description Operation	Set Data Memory Each bit of the specified Data Memory is set to 1. [m] ← FFH
Affected flag(s)	None
SET [m].i	Set bit of Data Memory
Description	Bit i of the specified Data Memory is set to 1.
Operation	[m].i ← 1
Affected flag(s)	None
SIZ [m]	Skip if increment Data Memory is 0
Description	The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	[m] ← [m] + 1 Skip if [m]=0
Affected flag(s)	None
SIZA [m]	Skip if increment Data Memory is zero with result in ACC
Description	The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	ACC ← [m] + 1 Skip if ACC=0
Affected flag(s)	None
SNZ [m].i	Skip if Data Memory is not 0
Description	If the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is 0 the program proceeds with the following instruction.
Operation	Skip if $[m]$ .i $\neq 0$
Affected flag(s)	None
SNZ [m]	Skip if Data Memory is not 0
Description	If the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is 0 the program proceeds with the following instruction.
Operation	Skip if [m]≠ 0
Affected flag(s)	None
SUB A,[m]	Subtract Data Memory from ACC
Description	The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - [m]$
Affected flag(s)	OV, Z, AC, C, SC, CZ



SUBM A,[m]	Subtract Data Memory from ACC with result in Data Memory
Description	The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$[m] \leftarrow ACC - [m]$
Affected flag(s)	OV, Z, AC, C, SC, CZ
SUB A,x	Subtract immediate data from ACC
Description	The immediate data specified by the code is subtracted from the contents of the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - x$
Affected flag(s)	OV, Z, AC, C, SC, CZ
SWAP [m]	Swap nibbles of Data Memory
Description	The low-order and high-order nibbles of the specified Data Memory are interchanged.
Operation	$[m].3 \sim [m].0 \leftrightarrow [m].7 \sim [m].4$
Affected flag(s)	None
SWAPA [m]	Swap nibbles of Data Memory with result in ACC
Description	The low-order and high-order nibbles of the specified Data Memory are interchanged. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	ACC.3~ACC.0 $\leftarrow$ [m].7~[m].4 ACC.7~ACC.4 $\leftarrow$ [m].3~[m].0
Affected flag(s)	None
SZ [m]	Skip if Data Memory is 0
Description	If the contents of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	Skip if [m]=0
Affected flag(s)	None
SZA [m]	Skip if Data Memory is 0 with data movement to ACC
Description	The contents of the specified Data Memory are copied to the Accumulator. If the value is zero, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$ACC \leftarrow [m]$ Skip if $[m]=0$
Affected flag(s)	None
SZ [m].i	Skip if bit i of Data Memory is 0
Description	If bit i of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0, the program proceeds with the following instruction.
Operation	Skip if [m].i=0
Affected flag(s)	None



TABRD [m]	Read table (specific page) to TBLH and Data Memory
Description	The low byte of the program code (specific page) addressed by the table pointer pair
Operation	<ul> <li>(TBLP and TBHP) is moved to the specified Data Memory and the high byte moved to TBLH.</li> <li>[m] ← program code (low byte)</li> </ul>
operation	TBLH $\leftarrow$ program code (high byte)
Affected flag(s)	None
TABRDL [m]	Read table (last page) to TBLH and Data Memory
Description	The low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
ITABRD [m]	Increment table pointer low byte first and read table to TBLH and Data Memory
Description	Increment table pointer low byte, TBLP, first and then the program code addressed by the table pointer (TBHP and TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte)
	$TBLH \leftarrow program \ code \ (high \ byte)$
Affected flag(s)	None
ITABRDL [m]	Increment table pointer low byte first and read table (last page) to TBLH and Data Memory
Description	Increment table pointer low byte, TBLP, first and then the low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte)
	TBLH ← program code (high byte)
Affected flag(s)	None
XOR A,[m]	Logical XOR Data Memory to ACC
Description	Data in the Accumulator and the specified Data Memory perform a bitwise logical XOR operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC "XOR" [m]$
Affected flag(s)	Z
XORM A,[m]	Logical XOR ACC to Data Memory
Description	Data in the specified Data Memory and the Accumulator perform a bitwise logical XOR operation. The result is stored in the Data Memory.
Operation	$[m] \leftarrow ACC "XOR" [m]$
Affected flag(s)	Z
XOR A,x	Logical XOR immediate data to ACC
Description	Data in the Accumulator and the specified immediate data perform a bitwise logical XOR operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC "XOR" x$
Affected flag(s)	Z



## **Extended Instruction Definition**

The extended instructions are used to directly access the data stored in any data memory sections.

