



High Voltage Touch A/D Flash MCU with HVIO

BS86DH12C

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Features

CPU Features

- Operating voltage
 - ♦ $f_{SYS}=8/12/16\text{MHz}$, using internal LDO: $V_{DD}=5\text{V}$ (Typ.)
 - ♦ High Voltage Driver: $V_{CC}=7\text{V}\sim 10\text{V}$
- Up to $0.25\mu\text{s}$ instruction cycle with 16MHz system clock at $V_{DD}=5\text{V}$
- Power down and wake-up functions to reduce power consumption
- Oscillator types
 - ♦ Internal High Speed 8/12/16MHz RC – HIRC
 - ♦ Internal Low Speed 32kHz RC – LIRC
 - ♦ External Low Speed 32.768kHz Crystal – LXT
- Multi-mode operation: FAST, SLOW, IDLE and SLEEP
- Fully integrated internal oscillators require no external components
- All instructions executed in 1~3 instruction cycles
- Table read instructions
- 115 powerful instructions
- 8-level subroutine nesting
- Bit manipulation instruction

Peripheral Features

- Flash Program Memory: $8\text{K}\times 16$
- Data Memory: 512×8
- True EEPROM Memory: 64×8
- 12 touch key functions – fully integrated without requiring external components
- Watchdog Timer function
- 22 bidirectional I/O lines
- 6 bidirectional High Voltage I/O lines with short circuit protection function
- Programmable I/O port source current and sink current for LED driving applications
- Single external interrupt line shared with I/O pin
- Multiple Timer Modules for time measurement, input capture, compare match output or PWM output or single pulse output function
- Single Time-Base function for generation of fixed time interrupt signals
- 8 external channels 12-bit resolution A/D converter with internal reference voltage V_{BG}
- Over Current Protection function – OCP
- Over Voltage Protection function – OVP
- Internal 5V LDO with driving current of up to 500mA, providing power supply for MCU and external components
- I²C interface
- Fully-duplex Universal Asynchronous Receiver and Transmitter Interface – UART
- Low voltage reset function

- Low voltage detect function
- Package types: 20/28-pin SOP, 44-pin LQFP

General Description

The device is a Flash Memory type 8-bit high performance RISC architecture microcontroller with fully integrated touch key functions. With all touch key functions provided internally and with the convenience of Flash Memory multi-programming features, the device has all the features to offer designers a reliable and easy means of implementing touch keys within their products applications.

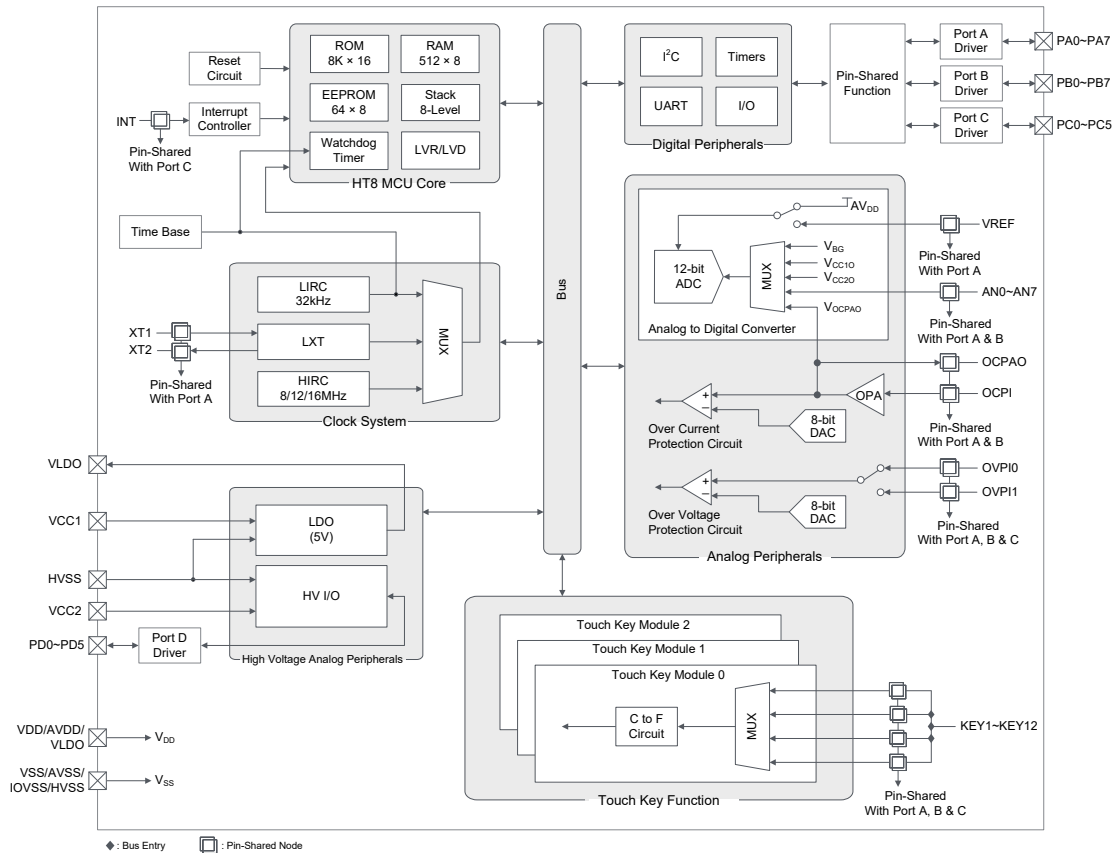
The touch key functions are fully integrated, completely eliminating the need for external components. In addition to the Flash program memory, other memory includes an area of RAM Data Memory as well as an area of true EEPROM memory for storage of non-volatile data such as serial numbers, calibration data etc. Analog feature includes a multi-channel 12-bit A/D converter and an internal LDO for power supply. Protective features such as an internal Watchdog Timer, Low Voltage Reset, Low Voltage Detector, Over Current Protection and Over Voltage Protection coupled with excellent noise immunity and ESD protection ensure that reliable operation is maintained in hostile electrical environments.

A full choice of external low, internal high and low speed oscillators are provided including fully integrated system oscillators which require no external components for its implementation. The ability to operate and switch dynamically between a range of operating modes using different clock sources gives users the ability to optimise microcontroller operation and minimise power consumption. Easy communication with the outside world is provided using the internal I²C and UART interfaces, while the inclusion of flexible I/O programming features, Time-Base function, Timer Modules and many other features further enhance device functionality and flexibility.

This device contains programmable I/O port source current and sink current functions which are used to implement LED driving function. The High Voltage I/O function specific to high voltage and high current applications is also fully integrated within the device.

The touch key device will find excellent use in a huge range of modern Touch Key product applications such as sensor signal processing, household appliances, health care products, industrial control, consumer products, subsystem control to name but a few.

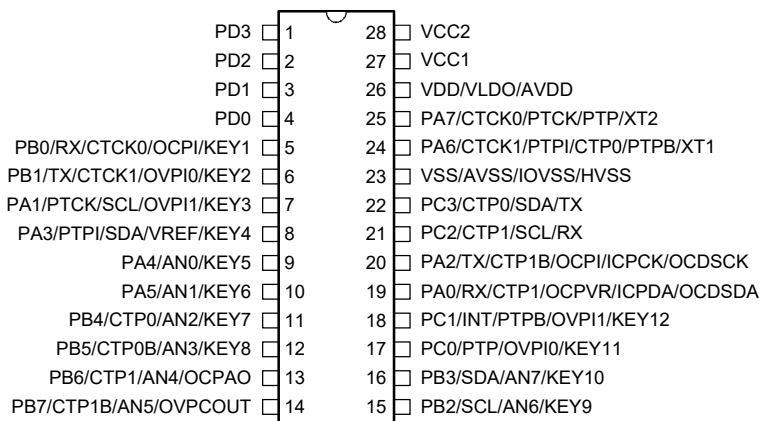
Block Diagram



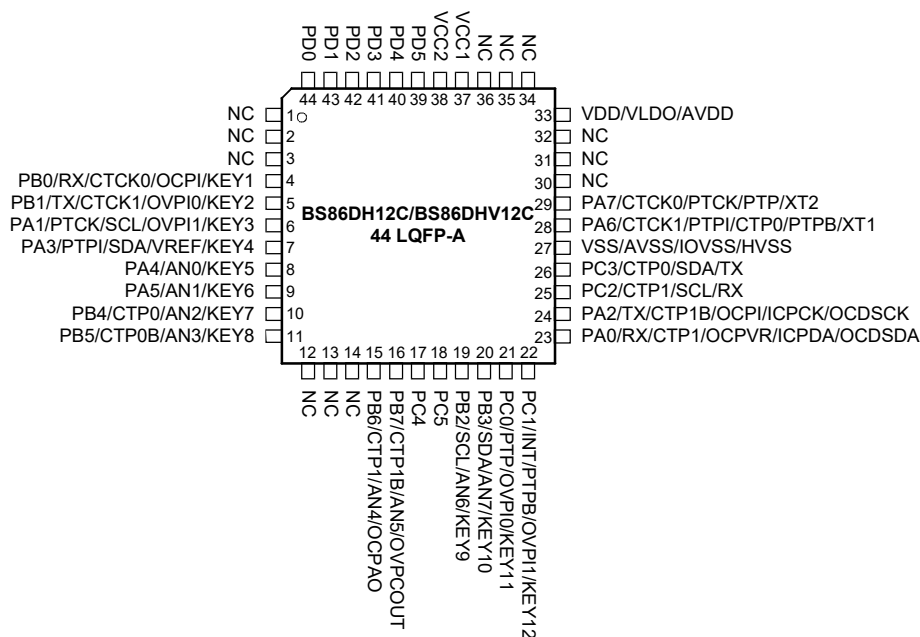
Pin Assignment

PD3	1	20	VCC2
PD2	2	19	VCC1
PD1	3	18	VDD/VLDO/AVDD
PD0	4	17	PA7/CTCK0/PTCK/PTP/XT2
PB0/RX/CTCK0/OCPI/KEY1	5	16	PA6/CTCK1/PTPI/CTP0/PTPB/XT1
PB1/TX/CTCK1/OVPI0/KEY2	6	15	VSS/AVSS/IOVSS/HVSS
PA1/PTCK/SCL/OVPI1/KEY3	7	14	PA2/TX/CTP1B/OCPI/ICPCK/OCDSCK
PA3/PTPI/SDA/VREF/KEY4	8	13	PA0/RX/CTP1/OCVPR/ICPDA/OCSDA
PA4/AN0/KEY5	9	12	PB3/SDA/AN7/KEY10
PA5/AN1/KEY6	10	11	PB2/SCL/AN6/KEY9

BS86DH12C/BS86DHV12C
20 SOP-A



BS86DH12C/BS86DHV12C
28 SOP-A



BS86DH12C/BS86DHV12C
44 LQFP-A

- Note: 1. If the pin-shared pin functions have multiple outputs, the desired pin-shared function is determined by the corresponding software control bits.
2. The OCDSCK and OCDSCK pins are supplied for the OCDS dedicated pins and as such only available for the BS86DHV12C device which is the OCDS EV chip for the BS86DH12C device.
3. For less pin-count package types there will be unbonded pins which should be properly configured to avoid unwanted current consumption resulting from floating input conditions. Refer to the “Standby Current Considerations” and “Input/Output Ports” sections.

Pin Descriptions

With the exception of the power pins, all pins on the device can be referenced by their Port name, e.g. PA0, PA1 etc., which refer to the digital I/O function of the pins. However these Port pins are also shared with other function such as the Touch Key function, Timer Module pins etc. The function of each pin is listed in the following table, however the details behind how each pin is configured is contained in other sections of the datasheet.

As the Pin Description table shows the situation for the package with the most pins, not all pins in the table will be available on smaller package sizes.

Pin Name	Function	OPT	I/T	O/T	Description
PA0/RX/CTP1/ OCPVR/ICPDA/ OCDSDA	PA0	PAPU PAWU PAS0	ST	CMOS	General purpose I/O. Register enabled pull-up and wake-up
	RX	PAS0 IFS0	ST	—	UART data receive pin
	CTP1	PAS0	—	CMOS	CTM1 output
	OCPVR	PAS0	AN	—	OCV D/A converter reference voltage input
	ICPDA	—	ST	CMOS	ICP data/address
	OCDSDA	—	ST	CMOS	OCDS data/address, for EV chip only
PA1/PTCK/SCL/ OVPI1/KEY3	PA1	PAPU PAWU PAS0	ST	CMOS	General purpose I/O. Register enabled pull-up and wake-up
	PTCK	PAS0 IFS1	ST	—	PTM clock input
	SCL	PAS0 IFS0	ST	NMOS	I ² C clock line
	OVPI1	PAS0	AN	—	OVP input 1
	KEY3	PAS0	NSI	—	Touch key input
PA2/TX/CTP1B/ OCPI/ICPCK/ OCDSCK	PA2	PAPU PAWU PAS0	ST	CMOS	General purpose I/O. Register enabled pull-up and wake-up
	TX	PAS0	—	CMOS	UART data transmit pin
	CTP1B	PAS0	—	CMOS	CTM1 inverted output
	OCPI	PAS0	AN	—	OCV input
	ICPCK	—	ST	—	ICP clock
	OCDSCK	—	ST	—	OCDS clock, for EV chip only
PA3/PTPI/SDA/ VREF/KEY4	PA3	PAPU PAWU PAS0	ST	CMOS	General purpose I/O. Register enabled pull-up and wake-up
	PTPI	PAS0 IFS1	ST	—	PTM capture input
	SDA	PAS0 IFS0	ST	NMOS	I ² C data line
	VREF	PAS0	AN	—	A/D converter external input channel
	KEY4	PAS0	NSI	—	Touch key input
PA4/AN0/KEY5	PA4	PAPU PAWU PAS1	ST	CMOS	General purpose I/O. Register enabled pull-up and wake-up
	AN0	PAS1	AN	—	A/D converter external input channel
	KEY5	PAS1	NSI	—	Touch key input

Pin Name	Function	OPT	I/T	O/T	Description
PA5/AN1/KEY6	PA5	PAPU PAWU PAS1	ST	CMOS	General purpose I/O. Register enabled pull-up and wake-up
	AN1	PAS1	AN	—	A/D converter external input channel
	KEY6	PAS1	NSI	—	Touch key input
PA6/CTCK1/PTPI/ CTP0/PTPB/XT1	PA6	PAPU PAWU PAS1	ST	CMOS	General purpose I/O. Register enabled pull-up and wake-up
	CTCK1	PAS1 IFS1	ST	—	CTM1 clock input
	PTPI	PAS1 IFS1	ST	—	PTM capture input
	CTP0	PAS1	—	CMOS	CTM0 output
	PTPB	PAS1	—	CMOS	PTM inverted output
	XT1	PAS1	LXT	—	LXT oscillator pin
PA7/CTCK0/PTCK/ PTP/XT2	PA7	PAPU PAWU PAS1	ST	CMOS	General purpose I/O. Register enabled pull-up and wake-up
	CTCK0	PAS1 IFS1	ST	—	CTM0 clock input
	PTCK	PAS1 IFS1	ST	—	PTM clock input
	PTP	PAS1	—	CMOS	PTM output
	XT2	PAS1	—	LXT	LXT oscillator pin
PB0/RX/CTCK0/ OCPI/KEY1	PB0	PBPU PBS0	ST	CMOS	General purpose I/O. Register enabled pull-up
	RX	PBS0 IFS0	ST	—	UART data receive pin
	CTCK0	PBS0 IFS1	ST	—	CTM0 clock input
	OCPI	PBS0	AN	—	OCP input
	KEY1	PBS0	NSI	—	Touch key input
PB1/TX/CTCK1/ OVPI0/KEY2	PB1	PBPU PBS0	ST	CMOS	General purpose I/O. Register enabled pull-up
	TX	PBS0	—	CMOS	UART data transmit pin
	CTCK1	PBS0 IFS1	ST	—	CTM1 clock input
	OVPI0	PBS0	AN	—	OVP input 0
	KEY2	PBS0	NSI	—	Touch key input
PB2/SCL/AN6/KEY9	PB2	PBPU PBS0	ST	CMOS	General purpose I/O. Register enabled pull-up
	SCL	PBS0 IFS0	ST	NMOS	I ² C clock line
	AN6	PBS0	AN	—	A/D converter external input channel
	KEY9	PBS0	NSI	—	Touch key input
PB3/SDA/AN7/ KEY10	PB3	PBPU PBS0	ST	CMOS	General purpose I/O. Register enabled pull-up
	SDA	PBS0 IFS0	ST	NMOS	I ² C data line
	AN7	PBS0	AN	—	A/D converter external input channel
	KEY10	PBS0	NSI	—	Touch key input

Pin Name	Function	OPT	I/T	O/T	Description
PB4/CTP0/AN2/ KEY7	PB4	PBPU PBS1	ST	CMOS	General purpose I/O. Register enabled pull-up
	CTP0	PBS1	—	CMOS	CTM0 output
	AN2	PBS1	AN	—	A/D converter external input channel
	KEY7	PBS1	NSI	—	Touch key input
PB5/CTP0B/AN3/ KEY8	PB5	PBPU PBS1	ST	CMOS	General purpose I/O. Register enabled pull-up
	CTP0B	PBS1	—	CMOS	CTM0 inverted output
	AN3	PBS1	AN	—	A/D converter external input channel
	KEY8	PBS1	NSI	—	Touch key input
PB6/CTP1/AN4/ OCPAO	PB6	PBPU PBS1	ST	CMOS	General purpose I/O. Register enabled pull-up
	CTP1	PBS1	—	CMOS	CTM1 output
	AN4	PBS1	AN	—	A/D converter external input channel
	OCPAO	PBS1	AN	—	OCP operational amplifier output
PB7/CTP1B/AN5/ OVPCOUT	PB7	PBPU PBS1	ST	CMOS	General purpose I/O. Register enabled pull-up
	CTP1B	PBS1	—	CMOS	CTM1 inverted output
	AN5	PBS1	AN	—	A/D converter external input channel
	OVPCOUT	PBS1	—	CMOS	OVP comparator output
PC0/PTP/OVPI0/ KEY11	PC0	PCPU PCS0	ST	CMOS	General purpose I/O. Register enabled pull-up
	PTP	PCS0	—	CMOS	PTM output
	OVPI0	PCS0	AN	—	OVP input 0
	KEY11	PCS0	NSI	—	Touch key input
PC1/INT/PTPB/ OVPI1/KEY12	PC1	PCPU PCS0	ST	CMOS	General purpose I/O. Register enabled pull-up
	INT	PCS0 INTC0 INTEG	ST	—	External interrupt input
	PTPB	PCS0	—	CMOS	PTM inverted output
	OVPI1	PCS0	AN	—	OVP input 1
	KEY12	PCS0	NSI	—	Touch key input
PC2/CTP1/SCL/RX	PC2	PCPU PCS0	ST	CMOS	General purpose I/O. Register enabled pull-up
	CTP1	PCS0	—	CMOS	CTM1 output
	SCL	PCS0 IFS0	ST	NMOS	I ² C clock line
	RX	PCS0 IFS0	ST	—	UART data receive pin
PC3/CTP0/SDA/TX	PC3	PCPU PCS0	ST	CMOS	General purpose I/O. Register enabled pull-up
	CTP0	PCS0	—	CMOS	CTM0 output
	SDA	PCS0 IFS0	ST	NMOS	I ² C data line
	TX	PCS0	—	CMOS	UART data transmit pin
PC4~PC5	PC4~PC5	PCPU	ST	CMOS	General purpose I/O. Register enabled pull-up
PD0~PD5	PD0~PD5	—	ST	CMOS	High voltage I/O port
VDD/AVDD/VLDO	VDD	—	PWR	—	Digital positive power supply
	AVDD	—	PWR	—	Analog positive power supply
	VLDO	—	—	PWR	LDO output voltage

Pin Name	Function	OPT	I/T	O/T	Description
VSS/AVSS/IOVSS/HVSS	VSS	—	PWR	—	Digital negative power supply, ground
	AVSS	—	PWR	—	Analog negative power supply, ground
	IOVSS	—	PWR	—	I/O port negative power supply, ground
	HVSS	—	PWR	—	High voltage negative power supply, ground
VCC1	VCC1	—	PWR	—	Provides high voltage positive power supply for LDO input
VCC2	VCC2	—	PWR	—	Provides high voltage positive power supply for High Voltage I/O and Level Shifter
NC	NC	—	—	—	Unused

Legend: I/T: Input type;

O/T: Output type;

OPT: Optional by register selection;

PWR: Power;

ST: Schmitt Trigger input;

CMOS: CMOS output;

NMOS: NMOS output;

AN: Analog signal;

NSI: Non-standard input;

LXT: Low frequency crystal oscillator.

Absolute Maximum Ratings

Supply Voltage (V_{CC}).....	$V_{SS}-0.3V$ to $10.0V$
Supply Voltage (V_{DD})	$V_{SS}-0.3V$ to $V_{SS}+6.0V$
High Voltage Input Voltage	$V_{SS}-0.3V$ to $V_{CC}+0.3V$
Input Voltage	$V_{SS}-0.3V$ to $V_{DD}+0.3V$
Storage Temperature.....	$-50^{\circ}C$ to $125^{\circ}C$
Operating Temperature.....	$-40^{\circ}C$ to $85^{\circ}C$
High Voltage I_{OH} Total.....	$-150mA$
I_{OH} Total	$-80mA$
High Voltage I_{OL} Total	$150mA$
I_{OL} Total	$80mA$
Total Power Dissipation	$500mW$

Note: These are stress ratings only. Stresses exceeding the range specified under “Absolute Maximum Ratings” may cause substantial damage to this device. Functional operation of this device at other conditions beyond those listed in the specification is not implied and prolonged exposure to extreme conditions may affect device reliability.

D.C. Electrical Characteristics

For data in the following tables, note that factors such as oscillator type, operating voltage, operating frequency, pin load conditions, temperature and program instruction type, can all exert an influence on the measured values.

Operating Voltage Characteristics

Ta=25°C

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
V _{DD}	Operating Voltage – HIRC	f _{SYS} =f _{HIRC} =8MHz	4.5	—	5.5	V
		f _{SYS} =f _{HIRC} =12MHz	4.5	—	5.5	
		f _{SYS} =f _{HIRC} =16MHz	4.5	—	5.5	
	Operating Voltage – LIRC	f _{SYS} =f _{LIRC} =32kHz	4.5	—	5.5	V
	Operating Voltage – LXT	f _{SYS} =f _{LXT} =32.768kHz	4.5	—	5.5	V

Operating Current Characteristics

Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
I _{DD}	SLOW Mode – LIRC	5V	f _{SYS} =f _{LIRC} =32kHz, LDO current consumption included	—	180	200	μA
	SLOW Mode – LXT	5V	f _{SYS} =f _{LXT} =32768Hz, LDO current consumption included	—	180	200	μA
	FAST Mode – HIRC	5V	f _{SYS} =f _{HIRC} =8MHz, LDO current consumption included	—	1.6	2.4	mA
			f _{SYS} =f _{HIRC} =12MHz, LDO current consumption included	—	2.4	3.6	
			f _{SYS} =f _{HIRC} =16MHz, LDO current consumption included	—	6.0	9.0	

Note: When using the characteristic table data, the following notes should be taken into consideration:

1. Any digital input is set in a non-floating condition.
2. All measurements are taken under conditions of no load and with all peripherals in an off state.
3. There are no DC current paths.
4. All Operating Current values are measured using a continuous NOP instruction program loop.

Standby Current Characteristics

Ta=25°C, unless otherwise specified

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Max. @85°C	Unit
		V _{DD}	Conditions					
I _{STB}	SLEEP Mode	5V	WDT on, LDO current consumption included	—	160	200	210	μA
	IDLE0 Mode – LIRC	5V	f _{SUB} on, LDO current consumption included	—	165	200	215	μA
	IDLE0 Mode – LXT	5V	f _{SUB} on, LDO current consumption included	—	165	200	215	μA
	IDLE1 Mode – HIRC	5V	f _{SUB} on, f _{SYS} =8MHz, LDO current consumption included	—	1.0	1.8	2.0	mA
			f _{SUB} on, f _{SYS} =12MHz, LDO current consumption included	—	1.5	2.6	3.0	
			f _{SUB} on, f _{SYS} =16MHz, LDO current consumption included	—	2.0	3.5	4.0	

Note: When using the characteristic table data, the following notes should be taken into consideration:

1. Any digital input is set in a non-floating condition.
2. All measurements are taken under conditions of no load and with all peripherals in an off state.
3. There are no DC current paths.
4. All Standby Current values are taken after a HALT instruction executed thus stopping all instruction execution.

A.C. Characteristics

For data in the following tables, note that factors such as oscillator type, operating voltage, operating frequency and temperature etc., can all exert an influence on the measured values.

High Speed Internal Oscillator – HIRC – Frequency Accuracy

During the program writing operation the writer will trim the HIRC oscillator at a user selected HIRC frequency and user selected voltage of 5V.

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Temp.				
f _{HIRC}	8MHz Writer Trimmed HIRC Frequency	5V	25°C	-1%	8	+1%	MHz
			-40°C~85°C	-2%	8	+2%	
		4.5V~5.5V	25°C	-2.5%	8	+2.5%	
			-40°C~85°C	-3%	8	+3%	
	12MHz Writer Trimmed HIRC Frequency	5V	25°C	-1%	12	+1%	MHz
			-40°C~85°C	-2%	12	+2%	
		4.5V~5.5V	25°C	-2.5%	12	+2.5%	
			-40°C~85°C	-3%	12	+3%	
	16MHz Writer Trimmed HIRC Frequency	5V	25°C	-1%	16	+1%	MHz
			-40°C~85°C	-2%	16	+2%	
		4.5V~5.5V	25°C	-2.5%	16	+2.5%	
			-40°C~85°C	-3%	16	+3%	

Note: 1. The 5V values for V_{DD} are provided as this is the fixed voltage at which the HIRC frequency is trimmed by the writer.

2. The row below the 5V trim voltage row is provided to show the values for the full V_{DD} range operating voltage. It is recommended that the trim voltage is fixed at 5V for application voltage ranges from 3.3V to 5.5V.

3. The minimum and maximum tolerance values provided in the table are only for the frequency at which the writer trims the HIRC oscillator. After trimming at this chosen specific frequency any change in HIRC oscillator frequency using the oscillator register control bits by the application program will give a frequency tolerance to within $\pm 20\%$.

Internal Low Speed Oscillator Characteristics – LIRC

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Temp.				
f _{LIRC}	LIRC Frequency	4.5V~5.5V	25°C	-10%	32	+10%	kHz
			-40°C~85°C	-50%	32	+60%	
t _{START}	LIRC Start Up Time	—	25°C	—	—	500	μs

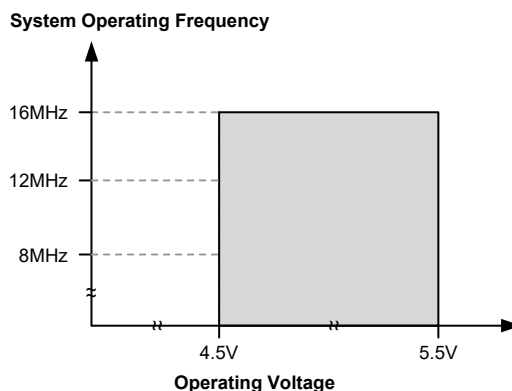
External Low Speed Oscillator Characteristics – LXT

Ta=25°C

C1=C2=10pF, R_P=10MΩ (C1, C2 and R_P are external components), C_L=7pF, ESR=30kΩ

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
f _{LXT}	LXT Frequency	4.5V~5.5V	—	—	32768	—	Hz
t _{START}	LXT Start Up Time	5V	—	—	—	1000	μs
Duty Cycle	Duty Cycle	—	—	40	—	60	%
R _{NEG}	Negative Resistance	5V	—	3×ESR	—	—	Ω

Operating Frequency Characteristic Curves



System Start Up Time Characteristics

Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
t _{SST}	System Start-up Time Wake-up from Condition where f _{SYS} is Off	—	f _{SYS} =f _H ~f _H /64, f _H =f _{HIRC}	—	16	—	t _{HIRC}
		—	f _{SYS} =f _{SUB} =f _{LXT}	—	1024	—	t _{LXT}
		—	f _{SYS} =f _{SUB} =f _{LIRC}	—	2	—	t _{LIRC}
	System Start-up Time Wake-up from Condition where f _{SYS} is On	—	f _{SYS} =f _H ~f _H /64, f _H =f _{HIRC}	—	2	—	t _H
		—	f _{SYS} =f _{SUB} =f _{LIRC} or f _{LXT}	—	2	—	t _{SUB}
	System Speed Switch Time FAST to SLOW Mode or SLOW to FAST Mode	—	f _{HIRC} switches from off → on	—	16	—	t _{HIRC}
		—	f _{LXT} switches from off → on	—	1024	—	t _{LXT}
t _{RSTD}	System Reset Delay Time Reset Source from Power-on Reset or LVR Hardware Reset	—	RR _{POR} =5V/ms	30	48	72	ms
	System Reset Delay Time LVRC/WDTC/RSTC Software Reset	—	—				
	System Reset Delay Time Reset Source from WDT Overflow	—	—	10	16	24	ms
t _{SRESET}	Minimum Software Reset Width to Reset	—	—	45	90	180	μs

- Note: 1. For the System Start-up time values, whether f_{SYS} is on or off depends upon the mode type and the chosen f_{SYS} system oscillator. Details are provided in the System Operating Modes section.
2. The time units, shown by the symbols t_{HIRC}, t_{SYS} etc. are the inverse of the corresponding frequency values as provided in the frequency tables. For example t_{HIRC}=1/f_{HIRC}, t_{SYS}=1/f_{SYS} etc.
3. If the LIRC is used as the system clock and if it is off when in the SLEEP Mode, then an additional LIRC start up time, t_{START}, as provided in the LIRC frequency table, must be added to the t_{SST} time in the table above.
4. The System Speed Switch Time is effectively the time taken for the newly activated oscillator to start up.