LADC A,[m]	Add Data Memory to ACC with Carry
Description	The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC + [m] + C$
Affected flag(s)	OV, Z, AC, C, SC
LADCM A,[m]	Add ACC to Data Memory with Carry
Description	The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the specified Data Memory.
Operation	$[m] \leftarrow ACC + [m] + C$
Affected flag(s)	OV, Z, AC, C, SC
LADD A,[m]	Add Data Memory to ACC
Description	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC + [m]$
Affected flag(s)	OV, Z, AC, C, SC
LADDM A,[m]	Add ACC to Data Memory
Description	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory.
Operation	$[m] \leftarrow ACC + [m]$
Affected flag(s)	OV, Z, AC, C, SC
LAND A,[m]	Logical AND Data Memory to ACC
Description	Data in the Accumulator and the specified Data Memory perform a bitwise logical AND operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC "AND" [m]$
Affected flag(s)	Z
LANDM A,[m]	Logical AND ACC to Data Memory
Description	Data in the specified Data Memory and the Accumulator perform a bitwise logical AND operation. The result is stored in the Data Memory.
Operation	$[m] \leftarrow ACC "AND" [m]$
Affected flag(s)	Z
LCLR [m]	Clear Data Memory
Description	Each bit of the specified Data Memory is cleared to 0.
Operation	[m] ← 00H
Affected flag(s)	None
LCLR [m].i	Clear bit of Data Memory
Description	Bit i of the specified Data Memory is cleared to 0.
Operation	$[m]$ .i $\leftarrow 0$
Affected flag(s)	None



LCPL [m] Description Operation	Complement Data Memory Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa. $[m] \leftarrow \overline{[m]}$
Affected flag(s)	Z
LCPLA [m] Description	Complement Data Memory with result in ACC Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa. The complemented result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation Affected flag(s)	$ACC \leftarrow [m]$ Z
LDAA [m]	Decimal-Adjust ACC for addition with result in Data Memory
Description	Convert the contents of the Accumulator value to a BCD (Binary Coded Decimal) value resulting from the previous addition of two BCD variables. If the low nibble is greater than 9 or if AC flag is set, then a value of 6 will be added to the low nibble. Otherwise the low nibble remains unchanged. If the high nibble is greater than 9 or if the C flag is set, then a value of 6 will be added to the high nibble. Essentially, the decimal conversion is performed by adding 00H, 06H, 60H or 66H depending on the Accumulator and flag conditions. Only the C flag may be affected by this instruction which indicates that if the original BCD sum is greater than 100, it allows multiple precision decimal addition.
Operation	$[m] \leftarrow ACC + 00H \text{ or}$ $[m] \leftarrow ACC + 06H \text{ or}$ $[m] \leftarrow ACC + 60H \text{ or}$ $[m] \leftarrow ACC + 66H$
Affected flag(s)	C
LDEC [m]	Decrement Data Memory
Description	Data in the specified Data Memory is decremented by 1.
Operation	[m] ← [m] - 1
Affected flag(s)	Z
LDECA [m]	Decrement Data Memory with result in ACC
Description	Data in the specified Data Memory is decremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	$ACC \leftarrow [m] - 1$
Affected flag(s)	Z
LINC [m] Description Operation Affected flag(s)	Increment Data Memory Data in the specified Data Memory is incremented by 1. [m] ← [m] + 1 Z
	In successful Data Manager with growthin ACC
LINCA [m] Description	Increment Data Memory with result in ACC Data in the specified Data Memory is incremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	$ACC \leftarrow [m] + 1$
Affected flag(s)	Z



<b>LMOV A,[m]</b> Description Operation Affected flag(s)	Move Data Memory to ACC The contents of the specified Data Memory are copied to the Accumulator. $ACC \leftarrow [m]$ None
<b>LMOV [m],A</b> Description Operation Affected flag(s)	Move ACC to Data Memory The contents of the Accumulator are copied to the specified Data Memory. [m] ← ACC None
LOR A,[m] Description Operation Affected flag(s)	Logical OR Data Memory to ACC Data in the Accumulator and the specified Data Memory perform a bitwise logical OR operation. The result is stored in the Accumulator. ACC ← ACC "OR" [m] Z
LORM A,[m] Description Operation Affected flag(s)	Logical OR ACC to Data Memory Data in the specified Data Memory and the Accumulator perform a bitwise logical OR operation. The result is stored in the Data Memory. [m] ← ACC "OR" [m] Z
LRL [m] Description Operation Affected flag(s)	Rotate Data Memory left The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0. $[m].(i+1) \leftarrow [m].i; (i=0\sim6)$ $[m].0 \leftarrow [m].7$ None
<b>LRLA [m]</b> Description Operation Affected flag(s)	Rotate Data Memory left with result in ACC The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged. ACC.(i+1) $\leftarrow$ [m].i; (i=0~6) ACC.0 $\leftarrow$ [m].7 None
LRLC [m] Description Operation Affected flag(s)	Rotate Data Memory left through Carry The contents of the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into bit 0. $[m].(i+1) \leftarrow [m].i; (i=0\sim6)$ $[m].0 \leftarrow C$ $C \leftarrow [m].7$ C
<b>LRLCA [m]</b> Description Operation	Rotate Data Memory left through Carry with result in ACC Data in the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into the bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged. ACC.(i+1) $\leftarrow$ [m].i; (i=0~6) ACC.0 $\leftarrow$ C C $\leftarrow$ [m].7
Affected flag(s)	C



LRR [m] Description	Rotate Data Memory right The contents of the specified Data Memory are rotated right by 1 bit with bit 0 rotated into bit 7.
Operation	[m].i $\leftarrow$ [m].(i+1); (i=0~6) [m].7 $\leftarrow$ [m].0
Affected flag(s)	None
LRRA [m]	Rotate Data Memory right with result in ACC
Description	Data in the specified Data Memory is rotated right by 1 bit with bit 0 rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.i \leftarrow [m].(i+1); (i=0\sim6)$ $ACC.7 \leftarrow [m].0$
Affected flag(s)	None
LRRC [m]	Rotate Data Memory right through Carry
Description	The contents of the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7.
Operation	$\begin{array}{l} [m].i \leftarrow [m].(i+1); \ (i=0\sim6) \\ [m].7 \leftarrow C \\ C \leftarrow [m].0 \end{array}$
Affected flag(s)	C
LRRCA [m]	Rotate Data Memory right through Carry with result in ACC
Description	Data in the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	ACC.i $\leftarrow$ [m].(i+1); (i=0~6) ACC.7 $\leftarrow$ C C $\leftarrow$ [m].0
Affected flag(s)	C
LSBC A,[m]	Subtract Data Memory from ACC with Carry
Description	The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - [m] - \overline{C}$
Affected flag(s)	OV, Z, AC, C, SC, CZ
LSBCM A,[m] Description	Subtract Data Memory from ACC with Carry and result in Data Memory The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation Affected flag(s)	$[m] \leftarrow ACC - [m] - \overline{C}$ OV, Z, AC, C, SC, CZ