Input/Output Characteristics

Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{IL}	Input Low Voltage for I/O Ports	5V	—	0	—	1.5	V
		—		0	—	0.2V _{DD}	
V _{IH}	Input High Voltage for I/O Ports	5V	—	3.5	—	5.0	V
		—		0.8V _{DD}	—	V _{DD}	
I _{OL}	Sink Current for I/O Ports (PA, PB, PC)	5V	V _{OL} =0.1V _{DD} , PxNS=0, x=A, B or C	32	64	—	mA
			V _{OL} =0.1V _{DD} , PxNS=1, x=A, B or C	50	100	—	
I _{OH}	Source Current for I/O Ports (PA, PB, PC)	5V	V _{OH} =0.9V _{DD} , SLEDCn[m+1:m]=00B (n=0, 1; m=0, 2, 4 or 6)	-1.5	-2.9	—	mA
			V _{OH} =0.9V _{DD} , SLEDCn[m+1:m]=01B (n=0, 1; m=0, 2, 4 or 6)	-2.5	-5.1	—	
			V _{OH} =0.9V _{DD} , SLEDCn[m+1:m]=10B (n=0, 1; m=0, 2, 4 or 6)	-3.6	-7.3	—	
			V _{OH} =0.9V _{DD} , SLEDCn[m+1:m]=11B (n=0, 1; m=0, 2, 4 or 6)	-8	-16	—	
R _{PH}	Pull-high Resistance for I/O Ports (Note)	5V	—	10	30	50	kΩ
I _{LEAK}	Input Leakage Current	5V	V _{IN} =V _{DD} or V _{IN} =V _{SS}	—	—	±1	μA
t _{TCK}	TM Clock Input Pin Minimum Pulse Width	—	—	0.3	—	—	μs
t _{TPI}	TM Capture Input Pin Minimum Pulse Width	—	—	0.3	—	—	μs
t _{INT}	External Interrupt Minimum Pulse Width	—	—	10	—	—	μs

Note: The R_{PH} internal pull high resistance value is calculated by connecting to ground and enabling the input pin with a pull-high resistor and then measuring the pin current at the specified supply voltage level. Dividing the voltage by this measured current provides the R_{PH} value.

Memory Characteristics

Ta=-40°C~85°C, unless otherwise specified

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{RW}	V _{DD} for Read / Write	—	—	V _{DDmin}	—	V _{DDmax}	V
Flash Program / Data EEPROM Memory							
t _{DEW}	Erase / Write Cycle Time – Flash Program Memory	—	—	—	2	3	ms
	Write Cycle Time – Data EEPROM Memory	—	—	—	4	6	ms
I _{DDPGM}	Programming / Erase Current on V _{DD}	—	—	—	—	5.0	mA
E _P	Cell Endurance	—	—	100K	—	—	E/W
t _{RETD}	ROM Data Retention Time	—	Ta=25°C	—	40	—	Year
RAM Data Memory							
V _{DR}	RAM Data Retention Voltage	—	Device in SLEEP Mode	1.0	—	—	V

LVD/LVR Electrical Characteristics

Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{LVR}	Low Voltage Reset Voltage	—	LVR enable, voltage select 2.1V	-5%	2.1	+5%	V
		—	LVR enable, voltage select 2.55V		2.55		
		—	LVR enable, voltage select 3.15V		3.15		
		—	LVR enable, voltage select 3.8V		3.8		
V _{LVD}	Low Voltage Detection Voltage	—	LVD enable, voltage select 2.0V	-5%	2.0	+5%	V
		—	LVD enable, voltage select 2.2V		2.2		
		—	LVD enable, voltage select 2.4V		2.4		
		—	LVD enable, voltage select 2.7V		2.7		
		—	LVD enable, voltage select 3.0V		3.0		
		—	LVD enable, voltage select 3.3V		3.3		
		—	LVD enable, voltage select 3.6V		3.6		
		—	LVD enable, voltage select 4.0V		4.0		
t _{LVDs}	LVDO Stable Time	—	For LVR enable, VBGEN=0, LVD off → on	—	—	15	μs
t _{LVR}	Minimum Low Voltage Width to Reset	—	—	120	240	480	μs
t _{LVD}	Minimum Low Voltage Width to Interrupt	—	—	60	120	240	μs

A/D Converter Electrical Characteristics

Ta=-40°C~85°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{DD}	A/D Converter Operating Voltage	—	—	4.5	—	5.5	V
V _{ADI}	A/D Converter Input Voltage	—	—	0	—	V _{REF}	V
V _{REF}	A/D Converter Reference Voltage	—	—	2	—	V _{DD}	V
DNL	A/D Converter Differential Non-linearity	5V	V _{REF} =V _{DD} , t _{ADCK} =0.5μs	-3	—	+3	LSB
			V _{REF} =V _{DD} , t _{ADCK} =10μs				
INL	A/D Converter Integral Non-linearity	5V	V _{REF} =V _{DD} , t _{ADCK} =0.5μs	-4	—	+4	LSB
			V _{REF} =V _{DD} , t _{ADCK} =10μs				
I _{ADC}	Additional Current Consumption for A/D Converter Enable	5V	No load, t _{ADCK} =0.5μs	—	1.5	3.0	mA
t _{ADCK}	A/D Converter Clock Period	—	—	0.5	—	10	μs
t _{ON2ST}	A/D Converter On-to-Start Time	—	—	4	—	—	μs
t _{ADS}	A/D Converter Sampling Time	—	—	—	4	—	t _{ADCK}
t _{ADC}	A/D Conversion Time (Including A/D Sampling and Hold Time)	—	—	—	16	—	t _{ADCK}

Internal Reference Voltage Characteristics

Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{BG}	Bandgap Reference Voltage	—	—	-5%	1.04	+5%	V
t _{BGS}	V _{BG} Turn-on Stable Time	—	No load	—	—	150	μs

- Note: 1. All the above parameters are measured under conditions of no load condition unless otherwise described.
2. A 0.1μF ceramic capacitor should be connected between VDD and GND.
3. The V_{BG} voltage is used as the A/D converter internal signal input.

High Voltage I/O Electrical Characteristics

V_{DD}=5V, Ta=25°C, unless otherwise specified

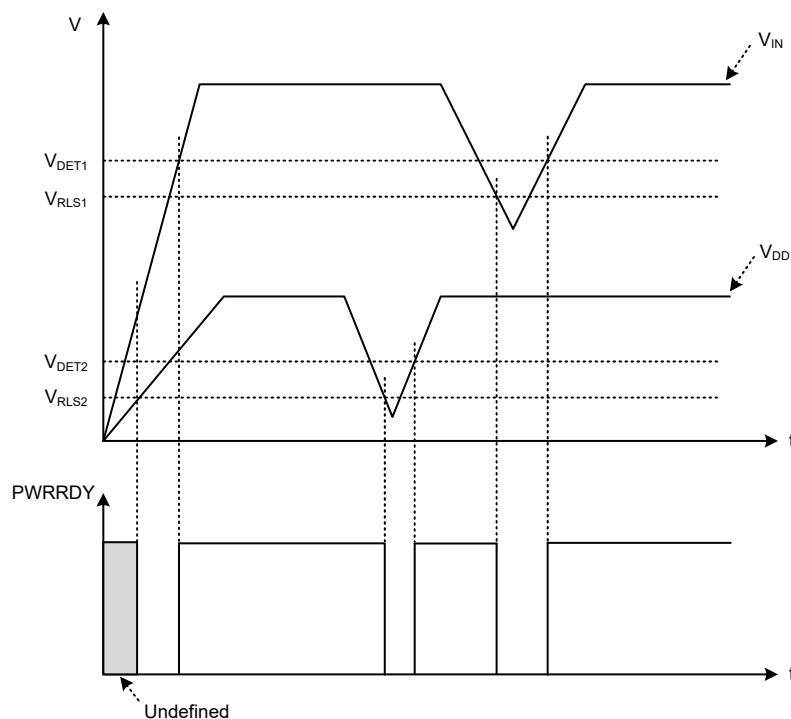
Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
V _{IN}	Input Voltage	—	V _{DET1}	—	10	V
V _{IH}	Input High Voltage for High Voltage I/O Ports	—	0.6V _{IN}	—	V _{IN}	V
V _{IL}	Input Low Voltage for High Voltage I/O Ports	—	0	—	0.3V _{IN}	V
I _{OH}	Source Current for High Voltage I/O Ports	V _{OH} =0.9×V _{IN} , V _{IN} =10V	-40	-70	—	mA
I _{OL}	Sink Current for High Voltage I/O Ports	V _{OL} =0.1×V _{IN} , V _{IN} =10V	50	80	—	mA
t _{SF}	Short Circuit Flag Response Time	SFRTC=0, Ta=25°C	2	—	3	ms
		SFRTC=0, Ta=-40°C~85°C	1.5	—	3.9	
		SFRTC=1, Ta=25°C	1.0	—	1.5	
		SFRTC=1, Ta=-40°C~85°C	0.75	—	1.85	

Voltage Detector Electrical Characteristics

Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{IN}	Input Voltage	—	—	V _{DET1}	—	10	V
V _{DET1}	V _{CC2} Detect Level (Note)	—	V _{IN} =0V → 10V	Typ.-0.5	7	Typ.+0.5	V
V _{RSL1}	V _{CC2} Release Level (Note)	—	V _{IN} =10V → 0V	V _{DET1} -V _{HYS1}			V
V _{HYS1}	Hysteresis	—	V _{IN} =10V ↔ 5V	100	750	1000	mV
V _{DET2}	V _{DD} Detect Level	—	V _{DD} =0V → 5V	Typ.-0.2	2.5	Typ.+0.2	V
V _{RSL2}	V _{DD} Release Level (Note)	—	V _{DD} =5V → 0V	V _{DET2} -V _{HYS2}			V
V _{HYS2}	Hysteresis	—	V _{DD} =0V ↔ 5V	100	250	500	mV

Note:



High Voltage I/O Other Electrical Characteristics

Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{IN}	Input Voltage	—	—	7	—	10	V
V _{CC10}	V _{CC10} Accuracy	—	V _{IN} =10V	-5%	0.2V _{IN}	+5%	V
V _{CC20}	V _{CC20} Accuracy	—	V _{IN} =10V	-5%	0.2V _{IN}	+5%	V

Note: Divider 1: R11:R12=4:1(12kΩ/3kΩ), V_{CC10}=R12/(R11+R12)×V_{CC1}=0.2V_{CC1}.

Divider 2: R21:R22=4:1(12kΩ/3kΩ), V_{CC20}=R22/(R21+R22)×V_{CC2}=0.2V_{CC2}.

Low Dropout Regulator Electrical Characteristics

C_{LOAD}=10μF+0.1μF, V_{IN}=V_{OUT}+1V, Ta=25°C, unless otherwise specified

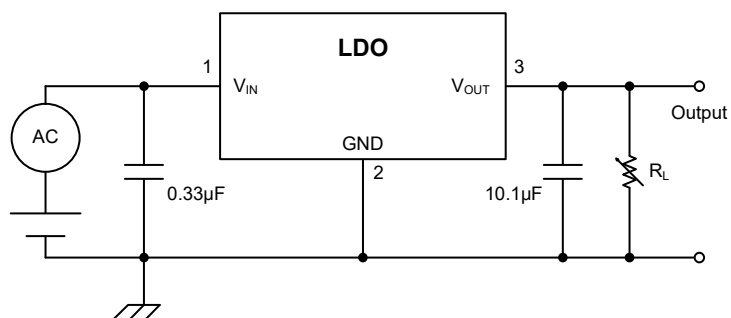
Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{IN}	Conditions				
V _{IN}	Input Voltage	—	—	6	—	10	V
V _{OUT}	Output Voltage	—	Ta=25°C, I _{LOAD} =1mA, V _{OUT} =5.0V	-2%	5.0	+2%	V
		—	Ta=-40°C~85°C, I _{LOAD} =1mA, V _{OUT} =5.0V	-5%	5.0	+5%	V
ΔV _{LOAD}	Load Regulation ⁽¹⁾	—	1mA≤I _{LOAD} ≤70mA, V _{IN} =V _{OUT} +1V	—	0.015	0.033	%/mA
V _{DROP}	Dropout Voltage ⁽²⁾	—	ΔV _{OUT} =2%, I _{LOAD} =1mA, V _{IN} =V _{OUT} +1V	—	—	100	mV
I _{OUT}	Output Current	—	V _{IN} =V _{OUT} +1V, ΔV _{OUT} =-3%	250	—	—	mA
		—	V _{IN} =V _{OUT} +2V, ΔV _{OUT} =-3%	500	—	—	mA

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{IN}	Conditions				
I _Q	Quiescent Current	10V	No load	—	120	200	μA
ΔV _{LINE}	Line Regulation	—	6V ≤ V _{IN} ≤ 10V, I _{LOAD} = 1mA	—	—	0.2	%/V
TC	Temperature Coefficient	—	T _a = -40°C ~ 85°C, I _{LOAD} = 10mA	—	±1.5	±2.0	mV/°C
ΔV _{OUT_RIPPLE}	Output Voltage Ripple	6V	I _{LOAD} = 10mA	—	—	40	mV
RR	Ripple Rejection ⁽³⁾	—	V _{IN} = 10V _{DC} + 2V _{P-P(AC)} , I _{LOAD} ≤ 50mA, f = 120Hz	35	—	—	dB
I _{LIMIT}	Current Limit	6V	ΔV _{OUT} = -10%	600	800	—	mA
t _{START}	LDO Start Up Time	6V	I _{LOAD} = 1mA, V _{OUT} settle to ±5%	—	—	10	ms

Note: 1. Load regulation is measured at a constant junction temperature, using pulse testing with a low ON time and is guaranteed up to the maximum power dissipation. Power dissipation is determined by the input/output differential voltage and the output current. Guaranteed maximum power dissipation will not be available over the full input/output range. The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(MAX)} - T_a) / \theta_{JA}$.

2. Dropout voltage is defined as the input voltage minus the output voltage that produces a 2% change in the output voltage from the value at appointed V_{IN}.

3. Ripple rejection ratio measurement circuit. $RR = 20 \times \log(\Delta V_{IN} / \Delta V_{OUT})$.



4. Application information for LDO load capacitor selection for stability:

Recommended Output Capacitor

T_a = 25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
C _{LOAD}	Output Load Capacitor	—	Capacitor	4.7	10	—	μF

In common with most regulators, the LDO requires an external capacitor connected between V_{OUT} and ground for regulator stability. If the ESR is less than 10Ω, capacitor values of 4.7μF or large are acceptable. Any aluminum electrolytic capacitor meeting the requirements described above is suitable.

For better load transient response purposes, use a combination of a C_{LOAD} 10μF and an extra 0.1μF capacitor on V_{OUT}. Note that the 0.1μF capacitor is always required on V_{OUT} and is strongly recommended to be a multi-layer ceramic capacitor. The internal regulator is designed to be stable with an output filter capacitor C_{LOAD} and ESR as recommended.

OCP Electrical Characteristics

Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
I _{OCP}	Operating Current	5V	OCPEN[1:0]=01B, DAC V _{REF} =2.5V	—	730	1250	μA
V _{OS_CMP}	Comparator Input Offset Voltage	5V	Without calibration (OCPCOF[4:0]=10000B)	-15	—	15	mV
		5V	With calibration	-4	—	4	
V _{HYS}	Hysteresis	5V	—	10	40	60	mV
V _{CM_CMP}	Comparator Common Mode Voltage Range	5V	—	V _{SS}	—	V _{DD} -1.4	V
V _{OS_OPA}	OPA Input Offset Voltage	5V	Without calibration (OCPOOF[5:0]=100000B)	-15	—	15	mV
		5V	With calibration	-4	—	4	
V _{CM_OPA}	OPA Common Mode Voltage Range	5V	—	V _{SS}	—	V _{DD} -1.4	V
V _{OR}	OPA Maximum Output Voltage Range	5V	—	V _{SS} +0.1	—	V _{DD} -0.1	V
G _a	PGA Gain Accuracy	5V	All gains	-5	—	+5	%
V _{REF}	D/A Converter Reference Voltage	5V	OCPVRS=1	2	—	V _{DD}	V
DNL	Differential Non-linearity	5V	DAC V _{REF} =V _{DD}	-1	—	+1	LSB
INL	Integral Non-linearity	5V	DAC V _{REF} =V _{DD}	-1.5	—	+1.5	LSB

OVP Electrical Characteristics

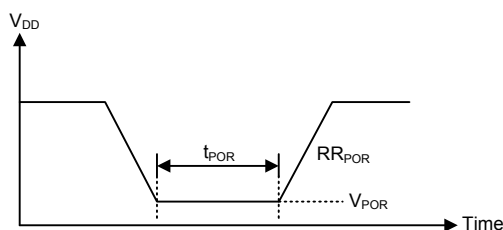
Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
I _{OVP}	Operating Current	5V	OVPEN=1, DAC V _{REF} =V _{DD}	—	500	750	μA
V _{OS}	Input Offset Voltage	5V	With calibration	-2	—	2	mV
V _{HYS}	Hysteresis	5V	HYS[1:0]=00B	0	0	5	mV
			HYS[1:0]=01B	15	30	45	
			HYS[1:0]=10B	40	60	80	
			HYS[1:0]=11B	60	80	100	
V _{CM}	Common Mode Voltage Range	5V	—	V _{SS}	—	V _{DD} -1.4	V
DNL	Differential Non-linearity	5V	DAC V _{REF} =V _{DD}	-1	—	+1	LSB
INL	Integral Non-linearity	5V	DAC V _{REF} =V _{DD}	-1.5	—	+1.5	LSB
t _{RP}	OVP Response Time	5V	OVPDA[7:0]=10000000B, OVPDEB[2:0]=000B, DAC V _{REF} =V _{DD} , OVP input=2.1V~3.6V	—	1.0	1.8	μs

Power-on Reset Characteristics

Ta=25°C

Symbol	Parameter	Test Conditions		Min.	Typ.	Max.	Unit
		V _{DD}	Conditions				
V _{POR}	V _{DD} Start Voltage to Ensure Power-on Reset	—	—	—	—	100	mV
RR _{POR}	V _{DD} Rising Rate to Ensure Power-on Reset	—	—	0.035	—	—	V/ms
t _{POR}	Minimum Time for V _{DD} Stays at V _{POR} to Ensure Power-on Reset	—	—	1	—	—	ms

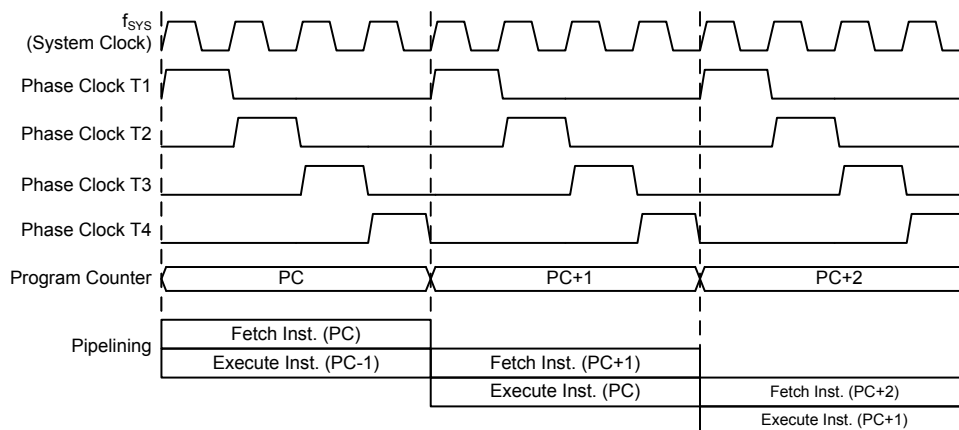


System Architecture

A key factor in the high-performance features of the range of microcontrollers is attributed to their internal system architecture. The device takes advantage of the usual features found within RISC microcontrollers providing increased speed of operation and enhanced performance. The pipelining scheme is implemented in such a way that instruction fetching and instruction execution are overlapped, hence instructions are effectively executed in one or two cycles for most of the standard or extended instructions respectively. The exceptions to this are branch or call instructions which need one more cycle. An 8-bit wide ALU is used in practically all instruction set operations, which carries out arithmetic operations, logic operations, rotation, increment, decrement, branch decisions, etc. The internal data path is simplified by moving data through the Accumulator and the ALU. Certain internal registers are implemented in the Data Memory and can be directly or indirectly addressed. The simple addressing methods of these registers along with additional architectural features ensure that a minimum of external components is required to provide a functional I/O and A/D control system with maximum reliability and flexibility. This makes the device suitable for low-cost, high-volume production for controller applications.

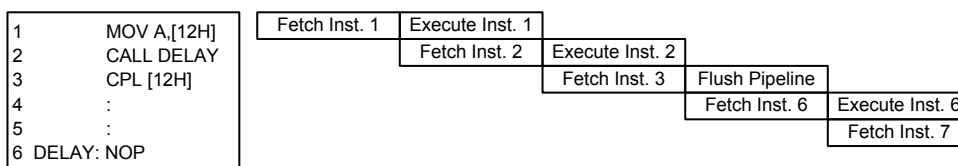
Clocking and Pipelining

The main system clock, derived from either a HIRC, LIRC or LXT oscillator is subdivided into four internally generated non-overlapping clocks, T1~T4. The Program Counter is incremented at the beginning of the T1 clock during which time a new instruction is fetched. The remaining T2~T4 clocks carry out the decoding and execution functions. In this way, one T1~T4 clock cycle forms one instruction cycle. Although the fetching and execution of instructions takes place in consecutive instruction cycles, the pipelining structure of the microcontroller ensures that instructions are effectively executed in one instruction cycle. The exception to this are instructions where the contents of the Program Counter are changed, such as subroutine calls or jumps, in which case the instruction will take one more instruction cycle to execute.



System Clocking and Pipelining

For instructions involving branches, such as jump or call instructions, two machine cycles are required to complete instruction execution. An extra cycle is required as the program takes one cycle to first obtain the actual jump or call address and then another cycle to actually execute the branch. The requirement for this extra cycle should be taken into account by programmers in timing sensitive applications.



Instruction Fetching

Program Counter

During program execution, the Program Counter is used to keep track of the address of the next instruction to be executed. It is automatically incremented by one each time an instruction is executed except for instructions, such as “JMP” or “CALL” that demand a jump to a non-consecutive Program Memory address. Only the lower 8 bits, known as the Program Counter Low Register, are directly addressable by the application program.

When executing instructions requiring jumps to non-consecutive addresses such as a jump instruction, a subroutine call, interrupt or reset, etc., the microcontroller manages program control by loading the required address into the Program Counter. For conditional skip instructions, once the condition has been met, the next instruction, which has already been fetched during the present instruction execution, is discarded and a dummy cycle takes its place while the correct instruction is obtained.

Program Counter	
High Byte	Low Byte (PCL)
PC12~PC8	PCL7~PCL0

Program Counter

The lower byte of the Program Counter, known as the Program Counter Low register or PCL, is available for program control and is a readable and writeable register. By transferring data directly into this register, a short program jump can be executed directly; however, as only this low byte is available for manipulation, the jumps are limited to the present page of memory that is 256 locations. When such program jumps are executed it should also be noted that a dummy cycle

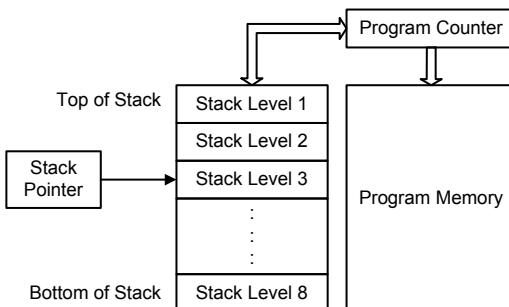
will be inserted. Manipulating the PCL register may cause program branching, so an extra cycle is needed to pre-fetch.

Stack

This is a special part of the memory which is used to save the contents of the Program Counter only. The stack is organized into 8 levels and neither part of the data nor part of the program space, and is neither readable nor writeable. The activated level is indexed by the Stack Pointer, and is neither readable nor writeable. At a subroutine call or interrupt acknowledge signal, the contents of the Program Counter are pushed onto the stack. At the end of a subroutine or an interrupt routine, signaled by a return instruction, RET or RETI, the Program Counter is restored to its previous value from the stack. After a device reset, the Stack Pointer will point to the top of the stack.

If the stack is full and an enabled interrupt takes place, the interrupt request flag will be recorded but the acknowledge signal will be inhibited. When the Stack Pointer is decremented, by RET or RETI, the interrupt will be serviced. This feature prevents stack overflow allowing the programmer to use the structure more easily. However, when the stack is full, a CALL subroutine instruction can still be executed which will result in a stack overflow. Precautions should be taken to avoid such cases which might cause unpredictable program branching.

If the stack is overflow, the first Program Counter save in the stack will be lost.



Arithmetic and Logic Unit – ALU

The arithmetic-logic unit or ALU is a critical area of the microcontroller that carries out arithmetic and logic operations of the instruction set. Connected to the main microcontroller data bus, the ALU receives related instruction codes and performs the required arithmetic or logical operations after which the result will be placed in the specified register. As these ALU calculation or operations may result in carry, borrow or other status changes, the status register will be correspondingly updated to reflect these changes. The ALU supports the following functions:

- Arithmetic operations:
ADD, ADDM, ADC, ADCM, SUB, SUBM, SBC, SBCM, DAA,
LADD, LADDM, LADC, LADCM, LSUB, LSUBM, LSBC, LSBCM, LDAA
- Logic operations:
AND, OR, XOR, ANDM, ORM, XORM, CPL, CPLA,
LAND, LOR, LXOR, LANDM, LORM, LXORM, LCPL, LCPLA
- Rotation:
RRA, RR, RRCA, RRC, RLA, RL, RLCA, RLC,
LRR, LRRCA, LRR, LRLA, LRL, LRLCA, LRLC
- Increment and Decrement:
INCA, INC, DECA, DEC,
LINCA, LINC, LDECA, LDEC

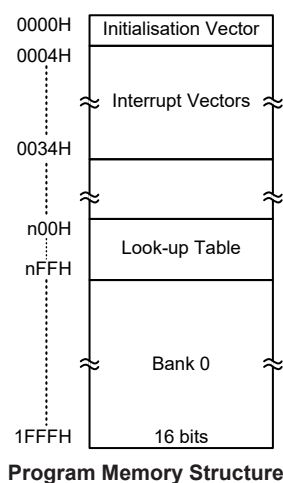
- Branch decision:
JMP, SZ, SZA, SNZ, SIZ, SDZ, SIZA, SDZA, CALL, RET, RETI,
LSZ, LSZA, LSNZ, LSIZ, LSDZ, LSIZA, LSDZA

Flash Program Memory

The Program Memory is the location where the user code or program is stored. For this device the Program Memory is Flash type, which means it can be programmed and re-programmed a large number of times, allowing the user the convenience of code modification on the same device. By using the appropriate programming tools, the Flash device offers users the flexibility to conveniently debug and develop their applications while also offering a means of field programming and updating.

Structure

The Program Memory has a capacity of 8K×16 bits. The Program Memory is addressed by the Program Counter and also contains data, table information and interrupt entries. Table data, which can be configured in any location within the Program Memory, is addressed by a separate table pointer register.



Program Memory Structure

Special Vectors

Within the Program Memory, certain locations are reserved for the reset and interrupts. The location 0000H is reserved for use by the device reset for program initialisation. After a device reset is initiated, the program will jump to this location and begin execution.

Look-up Table

Any location within the Program Memory can be defined as a look-up table where programmers can store fixed data. To use the look-up table, the table pointer must first be configured by placing the address of the look up data to be retrieved in the table pointer register, TBLP and TBHP. These registers define the total address of the look-up table.

After setting up the table pointer, the table data can be retrieved from the Program Memory using the “TABRD [m]” or “TABRDL [m]” instructions respectively when the memory [m] is located in sector 0. If the memory [m] is located in other sectors except sector 0, the data can be retrieved from the program memory using the corresponding extended table read instruction such as “LTABRD [m]” or “LTABRDL [m]” respectively. When the instruction is executed, the lower order table byte from the Program Memory will be transferred to the user defined Data Memory register [m] as specified

in the instruction. The higher order table data byte from the Program Memory will be transferred to the TBLH special register.

The accompanying diagram illustrates the addressing data flow of the look-up table.

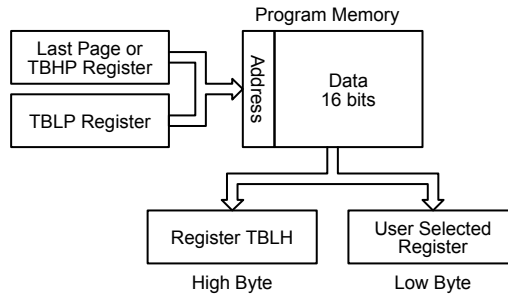


Table Program Example

The following example shows how the table pointer and table data is defined and retrieved from the microcontroller. This example uses raw table data located in the Program Memory which is stored there using the ORG statement. The value at this ORG statement is “1F00H” which refers to the start address of the last page within the 8K Program Memory of the microcontroller. The table pointer low byte register is set here to have an initial value of “06H”. This will ensure that the first data read from the data table will be at the Program Memory address “1F06H” or 6 locations after the start of the last page. Note that the value for the table pointer is referenced to the specific page pointed by the TBLP and TBHP registers if the “TABRD [m]” or “LTABRD [m]” instruction is being used. The high byte of the table data which in this case is equal to zero will be transferred to the TBLH register automatically when the “TABRD [m]” instruction is executed.

Because the TBLH register is a read/write register and can be restored, care should be taken to ensure its protection if both the main routine and Interrupt Service Routine use table read instructions. If using the table read instructions, the Interrupt Service Routines may change the value of the TBLH and subsequently cause errors if used again by the main routine. As a rule it is recommended that simultaneous use of the table read instructions should be avoided. However, in situations where simultaneous use cannot be avoided, the interrupts should be disabled prior to the execution of any main routine table-read instructions. Note that all table related instructions require two instruction cycles to complete their operation.

Table Read Program Example

```

tempreg1 db ?      ; temporary register #1
tempreg2 db ?      ; temporary register #2
:
:
mov a,06h          ; initialise low table pointer - note that this address
                  ; is referenced
mov tblp,a         ; to the last page or the page that tbhp pointed
mov a,1Fh          ; initialise high table pointer
mov tbhp,a
:
:
tabrd tempreg1     ; transfers value in table referenced by table pointer data at
                  ; program memory address "1F06H" transferred to tempreg1 and TBLH
dec tblp           ; reduce value of table pointer by one
tabrd tempreg2     ; transfers value in table referenced by table pointer
                  ; data at program memory address "1F05H" transferred to
                  ; tempreg2 and TBLH, in this example the data "1AH" is

```

```

; transferred to tempreg1 and data "0FH" to register tempreg2
:
:
org 1F00h          ; sets initial address of program memory
dc 00Ah, 00Bh, 00Ch, 00Dh, 00Eh, 00Fh, 01Ah, 01Bh
:
:

```

In Circuit Programming – ICP

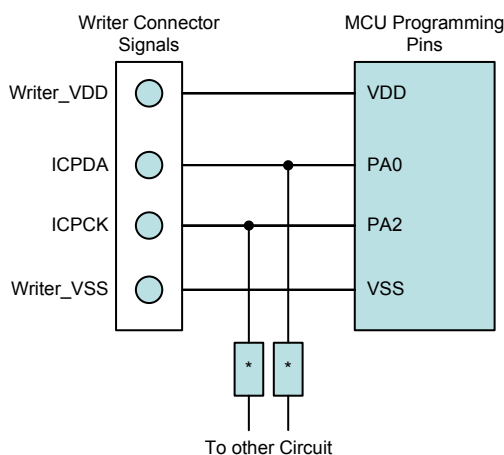
The provision of Flash type Program Memory provides the user with a means of convenient and easy upgrades and modifications to their programs on the same device.

As an additional convenience, Holtek has provided a means of programming the microcontroller in-circuit using a 4-pin interface. This provides manufacturers with the possibility of manufacturing their circuit boards complete with a programmed or un-programmed microcontroller, and then programming or upgrading the program at a later stage. This enables product manufacturers to easily keep their manufactured products supplied with the latest program releases without removal and re-insertion of the device.

Holtek Writer Pins	MCU Programming Pins	Pin Description
ICPDA	PA0	Programming Serial Data/Address
ICPCK	PA2	Programming Clock
VDD	VDD	Power Supply
VSS	VSS	Ground

The Program Memory can be programmed serially in-circuit using this 4-wire interface. Data is downloaded and uploaded serially on a single pin with an additional line for the clock. Two additional lines are required for the power supply. The technical details regarding the in-circuit programming of the device is beyond the scope of this document and will be supplied in supplementary literature.

During the programming process, the user must take care of the ICPDA and ICPCK pins for data and clock programming purposes to ensure that no other outputs are connected to these two pins.



Note: * may be resistor or capacitor. The resistance of * must be greater than 1kΩ or the capacitance of * must be less than 1nF.

On-Chip Debug Support – OCDS

There is an EV chip named BS86DHV12C which is used to emulate the real MCU device named BS86DH12C. The EV chip device also provides an “On-Chip Debug” function to debug the real MCU device during the development process. The EV chip and the real MCU device are almost functionally compatible except for “On-Chip Debug” function. Users can use the EV chip device to emulate the real chip device behavior by connecting the OCDSDA and OCDSCK pins to the Holtek HT-IDE development tools. The OCDSDA pin is the OCDS Data/Address input/output pin while the OCDSCK pin is the OCDS clock input pin. When users use the EV chip device for debugging, the corresponding pin functions shared with the OCDSDA and OCDSCK pins in the real MCU device will have no effect in the EV chip. However, the two OCDS pins which are pin-shared with the ICP programming pins are still used as the Flash Memory programming pins for ICP. For more detailed OCDS information, refer to the corresponding document named “Holtek e-Link for 8-bit MCU OCDS User’s Guide”.

Holtek e-Link Pins	EV Chip OCDS Pins	Pin Description
OCDSDA	OCDSDA	On-Chip Debug Support Data/Address input/output
OCDSCK	OCDSCK	On-Chip Debug Support Clock input
VDD	VDD	Power Supply
VSS	VSS	Ground

Data Memory

The Data Memory is a volatile area of 8-bit wide RAM internal memory and is the location where temporary information is stored.

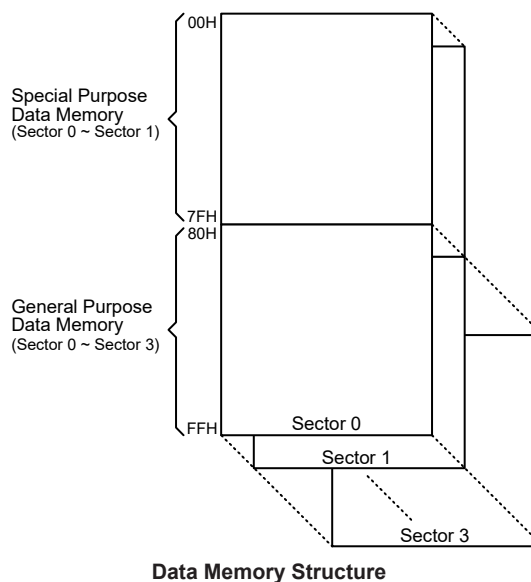
Categorized into two types, the first of these is an area of RAM where special function registers are located. These registers have fixed locations and are necessary for correct operation of the device. Many of these registers can be read from and written to directly under program control, however, some remain protected from user manipulation. The second area of Data Memory is reserved for general purpose use. All locations within this area are read and write accessible under program control.

Structure

The overall Data Memory is subdivided into several sectors, all of which are implemented in 8-bit wide RAM. Each of the Data Memory Sector is categorized into two types, the special Purpose Data Memory and the General Purpose Data Memory. The address range of the Special Purpose Data Memory for the device is from 00H to 7FH while the General Purpose Data Memory address range is from 80H to FFH. Switching between the different Data Memory sectors is achieved by setting the Memory Pointers to the correct value if using the indirect addressing method.

Special PurposeData Memory	General PurposeData Memory	
Located Sectors	Capacity	Sector: Address
Sector 0, Sector 1	512×8	Sector 0: 80H~FFH Sector 1: 80H~FFH Sector 2: 80H~FFH Sector 3: 80H~FFH

Data Memory Summary



Data Memory Addressing

For this device that supports the extended instructions, there is no Bank Pointer for Data Memory addressing. For Data Memory the desired Sector is pointed by the MP1H or MP2H register and the certain Data Memory address in the selected sector is specified by the MP1L or MP2L register when using indirect addressing access.

Direct Addressing can be used in all sectors using the extended instruction which can address all available data memory space. For the accessed data memory which is located in any data memory sectors except sector 0, the extended instructions can be used to access the data memory instead of using the indirect addressing access. The main difference between standard instructions and extended instructions is that the data memory address “m” in the extended instructions has 10 valid bits for the device, the high byte indicates a sector and the low byte indicates a specific address.

General Purpose Data Memory

All microcontroller programs require an area of read/write memory where temporary data can be stored and retrieved for use later. It is this area of RAM memory that is known as General Purpose Data Memory. This area of Data Memory is fully accessible by the user programing for both reading and writing operations. By using the bit operation instructions individual bits can be set or reset under program control giving the user a large range of flexibility for bit manipulation in the Data Memory.

Special Purpose Data Memory

This area of Data Memory is where registers, necessary for the correct operation of the microcontroller, are stored. Most of the registers are both readable and writeable but some are protected and are readable only, the details of which are located under the relevant Special Function Register section. Note that for locations that are unused, any read instruction to these addresses will return the value “00H”.

	Sector 0	Sector 1		Sector 0	Sector 1
00H	IAR0	PAS0	40H	PD0M	EEC
01H	MP0	PAS1	41H	PWRDET	
02H	IAR1	PBS0	42H	RSTC	
03H	MP1L	PBS1	43H		
04H	MP1H	PCS0	44H	TKTMR	
05H	ACC		45H	TKC0	
06H	PCL		46H	TK16DL	
07H	TBLP		47H	TK16DH	
08H	TBLH		48H	TKC1	
09H	TBHP		49H	TKM016DL	
0AH	STATUS		4AH	TKM016DH	
0BH			4BH	TKM0ROL	
0CH	IAR2		4CH	TKM0ROH	
0DH	MP2L		4DH	TKM0C0	
0EH	MP2H		4EH	TKM0C1	
0FH	RSTFC		4FH	TKM116DL	
10H	INTC0		50H	TKM116DH	
11H	INTC1		51H	TKM1ROL	
12H	INTC2		52H	TKM1ROH	
13H	INTC3		53H	TKM1C0	
14H	PA		54H	TKM1C1	
15H	PAC		55H	TKM216DL	
16H	PAPU		56H	TKM216DH	
17H	PAWU		57H	TKM2ROL	
18H	SLEDC0		58H	TKM2ROH	
19H	SLEDC1		59H	TKM2C0	
1AH	WDTA		5AH	TKM2C1	
1BH	TBC		5BH		
1CH	PSCR		5CH		
1DH	LVRC		5DH		
1EH	EEA		5EH		
1FH	EED		5FH		
20H	PB	SLEDCOM0	60H		
21H	PBC	SLEDCOM1	61H	CTM0C0	
22H	PBPV	SLEDCOM2	62H	CTM0C1	
23H	IICC0		63H	CTM0DL	
24H	IICC1		64H	CTM0DH	
25H	IICD		65H	CTM0AL	
26H	IICA		66H	CTM0AH	
27H	IICTOC		67H	PTMC0	
28H	USR		68H	PTMC1	
29H	UCR1		69H	PTMDL	
2AH	UCR2		6AH	PTMDH	
2BH	TXR_RXR		6BH	PTMAL	
2CH	BRG		6CH	PTMAH	
2DH	SADOL		6DH	PTMRPL	
2EH	SADOH		6EH	PTMRPH	
2FH	SADC0		6FH	CTM1C0	
30H	SADC1		70H	CTM1C1	
31H	INTEG		71H	CTM1DL	
32H			72H	CTM1DH	
33H	IFS0		73H	CTM1AL	
34H	IFS1		74H	CTM1AH	
35H	LVDC		75H		
36H	SCC		76H		
37H	HIRCC		77H	OVPC0	
38H	LXTC		78H	OVPC1	
39H	PC		79H	OVPC2	
3AH	PCC		7AH	OVDA	
3BH	PCPU		7BH	OCPC0	
3CH	MF10		7CH	OCPC1	
3DH	MF11		7DH	OCDA	
3EH	PD		7EH	OCPOCAL	
3FH	PDC		7FH	OCPCCAL	

 : Unused, read as 00H

Special Purpose Data Memory Structure

Special Function Register Description

Most of the Special Function Register details will be described in the relevant functional sections. However, several registers require a separate description in this section.

Indirect Addressing Registers – IAR0, IAR1, IAR2

The Indirect Addressing Registers, IAR0, IAR1 and IAR2, although having their locations in normal RAM register space, do not actually physically exist as normal registers. The method of indirect addressing for RAM data manipulation uses these Indirect Addressing Registers and Memory Pointers, in contrast to direct memory addressing, where the actual memory address is specified. Actions on the IAR0, IAR1 and IAR2 registers will result in no actual read or write operation to these registers but rather to the memory location specified by their corresponding Memory Pointers, MP0, MP1L/MP1H or MP2L/MP2H. Acting as a pair, IAR0 and MP0 can together access data only from Sector 0 while the IAR1 register together with the MP1L/MP1H register pair and IAR2 register together with the MP2L/MP2H register pair can access data from any Data Memory Sector. As the Indirect Addressing Registers are not physically implemented, reading the Indirect Addressing Registers will return a result of “00H” and writing to the registers will result in no operation.

Memory Pointers – MP0, MP1L/MP1H, MP2L/MP2H

Five Memory Pointers, known as MP0, MP1L, MP1H, MP2L, MP2H, are provided. These Memory Pointers are physically implemented in the Data Memory and can be manipulated in the same way as normal registers providing a convenient way with which to address and track data. When any operation to the relevant Indirect Addressing Registers is carried out, the actual address that the microcontroller is directed to is the address specified by the related Memory Pointer. MP0, together with Indirect Addressing Register, IAR0, are used to access data from Sector 0, while MP1L/MP1H together with IAR1 and MP2L/MP2H together with IAR2 are used to access data from all sectors according to the corresponding MP1H or MP2H register. Direct Addressing can be used in all sectors using the extended instruction which can address all available Data Memory space.

The following example shows how to clear a section of four Data Memory locations already defined as locations adres1 to adres4.

Indirect Addressing Program Example 1

```
data .section 'data'
adres1 db ?
adres2 db ?
adres3 db ?
adres4 db ?
block db ?
code .section at 0 'code'
org 00h
start:
    mov a, 04h                ; set size of block
    mov block, a
    mov a, offset adres1      ; Accumulator loaded with first RAM address
    mov mp0, a                ; set memory pointer with first RAM address
loop:
    clr IAR0                  ; clear the data at address defined by MP0
    inc mp0                   ; increase memory pointer
    sdz block                  ; check if last memory location has been cleared
    jmp loop
continue:
```

Indirect Addressing Program Example 2

```
data .section 'data'
adres1 db ?
adres2 db ?
adres3 db ?
adres4 db ?
block db ?
code .section at 0 'code'
org 00h
start:
    mov a, 04h            ; set size of block
    mov block, a
    mov a, 01h            ; set the memory sector
    mov mplh, a
    mov a, offset adres1  ; Accumulator loaded with first RAM address
    mov mp1l, a           ; set memory pointer with first RAM address
loop:
    clr IAR1              ; clear the data at address defined by MP1L
    inc mp1l               ; increase memory pointer MP1L
    sdz block              ; check if last memory location has been cleared
    jmp loop
continue:
```

The important point to note here is that in the example shown above, no reference is made to specific Data Memory addresses.

Direct Addressing Program Example using extended instructions

```
data .section 'data'
temp db ?
code .section at 0 'code'
org 00h
start:
    lmov a, [m]            ; move [m] data to acc
    lsub a, [m+1]          ; compare [m] and [m+1] data
    snz c                  ; [m]>[m+1]?
    jmp continue           ; no
    lmov a, [m]            ; yes, exchange [m] and [m+1] data
    mov temp, a
    lmov a, [m+1]
    lmov [m], a
    mov a, temp
    lmov [m+1], a
continue:
```

Note: here “m” is a data memory address located in any data memory sectors. For example, m=1F0H, it indicates address 0F0H in Sector 1.

Accumulator – ACC

The Accumulator is central to the operation of any microcontroller and is closely related with operations carried out by the ALU. The Accumulator is the place where all intermediate results from the ALU are stored. Without the Accumulator it would be necessary to write the result of each calculation or logical operation such as addition, subtraction, shift, etc., to the Data Memory resulting in higher programming and timing overheads. Data transfer operations usually involve the temporary storage function of the Accumulator; for example, when transferring data between one user-defined register and another, it is necessary to do this by passing the data through the Accumulator as no direct transfer between two registers is permitted.

Program Counter Low Register – PCL

To provide additional program control functions, the low byte of the Program Counter is made accessible to programmers by locating it within the Special Purpose area of the Data Memory. By manipulating this register, direct jumps to other program locations are easily implemented. Loading a value directly into this PCL register will cause a jump to the specified Program Memory location, however, as the register is only 8-bit wide, only jumps within the current Program Memory page are permitted. When such operations are used, note that a dummy cycle will be inserted.

Look-up Table Registers – TBLP, TBHP, TBLH

These three special function registers are used to control operation of the look-up table which is stored in the Program Memory. TBLP and TBHP are the table pointers and indicate the location where the table data is located. Their value must be set before any table read commands are executed. Their value can be changed, for example using the “INC” or “DEC” instructions, allowing for easy table data pointing and reading. TBLH is the location where the high order byte of the table data is stored after a table read data instruction has been executed. Note that the lower order table data byte is transferred to a user defined location.

Status Register – STATUS

This 8-bit register contains the SC flag, CZ flag, zero flag (Z), carry flag (C), auxiliary carry flag (AC), overflow flag (OV), power down flag (PDF), and watchdog time-out flag (TO). These arithmetic/logical operation and system management flags are used to record the status and operation of the microcontroller.

With the exception of the TO and PDF flags, bits in the status register can be altered by instructions like most other registers. Any data written into the status register will not change the TO or PDF flag. In addition, operations related to the status register may give different results due to the different instruction operations. The TO flag can be affected only by a system power-up, a WDT time-out or by executing the “CLR WDT” or “HALT” instruction. The PDF flag is affected only by executing the “HALT” or “CLR WDT” instruction or during a system power-up.

The Z, OV, AC, C, SC and CZ flags generally reflect the status of the latest operations.

- C is set if an operation results in a carry during an addition operation or if a borrow does not take place during a subtraction operation; otherwise C is cleared. C is also affected by a rotate through carry instruction.
- AC is set if an operation results in a carry out of the low nibbles in addition, or no borrow from the high nibble into the low nibble in subtraction; otherwise AC is cleared.
- Z is set if the result of an arithmetic or logical operation is zero; otherwise Z is cleared.
- OV is set if an operation results in a carry into the highest-order bit but not a carry out of the highest-order bit, or vice versa; otherwise OV is cleared.
- PDF is cleared by a system power-up or executing the “CLR WDT” instruction. PDF is set by executing the “HALT” instruction.
- TO is cleared by a system power-up or executing the “CLR WDT” or “HALT” instruction. TO is set by a WDT time-out.
- CZ is the operational result of different flags for different instructions. Refer to register definitions for more details.
- SC is the result of the “XOR” operation which is performed by the OV flag and the MSB of the current instruction operation result.

In addition, on entering an interrupt sequence or executing a subroutine call, the status register will not be pushed onto the stack automatically. If the contents of the status register are important and if the subroutine can corrupt the status register, precautions must be taken to correctly save it.

• **STATUS Register**

Bit	7	6	5	4	3	2	1	0
Name	SC	CZ	TO	PDF	OV	Z	AC	C
R/W	R/W	R/W	R	R	R/W	R/W	R/W	R/W
POR	x	x	0	0	x	x	x	x

"x": unknown

- Bit 7 **SC**: The result of the "XOR" operation which is performed by the OV flag and the MSB of the instruction operation result.
- Bit 6 **CZ**: The operational result of different flags for different instructions.
For SUB/SUBM/LSUB/LSUBM instructions, the CZ flag is equal to the Z flag.
For SBC/SBCM/LSBC/LSBCM instructions, the CZ flag is the "AND" operation result which is performed by the previous operation CZ flag and current operation zero flag.
For other instructions, the CZ flag will not be affected.
- Bit 5 **TO**: Watchdog Time-out flag
0: After power up or executing the "CLR WDT" or "HALT" instruction
1: A watchdog time-out occurred.
- Bit 4 **PDF**: Power down flag
0: After power up or executing the "CLR WDT" instruction
1: By executing the "HALT" instruction
- Bit 3 **OV**: Overflow flag
0: No overflow
1: An operation results in a carry into the highest-order bit but not a carry out of the highest-order bit or vice versa.
- Bit 2 **Z**: Zero flag
0: The result of an arithmetic or logical operation is not zero
1: The result of an arithmetic or logical operation is zero
- Bit 1 **AC**: Auxiliary flag
0: No auxiliary carry
1: An operation results in a carry out of the low nibbles in addition, or no borrow from the high nibble into the low nibble in subtraction
- Bit 0 **C**: Carry flag
0: No carry-out
1: An operation results in a carry during an addition operation or if a borrow does not take place during a subtraction operation
The "C" flag is also affected by a rotate through carry instruction.

EEPROM Data Memory

The device contains an area of internal EEPROM Data Memory. EEPROM is by its nature a non-volatile form of re-programmable memory, with data retention even when its power supply is removed. By incorporating this kind of data memory, a whole new host of application possibilities are made available to the designer. The availability of EEPROM storage allows information such as product identification numbers, calibration values, specific user data, system setup data or other product information to be stored directly within the product microcontroller. The process of reading and writing data to the EEPROM memory has been reduced to a very trivial affair.

Capacity	Address
64×8	00H~3FH

EEPROM Data Memory Structure

The EEPROM Data Memory capacity is 64×8 bits for the device. Unlike the Program Memory and RAM Data Memory, the EEPROM Data Memory is not directly mapped into memory space and is therefore not directly addressable in the same way as the other types of memory. Read and Write operations to the EEPROM are carried out in single byte operations using an address and a data register in Sector 0 and a single control register in Sector 1.

EEPROM Registers

Three registers control the overall operation of the internal EEPROM Data Memory. These are the address register, EEA, the data register, EED and a single control register, EEC. As both the EEA and EED registers are located in Sector 0, they can be directly accessed in the same way as any other Special Function Register. The EEC register however, being located in only Sector 1, can only be read from or written to indirectly using the MP1L/MP1H or MP2L/MP2H Memory Pointer and Indirect Addressing Register, IAR1/IAR2. As it is located at address 40H in Sector 1, the MP1L or MP2L Memory Pointer must first be set to the value 40H and the MP1H or MP2H Memory Pointer high byte set to the value, 01H, before any operations on the EEC register are executed.

Register Name	Bit							
	7	6	5	4	3	2	1	0
EEA	—	—	EEA5	EEA4	EEA3	EEA2	EEA1	EEA0
EED	D7	D6	D5	D4	D3	D2	D1	D0
EEC	—	—	—	—	WREN	WR	RDEN	RD

EEPROM Register List

• EEA Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	EEA5	EEA4	EEA3	EEA2	EEA1	EEA0
R/W	—	—	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	—	0	0	0	0	0	0

Bit 7~6 Unimplemented, read as “0”

Bit 5~0 **EEA5~EEA0**: Data EEPROM address bit 5 ~ bit 0

• **EED Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: Data EEPROM data bit 7 ~ bit 0

• **EEC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	WREN	WR	RDEN	RD
R/W	—	—	—	—	R/W	R/W	R/W	R/W
POR	—	—	—	—	0	0	0	0

Bit 7~4 Unimplemented, read as “0”

Bit 3 **WREN**: Data EEPROM Write Enable

0: Disable

1: Enable

This is the Data EEPROM Write Enable Bit which must be set high before Data EEPROM write operations are carried out. Clearing this bit to zero will inhibit Data EEPROM write operations.

Bit 2 **WR**: EEPROM Write Control

0: Write cycle has finished

1: Activate a write cycle

This is the Data EEPROM Write Control Bit and when set high by the application program will activate a write cycle. This bit will be automatically reset to zero by the hardware after the write cycle has finished. Setting this bit high will have no effect if the WREN has not first been set high.

Bit 1 **RDEN**: Data EEPROM Read Enable

0: Disable

1: Enable

This is the Data EEPROM Read Enable Bit which must be set high before Data EEPROM read operations are carried out. Clearing this bit to zero will inhibit Data EEPROM read operations.

Bit 0 **RD**: EEPROM Read Control

0: Read cycle has finished

1: Activate a read cycle

This is the Data EEPROM Read Control Bit and when set high by the application program will activate a read cycle. This bit will be automatically reset to zero by the hardware after the read cycle has finished. Setting this bit high will have no effect if the RDEN has not first been set high.

Note: 1. The WREN, WR, RDEN and RD bits cannot be set high at the same time in one instruction.

The WR and RD bits cannot be set high at the same time.

2. Ensure that the f_{SUB} clock is stable before executing the write operation.

3. Ensure that the write operation is totally complete before changing the EEC register content.

Reading Data from the EEPROM

To read data from the EEPROM, the EEPROM address of the data to be read must first be placed in the EEA register. The read enable bit, RDEN, in the EEC register must then be set high to enable the read function. If the RD bit in the EEC register is now set high, a read cycle will be initiated. Setting the RD bit high will not initiate a read operation if the RDEN bit has not been set. When the read cycle terminates, the RD bit will be automatically cleared to zero, after which the data can be read from

the EED register. The data will remain in the EED register until another read or write operation is executed. The application program can poll the RD bit to determine when the data is valid for reading.

Writing Data to the EEPROM

To write data to the EEPROM, the EEPROM address of the data to be written must first be placed in the EEA register and the data placed in the EED register. To initiate a write cycle the write enable bit, WREN, in the EEC register must first be set high to enable the write function. After this, the WR bit in the EEC register must be immediately set high to initiate a write cycle successfully. These two instructions must be executed in two consecutive instruction cycles. The global interrupt bit EMI should also first be cleared before implementing any write operations, and then set high again after the write cycle has started. Note that setting the WR bit high will not initiate a write cycle if the WREN bit has not been set. As the EEPROM write cycle is controlled using an internal timer whose operation is asynchronous to microcontroller system clock, a certain time will elapse before the data will have been written into the EEPROM. Detecting when the write cycle has finished can be implemented either by polling the WR bit in the EEC register or by using the EEPROM interrupt. When the write cycle terminates, the WR bit will be automatically cleared to zero by the microcontroller, informing the user that the data has been written to the EEPROM. The application program can therefore poll the WR bit to determine when the write cycle has ended.