LSDZ [m]	Skip if decrement Data Memory is 0
Description	The contents of the specified Data Memory are first decremented by 1. If the result is 0 the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program
Operation	proceeds with the following instruction. $[m] \leftarrow [m] - 1$ String if $[m] = 0$
Affected flag(s)	Skip if [m]=0 None
LSDZA [m]	Skip if decrement Data Memory is zero with result in ACC
Description	The contents of the specified Data Memory are first decremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0, the program proceeds with the following instruction. ACC $\leftarrow$ [m] – 1
Operation	Skip if ACC=0
Affected flag(s)	None
LSET [m]	Set Data Memory
Description	Each bit of the specified Data Memory is set to 1.
Operation	$[m] \leftarrow FFH$
Affected flag(s)	None
LSET [m].i	Set bit of Data Memory
Description	Bit i of the specified Data Memory is set to 1.
Operation	[m].i ← 1
Affected flag(s)	None
LSIZ [m]	Skip if increment Data Memory is 0
Description	
Description	The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction. $[m] \leftarrow [m] + 1$
-	following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation Affected flag(s)	following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction. $[m] \leftarrow [m] + 1$ Skip if $[m]=0$
Operation	<ul> <li>following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.</li> <li>[m] ← [m] + 1</li> <li>Skip if [m]=0</li> <li>None</li> <li>Skip if increment Data Memory is zero with result in ACC</li> <li>The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction. If the result is not</li> </ul>
Operation Affected flag(s) LSIZA [m]	<ul> <li>following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.</li> <li>[m] ← [m] + 1 Skip if [m]=0 None</li> <li>Skip if increment Data Memory is zero with result in ACC</li> <li>The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy</li> </ul>
Operation Affected flag(s) <b>LSIZA [m]</b> Description	following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction. $[m] \leftarrow [m] + 1$ Skip if $[m]=0$ None Skip if increment Data Memory is zero with result in ACC The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction. ACC $\leftarrow [m] + 1$
Operation Affected flag(s) <b>LSIZA [m]</b> Description Operation Affected flag(s)	following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction. $[m] \leftarrow [m] + 1$ Skip if $[m]=0$ None Skip if increment Data Memory is zero with result in ACC The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction. ACC $\leftarrow [m] + 1$ Skip if ACC=0 None
Operation Affected flag(s) LSIZA [m] Description	following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction. $[m] \leftarrow [m] + 1$ Skip if $[m]=0$ None Skip if increment Data Memory is zero with result in ACC The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction. ACC $\leftarrow [m] + 1$ Skip if ACC=0
Operation Affected flag(s) LSIZA [m] Description Operation Affected flag(s) LSNZ [m].i	following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction. $[m] \leftarrow [m] + 1$ Skip if $[m]=0$ None Skip if increment Data Memory is zero with result in ACC The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction. ACC $\leftarrow [m] + 1$ Skip if ACC=0 None Skip if Data Memory is not 0 If the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle

# BA45F5740/BA45F5740-2 Smoke Detector Flash MCU with Calendar



LSNZ [m]	Skip if Data Memory is not 0
Description	If the content of the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is 0 the program proceeds with the following instruction.
Operation	Skip if $[m] \neq 0$
Affected flag(s)	None
LSUB A,[m]	Subtract Data Memory from ACC
Description	The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - [m]$
Affected flag(s)	OV, Z, AC, C, SC, CZ
LSUBM A,[m]	Subtract Data Memory from ACC with result in Data Memory
Description	The specified Data Memory is subtracted from the contents of the Accumulator. The result is
Description	stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$[m] \leftarrow ACC - [m]$
Affected flag(s)	OV, Z, AC, C, SC, CZ
LSWAP [m]	Swap nibbles of Data Memory
Description	The low-order and high-order nibbles of the specified Data Memory are interchanged.
Operation	$[m].3\sim[m].0 \leftrightarrow [m].7\sim[m].4$
Affected flag(s)	None
LSWAPA [m]	Swap nibbles of Data Memory with result in ACC
Description	The low-order and high-order nibbles of the specified Data Memory are interchanged. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	$ACC.3 \sim ACC.0 \leftarrow [m].7 \sim [m].4$ $ACC.7 \sim ACC.4 \leftarrow [m].3 \sim [m].0$
Affected flag(s)	None
LSZ [m]	Skip if Data Memory is 0
Description	If the contents of the specified Data Memory is 0, the following instruction is skipped. As this
I	requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	Skip if [m]=0
Affected flag(s)	None
LSZA [m]	Skip if Data Memory is 0 with data movement to ACC
Description	The contents of the specified Data Memory are copied to the Accumulator. If the value is zero,
Description	the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$ACC \leftarrow [m]$ Skip if [m]=0
Affected flag(s)	None