Write Protection

Protection against inadvertent write operation is provided in several ways. After the device is powered-on the Write Enable bit in the control register will be cleared preventing any write operations. Also at power-on the Memory Pointer high byte register, MP1H or MP2H, will be reset to zero, which means that Data Memory Sector 0 will be selected. As the EEPROM control register is located in Sector 1, this adds a further measure of protection against spurious write operations. During normal program operation, ensuring that the Write Enable bit in the control register is cleared will safeguard against incorrect write operations.

EEPROM Interrupt

The EEPROM write interrupt is generated when an EEPROM write cycle has ended. The EEPROM interrupt must first be enabled by setting the DEE bit in the relevant interrupt register. When an EEPROM write cycle ends, the DEF request flag will be set. If the global, EEPROM interrupts are enabled and the stack is not full, a jump to the associated EEPROM Interrupt vector will take place. When the interrupt is serviced, the EEPROM Interrupt flag, DEF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts. More details can be obtained in the Interrupt section.

Programming Considerations

Care must be taken that data is not inadvertently written to the EEPROM. Protection can be enhanced by ensuring that the Write Enable bit is normally cleared to zero when not writing. Also the Memory Pointer high byte register, MP1H or MP2H, could be normally cleared to zero as this would inhibit access to Sector 1 where the EEPROM control register exists. Although certainly not necessary, consideration might be given in the application program to the checking of the validity of new write data by a simple read back process.

When writing data the WR bit must be set high immediately after the WREN bit has been set high, to ensure the write cycle executes correctly. The global interrupt bit EMI should also be cleared before a write cycle is executed and then re-enabled after the write cycle starts. Note that the device should not enter the IDLE or SLEEP mode until the EEPROM read or write operation is totally complete. Otherwise, the EEPROM read or write operation will fail.

Programming Examples

Reading data from the EEPROM – polling method

```
MOV A, EEPROM_ADRES      ; user defined address
MOV EEA, A
MOV A, 040H               ; set memory pointer MP1L
MOV MP1L, A               ; MP1L points to EEC register
MOV A, 01H                ; set memory pointer MP1H
MOV MP1H, A
SET IAR1.1                ; set RDEN bit, enable read operations
SET IAR1.0                ; start Read Cycle - set RD bit
BACK:
SZ IAR1.0                 ; check for read cycle end
JMP BACK
CLR IAR1                  ; disable EEPROM read if no more read operations are required
CLR MP1H
MOV A, EED                ; move read data to register
MOV READ_DATA, A
```

Note: For each read operation, the address register should be re-specified followed by setting the RD bit high to activate a read cycle even if the target address is consecutive.

Writing Data to the EEPROM – polling method

```
MOV A, EEPROM_ADRES      ; user defined address
MOV EEA, A
MOV A, EEPROM_DATA       ; user defined data
MOV EED, A
MOV A, 040H               ; set memory pointer MP1L
MOV MP1L, A               ; MP1L points to EEC register
MOV A, 01H                ; set memory pointer MP1H
MOV MP1H, A
CLR EMI
SET IAR1.3                ; set WREN bit, enable write operations
SET IAR1.2                ; start Write Cycle - set WR bit - executed immediately
                        ; after setting WREN bit

SET EMI
BACK:
SZ IAR1.2                 ; check for write cycle end
JMP BACK
CLR MP1H
```

Oscillators

Various oscillator types offer the user a wide range of functions according to their various application requirements. The flexible features of the oscillator functions ensure that the best optimisation can be achieved in terms of speed and power saving. Oscillator selections and operation are selected through a combination of configuration options and relevant control registers.

Oscillator Overview

In addition to being the source of the main system clock the oscillators also provide clock sources for the Watchdog Timer and Time Base Interrupts. External oscillator requiring some external components and fully integrated internal oscillators requiring no external components, are provided to form a wide range of both fast and slow system oscillators. The higher frequency oscillators provide higher performance but carry with it the disadvantage of higher power requirements, while the opposite is of course true for the lower frequency oscillators. With the capability of dynamically switching between fast and slow system clock, the device has the flexibility to optimize the performance/power ratio, a feature especially important in power sensitive portable applications.

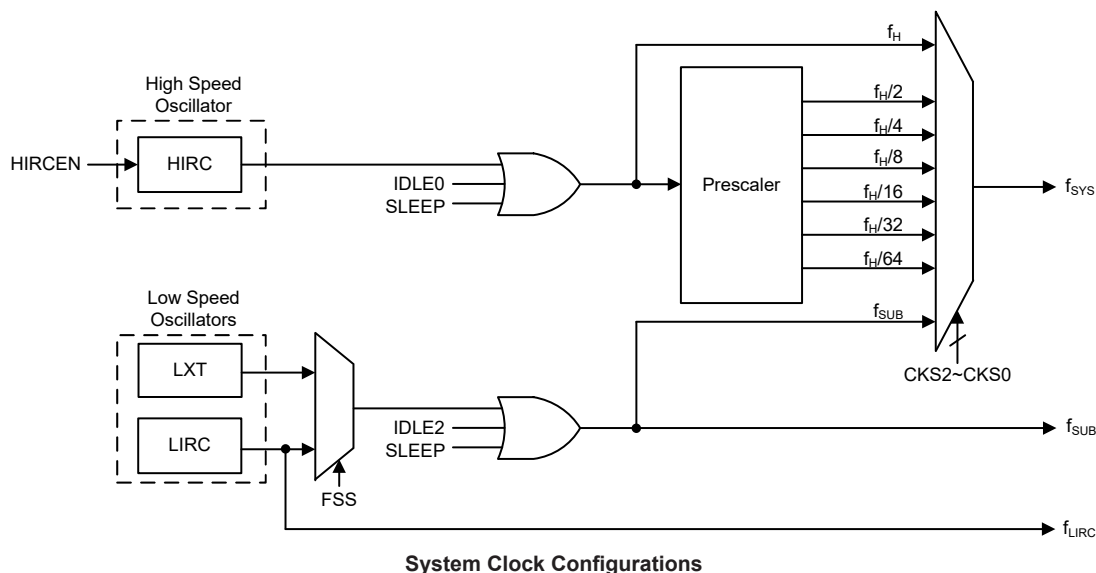
Type	Name	Frequency	Pins
Internal High Speed RC	HIRC	8/12/16MHz	—
Internal Low Speed RC	LIRC	32kHz	—
External Low Speed Crystal	LXT	32.768kHz	XT1/XT2

Oscillator Types

System Clock Configurations

There are three methods of generating the system clock, one high speed oscillator and two low speed oscillators. The high speed oscillator is the internal 8/12/16MHz RC oscillator, HIRC. The two low speed oscillators are the internal 32kHz RC oscillator, LIRC, and the external 32.768kHz crystal oscillator, LXT. Selecting whether the low or high speed oscillator is used as the system oscillator is implemented using the CKS2~CKS0 bits in the SCC register and as the system clock can be dynamically selected.

The actual source clock used for the low speed oscillators is chosen via the FSS bit in the SCC register. The frequency of the slow speed or high speed system clock is determined using the CKS2~CKS0 bits in the SCC register. Note that two oscillator selections must be made namely one high speed and one low speed system oscillators. It is not possible to choose a no-oscillator selection for either the high or low speed oscillator.



Internal High Speed RC Oscillator – HIRC

The internal RC oscillator is a fully integrated system oscillator requiring no external components. The internal RC oscillator has three fixed frequencies of 8MHz, 12MHz and 16MHz, which is selected using a configuration option. The HIRC1~HIRC0 bits in the HIRCC register must also be configured to match the selected configuration option frequency. Setting up these bits is necessary to ensure that the HIRC frequency accuracy specified in the A.C. Characteristics is achieved. Device trimming during the manufacturing process and the inclusion of internal frequency compensation circuits are used to ensure that the influence of the power supply voltage, temperature and process variations on the oscillation frequency are minimised. Note that this internal system clock option requires no external pins for its operation.

Internal 32kHz Oscillator – LIRC

The Internal 32 kHz System Oscillator is one of the low frequency oscillator choices, which is selected by the FSS bit in the SCC register. It is a fully integrated RC oscillator with a typical frequency of 32kHz, requiring no external components for its implementation. Device trimming during the manufacturing process and the inclusion of internal frequency compensation circuits are used to ensure that the influence of the power supply voltage, temperature and process variations on the oscillation frequency are minimised.

External 32.768 kHz Crystal Oscillator – LXT

The External 32.768 kHz Crystal System Oscillator is one of the low frequency oscillator choices, which is selected by the FSS bit in the SCC register. This clock source has a fixed frequency of 32.768 kHz and requires a 32.768 kHz crystal to be connected between pins XT1 and XT2. The external resistor and capacitor components connected to the 32.768 kHz crystal are necessary to provide oscillation. For applications where precise frequencies are essential, these components may be required to provide frequency compensation due to different crystal manufacturing tolerances. During power-up there is a time delay associated with the LXT oscillator waiting for it to start-up.

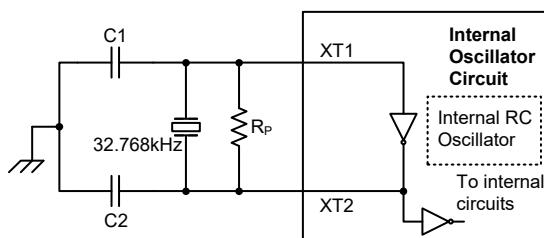
When the microcontroller enters the SLEEP or IDLE Mode, the system clock is switched off to stop microcontroller activity and to conserve power. However, in many microcontroller applications it may be necessary to keep the internal timers operational even when the microcontroller is in the SLEEP or IDLE Mode. To do this, another clock, independent of the system clock, must be provided.

However, for some crystals, to ensure oscillation and accurate frequency generation, it is necessary to add two small value external capacitors, C1 and C2. The exact values of C1 and C2 should be selected in consultation with the crystal or resonator manufacturer's specification. The external parallel feedback resistor, R_p, is required.

The pin-shared function selection bits determine if the XT1/XT2 pins are used for the LXT oscillator or as I/O or other pin-shared functions.

- If the LXT oscillator is not used for any clock source, the XT1/XT2 pins can be used as normal I/O or other pin-shared functions.
- If the LXT oscillator is used for any clock source, the 32.768 kHz crystal should be connected to the XT1/XT2 pins.

For oscillator stability and to minimise the effects of noise and crosstalk, it is important to ensure that the crystal and any associated resistors and capacitors along with interconnecting lines are all located as close to the MCU as possible.



Note: 1. R_p, C1 and C2 are required.
2. Although not shown pins have a parasitic capacitance of around 7pF.

External LXT Oscillator

LXT Oscillator C1 and C2 Values		
Crystal Frequency	C1	C2
32.768kHz	10pF	10pF

Note: 1. C1 and C2 values are for guidance only.
2. R_p=5M~10MΩ is recommended.

32.768kHz Crystal Recommended Capacitor Values

LXT Oscillator Low Power Function

The LXT oscillator can function in one of two modes, the Quick Start Mode and the Low Power Mode. The mode selection is executed using the LXTSP bit in the LXTC register.

LXTSP	LXT Operating Mode
0	Low Power
1	Quick Start

Setting the LXTSP bit high will enable the LXT Quick Start mode. In the Quick Start Mode the LXT oscillator will power up and stabilise quickly. However, after the LXT oscillator has fully powered up it can be placed into the Low power mode by clearing the LXTSP bit to zero. The oscillator will continue to run but with reduced current consumption, as the higher current consumption is only required during the LXT oscillator start-up. It is important to note that the LXT operating mode switching must be properly controlled before the LXT oscillator clock is selected as the system clock source. Once the LXT oscillator clock is selected as the system clock source using the CKS bit field and FSS bit in the SCC register, the LXT oscillator operating mode can not be changed.

It should be noted that, no matter what condition the LXTSP bit is set to, the LXT oscillator will always function normally. The only difference is that it will take more time to start up if in the Low-power mode.

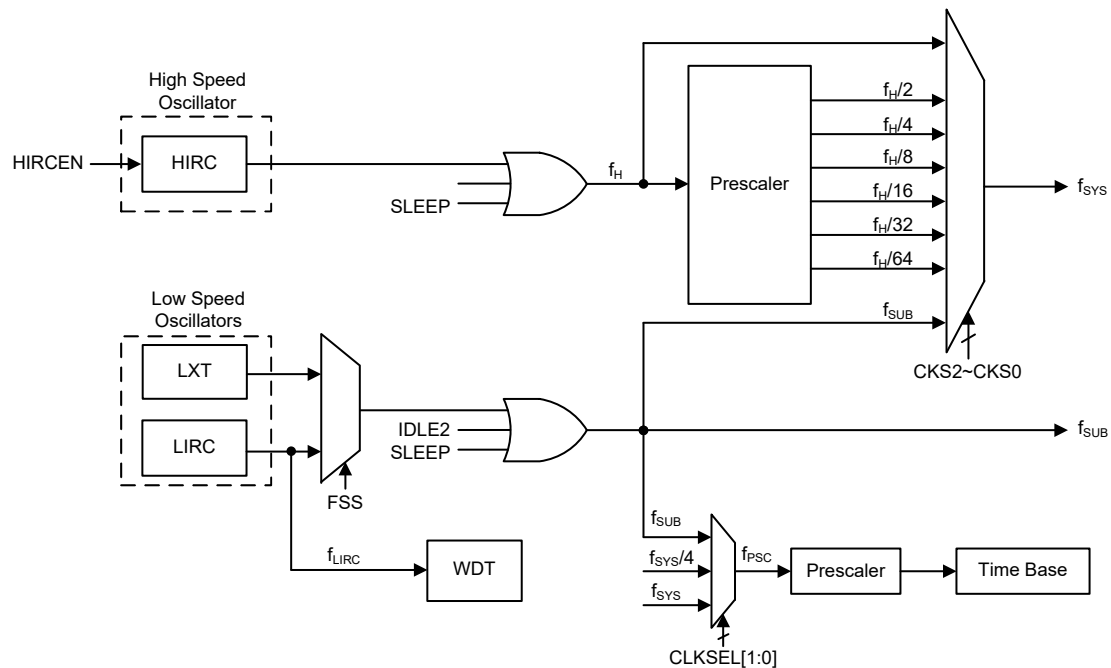
Operating Modes and System Clocks

Present day applications require that their microcontrollers have high performance but often still demand that they consume as little power as possible, conflicting requirements that are especially true in battery powered portable applications. The fast clocks required for high performance will by their nature increase current consumption and of course vice versa, lower speed clocks reduce current consumption. As Holtek has provided the device with both high and low speed clock sources and the means to switch between them dynamically, the user can optimise the operation of their microcontroller to achieve the best performance/power ratio.

System Clocks

The device has many different clock sources for both the CPU and peripheral function operation. By providing the user with a wide range of clock options using register programming, a clock system can be configured to obtain maximum application performance.

The main system clock can come from a high frequency f_H or low frequency f_{SUB} source, and is selected using the CKS2~CKS0 bits in the SCC register. The high speed system clock is sourced from the HIRC oscillator. The low speed system clock source can be sourced from the internal clock f_{SUB} . If f_{SUB} is selected then it can be sourced by either the LXT or LIRC oscillator, selected via configuring the FSS bit in the SCC register. The other choice, which is a divided version of the high speed system oscillator has a range of $f_H/2 \sim f_H/64$.



Device Clock Configurations

Note: When the system clock source f_{SYS} is switched to f_{SUB} from f_H , the high speed oscillator can be stopped to conserve the power or continue to oscillate to provide the clock source, $f_H \sim f_H/64$, for peripheral circuits to use, which is determined by configuring the corresponding high speed oscillator enable control bit.

System Operation Modes

There are six different modes of operation for the microcontroller, each one with its own special characteristics and which can be chosen according to the specific performance and power requirements of the application. There are two modes allowing normal operation of the microcontroller, the FAST Mode and SLOW Mode. The remaining four modes, the SLEEP, IDLE0, IDLE1 and IDLE2 Mode are used when the microcontroller CPU is switched off to conserve power.

Operation Mode	CPU	Register Setting			f _{sys}	f _H	f _{SUB}	f _{LIRC}
		FHIDEN	FSIDEN	CKS2~CKS0				
FAST	On	x	x	000~110	f _H ~f _H /64	On	On	On
SLOW	On	x	x	111	f _{SUB}	On/Off ⁽¹⁾	On	On
IDLE0	Off	0	1	000~110	Off	Off	On	On
				111	On			
IDLE1	Off	1	1	xxx	On	On	On	On
IDLE2	Off	1	0	000~110	On	On	Off	On
				111	Off			
SLEEP	Off	0	0	xxx	Off	Off	Off	On ⁽²⁾

“x”: don't care

- Note: 1. The f_H clock will be switched on or off by configuring the corresponding oscillator enable bit in the SLOW mode.
2. In the SLEEP mode, the f_{LIRC} clock is on as the WDT function is always enabled.

FAST Mode

This is one of the main operating modes where the microcontroller has all of its functions operational and where the system clock is provided by the high speed oscillator. This mode operates allowing the microcontroller to operate normally with a clock source coming from the HIRC oscillator. The high speed oscillator will however first be divided by a ratio ranging from 1 to 64, the actual ratio being selected by the CKS2~CKS0 bits in the SCC register. Although a high speed oscillator is used, running the microcontroller at a divided clock ratio reduces the operating current.

SLOW Mode

This is also a mode where the microcontroller operates normally although now with a slower speed clock source. The clock source used will be from f_{SUB}. The f_{SUB} clock is derived from either the LIRC or LXT oscillator determined by the FSS bit in the SCC register.

SLEEP Mode

The SLEEP Mode is entered when an HALT instruction is executed and when the FHIDEN and FSIDEN bit are low. In the SLEEP mode the CPU will be stopped. The f_{SUB} clock provided to the peripheral function will also be stopped, too. However the f_{LIRC} clock continues to operate as the WDT function is enabled.

IDLE0 Mode

The IDLE0 Mode is entered when an HALT instruction is executed and when the FHIDEN bit in the SCC register is low and the FSIDEN bit in the SCC register is high. In the IDLE0 Mode the CPU will be switched off but the low speed oscillator will be on to drive some peripheral functions.

IDLE1 Mode

The IDLE1 Mode is entered when an HALT instruction is executed and when the FHIDEN bit in the SCC register is high and the FSIDEN bit in the SCC register is high. In the IDLE1 Mode the CPU will be switched off but both the high and low speed oscillators will be on to provide a clock source to keep some peripheral functions operational.

IDLE2 Mode

The IDLE2 Mode is entered when an HALT instruction is executed and when the FHIDEN bit in the SCC register is high and the FSIDEN bit in the SCC register is low. In the IDLE2 Mode the CPU will be switched off but the high speed oscillator will be on to provide a clock source to keep some peripheral functions operational.

Control Registers

The registers, SCC, HIRCC and LXTC, are used to control the system clock and the corresponding oscillator configurations.

Register Name	Bit							
	7	6	5	4	3	2	1	0
SCC	CKS2	CKS1	CKS0	—	—	FSS	FHIDEN	FSIDEN
HIRCC	—	—	—	—	HIRC1	HIRC0	HIRCF	HIRCEN
LXTC	—	—	—	—	—	LXTSP	LXTF	LXTEN

System Operating Mode Control Register List

• SCC Register

Bit	7	6	5	4	3	2	1	0
Name	CKS2	CKS1	CKS0	—	—	FSS	FHIDEN	FSIDEN
R/W	R/W	R/W	R/W	—	—	R/W	R/W	R/W
POR	0	0	0	—	—	0	0	0

Bit 7~5 **CKS2~CKS0**: System clock selection

000: f_H
 001: $f_H/2$
 010: $f_H/4$
 011: $f_H/8$
 100: $f_H/16$
 101: $f_H/32$
 110: $f_H/64$
 111: f_{SUB}

These three bits are used to select which clock is used as the system clock source. In addition to the system clock source directly derived from f_H or f_{SUB} , a divided version of the high speed system oscillator can also be chosen as the system clock source.

Bit 4~3 Unimplemented, read as “0”

Bit 2 **FSS**: Low frequency clock selection

0: LIRC
 1: LXT

Bit 1 **FHIDEN**: High frequency oscillator control when CPU is switched off

0: Disable
 1: Enable

This bit is used to control whether the high speed oscillator is activated or stopped when the CPU is switched off by executing an “HALT” instruction.

Bit 0 **FSIDEN**: Low frequency oscillator control when CPU is switched off

0: Disable
 1: Enable

This bit is used to control whether the low speed oscillator is activated or stopped when the CPU is switched off by executing an “HALT” instruction.

• **HIRCC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	HIRC1	HIRC0	HIRCF	HIRCEN
R/W	—	—	—	—	R/W	R/W	R	R/W
POR	—	—	—	—	0	0	0	1

Bit 7~4 Unimplemented, read as “0”

Bit 3~2 **HIRC1~HIRC0**: HIRC frequency selection

00: 8MHz

01: 12MHz

10: 16MHz

11: 8MHz

When the HIRC oscillator is enabled or the HIRC frequency selection is changed by application program, the clock frequency will automatically be changed after the HIRCF flag is set to 1.

It is recommended that the HIRC frequency selected by these two bits should be the same with the frequency determined by the configuration option to achieve the HIRC frequency accuracy specified in the A.C. Characteristics.

Bit 1 **HIRCF**: HIRC oscillator stable flag

0: Unstable

1: Stable

This bit is used to indicate whether the HIRC oscillator is stable or not. When the HIRCEN bit is set to 1 to enable the HIRC oscillator, the HIRCF bit will first be cleared to 0 and then set to 1 after the HIRC oscillator is stable.

Bit 0 **HIRCEN**: HIRC oscillator enable control

0: Disable

1: Enable

• **LXTC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	LXTSP	LXTF	LXTEN
R/W	—	—	—	—	—	R/W	R	R/W
POR	—	—	—	—	—	0	0	0

Bit 7~3 Unimplemented, read as “0”

Bit 2 **LXTSP**: LXT oscillator quick start control

0: Disable – Low Power

1: Enable – Quick Start

This bit is used to control whether the LXT oscillator is operating in the low power or quick start mode. When the LXTSP bit is set to 1, the LXT oscillator will oscillate quickly but consume more power. If the LXTSP bit is cleared to 0, the LXT oscillator will consume less power but take longer time to stabilise. It is important to note that this bit can not be changed after the LXT oscillator is selected as the system clock source using the CKS2~CKS0 and FSS bits in the SCC register.

Bit 1 **LXTF**: LXT oscillator stable flag

0: Unstable

1: Stable

This bit is used to indicate whether the LXT oscillator is stable or not. When the LXTEN bit is set to 1 to enable the LXT oscillator, the LXTF bit will first be cleared to 0 and then set to 1 after the LXT oscillator is stable.

Bit 0 **LXTEN**: LXT oscillator enable control

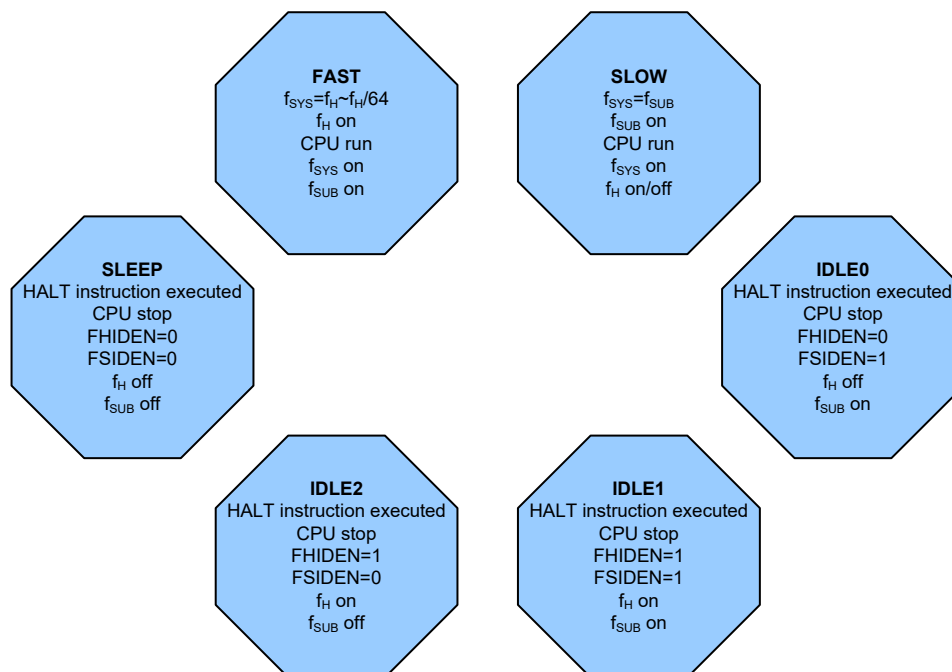
0: Disable

1: Enable

Operating Mode Switching

The device can switch between operating modes dynamically allowing the user to select the best performance/power ratio for the present task in hand. In this way microcontroller operations that do not require high performance can be executed using slower clocks thus requiring less operating current and prolonging battery life in portable applications.

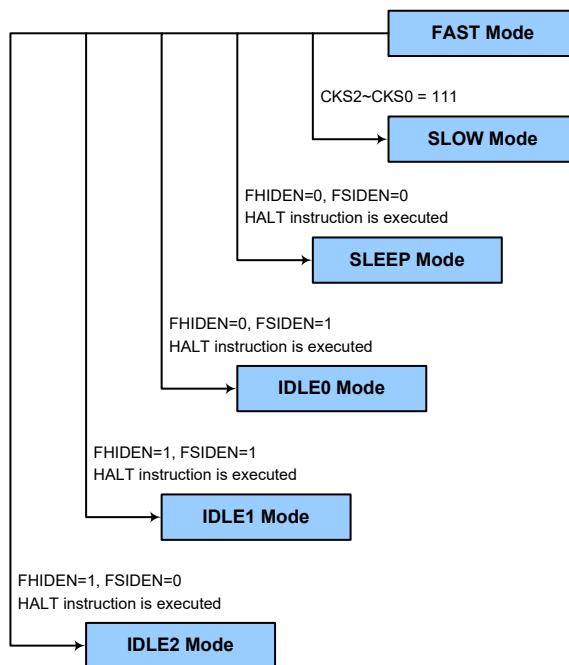
In simple terms, mode switching between the FAST Mode and SLOW Mode is executed using the CKS2~CKS0 bits in the SCC register while mode switching from the FAST/SLOW Mode to the SLEEP/IDLE Mode is executed via the HALT instruction. When an HALT instruction is executed, whether the device enters the IDLE Mode or the SLEEP Mode is determined by the condition of the FHIDEN and FSIDEN bits in the SCC register.



FAST Mode to SLOW Mode Switching

When running in the FAST Mode, which uses the high speed system oscillator, and therefore consumes more power, the system clock can switch to run in the SLOW Mode by set the CKS2~CKS0 bits to “111” in the SCC register. This will then use the low speed system oscillator which will consume less power. Users may decide to do this for certain operations which do not require high performance and can subsequently reduce power consumption.

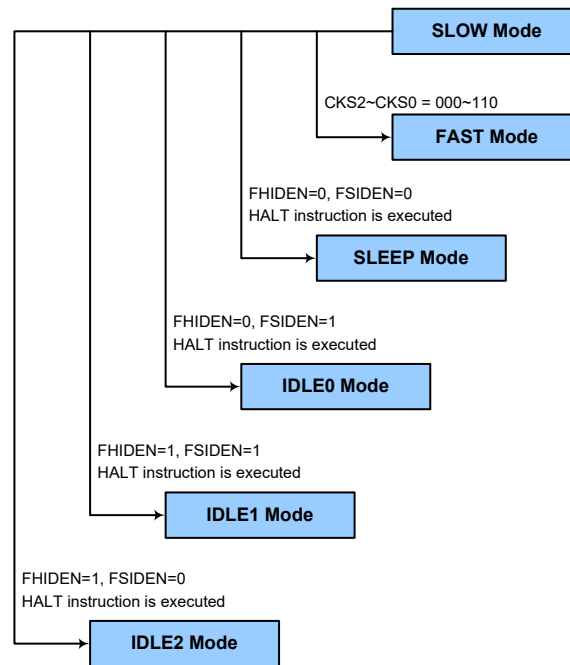
The SLOW Mode is sourced from the LXT or LIRC oscillator determined by the FSS bit in the SCC register and therefore requires the selected oscillator to be stable before full mode switching occurs.



SLOW Mode to FAST Mode Switching

In SLOW mode the system clock is derived from f_{SUB} . When system clock is switched back to the FAST mode from f_{SUB} , the CKS2~CKS0 bits should be set to “000”~“110” and then the system clock will respectively be switched to f_H ~ $f_H/64$.

However, if f_H is not used in SLOW mode and thus switched off, it will take some time to re-oscillate and stabilise when switching to the FAST mode from the SLOW Mode. This is monitored using the HIRCF bit in the HIRCC register. The time duration required for the high speed system oscillator stabilization is specified in the System Start Up Time Characteristics.



Entering the SLEEP Mode

There is only one way for the device to enter the SLEEP Mode and that is to execute the “HALT” instruction in the application program with both the FHIDEN and FSIDEN bits in the SCC register equal to “0”. When this instruction is executed under the conditions described above, the following will occur:

- The system clock will be stopped and the application program will stop at the “HALT” instruction.
- The Data Memory contents and registers will maintain their present condition.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag, PDF, will be set and the Watchdog time-out flag, TO, will be cleared.
- The WDT will be cleared and resume counting as the WDT function is always enabled.

Entering the IDLE0 Mode

There is only one way for the device to enter the IDLE0 Mode and that is to execute the “HALT” instruction in the application program with the FHIDEN bit in the SCC register equal to “0” and the FSIDEN bit in the SCC register equal to “1”. When this instruction is executed under the conditions described above, the following will occur:

- The f_H clock will be stopped and the application program will stop at the “HALT” instruction, but the f_{SUB} clock will be on.
- The Data Memory contents and registers will maintain their present condition.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag, PDF, will be set and the Watchdog time-out flag, TO, will be cleared.
- The WDT will be cleared and resume counting as the WDT function is always enabled.

Entering the IDLE1 Mode

There is only one way for the device to enter the IDLE1 Mode and that is to execute the “HALT” instruction in the application program with both the FHIDEN and FSIDEN bits in the SCC register equal to “1”. When this instruction is executed under the conditions described above, the following will occur:

- The f_H and f_{SUB} clocks will be on but the application program will stop at the “HALT” instruction.
- The Data Memory contents and registers will maintain their present condition.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag, PDF, will be set and the Watchdog time-out flag, TO, will be cleared.
- The WDT will be cleared and resume counting as the WDT function is always enabled.

Entering the IDLE2 Mode

There is only one way for the device to enter the IDLE2 Mode and that is to execute the “HALT” instruction in the application program with the FHIDEN bit in the SCC register equal to “1” and the FSIDEN bit in the SCC register equal to “0”. When this instruction is executed under the conditions described above, the following will occur:

- The f_H clock will be on but the f_{SUB} clock will be off and the application program will stop at the “HALT” instruction.
- The Data Memory contents and registers will maintain their present condition.
- The I/O ports will maintain their present conditions.
- In the status register, the Power Down flag, PDF, will be set and the Watchdog time-out flag, TO, will be cleared.
- The WDT will be cleared and resume counting as the WDT function is always enabled.

Standby Current Considerations

As the main reason for entering the SLEEP or IDLE Mode is to keep the current consumption of the device to as low a value as possible, perhaps only in the order of several micro-amps except in the IDLE1 and IDLE2 Mode, there are other considerations which must also be taken into account by the circuit designer if the power consumption is to be minimised. Special attention must be made to the I/O pins on the device. All high-impedance input pins must be connected to either a fixed high or low level as any floating input pins could create internal oscillations and result in increased current consumption. This also applies to the device which has different package types, as there may be unbonded pins. These must either be set as outputs or if set as inputs must have pull-high resistors connected.

Care must also be taken with the loads, which are connected to I/O pins, which are set as outputs. These should be placed in a condition in which minimum current is drawn or connected only to external circuits that do not draw current, such as other CMOS inputs. Also note that additional standby current will also be required if the LXT or LIRC oscillator has enabled.

In the IDLE1 and IDLE2 Mode the high speed oscillator is on and if the system clock is from the high speed system oscillator, the additional standby current will also be perhaps in the order of several hundred micro-amps.

Wake-up

To minimise power consumption the device can enter the SLEEP or any IDLE Mode, where the CPU will be switched off. However, when the device is woken up again, it will take a considerable time for the original system oscillator to restart, stabilise and allow normal operation to resume.

After the system enters the SLEEP or IDLE Mode, it can be woken up from one of various sources listed as follows:

- An external falling edge on Port A
- A system interrupt
- A WDT overflow

When the device executes the “HALT” instruction, it will enter the SLEEP or IDLE mode and the PDF flag will be set high. The PDF flag will be cleared to 0 if the device experiences a system power-up or executes the clear Watchdog Timer instruction. If the system is woken up by a WDT overflow, a Watchdog Time-out hardware reset will be initiated and the TO flag will be set to 1. The TO flag is set if a WDT time-out occurs and causes a wake-up that only resets the Program Counter and Stack Pointer, other flags remain in their original status.

Each pin on Port A can be set using the PAWU register to permit a negative transition on the pin to wake-up the system. When a pin wake-up occurs, the program will resume execution at the instruction following the “HALT” instruction. If the system is woken up by an interrupt, then two possible situations may occur. The first is where the related interrupt is disabled or the interrupt is enabled but the stack is full, in which case the program will resume execution at the instruction following the “HALT” instruction. In this situation, the interrupt which woke-up the device will not be immediately serviced, but will rather be serviced later when the related interrupt is finally enabled or when a stack level becomes free. The other situation is where the related interrupt is enabled and the stack is not full, in which case the regular interrupt response takes place. If an interrupt request flag is set high before entering the SLEEP or IDLE Mode, the wake-up function of the related interrupt will be disabled.

Watchdog Timer

The Watchdog Timer is provided to prevent program malfunctions or sequences from jumping to unknown locations, due to certain uncontrollable external events such as electrical noise.

Watchdog Timer Clock Source

The Watchdog Timer clock source is provided by the internal clock, f_{LIRC} which is sourced from the LIRC oscillator. The LIRC internal oscillator has an approximate frequency of 32kHz and this specified internal clock period can vary with V_{DD} , temperature and process variations. The Watchdog Timer source clock is then subdivided by a ratio of 2^8 to 2^{18} to give longer timeouts, the actual value being chosen using the WS2~WS0 bits in the WDTC register.

Watchdog Timer Control Register

A single register, WDTC, controls the required time-out period as well as the enable and reset MCU operation.

• WDTC Register

Bit	7	6	5	4	3	2	1	0
Name	WE4	WE3	WE2	WE1	WE0	WS2	WS1	WS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	1	0	1	0	0	1	1

Bit 7~3 **WE4~WE0**: WDT function control

10101 or 01010: Enable

Other values: Reset MCU

When these bits are changed to any other values due to environmental noise the microcontroller will be reset; this reset operation will be activated after a delay time, t_{SRESET} , and the WRF bit in the RSTFC register will be set high.

Bit 2~0 **WS2~WS0**: WDT time-out period selection

000: $2^8/f_{LIRC}$

001: $2^{10}/f_{LIRC}$

010: $2^{12}/f_{LIRC}$

011: $2^{14}/f_{LIRC}$

100: $2^{15}/f_{LIRC}$

101: $2^{16}/f_{LIRC}$

110: $2^{17}/f_{LIRC}$

111: $2^{18}/f_{LIRC}$

These three bits determine the division ratio of the Watchdog Timer source clock, which in turn determines the time-out period.

• RSTFC Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	RSTF	LVRF	LRF	WRF
R/W	—	—	—	—	R/W	R/W	R/W	R/W
POR	—	—	—	—	0	x	0	0

"x": unknown

Bit 7~4 Unimplemented, read as "0"

Bit 3 **RSTF**: Reset control register software reset flag

Described elsewhere

Bit 2 **LVRF**: LVR function reset flag

Described elsewhere

- Bit 1 **LRF**: LVRC register software reset flag
Described elsewhere
- Bit 0 **WRF**: WDTC register software reset flag
0: Not occurred
1: Occurred
- This bit is set high by the WDTC register software reset and cleared to zero by the application program. Note that this bit can only be cleared to zero by the application program.

Watchdog Timer Operation

The Watchdog Timer operates by providing a device reset when its timer overflows. This means that in the application program and during normal operation the user has to strategically clear the Watchdog Timer before it overflows to prevent the Watchdog Timer from executing a reset. This is done using the clear watchdog instruction. If the program malfunctions for whatever reason, jumps to an unknown location, or enters an endless loop, the clear instruction will not be executed in the correct manner, in which case the Watchdog Timer will overflow and reset the device. There are five bits, WE4~WE0, in the WDTC register to offer the enable control and reset control of the Watchdog Timer. The WDT function will be enabled if the WE4~WE0 bits are equal to 10101B or 01010B. If the WE4~WE0 bits are set to any other values, other than 01010B and 10101B, it will reset the device after a delay time, t_{SRESET} . After power on these bits will have a value of 01010B.

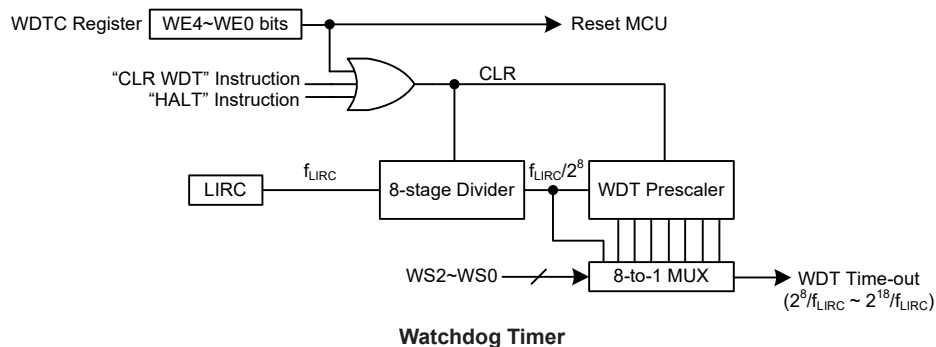
WE4~WE0 Bits	WDT Function
01010B or 10101B	Enable
Any other value	Reset MCU

Watchdog Timer Enable/Reset Control

Under normal program operation, a Watchdog Timer time-out will initialise a device reset and set the status bit TO. However, if the system is in the SLEEP or IDLE Mode, when a Watchdog Timer time-out occurs, the TO bit in the STATUS register will be set and only the Program Counter and Stack Pointer will be reset. Three methods can be adopted to clear the contents of the Watchdog Timer. The first is a WDTC register software reset, which means a certain value except 01010B and 10101B written into the WE4~WE0 bits, the second is using the Watchdog Timer software clear instruction and the third is via a HALT instruction.

There is only one method of using software instruction to clear the Watchdog Timer. That is to use the single “CLR WDT” instruction to clear the WDT.

The maximum time-out period is when the 2^{18} division ratio is selected. As an example, with a 32kHz LIRC oscillator as its source clock, this will give a maximum watchdog period of around 8s for the 2^{18} division ratio, and a minimum timeout of 8ms for the 2^8 division ratio.



Reset and Initialisation

A reset function is a fundamental part of any microcontroller ensuring that the device can be set to some predetermined condition irrespective of outside parameters. The most important reset condition is after power is first applied to the microcontroller. In this case, internal circuitry will ensure that the microcontroller, after a short delay, will be in a well-defined state and ready to execute the first program instruction. After this power-on reset, certain important internal registers will be set to defined states before the program commences. One of these registers is the Program Counter, which will be reset to zero forcing the microcontroller to begin program execution from the lowest Program Memory address.

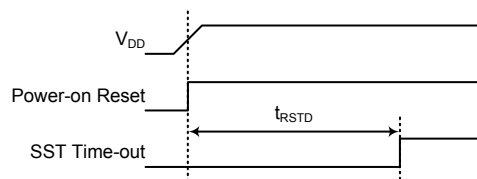
In addition to the power-on reset, another reset exists in the form of a Low Voltage Reset, LVR, where a full reset is implemented in situations where the power supply voltage falls below a certain threshold. Another type of reset is when the Watchdog Timer overflows and resets the microcontroller. All types of reset operations result in different register conditions being set.

Reset Functions

There are several ways in which a microcontroller reset can occur, through events occurring internally.

Power-on Reset

The most fundamental and unavoidable reset is the one that occurs after power is first applied to the microcontroller. As well as ensuring that the Program Memory begins execution from the first memory address, a power-on reset also ensures that certain other registers are preset to known conditions. All the I/O port and port control registers will power up in a high condition ensuring that all pins will be first set to inputs.



Power-On Reset Timing Chart

Internal Reset Control

There is an internal reset control register, RSTC, which is used to provide a reset when the device operates abnormally due to the environmental noise interference. If the content of the RSTC register is set to any value other than 01010101B or 10101010B, it will reset the device after a delay time, t_{SRESET} . After power on the register will have a value of 01010101B.

RSTC7~RSTC0 Bits	Reset Function
01010101B	No operation
10101010B	No operation
Any other value	Reset MCU

Internal Reset Function Control

• **RSTC Register**

Bit	7	6	5	4	3	2	1	0
Name	RSTC7	RSTC6	RSTC5	RSTC4	RSTC3	RSTC2	RSTC1	RSTC0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	1	0	1	0	1	0	1

Bit 7~0 **RSTC7~RSTC0**: Reset function control

01010101: No operation

10101010: No operation

Other values: Reset MCU

If these bits are changed due to adverse environmental conditions, the microcontroller will be reset. The reset operation will be activated after a delay time, t_{SRESET} , and the RSTF bit in the RSTFC register will be set to 1.

• **RSTFC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	RSTF	LVRF	LRF	WRF
R/W	—	—	—	—	R/W	R/W	R/W	R/W
POR	—	—	—	—	0	x	0	0

"x": unknown

Bit 7~4 Unimplemented, read as "0"

Bit 3 **RSTF**: RSTC register software reset flag

0: Not occurred

1: Occurred

This bit is set to 1 by the RSTC control register software reset and cleared to zero by the application program. Note that this bit can only be cleared to zero by the application program.

Bit 2 **LVRF**: LVR function reset flag

Described elsewhere

Bit 1 **LRF**: LVRC register software reset flag

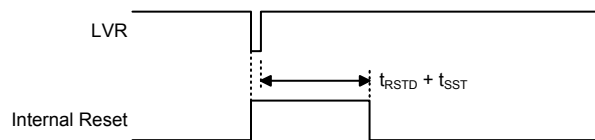
Described elsewhere

Bit 0 **WRF**: WDTC register software reset flag

Described elsewhere

Low Voltage Reset – LVR

The microcontroller contains a low voltage reset circuit in order to monitor the supply voltage of the device. The LVR function is always enabled in the FAST and SLOW modes with a specific LVR voltage, V_{LVR} . If the supply voltage of the device drops to within a range of $0.9V \sim V_{LVR}$ such as might occur when changing the battery, the LVR will automatically reset the device internally and the LVRF bit in the RSTFC register will also be set to 1. For a valid LVR signal, a low supply voltage, i.e., a voltage in the range between $0.9V \sim V_{LVR}$ must exist for a time greater than that specified by t_{LVR} in the LVD/LVR Electrical Characteristics. If the low supply voltage state does not exceed this value, the LVR will ignore the low supply voltage and will not perform a reset function. The actual V_{LVR} value can be selected by the LVS bits in the LVRC register. If the LVS7~LVS0 bits have any other value, which may perhaps occur due to adverse environmental conditions such as noise, the LVR will reset the device after a delay time, t_{SRESET} . When this happens, the LRF bit in the RSTFC register will be set to 1. Note that the LVR function will be automatically disabled when the device enters the SLEEP/IDLE mode.


Low Voltage Reset Timing Chart
• LVRC Register

Bit	7	6	5	4	3	2	1	0
Name	LVS7	LVS6	LVS5	LVS4	LVS3	LVS2	LVS1	LVS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	1	0	1	0	1	0	1

Bit 7~0 **LVS7~LVS0**: LVR voltage selection

01010101: 2.1V

00110011: 2.55V

10011001: 3.15V

10101010: 3.8V

Other values: Generates a MCU reset – register is reset to POR value

When an actual low voltage condition occurs, as specified by the LVR voltage value above, an MCU reset will be generated. The reset operation will be activated after the low voltage condition keeps for greater than a t_{LVR} time. In this situation the register contents will remain the same after such a reset occurs.

Any register value, other than the four defined register values above, will also result in the generation of an MCU reset. The reset operation will be activated after a delay time, t_{SRESET} . However in this situation the register contents will be reset to the POR value.

• RSTFC Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	RSTF	LVRF	LRF	WRF
R/W	—	—	—	—	R/W	R/W	R/W	R/W
POR	—	—	—	—	0	x	0	0

“x”: unknown

Bit 7~4 Unimplemented, read as “0”

Bit 3 **RSTF**: RSTC register software reset flag

Described elsewhere

Bit 2 **LVRF**: LVR function reset flag

0: Not occurred

1: Occurred

This bit is set to 1 when a specific low voltage reset condition occurs. Note that this bit can only be cleared to 0 by the application program.

Bit 1 **LRF**: LVR control register software reset flag

0: Not occurred

1: Occurred

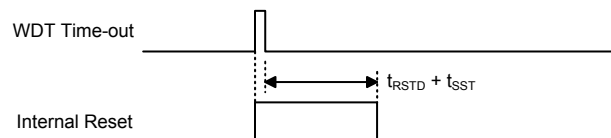
This bit is set to 1 by the LVRC control register contains any undefined LVR voltage register values. This in effect acts like a software-reset function. Note that this bit can only be cleared to 0 by the application program.

Bit 0 **WRF**: WDT control register software reset flag

Described elsewhere

Watchdog Time-out Reset during Normal Operation

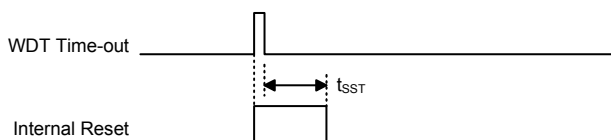
The Watchdog time-out flag TO will be set to “1” when Watchdog time-out Reset during normal operation.



WDT Time-out Reset during Normal Operation Timing Chart

Watchdog Time-out Reset during SLEEP or IDLE Mode

The Watchdog time-out Reset during SLEEP or IDLE Mode is a little different from other kinds of reset. Most of the conditions remain unchanged except that the Program Counter and the Stack Pointer will be cleared to “0” and the TO flag will be set to “1”. Refer to the System Start Up Time Characteristics for t_{SST} details.



WDT Time-out Reset during SLEEP or IDLE Mode Timing Chart

Reset Initial Conditions

The different types of reset described affect the reset flags in different ways. These flags, known as PDF and TO are located in the status register and are controlled by various microcontroller operations, such as the SLEEP or IDLE Mode function or Watchdog Timer. The reset flags are shown in the table:

TO	PDF	Reset Conditions
0	0	Power-on reset
u	u	LVR reset during FAST or SLOW Mode operation
1	u	WDT time-out reset during FAST or SLOW Mode operation
1	1	WDT time-out reset during IDLE or SLEEP Mode operation

“u” stands for unchanged

The following table indicates the way in which the various components of the microcontroller are affected after a power-on reset occurs.

Item	Condition after Reset
Program Counter	Reset to zero
Interrupts	All interrupts will be disabled
WDT, Time Bases	Cleared after reset, WDT begins counting
Timer Modules	Timer Modules will be turned off
Input/Output Ports	I/O ports will be set as inputs
Stack Pointer	Stack Pointer will point to the top of the stack

The different kinds of resets all affect the internal registers of the microcontroller in different ways. To ensure reliable continuation of normal program execution after a reset occurs, it is important to know what condition the microcontroller is in after a particular reset occurs. The following table describes how each type of reset affects the microcontroller internal registers. Note that where more than one package type exists the table will reflect the situation for the larger package type.

Register	Power On Reset	LVR Reset (Normal Operation)	WDT Time-out (Normal Operation)	WDT Time-out (IDLE/SLEEP)
IAR0	0000 0000	0000 0000	0000 0000	uuuu uuuu
MP0	0000 0000	0000 0000	0000 0000	uuuu uuuu
IAR1	0000 0000	0000 0000	0000 0000	uuuu uuuu
MP1L	0000 0000	0000 0000	0000 0000	uuuu uuuu
MP1H	0000 0000	0000 0000	0000 0000	uuuu uuuu
ACC	xxxx xxxx	uuuu uuuu	uuuu uuuu	uuuu uuuu
PCL	0000 0000	0000 0000	0000 0000	0000 0000
TBLP	xxxx xxxx	uuuu uuuu	uuuu uuuu	uuuu uuuu
TBLH	xxxx xxxx	uuuu uuuu	uuuu uuuu	uuuu uuuu
TBHP	---x xxxx	---u uuuu	---u uuuu	---u uuuu
STATUS	xx00 xxxx	uuuu uuuu	uu1u uuuu	uu11 uuuu
IAR2	0000 0000	0000 0000	0000 0000	uuuu uuuu
MP2L	0000 0000	0000 0000	0000 0000	uuuu uuuu
MP2H	0000 0000	0000 0000	0000 0000	uuuu uuuu
RSTFC	---- 0x00	---- u1uu	---- uuuu	---- uuuu
INTC0	-000 0000	-000 0000	-000 0000	-uuu uuuu
INTC1	0000 0000	0000 0000	0000 0000	uuuu uuuu
INTC2	0000 0000	0000 0000	0000 0000	uuuu uuuu
INTC3	--00 --00	--00 --00	--00 --00	--uu --uu
PA	1111 1111	1111 1111	1111 1111	uuuu uuuu
PAC	1111 1111	1111 1111	1111 1111	uuuu uuuu
PAPU	0000 0000	0000 0000	0000 0000	uuu uuuu
PAWU	0000 0000	0000 0000	0000 0000	uuu uuuu
SLEDC0	0000 0000	0000 0000	0000 0000	uuuu uuuu
SLEDC1	---- 0000	---- 0000	---- 0000	---- uuuu
WDT0	0101 0011	0101 0011	0101 0011	uuuu uuuu
TBC	0--- -000	0--- -000	0--- -000	u--- -uuu
PSCR	---- --00	---- --00	---- --00	---- --uu
LVRC	0101 0101	uuuu uuuu	0101 0101	uuuu uuuu
EEA	--00 0000	--00 0000	--00 0000	--uu uuuu
EED	0000 0000	0000 0000	0000 0000	uuuu uuuu
PB	1111 1111	1111 1111	1111 1111	uuuu uuuu
PBC	1111 1111	1111 1111	1111 1111	uuuu uuuu
PBPU	0000 0000	0000 0000	0000 0000	uuuu uuuu
IICC0	---- 000-	---- 000-	---- 000-	---- uuu-
IICC1	1000 0001	1000 0001	1000 0001	uuuu uuuu
IICD	xxxx xxxx	xxxx xxxx	xxxx xxxx	uuuu uuuu
IICA	0000 000-	0000 000-	0000 000-	uuuu uu--
IICTOC	0000 0000	0000 0000	0000 0000	uuuu uuuu
USR	0000 1011	0000 1011	0000 1011	uuuu uuuu
UCR1	0000 00x0	0000 00x0	0000 00x0	uuuu uuuu
UCR2	0000 0000	0000 0000	0000 0000	uuuu uuuu

Register	Power On Reset	LVR Reset (Normal Operation)	WDT Time-out (Normal Operation)	WDT Time-out (IDLE/SLEEP)
TXR_RXR	x x x x x x x x	x x x x x x x x	x x x x x x x x	u u u u u u u u
BRG	x x x x x x x x	x x x x x x x x	x x x x x x x x	u u u u u u u u
SADOL	x x x x - - - -	x x x x - - - -	x x x x - - - -	u u u u - - - (ADRFs=0)
				u u u u u u u u (ADRFs=1)
SADOH	x x x x x x x x	x x x x x x x x	x x x x x x x x	u u u u u u u u (ADRFs=0)
				- - - - u u u u (ADRFs=1)
SADC0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	u u u u u u u u
SADC1	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	u u u u u u u u
INTEG	- - - - - 0 0	- - - - - 0 0	- - - - - 0 0	- - - - - u u
IFS0	- - 0 0 0 0 0 0	- - 0 0 0 0 0 0	- - 0 0 0 0 0 0	- - u u u u u u
IFS1	- - - - 0 0 0 0	- - - - 0 0 0 0	- - - - 0 0 0 0	- - - - u u u u
LVDC	- - 0 0 0 0 0 0	- - 0 0 0 0 0 0	- - 0 0 0 0 0 0	- - u u u u u u
SCC	0 0 0 - 0 0 0 0	0 0 0 - 0 0 0 0	0 0 0 - 0 0 0 0	u u u u - u u u
HIRCC	- - - - 0 0 0 1	- - - - 0 0 0 1	- - - - 0 0 0 1	- - - - u u u u
LXTC	- - - - - 0 0 0	- - - - - 0 0 0	- - - - - 0 0 0	- - - - - u u u
PC	- - 1 1 1 1 1 1	- - 1 1 1 1 1 1	- - 1 1 1 1 1 1	- - u u u u u u
PCC	- - 1 1 1 1 1 1	- - 1 1 1 1 1 1	- - 1 1 1 1 1 1	- - u u u u u u
PCPU	- - 0 0 0 0 0 0	- - 0 0 0 0 0 0	- - 0 0 0 0 0 0	- - u u u u u u
MFIO	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	u u u u u u u u
MF11	- - 0 0 - - 0 0	- - 0 0 - - 0 0	- - 0 0 - - 0 0	- - u u - - u u
PD	- - 1 1 1 1 1 1	- - 1 1 1 1 1 1	- - 1 1 1 1 1 1	- - u u u u u u
PDC	- - 1 1 1 1 1 1	- - 1 1 1 1 1 1	- - 1 1 1 1 1 1	- - u u u u u u
PDOM	- - 0 0 0 0 0 0	- - 0 0 0 0 0 0	- - 0 0 0 0 0 0	- - u u u u u u
PWRDET	x - - - - 0	u - - - - 0	u - - - - 0	u - - - - u
RSTC	0 1 0 1 0 1 0 1	0 1 0 1 0 1 0 1	0 1 0 1 0 1 0 1	u u u u u u u u
TKTMR	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	u u u u u u u u
TKC0	- 0 0 0 0 0 0 0	- 0 0 0 0 0 0 0	- 0 0 0 0 0 0 0	- u u u u u u u
TK16DL	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	u u u u u u u u
TK16DH	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	u u u u u u u u
TKC1	- - - - - 1 1	- - - - - 1 1	- - - - - 1 1	- - - - - u u
TKM016DL	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	u u u u u u u u
TKM016DH	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	u u u u u u u u
TKM0ROL	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	u u u u u u u u
TKM0ROH	- - - - - 0 0	- - - - - 0 0	- - - - - 0 0	- - - - - u u
TKM0C0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	u u u u u u u u
TKM0C1	0 - 0 0 0 0 0 0	0 - 0 0 0 0 0 0	0 - 0 0 0 0 0 0	u - u u u u u u
TKM116DL	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	u u u u u u u u
TKM116DH	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	u u u u u u u u
TKM1ROL	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	u u u u u u u u
TKM1ROH	- - - - - 0 0	- - - - - 0 0	- - - - - 0 0	- - - - - u u
TKM1C0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	u u u u u u u u
TKM1C1	0 - 0 0 0 0 0 0	0 - 0 0 0 0 0 0	0 - 0 0 0 0 0 0	u - u u u u u u
TKM216DL	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	u u u u u u u u
TKM216DH	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	0 0 0 0 0 0 0 0	u u u u u u u u

Register	Power On Reset	LVR Reset (Normal Operation)	WDT Time-out (Normal Operation)	WDT Time-out (IDLE/SLEEP)
TKM2ROL	0000 0000	0000 0000	0000 0000	uuuu uuuu
TKM2ROH	---- --00	---- --00	---- --00	---- --uu
TKM2C0	0000 0000	0000 0000	0000 0000	uuuu uuuu
TKM2C1	0-00 0000	0-00 0000	0-00 0000	u-uu uuuu
CTM0C0	0000 0000	0000 0000	0000 0000	uuuu uuuu
CTM0C1	0000 0000	0000 0000	0000 0000	uuuu uuuu
CTM0DL	0000 0000	0000 0000	0000 0000	uuuu uuuu
CTM0DH	---- --00	---- --00	---- --00	---- --uu
CTM0AL	0000 0000	0000 0000	0000 0000	uuuu uuuu
CTM0AH	---- --00	---- --00	---- --00	---- --uu
PTMC0	0000 0---	0000 0---	0000 0---	uuuu u---
PTMC1	0000 0000	0000 0000	0000 0000	uuuu uuuu
PTMDL	0000 0000	0000 0000	0000 0000	uuuu uuuu
PTMDH	---- --00	---- --00	---- --00	---- --uu
PTMAL	0000 0000	0000 0000	0000 0000	uuuu uuuu
PTMAH	---- --00	---- --00	---- --00	---- --uu
PTMRPL	0000 0000	0000 0000	0000 0000	uuuu uuuu
PTMRPH	---- --00	---- --00	---- --00	---- --uu
CTM1C0	0000 0000	0000 0000	0000 0000	uuuu uuuu
CTM1C1	0000 0000	0000 0000	0000 0000	uuuu uuuu
CTM1DL	0000 0000	0000 0000	0000 0000	uuuu uuuu
CTM1DH	---- --00	---- --00	---- --00	---- --uu
CTM1AL	0000 0000	0000 0000	0000 0000	uuuu uuuu
CTM1AH	---- --00	---- --00	---- --00	---- --uu
OVPC0	000- -000	000- -000	000- -000	uuu- -uuu
OVPC1	0001 0000	0001 0000	0001 0000	uuuu uuuu
OVPC2	---- 0000	---- 0000	---- 0000	---- uuuu
OVPCA	0000 0000	0000 0000	0000 0000	uuuu uuuu
OCPC0	0000 ---0	0000 ---0	0000 ---0	uuuu ---u
OCPC1	--00 0000	--00 0000	--00 0000	--uu uuuu
OCPCA	0000 0000	0000 0000	0000 0000	uuuu uuuu
OCPOCAL	0010 0000	0010 0000	0010 0000	uuuu uuuu
OCPCCAL	0001 0000	0001 0000	0001 0000	uuuu uuuu
PAS0	0000 0000	0000 0000	0000 0000	uuuu uuuu
PAS1	0000 0000	0000 0000	0000 0000	uuuu uuuu
PBS0	0000 0000	0000 0000	0000 0000	uuuu uuuu
PBS1	0000 0000	0000 0000	0000 0000	uuuu uuuu
PCS0	0000 0000	0000 0000	0000 0000	uuuu uuuu
SLEDCOM0	0000 0000	0000 0000	0000 0000	uuuu uuuu
SLEDCOM1	0000 0000	0000 0000	0000 0000	uuuu uuuu
SLEDCOM2	--00 0000	--00 0000	--00 0000	--uu uuuu
EEC	---- 0000	---- 0000	---- 0000	---- uuuu

Note: “u” stands for unchanged
“x” stands for unknown
“-” stands for unimplemented

Input/Output Ports

Holtek microcontrollers offer considerable flexibility on their I/O ports. With the input or output designation of every pin fully under user program control, pull-high selections for all ports and wake-up selections on certain pins, the user is provided with an I/O structure to meet the needs of a wide range of application possibilities.