LSZ [m].i	Skip if bit i of Data Memory is 0
Description	If bit i of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0, the program proceeds with the following instruction.
Operation	Skip if [m].i=0
Affected flag(s)	None
LTABRD [m]	Read table (current page) to TBLH and Data Memory
Description	The low byte of the program code (current page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
LTABRDL [m]	Read table (last page) to TBLH and Data Memory
Description	The low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
LITABRD [m]	Increment table pointer low byte first and read table to TBLH and Data Memory
Description	Increment table pointer low byte, TBLP, first and then the program code addressed by the table pointer (TBHP and TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
LITABRDL [m]	Increment table pointer low byte first and read table (last page) to TBLH and Data Memory
Description	Increment table pointer low byte, TBLP, first and then the low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
LXOR A,[m]	Logical XOR Data Memory to ACC
Description	Data in the Accumulator and the specified Data Memory perform a bitwise logical XOR operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC "XOR" [m]$
Affected flag(s)	Z
LXORM A,[m]	Logical XOR ACC to Data Memory
Description	Data in the specified Data Memory and the Accumulator perform a bitwise logical XOR operation. The result is stored in the Data Memory.
Operation	$[m] \leftarrow ACC "XOR" [m]$
Affected flag(s)	Z



# Package Information

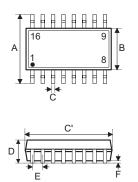
Note that the package information provided here is for consultation purposes only. As this information may be updated at regular intervals users are reminded to consult the <u>Holtek website</u> for the latest version of the <u>Package/Carton Information</u>.

Additional supplementary information with regard to packaging is listed below. Click on the relevant section to be transferred to the relevant website page.

- Package Information (include Outline Dimensions, Product Tape and Reel Specifications)
- The Operation Instruction of Packing Materials
- Carton information



# 16-pin NSOP (150mil) Outline Dimensions



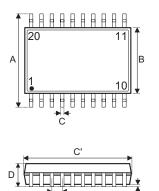


Symbol	Dimensions in inch		
	Min.	Nom.	Max.
A	—	0.236 BSC	—
В	—	0.154 BSC	—
С	0.012	—	0.020
C'	—	0.390 BSC	—
D	—	—	0.069
E	—	0.050 BSC	—
F	0.004	—	0.010
G	0.016	—	0.050
Н	0.004	_	0.010
α	0°	_	8°

Symbol	Dimensions in mm		
	Min.	Nom.	Max.
A	—	6.00 BSC	—
В	—	3.90 BSC	_
С	0.31	—	0.51
C'	—	9.90 BSC	_
D	—	—	1.75
E	_	1.27 BSC	_
F	0.10	_	0.25
G	0.40	—	1.27
Н	0.10	—	0.25
α	0°	—	8°



# 20-pin SOP (300mil) Outline Dimensions

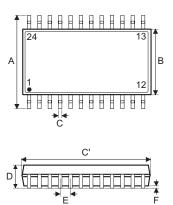




Symbol	Dimensions in inch		
	Min.	Nom.	Max.
А	_	0.406 BSC	_
В	_	0.295 BSC	_
С	0.012	_	0.020
C'	_	0.504 BSC	_
D	_	_	0.104
E	—	0.050 BSC	_
F	0.004	_	0.012
G	0.016	_	0.050
Н	0.008	_	0.013
α	0°	_	8°

Symbol	Dimensions in mm		
	Min.	Nom.	Max.
A		10.30 BSC	—
В	_	7.50 BSC	—
С	0.31	—	0.51
C'	_	12.80 BSC	—
D	—	_	2.65
E	—	1.27 BSC	—
F	0.10	—	0.30
G	0.40	—	1.27
Н	0.20	_	0.33
α	0°		8°

# 24-pin SOP (300mil) Outline Dimensions





Symbol	Dimensions in inch		
	Min.	Nom.	Max.
А	_	0.406 BSC	—
В	_	0.295 BSC	_
С	0.012	_	0.020
C'	_	0.606 BSC	
D	_	—	0.104
E	_	0.050 BSC	
F	0.004	_	0.012
G	0.016	—	0.050
Н	0.008	—	0.013
α	0°	_	8°

Symbol	Dimensions in mm		
	Min.	Nom.	Max.
A	—	10.30 BSC	—
В	—	7.50 BSC	_
С	0.31	—	0.51
C'	—	15.40 BSC	—
D	—	_	2.65
E	—	1.27 BSC	—
F	0.10	_	0.30
G	0.40	—	1.27
Н	0.20	_	0.33
α	0°	_	8°



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