This device provides bidirectional input/output lines. These I/O ports are mapped to the RAM Data Memory with specific addresses as shown in the Special Purpose Data Memory table. All of these I/O ports can be used for input and output operations. For input operation, these ports are non-latching, which means the inputs must be ready at the T2 rising edge of instruction “MOV A, [m]”, where “m” denotes the port address. For output operation, all the data is latched and remains unchanged until the output latch is rewritten.

Register Name	Bit							
	7	6	5	4	3	2	1	0
PA	PA7	PA6	PA5	PA4	PA3	PA2	PA1	PA0
PAC	PAC7	PAC6	PAC5	PAC4	PAC3	PAC2	PAC1	PAC0
PAPU	PAPU7	PAPU6	PAPU5	PAPU4	PAPU3	PAPU2	PAPU1	PAPU0
PAWU	PAWU7	PAWU6	PAWU5	PAWU4	PAWU3	PAWU2	PAWU1	PAWU0
PB	PB7	PB6	PB5	PB4	PB3	PB2	PB1	PB0
PBC	PBC7	PBC6	PBC5	PBC4	PBC3	PBC2	PBC1	PBC0
PBPU	PBPU7	PBPU6	PBPU5	PBPU4	PBPU3	PBPU2	PBPU1	PBPU0
PC	—	—	PC5	PC4	PC3	PC2	PC1	PC0
PCC	—	—	PCC5	PCC4	PCC3	PCC2	PCC1	PCC0
PCPU	—	—	PCPU5	PCPU4	PCPU3	PCPU2	PCPU1	PCPU0

“—”: Unimplemented, read as “0”

I/O Logic Function Register List

Pull-high Resistors

Many product applications require pull-high resistors for their switch inputs usually requiring the use of an external resistor. To eliminate the need for these external resistors, all I/O pins, when configured as a digital input have the capability of being connected to an internal pull-high resistor. These pull-high resistors are selected using the relevant pull-high control registers and are implemented using weak PMOS transistors.

Note that the pull-high resistor can be controlled by the relevant pull-high control register only when the pin-shared functional pin is selected as a digital input or NMOS output. Otherwise, the pull-high resistors cannot be enabled.

• PxPU Register

Bit	7	6	5	4	3	2	1	0
Name	PxPU7	PxPU6	PxPU5	PxPU4	PxPU3	PxPU2	PxPU1	PxPU0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

PxPU_n: I/O Port x Pin pull-high function control

0: Disable

1: Enable

The PxPU_n bit is used to control the pin pull-high function. Here the “x” is the Port name which can be A, B or C. However, the actual available bits for each I/O Port may be different.

Port A Wake-up

The HALT instruction forces the microcontroller into the SLEEP or IDLE Mode which preserves power, a feature that is important for battery and other low-power applications. Various methods exist to wake-up the microcontroller, one of which is to change the logic condition on one of the Port A pins from high to low. This function is especially suitable for applications that can be woken up via external switches. Each pin on Port A can be selected individually to have this wake-up feature using the PAWU register.

Note that the wake-up function can be controlled by the wake-up control registers only when the pin is selected as a general purpose input and the MCU enters the IDLE or SLEEP mode.

• PAWU Register

Bit	7	6	5	4	3	2	1	0
Name	PAWU7	PAWU6	PAWU5	PAWU4	PAWU3	PAWU2	PAWU1	PAWU0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **PAWU7~PAWU0**: PA7~PA0 pin Wake-up function control
 0: Disable
 1: Enable

I/O Port Control Registers

Each I/O Port has its own control register which controls the input/output configuration. With this control register, each I/O pin with or without pull-high resistors can be reconfigured dynamically under software control. For the I/O pin to function as an input, the corresponding bit of the control register must be written as a “1”. This will then allow the logic state of the input pin to be directly read by instructions. When the corresponding bit of the control register is written as a “0”, the I/O pin will be set as a CMOS output. If the pin is currently set as an output, instructions can still be used to read the output register. However, it should be noted that the program will in fact only read the status of the output data latch and not the actual logic status of the output pin.

• PxC Register

Bit	7	6	5	4	3	2	1	0
Name	PxC7	PxC6	PxC5	PxC4	PxC3	PxC2	PxC1	PxC0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	1	1	1	1	1	1	1	1

PxCn: I/O Port x Pin type selection
 0: Output
 1: Input

The PxCn bit is used to control the pin type selection. Here the “x” is the Port name which can be A, B or C. However, the actual available bits for each I/O Port may be different.

I/O Port Source Current Selection

The device supports different output source current driving capability for each I/O port. With the selection register, SLEDCn, specific I/O port can support four levels of the source current driving capability. These source current selection bits are available only when the corresponding pin is configured as a CMOS output. Otherwise, these select bits have no effect. Users should refer to the Input/Output Characteristics section to select the desired output source current for different applications.

Register Name	Bit							
	7	6	5	4	3	2	1	0
SLEDC0	SLEDC07	SLEDC06	SLEDC05	SLEDC04	SLEDC03	SLEDC02	SLEDC01	SLEDC00
SLEDC1	—	—	—	—	SLEDC13	SLEDC12	SLEDC11	SLEDC10

I/O Port Source Current Selection Register List

• **SLEDC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	SLEDC07	SLEDC06	SLEDC05	SLEDC04	SLEDC03	SLEDC02	SLEDC01	SLEDC00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7~6 **SLEDC07~SLEDC06**: PB7~PB4 source current selection
00: Source current=Level 0 (Min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (Max.)
- Bit 5~4 **SLEDC05~SLEDC04**: PB3~PB0 source current selection
00: Source current=Level 0 (Min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (Max.)
- Bit 3~2 **SLEDC03~SLEDC02**: PA7~PA4 source current selection
00: Source current=Level 0 (Min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (Max.)
- Bit 1~0 **SLEDC01~SLEDC00**: PA3~PA0 source current selection
00: Source current=Level 0 (Min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (Max.)

• **SLEDC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	SLEDC13	SLEDC12	SLEDC11	SLEDC10
R/W	—	—	—	—	R/W	R/W	R/W	R/W
POR	—	—	—	—	0	0	0	0

- Bit 7~4 Unimplemented, read as “0”
- Bit 3~2 **SLEDC13~SLEDC12**: PC5~PC4 source current selection
00: Source current=Level 0 (Min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (Max.)
- Bit 1~0 **SLEDC11~SLEDC10**: PC3~PC0 source current selection
00: Source current=Level 0 (Min.)
01: Source current=Level 1
10: Source current=Level 2
11: Source current=Level 3 (Max.)

I/O Port Sink Current Selection

The device supports different output sink current driving capability for each I/O port. With the selection register, SLEDCOMn, specific I/O port can support two levels of the sink current driving capability. These sink current selection bits are available when the corresponding pin is configured as a CMOS output. Otherwise, these select bits have no effect. Users should refer to the Input/Output Characteristics section to select the desired output sink current for different applications.

Register Name	Bit							
	7	6	5	4	3	2	1	0
SLEDCOM0	PANS7	PANS6	PANS5	PANS4	PANS3	PANS2	PANS1	PANS0
SLEDCOM1	PBNS7	PBNS6	PBNS5	PBNS4	PBNS3	PBNS2	PBNS1	PBNS0
SLEDCOM2	—	—	PCNS5	PCNS4	PCNS3	PCNS2	PCNS1	PCNS0

I/O Port Sink Current Selection Register List

• SLEDCOM0 Register

Bit	7	6	5	4	3	2	1	0
Name	PANS7	PANS6	PANS5	PANS4	PANS3	PANS2	PANS1	PANS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7 **PANS7:** PA7 sink current selection
 0: Sink current=Level 0 (Min.)
 1: Sink current=Level 1 (Max.)
- Bit 6 **PANS6:** PA6 sink current selection
 0: Sink current=Level 0 (Min.)
 1: Sink current=Level 1 (Max.)
- Bit 5 **PANS5:** PA5 sink current selection
 0: Sink current=Level 0 (Min.)
 1: Sink current=Level 1 (Max.)
- Bit 4 **PANS4:** PA4 sink current selection
 0: Sink current=Level 0 (Min.)
 1: Sink current=Level 1 (Max.)
- Bit 3 **PANS3:** PA3 sink current selection
 0: Sink current=Level 0 (Min.)
 1: Sink current=Level 1 (Max.)
- Bit 2 **PANS2:** PA2 sink current selection
 0: Sink current=Level 0 (Min.)
 1: Sink current=Level 1 (Max.)
- Bit 1 **PANS1:** PA1 sink current selection
 0: Sink current=Level 0 (Min.)
 1: Sink current=Level 1 (Max.)
- Bit 0 **PANS0:** PA0 sink current selection
 0: Sink current=Level 0 (Min.)
 1: Sink current=Level 1 (Max.)

• **SLEDCOM1 Register**

Bit	7	6	5	4	3	2	1	0
Name	PBNS7	PBNS6	PBNS5	PBNS4	PBNS3	PBNS2	PBNS1	PBNS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7 **PBNS7**: PB7 sink current selection
0: Sink current=Level 0 (Min.)
1: Sink current=Level 1 (Max.)
- Bit 6 **PBNS6**: PB6 sink current selection
0: Sink current=Level 0 (Min.)
1: Sink current=Level 1 (Max.)
- Bit 5 **PBNS5**: PB5 sink current selection
0: Sink current=Level 0 (Min.)
1: Sink current=Level 1 (Max.)
- Bit 4 **PBNS4**: PB4 sink current selection
0: Sink current=Level 0 (Min.)
1: Sink current=Level 1 (Max.)
- Bit 3 **PBNS3**: PB3 sink current selection
0: Sink current=Level 0 (Min.)
1: Sink current=Level 1 (Max.)
- Bit 2 **PBNS2**: PB2 sink current selection
0: Sink current=Level 0 (Min.)
1: Sink current=Level 1 (Max.)
- Bit 1 **PBNS1**: PB1 sink current selection
0: Sink current=Level 0 (Min.)
1: Sink current=Level 1 (Max.)
- Bit 0 **PBNS0**: PB0 sink current selection
0: Sink current=Level 0 (Min.)
1: Sink current=Level 1 (Max.)

• **SLEDCOM2 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	PCNS5	PCNS4	PCNS3	PCNS2	PCNS1	PCNS0
R/W	—	—	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	—	0	0	0	0	0	0

- Bit 7~6 Unimplemented, read as “0”
- Bit 5 **PCNS5**: PC5 sink current selection
0: Sink current=Level 0 (Min.)
1: Sink current=Level 1 (Max.)
- Bit 4 **PCNS4**: PC4 sink current selection
0: Sink current=Level 0 (Min.)
1: Sink current=Level 1 (Max.)
- Bit 3 **PCNS3**: PC3 sink current selection
0: Sink current=Level 0 (Min.)
1: Sink current=Level 1 (Max.)
- Bit 2 **PCNS2**: PC2 sink current selection
0: Sink current=Level 0 (Min.)
1: Sink current=Level 1 (Max.)
- Bit 1 **PCNS1**: PC1 sink current selection
0: Sink current=Level 0 (Min.)
1: Sink current=Level 1 (Max.)

Bit 0 **PCNS0**: PC0 sink current selection
 0: Sink current=Level 0 (Min.)
 1: Sink current=Level 1 (Max.)

Pin-shared Functions

The flexibility of the microcontroller range is greatly enhanced by the use of pins that have more than one function. Limited numbers of pins can force serious design constraints on designers but by supplying pins with multi-functions, many of these difficulties can be overcome. For these pins, the desired function of the multi-function I/O pins is selected by a series of registers via the application program control.

Pin-shared Function Selection Registers

The limited number of supplied pins in a package can impose restrictions on the amount of functions a certain device can contain. However by allowing the same pins to share several different functions and providing a means of function selection, a wide range of different functions can be incorporated into even relatively small package sizes. The device includes Port “x” Output Function Selection register “n”, labeled as PxSn, and Input Function Selection register, labeled as IFSi, which can select the desired functions of the multi-function pin-shared pins.

The most important point to note is to make sure that the desired pin-shared function is properly selected and also deselected. For most pin-shared functions, to select the desired pin-shared function, the pin-shared function should first be correctly selected using the corresponding pin-shared control register. After that the corresponding peripheral functional setting should be configured and then the peripheral function can be enabled. However, a special point must be noted for some digital input pins, such as INT, xTCKn, PTPI, etc, which share the same pin-shared control configuration with their corresponding general purpose I/O functions when setting the relevant pin-shared control bit fields. To select these pin functions, in addition to the necessary pin-shared control and peripheral functional setup aforementioned, they must also be set as an input by setting the corresponding bit in the I/O port control register. To correctly deselect the pin-shared function, the peripheral function should first be disabled and then the corresponding pin-shared function control register can be modified to select other pin-shared functions.

Register Name	Bit							
	7	6	5	4	3	2	1	0
PAS0	PAS07	PAS06	PAS05	PAS04	PAS03	PAS02	PAS01	PAS00
PAS1	PAS17	PAS16	PAS15	PAS14	PAS13	PAS12	PAS11	PAS10
PBS0	PBS07	PBS06	PBS05	PBS04	PBS03	PBS02	PBS01	PBS00
PBS1	PBS17	PBS16	PBS15	PBS14	PBS13	PBS12	PBS11	PBS10
PCS0	PCS07	PCS06	PCS05	PCS04	PCS03	PCS02	PCS01	PCS00
IFS0	—	—	SDAPS1	SDAPS0	SCLPS1	SCLPS0	RXPS1	RXPS0
IFS1	—	—	—	—	PTPIPS	PTCKPS	CTCK1PS	CTCK0PS

Pin-shared Function Selection Register List

• **PAS0 Register**

Bit	7	6	5	4	3	2	1	0
Name	PAS07	PAS06	PAS05	PAS04	PAS03	PAS02	PAS01	PAS00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7~6 **PAS07~PAS06:** PA3 pin-shared function selection
 00: PA3/PTPI
 01: SDA
 10: VREF
 11: KEY4
- Bit 5~4 **PAS05~PAS04:** PA2 pin-shared function selection
 00: PA2
 01: TX
 10: CTP1B
 11: OCPI
- Bit 3~2 **PAS03~PAS02:** PA1 pin-shared function selection
 00: PA1/PTCK
 01: SCL
 10: OVPI1
 11: KEY3
- Bit 1~0 **PAS01~PAS00:** PA0 pin-shared function selection
 00: PA0
 01: RX
 10: CTP1
 11: OCPVR

• **PAS1 Register**

Bit	7	6	5	4	3	2	1	0
Name	PAS17	PAS16	PAS15	PAS14	PAS13	PAS12	PAS11	PAS10
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7~6 **PAS17~PAS16:** PA7 pin-shared function selection
 00: PA7/CTCK0/PTCK
 01: PA7/CTCK0/PTCK
 10: PTP
 11: XT2
- Bit 5~4 **PAS15~PAS14:** PA6 pin-shared function selection
 00: PA6/CTCK1/PTPI
 01: CTP0
 10: PTPB
 11: XT1
- Bit 3~2 **PAS13~PAS12:** PA5 pin-shared function selection
 00: PA5
 01: PA5
 10: AN1
 11: KEY6
- Bit 1~0 **PAS11~PAS10:** PA4 pin-shared function selection
 00: PA4
 01: PA4
 10: AN0
 11: KEY5

• **PBS0 Register**

Bit	7	6	5	4	3	2	1	0
Name	PBS07	PBS06	PBS05	PBS04	PBS03	PBS02	PBS01	PBS00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PBS07~PBS06:** PB3 pin-shared function selection

00: PB3
01: SDA
10: AN7
11: KEY10

Bit 5~4 **PBS05~PBS04:** PB2 pin-shared function selection

00: PB2
01: SCL
10: AN6
11: KEY9

Bit 3~2 **PBS03~PBS02:** PB1 pin-shared function selection

00: PB1/CTCK1
01: TX
10: OVPI0
11: KEY2

Bit 1~0 **PBS01~PBS00:** PB0 pin-shared function selection

00: PB0/CTCK0
01: RX
10: OCPI
11: KEY1

• **PBS1 Register**

Bit	7	6	5	4	3	2	1	0
Name	PBS17	PBS16	PBS15	PBS14	PBS13	PBS12	PBS11	PBS10
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PBS17~PBS16:** PB7 pin-shared function selection

00: PB7
01: CTP1B
10: AN5
11: OVPCOUT

Bit 5~4 **PBS15~PBS14:** PB6 pin-shared function selection

00: PB6
01: CTP1
10: AN4
11: OCPAO

Bit 3~2 **PBS13~PBS12:** PB5 pin-shared function selection

00: PB5
01: CTP0B
10: AN3
11: KEY8

Bit 1~0 **PBS11~PBS10:** PB4 pin-shared function selection

00: PB4
01: CTP0
10: AN2
11: KEY7

• **PCS0 Register**

Bit	7	6	5	4	3	2	1	0
Name	PCS07	PCS06	PCS05	PCS04	PCS03	PCS02	PCS01	PCS00
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PCS07~PCS06**: PC3 pin-shared function selection

00: PC3
01: CTP0
10: SDA
11: TX

Bit 5~4 **PCS05~PCS04**: PC2 pin-shared function selection

00: PC2
01: CTP1
10: SCL
11: RX

Bit 3~2 **PCS03~PCS02**: PC1 pin-shared function selection

00: PC1/INT
01: PTPB
10: OVPI1
11: KEY12

Bit 1~0 **PCS01~PCS00**: PC0 pin-shared function selection

00: PC0
01: PTP
10: OVPI0
11: KEY11

• **IFS0 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	SDAPS1	SDAPS0	SCLPS1	SCLPS0	RXPS1	RXPS0
R/W	—	—	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	—	0	0	0	0	0	0

Bit 7~6 Unimplemented, read as “0”

Bit 5~4 **SDAPS1~SDAPS0**: SDA input source pin selection

00: PB3
01: PB3
10: PC3
11: PA3

Bit 3~2 **SCLPS1~SCLPS0**: SCL input source pin selection

00: PB2
01: PB2
10: PC2
11: PA1

Bit 1~0 **RXPS1~RXPS0**: RX input source pin selection

00: PB0
01: PB0
10: PC2
11: PA0

• IFS1 Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	PTPIPS	PTCKPS	CTCK1PS	CTCK0PS
R/W	—	—	—	—	R/W	R/W	R/W	R/W
POR	—	—	—	—	0	0	0	0

Bit 7~4 Unimplemented, read as “0”

Bit 3 **PTPIPS**: PTPI input source pin selection
 0: PA6
 1: PA3

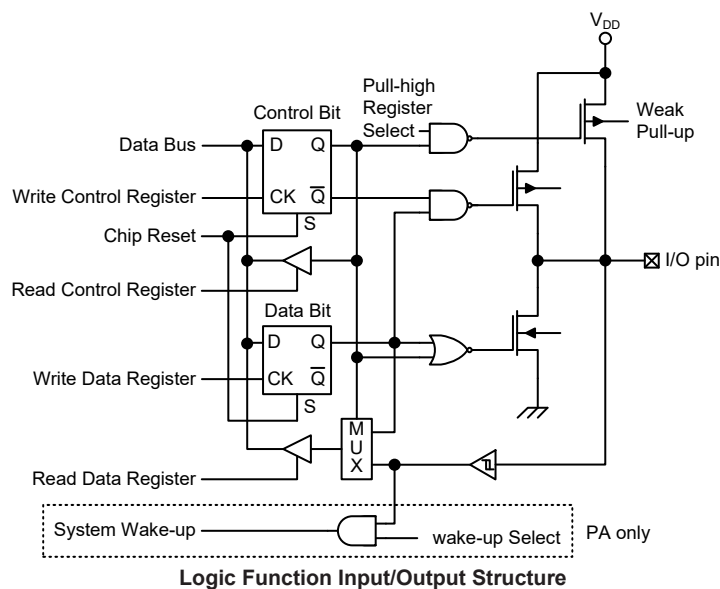
Bit 2 **PTCKPS**: PTCK input source pin selection
 0: PA1
 1: PA7

Bit 1 **CTCK1PS**: CTCK1 input source pin selection
 0: PA6
 1: PB1

Bit 0 **CTCK0PS**: CTCK0 input source pin selection
 0: PA7
 1: PB0

I/O Pin Structures

The accompanying diagram illustrates the internal structure of the I/O logic function. As the exact logical construction of the I/O pin will differ from this drawing, it is supplied as a guide only to assist with the functional understanding of the I/O logic function. The wide range of pin-shared structures does not permit all types to be shown.


Programming Considerations

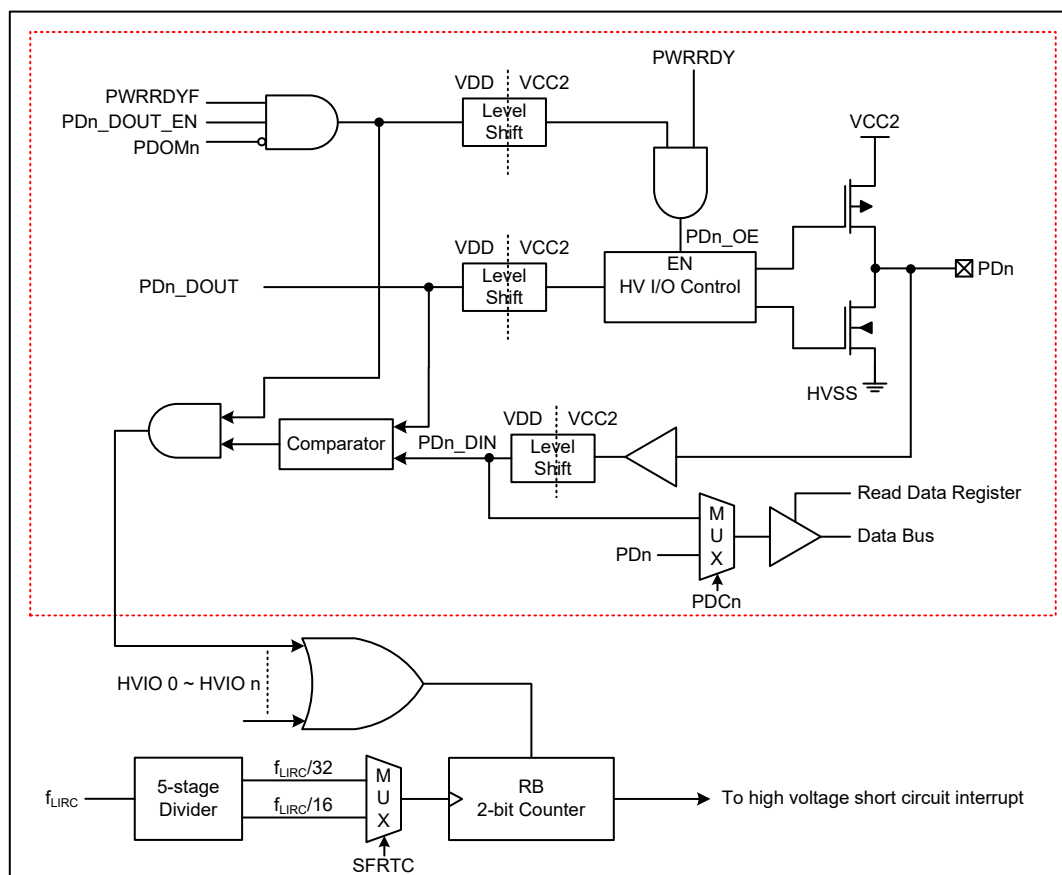
Within the user program, one of the things first to consider is port initialisation. After a reset, all of the I/O data and port control registers will be set high. This means that all I/O pins will be defaulted to an input state, the level of which depends on the other connected circuitry and whether pull-high selections have been chosen. If the port control registers are then programmed to set some pins as outputs, these output pins will have an initial high output value unless the associated port

data registers are first programmed. Selecting which pins are inputs and which are outputs can be achieved byte-wide by loading the correct values into the appropriate port control register or by programming individual bits in the port control register using the “SET [m].i” and “CLR [m].i” instructions. Note that when using these bit control instructions, a read-modify-write operation takes place. The microcontroller must first read in the data on the entire port, modify it to the required new bit values and then rewrite this data back to the output ports.

Port A has the additional capability of providing wake-up functions. When the device is in the SLEEP or IDLE Mode, various methods are available to wake the device up. One of these is a high to low transition of any of the Port A pins. Single or multiple pins on Port A can be set to have this function.

High Voltage I/O Port

The device provides several 10V high voltage input/output lines, known as PD0~PD5. These high voltage I/O ports can convert 5V logic output signals to 10V voltage outputs to directly drive TRIACs, relays, and buzzers.



High Voltage I/O Block Diagram (n=0~5)

Note: 1. The structure contained in the dash line is identical for each PDn HVIO, and the structure contained in the solid line is shared by all HVIO lines.

2. Each symbol name with a “_” sign in the figure is a circuit node name and not the Special Function Register bit.

PDn_DOUT_EN: PDn data output enable signal

PDn_DOUT: PDn output data

PDn_OE: PDn output global enable signal

PDn_DIN: PDn input data

3. When the comparison result between PDn_DOUT and PDn_DIN is different, the LIRC oscillator will be enabled by the hardware until the short circuit condition is released even if the CPU and LIRC are both off. After the LIRC clock is stable, the f_{LIRC} clock can be used as the clock source of the peripheral functions. If the MCU is still in CPU off mode, the LIRC will be turned off after the short circuit condition has been released.
4. The PDn_OE truth table is shown as follows:

PWRRDY	PWRRDYF	PD0Mn	PDn_DOUT_EN	PDn_OE
0	0	0	0	0
1	1	0	0	0
1	1	0	1	1
1	1	1	0	0
1	1	1	1	0

5. The PDn truth table is shown as follows:

PDn_OE	PDn_DOUT	PDn	PDn Mode
0	0	Floating	Input mode
0	1		
1	0	V _{SS}	Output mode (Short-circuit protection is enabled when PWRRDYF=1)
1	1	V _{CC2}	

High Voltage I/O Registers

Overall operation of high voltage I/O port is controlled using a series of registers. The PD register is the data register. The PDC register is used to select the input/output type. The PD0M register is used for output mask control. The remaining register PWRDET is used to monitor the power supply status and select short flag response time.

Register Name	Bit							
	7	6	5	4	3	2	1	0
PD	—	—	PD5	PD4	PD3	PD2	PD1	PD0
PDC	—	—	PDC5	PDC4	PDC3	PDC2	PDC1	PDC0
PD0M	—	—	PD0M5	PD0M4	PD0M3	PD0M2	PD0M1	PD0M0
PWRDET	PWRRDYF	—	—	—	—	—	—	SFRTC

High Voltage I/O Register List

• PD Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	PD5	PD4	PD3	PD2	PD1	PD0
R/W	—	—	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	—	1	1	1	1	1	1

Bit 7~6 Unimplemented, read as “0”

Bit 5~0 **PD5~PD0**: HVIO PD5~PD0 Data bit

• **PDC Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	PDC5	PDC4	PDC3	PDC2	PDC1	PDC0
R/W	—	—	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	—	1	1	1	1	1	1

Bit 7~6 Unimplemented, read as “0”

Bit 5~0 **PDC5~PDC0**: HVIO PD5~PD0 pin input/output type selection
 0: Output
 1: Input

• **PDOM Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	PDOM5	PDOM4	PDOM3	PDOM2	PDOM1	PDOM0
R/W	—	—	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	—	0	0	0	0	0	0

Bit 7~6 Unimplemented, read as “0”

Bit 5~0 **PDOM5~PDOM0**: HVIO PD5~PD0 output mask control
 0: No output mask
 1: Output mask

• **PWRDET Register**

Bit	7	6	5	4	3	2	1	0
Name	PWRRDYF	—	—	—	—	—	—	SFRTC
R/W	R	—	—	—	—	—	—	R/W
POR	x	—	—	—	—	—	—	0

“x”: unknown

Bit 7 **PWRRDYF**: V_{CC2} and V_{DD} Power ready flag
 0: Power not ready – $V_{CC2} < V_{DET1}$ or $V_{DD} < V_{DET2}$
 1: Power ready – $V_{CC2} \geq V_{DET1}$ and $V_{DD} \geq V_{DET2}$

During the power-on voltage rising process, when the MCU operates normally and the V_{DD} voltage is equal to or greater than the V_{DET2} voltage, it is not yet determined whether the V_{CC2} voltage is equal to or greater than the V_{DET1} voltage, then the PWRRDYF read value may be 0 or 1 and therefore in an unknown condition.

Bit 6~1 Unimplemented, read as “0”

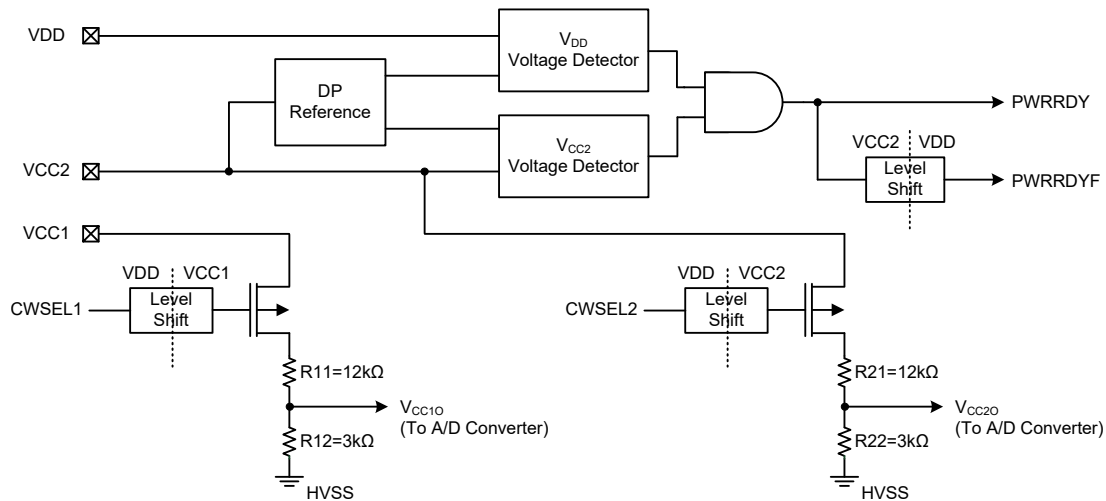
Bit 0 **SFRTC**: HVIO short flag response time selection
 0: $64 \times t_{LIRC} \sim 96 \times t_{LIRC}$
 1: $32 \times t_{LIRC} \sim 64 \times t_{LIRC}$

Voltage Detector

An internal voltage detector circuit is used to monitor the V_{CC2} and V_{DD} voltage levels. It provides a power ready flag, PWRRDYF, which can be read by the MCU to indicate the power status. If the V_{CC2} is equal to or greater than the V_{DET1} , and the V_{DD} is equal to or greater than the V_{DET2} , then PWRRDYF=1, otherwise PWRRDYF=0.

In addition, there is also a V_{CC2O} voltage output with a value of $0.2V_{CC2}$, which can be measured by the internal A/D converter for the power good detection purpose.

The VCC1 is the LDO high voltage power input pin. A power divided voltage V_{CC1O} is generated using a divider resistor, this voltage can be externally connected to the A/D converter internal input channel for measurement.



Note: The CWSEL1 and CWSEL2 are generated by the internal A/D converter input channel selection. When the A/D converter selects the VCC10 or VCC20 signal as its internal input, then CWSEL1=0 or CWSEL2=0, otherwise CWSEL1=1 or CWSEL2=1.

Voltage Detector Circuit

Short-circuit Protection Function

All high voltage I/O pins share the same short-circuit protection circuit which contains a 2-bit clock counter. The counter, with an initial value of 0, has a clock source derived from $f_{LIRC}/32$ or $f_{LIRC}/16$, which is selected by the SFRTC bit in the PWRDET register.

The high voltage I/O short-circuit protection circuit compares the output signal PDn_DOUT with the input signal, PDn_DIN. If the PDn_DOUT has the same value as the PDn_DIN, it indicates the high voltage I/O is in a normal condition, and then the clock counter will be cleared. If the comparison result of any high voltage I/O pins is different, the common clock counter will not be cleared.

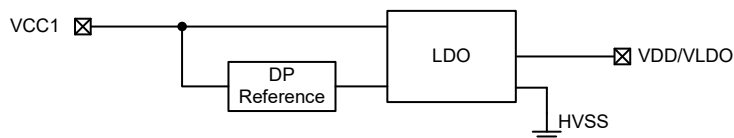
When the comparison result is different and the count value of the clock counter is more than 2~3, the short-circuit protection circuit will determine it as a short-circuit condition. After the short-circuit condition occurs, the high voltage short circuit interrupt flag, HVSCF, will be set high.

The short-circuit protection circuits use the same interrupt vector for all high voltage I/O pins. When a short-circuit situation occurs on any one of the high voltage I/O pins, a high voltage short circuit interrupt will be triggered. When this condition appears, if the global interrupt enable bit and its corresponding interrupt control bit are enabled and the stack is not full, a subroutine call to the high voltage short circuit interrupt vector will take place.

However, it must be noted that the short-circuit protection function is only available when the voltage detector detects the condition of PWRDYF=1.

Low Dropout Regulator – LDO

The device includes an internal low dropout regulator, LDO. The LDO can reduce the higher input voltage on the VCC1 pin, which ranges from 6V to 10V, to a stable 5V voltage and then output on the VLDO pin. This 5V voltage can be used as a power for external or internal circuits.



LDO Block Diagram

Timer Modules – TM

One of the most fundamental functions in any microcontroller device is the ability to control and measure time. To implement time related functions the device includes several Timer Modules, generally abbreviated to the name TM. The TMs are multi-purpose timing units and serve to provide operations such as Timer/Counter, Input Capture, Compare Match Output and Single Pulse Output as well as being the functional unit for the generation of PWM signals. Each of the TMs has two interrupts. The addition of input and output pins for each TM ensures that users are provided with timing units with a wide and flexible range of features.

The common features of the different TM types are described here with more detailed information provided in the individual Compact and Periodic TM sections.

Introduction

The device contains several TMs and each individual TM can be categorised as a certain type, namely Compact Type TM or Periodic Type TM. Although similar in nature, the different TM types vary in their feature complexity. The common features to all of the Compact and Periodic TMs will be described in this section and the detailed operation regarding each of the TM types will be described in separate sections. The main features and differences between the two types of TMs are summarised in the accompanying table.

TM Function	CTM	PTM
Timer/Counter	√	√
Input Capture	—	√
Compare Match Output	√	√
PWM Output	√	√
Single Pulse Output	—	√
PWM Alignment	Edge	Edge
PWM Adjustment Period & Duty	Duty or Period	Duty or Period

TM Function Summary

CTM0	CTM1	PTM
10-bit CTM	10-bit CTM	10-bit PTM

TM Name/Type Summary

TM Operation

The different types of TM offer a diverse range of functions, from simple timing operations to PWM signal generation. The key to understanding how the TM operates is to see it in terms of a free running count-up counter whose value is then compared with the value of pre-programmed internal

comparators. When the free running count-up counter has the same value as the pre-programmed comparator, known as a compare match situation, a TM interrupt signal will be generated which can clear the counter and perhaps also change the condition of the TM output pin. The internal TM counter is driven by a user selectable clock source, which can be an internal clock or an external pin.

TM Clock Source

The clock source which drives the main counter in each TM can originate from various sources. The selection of the required clock source is implemented using the xTnCK2~xTnCK0 bits in the xTMn control registers, where “x” stands for C or P type TM and “n” stands for the specific TM serial number. For the PTM there is no serial number “n” in the relevant pins, registers and control bits since there is only one PTM in the device. The clock source can be a ratio of the system clock, f_{SYS} , or the internal high clock, f_{IH} , the f_{SUB} clock source or the external xTCKn pin. The xTCKn pin clock source is used to allow an external signal to drive the TM as an external clock source for event counting.

TM Interrupts

The Compact or Periodic type TM each has two internal interrupt, one for each of the internal comparator A or comparator P, which generate a TM interrupt when a compare match condition occurs. When a TM interrupt is generated, it can be used to clear the counter and also to change the state of the TM output pin.

TM External Pins

Each of the TMs, irrespective of what type, has one input pin with the label xTCKn while the Periodic TM has another input pin with the label PTPI. The xTMn input pin, xTCKn, is essentially a clock source for the xTMn and is selected using the xTnCK2~xTnCK0 bits in the xTMnC0 register. This external TM input pin allows an external clock source to drive the internal TM. The xTCKn input pin can be chosen to have either a rising or falling active edge. The PTCK pin is also used as the external trigger input pin in single pulse output mode for the PTM.

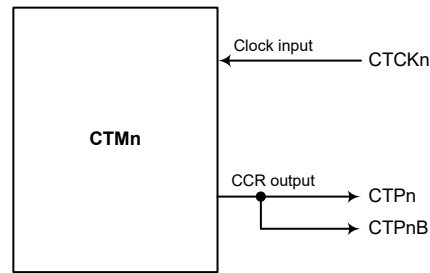
The other PTM input pin, PTPI, is the capture input whose active edge can be a rising edge, a falling edge or both rising and falling edges and the active edge transition type is selected using the PTIO1~PTIO0 bits in the PTMC1 register. There is another capture input, PTCK, for PTM capture input mode, which can be used as the external trigger input source except the PTPI pin.

The TMs each have two output pins, xTPn and xTPnB. When the TM is in the Compare Match Output Mode, these pins can be controlled by the TM to switch to a high or low level or to toggle when a compare match situation occurs. The external xTPn and xTPnB output pins are also the pins where the xTMn generates the PWM output waveform.

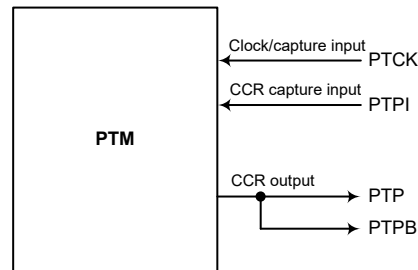
As the TM input and output pins are pin-shared with other functions, the TM input and output functions must first be configured using the relevant pin-shared function selection bits described in the Pin-shared Function section. The details of the pin-shared function selection are described in the pin-shared function section.

CTM0		CTM1		PTM	
Input	Output	Input	Output	Input	Output
CTCK0	CTP0, CTP0B	CTCK1	CTP1, CTP1B	PTCK, PTPI	PTP, PTPB

TM External Pins



CTM Function Pin Block Diagram (n=0~1)

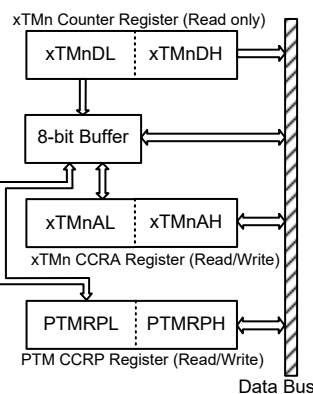


PTM Function Pin Block Diagram

Programming Considerations

The TM Counter Registers and the Capture/Compare CCRA and CCRP registers, all have a low and high byte structure. The high bytes can be directly accessed, but as the low bytes can only be accessed via an internal 8-bit buffer, reading or writing to these register pairs must be carried out in a specific way. The important point to note is that data transfer to and from the 8-bit buffer and its related low byte only takes place when a write or read operation to its corresponding high byte is executed.

As the CCRA and CCRP registers are implemented in the way shown in the following diagram and accessing these register pairs is carried out in a specific way as described above, it is recommended to use the “MOV” instruction to access the CCRA and CCRP low byte registers, named xTMnAL and PTMRPL, using the following access procedures. Accessing the CCRA or CCRP low byte registers without following these access procedures will result in unpredictable values.



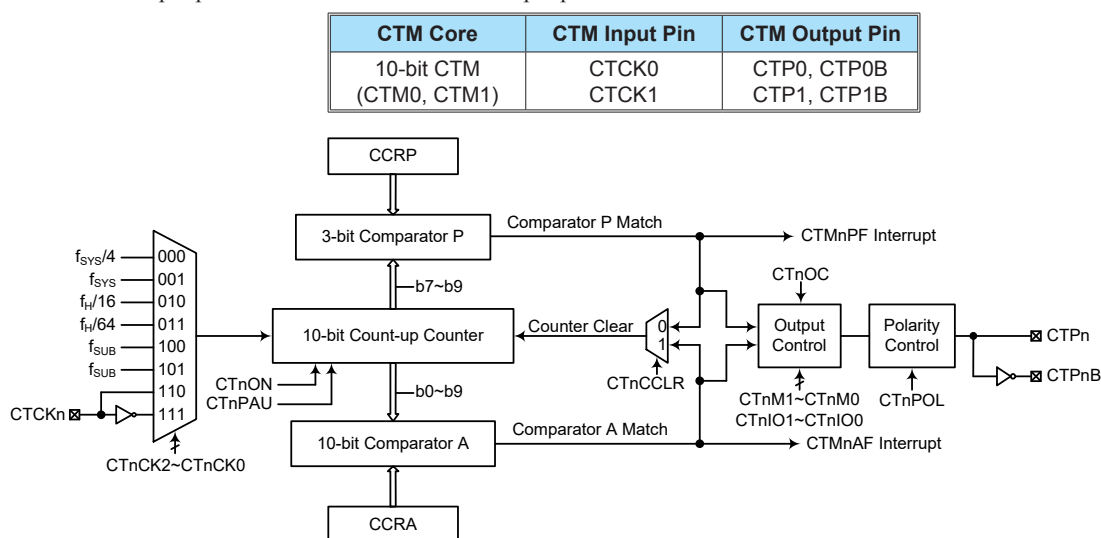
The following steps show the read and write procedures:

- Writing Data to CCRA or CCRP
 - ♦ Step 1. Write data to Low Byte xTMnAL or PTMRPL
 - Note that here data is only written to the 8-bit buffer.

- ♦ Step 2. Write data to High Byte xTMnAH or PTMRPH
 - Here data is written directly to the high byte registers and simultaneously data is latched from the 8-bit buffer to the Low Byte registers.
- ♦ Reading Data from the Counter Registers and CCRA or CCRP
 - ♦ Step 1. Read data from the High Byte xTMnDH, xTMnAH or PTMRPH
 - Here data is read directly from the High Byte registers and simultaneously data is latched from the Low Byte register into the 8-bit buffer.
 - ♦ Step 2. Read data from the Low Byte xTMnDL, xTMnAL or PTMRPL
 - This step reads data from the 8-bit buffer.

Compact Type TM – CTM

The Compact Type TM contains three operating modes, which are Compare Match Output, Timer/Event Counter and PWM Output modes. The Compact TM can also be controlled with an external input pin and can drive two external output pins.



Note: The CTMn external pins are pin-shared with other functions, so before using the CTMn function the relevant pin-shared function registers must be set properly.

Compact Type TM Block Diagram (n=0~1)

Compact Type TM Operation

The size of Compact TM is 10-bit wide and its core is a 10-bit count-up counter which is driven by a user selectable internal or external clock source. There are also two internal comparators with the names, Comparator A and Comparator P. These comparators will compare the value in the counter with CCRP and CCRA registers. The CCRP comparator is 3-bit wide whose value is compared with the highest 3 bits in the counter while the CCRA is the 10 bits and therefore compares all counter bits.

The only way of changing the value of the 10-bit counter using the application program, is to clear the counter by changing the CTnON bit from low to high. The counter will also be cleared automatically by a counter overflow or a compare match with one of its associated comparators. When these conditions occur, a CTMn interrupt signal will also usually be generated. The Compact Type TM can operate in a number of different operational modes, can be driven by different clock sources including an input pin and can also control two output pins. All operating setup conditions are selected using relevant internal registers.

Compact Type TM Register Description

Overall operation of the Compact TM is controlled using a series of registers. A read only register pair exists to store the internal counter 10-bit value, while a read/write register pair exists to store the internal 10-bit CCRA value. The remaining two registers are control registers which set the different operating and control modes as well as the three CCRP bits.

Register Name	Bit							
	7	6	5	4	3	2	1	0
CTMnC0	CTnPAU	CTnCK2	CTnCK1	CTnCK0	CTnON	CTnRP2	CTnRP1	CTnRP0
CTMnC1	CTnM1	CTnM0	CTnIO1	CTnIO0	CTnOC	CTnPOL	CTnDPX	CTnCCLR
CTMnDL	D7	D6	D5	D4	D3	D2	D1	D0
CTMnDH	—	—	—	—	—	—	D9	D8
CTMnAL	D7	D6	D5	D4	D3	D2	D1	D0
CTMnAH	—	—	—	—	—	—	D9	D8

10-bit Compact Type TM Register List (n=0~1)

• CTMnC0 Register

Bit	7	6	5	4	3	2	1	0
Name	CTnPAU	CTnCK2	CTnCK1	CTnCK0	CTnON	CTnRP2	CTnRP1	CTnRP0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 CTnPAU: CTMn counter pause control
0: Run
1: Pause

The counter can be paused by setting this bit high. Clearing the bit to zero restores normal counter operation. When in a Pause condition the CTMn will remain powered up and continue to consume power. The counter will retain its residual value when this bit changes from low to high and resume counting from this value when the bit changes to a low value again.

Bit 6~4 CTnCK2~CTnCK0: CTMn counter clock selection
000: $f_{SYS}/4$
001: f_{SYS}
010: $f_H/16$
011: $f_H/64$
100: f_{SUB}
101: f_{SUB}
110: CTCKn rising edge clock
111: CTCKn falling edge clock

These three bits are used to select the clock source for the CTMn. The external pin clock source can be chosen to be active on the rising or falling edge. The clock source f_{SYS} is the system clock, while f_H and f_{SUB} are other internal clocks, the details of which can be found in the oscillator section.

Bit 3 CTnON: CTMn counter on/off control
0: Off
1: On

This bit controls the overall on/off function of the CTMn. Setting the bit high enables the counter to run while clearing the bit disables the CTMn. Clearing this bit to zero will stop the counter from counting and turn off the CTMn which will reduce its power consumption. When the bit changes state from low to high the internal counter value will be reset to zero, however when the bit changes from high to low, the internal counter will retain its residual value until the bit returns high again.

If the CTMn is in the Compare Match Output Mode or the PWM Output Mode, then the CTMn output pin will be reset to its initial condition, as specified by the CTnOC bit, when the CTnON bit changes from low to high.

Bit 2~0 **CTnRP2~CTnRP0**: CTMn CCRP 3-bit register, compared with the CTMn counter bit 9 ~ bit 7

Comparator P match period=
0: 1024 CTMn clocks
1~7: (1~7)×128 CTMn clocks

These three bits are used to set the value on the internal CCRP 3-bit register, which are then compared with the internal counter's highest three bits. The result of this comparison can be selected to clear the internal counter if the CTnCCLR bit is set to zero. Setting the CTnCCLR bit to zero ensures that a compare match with the CCRP values will reset the internal counter. As the CCRP bits are only compared with the highest three counter bits, the compare values exist in 128 clock cycle multiples. Clearing all three bits to zero is in effect allowing the counter to overflow at its maximum value.

• **CTMnC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	CTnM1	CTnM0	CTnIO1	CTnIO0	CTnOC	CTnPOL	CTnDPX	CTnCCLR
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **CTnM1~CTnM0**: CTMn operating mode selection

00: Compare Match Output Mode
01: Undefined
10: PWM Output Mode
11: Timer/Counter Mode

These bits set the required operating mode for the CTMn. To ensure reliable operation the CTMn should be switched off before any changes are made to the CTnM1 and CTnM0 bits. In the Timer/Counter Mode, the CTMn output pin pin state is undefined.

Bit 5~4 **CTnIO1~CTnIO0**: CTMn external pin CTPn function selection

Compare Match Output Mode

00: No change
01: Output low
10: Output high
11: Toggle output

PWM Output Mode

00: PWM output inactive state
01: PWM output active state
10: PWM output
11: Undefined

Timer/Counter Mode

Unused

These two bits are used to determine how the CTMn external pin changes state when a certain condition is reached. The function that these bits select depends upon in which mode the CTMn is running.

In the Compare Match Output Mode, the CTnIO1 and CTnIO0 bits determine how the CTMn output pin changes state when a compare match occurs from the Comparator A. The CTMn output pin can be set to switch high, switch low or to toggle its present state when a compare match occurs from the Comparator A. When the bits are both zero, then no change will take place on the output. The initial value of the CTMn output pin should be set using the CTnOC bit in the CTMnC1 register. Note that the output level requested by the CTnIO1 and CTnIO0 bits must be different from the initial value setup using the CTnOC bit otherwise no change will occur on the CTMn output pin

when a compare match occurs. After the CTMn output pin changes state, it can be reset to its initial level by changing the level of the CTnON bit from low to high.

In the PWM Output Mode, the CTnIO1 and CTnIO0 bits determine how the CTMn output pin changes state when a certain compare match condition occurs. The PWM output function is modified by changing these two bits. It is necessary to only change the values of the CTnIO1 and CTnIO0 bits only after the CTMn has been switched off. Unpredictable PWM outputs will occur if the CTnIO1 and CTnIO0 bits are changed when the CTMn is running.

Bit 3 **CTnOC**: CTMn CTPn output control

Compare Match Output Mode

0: Initial low

1: Initial high

PWM Output Mode/Single Pulse Output Mode

0: Active low

1: Active high

This is the output control bit for the CTMn output pin. Its operation depends upon whether CTMn is being used in the Compare Match Output Mode or in the PWM Output Mode. It has no effect if the CTMn is in the Timer/Counter Mode. In the Compare Match Output Mode it determines the logic level of the CTMn output pin before a compare match occurs. In the PWM Output Mode it determines if the PWM signal is active high or active low.

Bit 2 **CTnPOL**: CTMn CTPn output polarity control

0: Non-invert

1: Invert

This bit controls the polarity of the CTPn output pin. When the bit is set high the CTMn output pin will be inverted and not inverted when the bit is zero. It has no effect if the CTMn is in the Timer/Counter Mode.

Bit 1 **CTnDPX**: CTMn PWM duty/period control

0: CCRP – period; CCRA – duty

1: CCRP – duty; CCRA – period

This bit determines which of the CCRA and CCRP registers are used for period and duty control of the PWM waveform.

Bit 0 **CTnCCLR**: CTMn counter clear condition selection

0: Comparator P match

1: Comparator A match

This bit is used to select the method which clears the counter. Remember that the CTMn contains two comparators, Comparator A and Comparator P, either of which can be selected to clear the internal counter. With the CTnCCLR bit set high, the counter will be cleared when a compare match occurs from the Comparator A. When the bit is low, the counter will be cleared when a compare match occurs from the Comparator P or with a counter overflow. A counter overflow clearing method can only be implemented if the CCRP bits are all cleared to zero. The CTnCCLR bit is not used in the PWM Output mode.

• CTMnDL Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R	R	R	R	R	R	R	R
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: CTMn Counter Low Byte Register bit 7 ~ bit 0

CTMn 10-bit Counter bit 7 ~ bit 0

• **CTMnDH Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	D9	D8
R/W	—	—	—	—	—	—	R	R
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as “0”

Bit 1~0 **D9~D8**: CTMn Counter High Byte Register bit 1 ~ bit 0
CTMn 10-bit Counter bit 9 ~ bit 8

• **CTMnAL Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: CTMn CCRA Low Byte Register bit 7 ~ bit 0
CTMn 10-bit CCRA bit 7 ~ bit 0

• **CTMnAH Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	D9	D8
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as “0”

Bit 1~0 **D9~D8**: CTMn CCRA High Byte Register bit 7 ~ bit 0
CTMn 10-bit CCRA bit 9 ~ bit 8

Compact Type TM Operation Modes

The Compact Type TM can operate in one of three operating modes, Compare Match Output Mode, PWM Output Mode or Timer/Counter Mode. The operating mode is selected using the CTnM1 and CTnM0 bits in the CTMnC1 register.

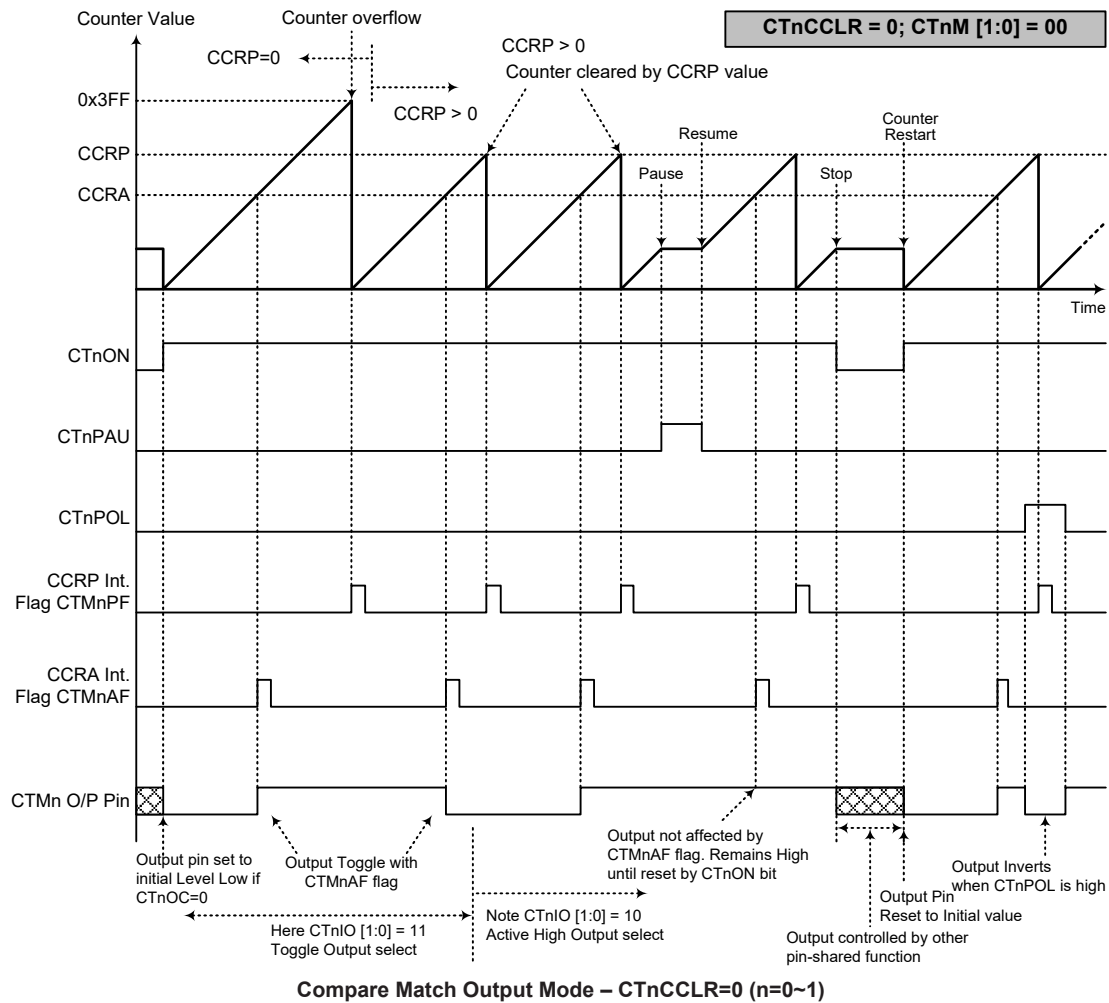
Compare Match Output Mode

To select this mode, bits CTnM1 and CTnM0 in the CTMnC1 register, should be set to “00” respectively. In this mode once the counter is enabled and running it can be cleared by three methods. These are a counter overflow, a compare match from Comparator A and a compare match from Comparator P. When the CTnCCLR bit is low, there are two ways in which the counter can be cleared. One is when a compare match from Comparator P, the other is when the CCRP bits are all zero which allows the counter to overflow. Here both CTMnAF and CTMnPF interrupt request flags for Comparator A and Comparator P respectively, will both be generated.

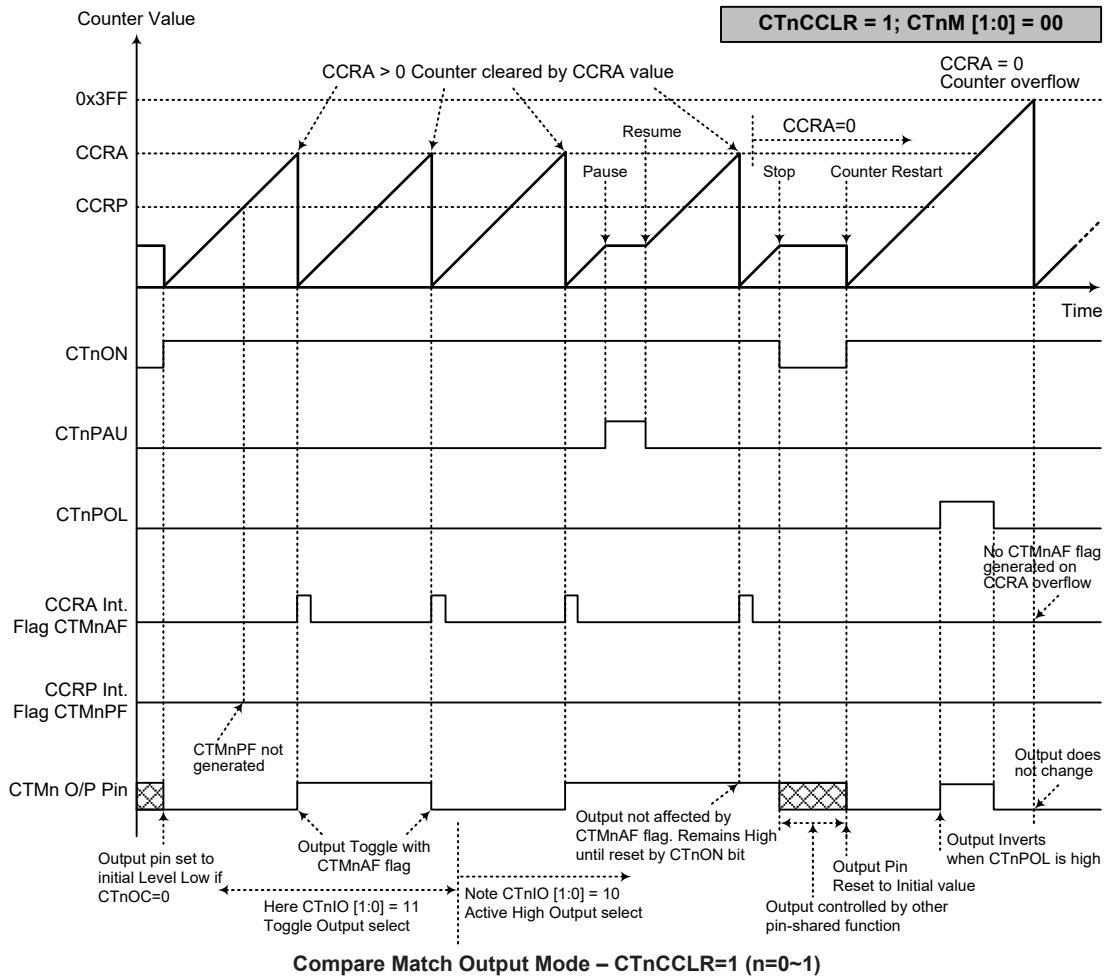
If the CTnCCLR bit in the CTMnC1 register is high then the counter will be cleared when a compare match occurs from Comparator A. However, here only the CTMnAF interrupt request flag will be generated even if the value of the CCRP bits is less than that of the CCRA registers. Therefore when CTnCCLR is high no CTMnPF interrupt request flag will be generated. If the CCRA bits are all zero, the counter will overflow when it reaches its maximum 10-bit, 3FF Hex, value. However, here the CTMnAF interrupt request flag will not be generated.

As the name of the mode suggests, after a comparison is made, the CTMn output pin, will change state. The CTMn output pin condition however only changes state when a CTMnAF interrupt request flag is generated after a compare match occurs from Comparator A. The CTMnPF interrupt

request flag, generated from a compare match occurs from Comparator P, will have no effect on the CTMn output pin. The way in which the CTMn output pin changes state are determined by the condition of the CTnIO1 and CTnIO0 bits in the CTMnC1 register. The CTMn output pin can be selected using the CTnIO1 and CTnIO0 bits to go high, to go low or to toggle from its present condition when a compare match occurs from Comparator A. The initial condition of the CTMn output pin, which is set after the CTnON bit changes from low to high, is set using the CTnOC bit. Note that if the CTnIO1 and CTnIO0 bits are zero then no pin change will take place.



- Note: 1. With CTnCCLR=0 a Comparator P match will clear the counter
2. The CTMn output pin is controlled only by the CTMnAF flag
3. The output pin is reset to its initial state by a CTnON bit rising edge



- Note: 1. With CTnCCLR=1 a Comparator A match will clear the counter
 2. The CTMn output pin is controlled only by the CTMnAF flag
 3. The output pin is reset to its initial state by a CTnON bit rising edge
 4. A CTMnPF flag is not generated when CTnCCLR=1

Timer/Counter Mode

To select this mode, bits CTnM1 and CTnM0 in the CTMnC1 register should be set to “11” respectively. The Timer/Counter Mode operates in an identical way to the Compare Match Output Mode generating the same interrupt flags. The exception is that in the Timer/Counter Mode the CTMn output pin is not used. Therefore the above description and Timing Diagrams for the Compare Match Output Mode can be used to understand its function. As the CTMn output pin is not used in this mode, the pin can be used as a normal I/O pin or other pin-shared functions.

PWM Output Mode

To select this mode, bits CTnM1 and CTnM0 in the CTMnC1 register should be set to “10” respectively. The PWM function within the CTMn is useful for applications which require functions such as motor control, heating control, illumination control, etc. By providing a signal of fixed frequency but of varying duty cycle on the CTMn output pin, a square wave AC waveform can be generated with varying equivalent DC RMS values.

As both the period and duty cycle of the PWM waveform can be controlled, the choice of generated waveform is extremely flexible. In the PWM Output Mode, the CTnCCLR bit has no effect as the PWM period. Both of the CCRA and CCRP registers are used to generate the PWM waveform, one register is used to clear the internal counter and thus control the PWM waveform frequency, while the other one is used to control the duty cycle. Which register is used to control either frequency or duty cycle is determined using the CTnDPX bit in the CTMnC1 register. The PWM waveform frequency and duty cycle can therefore be controlled by the values in the CCRA and CCRP registers.

An interrupt flag, one for each of the CCRA and CCRP, will be generated when a compare match occurs from either Comparator A or Comparator P. The CTnOC bit in the CTMnC1 register is used to select the required polarity of the PWM waveform while the two CTnIO1 and CTnIO0 bits are used to enable the PWM output or to force the CTMn output pin to a fixed high or low level. The CTnPOL bit is used to reverse the polarity of the PWM output waveform.

• 10-bit CTMn, PWM Output Mode, Edge-aligned Mode, CTnDPX=0

CCRP	1~7	0
Period	CCRP×128	1024
Duty	CCRA	

If $f_{SYS}=16\text{MHz}$, CTMn clock source is $f_{SYS}/4$, CCRP=4 and CCRA=128,

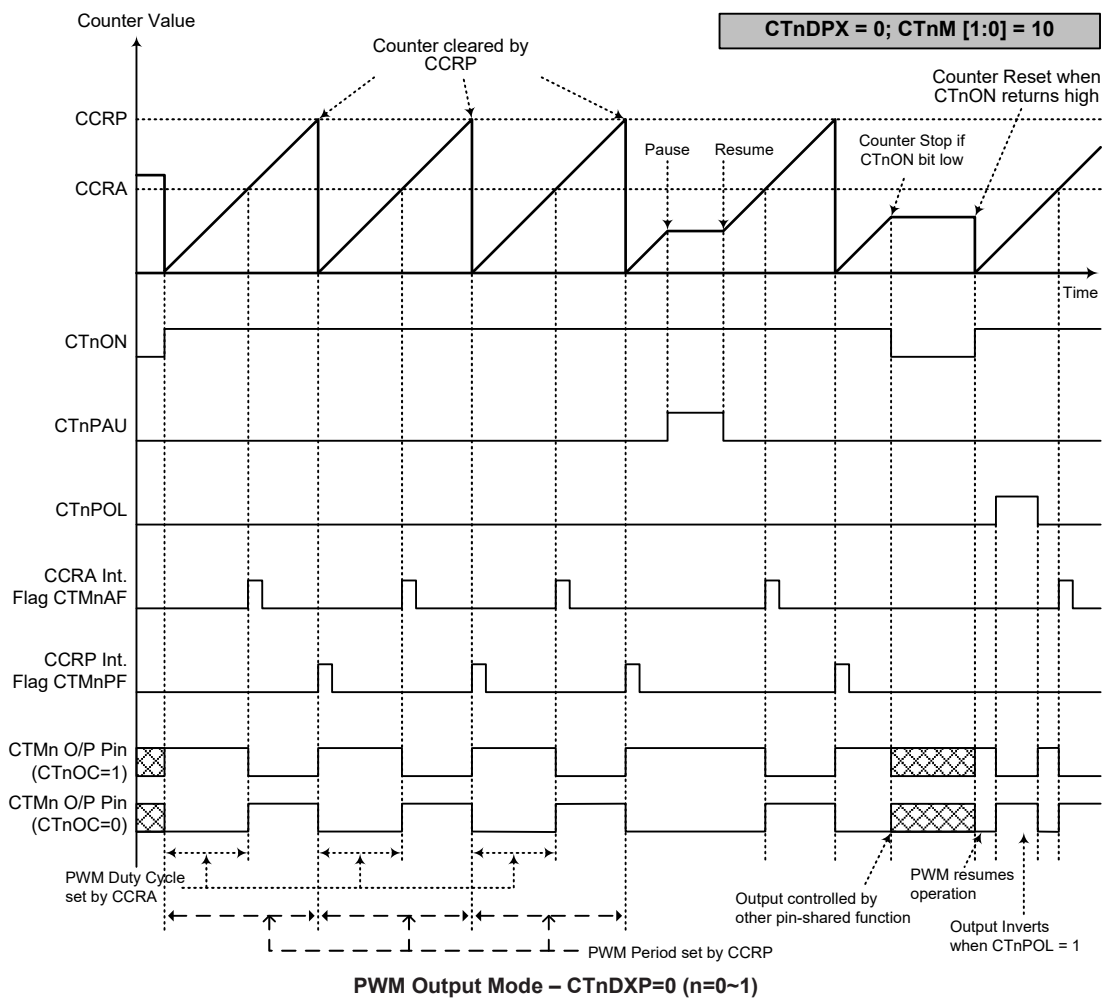
The CTMn PWM output frequency= $(f_{SYS}/4)/(4 \times 128)=f_{SYS}/2048=8\text{kHz}$, duty= $128/(4 \times 128)=25\%$.

If the Duty value defined by the CCRA register is equal to or greater than the Period value, then the PWM output duty is 100%.

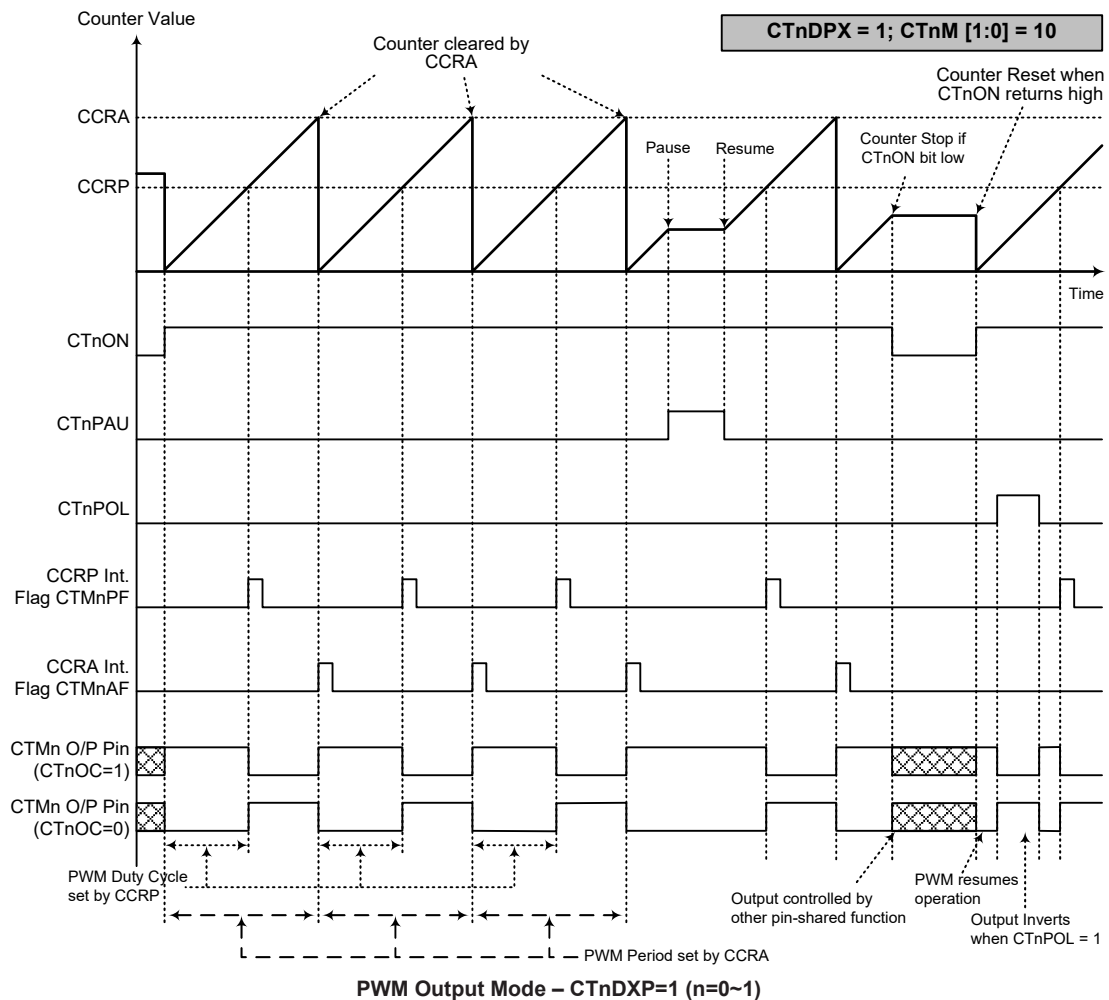
• 10-bit CTMn, PWM Output Mode, Edge-aligned Mode, CTnDPX=1

CCRP	1~7	0
Period	CCRA	
Duty	CCRP×128	1024

The PWM output period is determined by the CCRA register value together with the CTMn clock while the PWM duty cycle is defined by the CCRP register value except when the CCRP value is equal to 0.

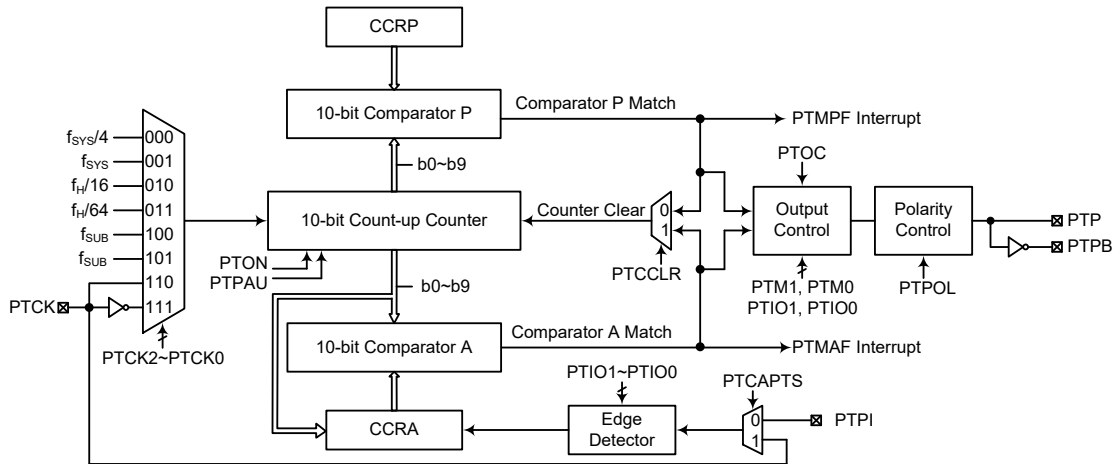


- Note: 1. Here CTnDPX=0 – Counter cleared by CCRP
 2. A counter clear sets the PWM Period
 3. The internal PWM function continues running even when CTnIO[1:0]=00 or 01
 4. The CTnCCLR bit has no influence on PWM operation



- Note: 1. Here CTnDPX=1 – Counter cleared by CCRA
2. A counter clear sets the PWM Period
3. The internal PWM function continues even when CTnIO[1:0]=00 or 01
4. The CTnCCLR bit has no influence on PWM operation

The Periodic Type TM contains five operating modes, which are Compare Match Output, Timer/Event Counter, Capture Input, Single Pulse Output and PWM Output modes. The Periodic TM can also be controlled with two external input pins and can drive two external output pins.



Periodic Type TM Block Diagram

The only way of changing the value of the 10-bit counter using the application program is to clear the counter by changing the PTON bit from low to high. The counter will also be cleared automatically by a counter overflow or a compare match with one of its associated comparators. When these conditions occur, a PTM interrupt signal will also usually be generated. The Periodic Type TM can operate in a number of different operational modes, can be driven by different clock sources including an input pin and can also control the output pins. All operating setup conditions are selected using relevant internal registers.

Overall operation of the Periodic TM is controlled using a series of registers. A read only register pair exists to store the internal counter 10-bit value, while two read/write register pairs exist to store the internal 10-bit CCRA and CCRP value. The remaining two registers are control registers which set the different operating and control modes.

Register Name	Bit							
	7	6	5	4	3	2	1	0
PTMC0	PTPAU	PTCK2	PTCK1	PTCK0	PTON	—	—	—
PTMC1	PTM1	PTM0	PTIO1	PTIO0	PTOC	PTPOL	PTCAPTS	PTCCLR
PTMDL	D7	D6	D5	D4	D3	D2	D1	D0
PTMDH	—	—	—	—	—	—	D9	D8
PTMAL	D7	D6	D5	D4	D3	D2	D1	D0
PTMAH	—	—	—	—	—	—	D9	D8
PTMRPL	PTRP7	PTRP6	PTRP5	PTRP4	PTRP3	PTRP2	PTRP1	PTRP0
PTMRPH	—	—	—	—	—	—	PTRP9	PTRP8

10-bit Periodic TM Register List

• **PTMC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	PTPAU	PTCK2	PTCK1	PTCK0	PTON	—	—	—
R/W	R/W	R/W	R/W	R/W	R/W	—	—	—
POR	0	0	0	0	0	—	—	—

Bit 7 **PTPAU**: PTM counter pause control

0: Run
1: Pause

The counter can be paused by setting this bit high. Clearing the bit to zero restores normal counter operation. When in a Pause condition the PTM will remain powered up and continue to consume power. The counter will retain its residual value when this bit changes from low to high and resume counting from this value when the bit changes to a low value again.

Bit 6~4 **PTCK2~PTCK0**: PTM Counter clock selection

000: $f_{SYS}/4$
001: f_{SYS}
010: $f_{IH}/16$
011: $f_{IH}/64$
100: f_{SUB}
101: f_{SUB}
110: PTCK rising edge clock
111: PTCK falling edge clock

These three bits are used to select the clock source for the PTM. The external pin clock source can be chosen to be active on the rising or falling edge. The clock source f_{SYS} is the system clock, while f_{IH} and f_{SUB} are other internal clocks, the details of which can be found in the oscillator section.

Bit 3 **PTON**: PTM counter on/off control

0: Off
1: On

This bit controls the overall on/off function of the PTM. Setting the bit high enables the counter to run while clearing the bit disables the PTM. Clearing this bit to zero will stop the counter from counting and turn off the PTM which will reduce its power consumption. When the bit changes state from low to high the internal counter value will be reset to zero, however when the bit changes from high to low, the internal counter will retain its residual value until the bit returns high again. If the PTM is in the Compare Match Output Mode, PWM output Mode or Single Pulse Output Mode then the PTM output pin will be reset to its initial condition, as specified by the PTOC bit, when the PTON bit changes from low to high.

Bit 2~0 Unimplemented, read as “0”

• **PTMC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	PTM1	PTM0	PTIO1	PTIO0	PTOC	PTPOL	PTCAPTS	PTCCLR
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **PTM1~PTM0**: Select PTM Operating Mode
 00: Compare Match Output Mode
 01: Capture Input Mode
 10: PWM Output Mode or Single Pulse Output Mode
 11: Timer/Counter Mode

These bits set the required operating mode for the PTM. To ensure reliable operation the PTM should be switched off before any changes are made to the PTM1 and PTM0 bits. In the Timer/Counter Mode, the PTM output pin state is undefined.

Bit 5~4 **PTIO1~PTIO0**: PTM external pin PTP, PTPI or PTCK function selection

Compare Match Output Mode

00: No change
 01: Output low
 10: Output high
 11: Toggle output

PWM Output Mode/Single Pulse Output Mode

00: PWM output inactive state
 01: PWM output active state
 10: PWM output
 11: Single Pulse Output

Capture Input Mode

00: Input capture at rising edge of PTPI or PTCK
 01: Input capture at falling edge of PTPI or PTCK
 10: Input capture at rising/falling edge of PTPI or PTCK
 11: Input capture disabled

Timer/Counter Mode

Unused

These two bits are used to determine how the PTM external pin changes state when a certain condition is reached. The function that these bits select depends upon in which mode the PTM is running.

In the Compare Match Output Mode, the PTIO1 and PTIO0 bits determine how the PTM output pin changes state when a compare match occurs from the Comparator A. The PTM output pin can be set to switch high, switch low or to toggle its present state when a compare match occurs from the Comparator A. When the bits are both zero, then no change will take place on the output. The initial value of the PTM output pin should be set using the PTOC bit in the PTMC1 register. Note that the output level requested by the PTIO1 and PTIO0 bits must be different from the initial value setup using the PTOC bit otherwise no change will occur on the PTM output pin when a compare match occurs. After the PTM output pin changes state, it can be reset to its initial level by changing the level of the PTON bit from low to high.

In the PWM Output Mode, the PTIO1 and PTIO0 bits determine how the TM output pin changes state when a certain compare match condition occurs. The PTM output function is modified by changing these two bits. It is necessary to only change the values of the PTIO1 and PTIO0 bits only after the PTM has been switched off. Unpredictable PWM outputs will occur if the PTIO1 and PTIO0 bits are changed when the PTM is running.

- Bit 3 **PTOC**: PTM PTP output control
 Compare Match Output Mode
 0: Initial low
 1: Initial high
 PWM Output Mode/Single Pulse Output Mode
 0: Active low
 1: Active high
 This is the output control bit for the PTM output pin. Its operation depends upon whether PTM is being used in the Compare Match Output Mode or in the PWM Output Mode/Single Pulse Output Mode. It has no effect if the PTM is in the Timer/Counter Mode. In the Compare Match Output Mode it determines the logic level of the PTM output pin before a compare match occurs. In the PWM Output Mode it determines if the PWM signal is active high or active low. In the Single Pulse Output Mode it determines the logic level of the PTM output pin when the PTON bit changes from low to high.
- Bit 2 **PTPOL**: PTM PTP output polarity control
 0: Non-invert
 1: Invert
 This bit controls the polarity of the PTP output pin. When the bit is set high the PTM output pin will be inverted and not inverted when the bit is zero. It has no effect if the PTM is in the Timer/Counter Mode.
- Bit 1 **PTCAPTS**: PTM capture trigger source selection
 0: From PTPI pin
 1: From PTCK pin
- Bit 0 **PTCCLR**: PTM counter clear condition selection
 0: Comparator P match
 1: Comparator A match
 This bit is used to select the method which clears the counter. Remember that the Periodic TM contains two comparators, Comparator A and Comparator P, either of which can be selected to clear the internal counter. With the PTCCLR bit set high, the counter will be cleared when a compare match occurs from the Comparator A. When the bit is low, the counter will be cleared when a compare match occurs from the Comparator P or with a counter overflow. A counter overflow clearing method can only be implemented if the CCRP bits are all cleared to zero. The PTCCLR bit is not used in the PWM Output, Single Pulse Output or Capture Input Mode.

• **PTMDL Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R	R	R	R	R	R	R	R
POR	0	0	0	0	0	0	0	0

- Bit 7~0 **D7~D0**: PTM Counter Low Byte Register bit 7 ~ bit 0
 PTM 10-bit Counter bit 7 ~ bit 0

• **PTMDH Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	D9	D8
R/W	—	—	—	—	—	—	R	R
POR	—	—	—	—	—	—	0	0

- Bit 7~2 Unimplemented, read as “0”
 Bit 1~0 **D9~D8**: PTM Counter High Byte Register bit 1 ~ bit 0
 PTM 10-bit Counter bit 9 ~ bit 8

• **PTMAL Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: PTM CCRA Low Byte Register bit 7 ~ bit 0
PTM 10-bit CCRA bit 7 ~ bit 0

• **PTMAH Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	D9	D8
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as “0”
Bit 1~0 **D9~D8**: PTM CCRA High Byte Register bit 1 ~ bit 0
PTM 10-bit CCRA bit 9 ~ bit 8

• **PTMRPL Register**

Bit	7	6	5	4	3	2	1	0
Name	PTRP7	PTRP6	PTRP5	PTRP4	PTRP3	PTRP2	PTRP1	PTRP0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **PTRP7~PTRP0**: PTM CCRP Low Byte Register bit 7 ~ bit 0
PTM 10-bit CCRP bit 7 ~ bit 0

• **PTMRPH Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	PTRP9	PTRP8
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as “0”
Bit 1~0 **PTRP9~PTRP8**: PTM CCRP High Byte Register bit 1 ~ bit 0
PTM 10-bit CCRP bit 9 ~ bit 8

Periodic Type TM Operation Modes

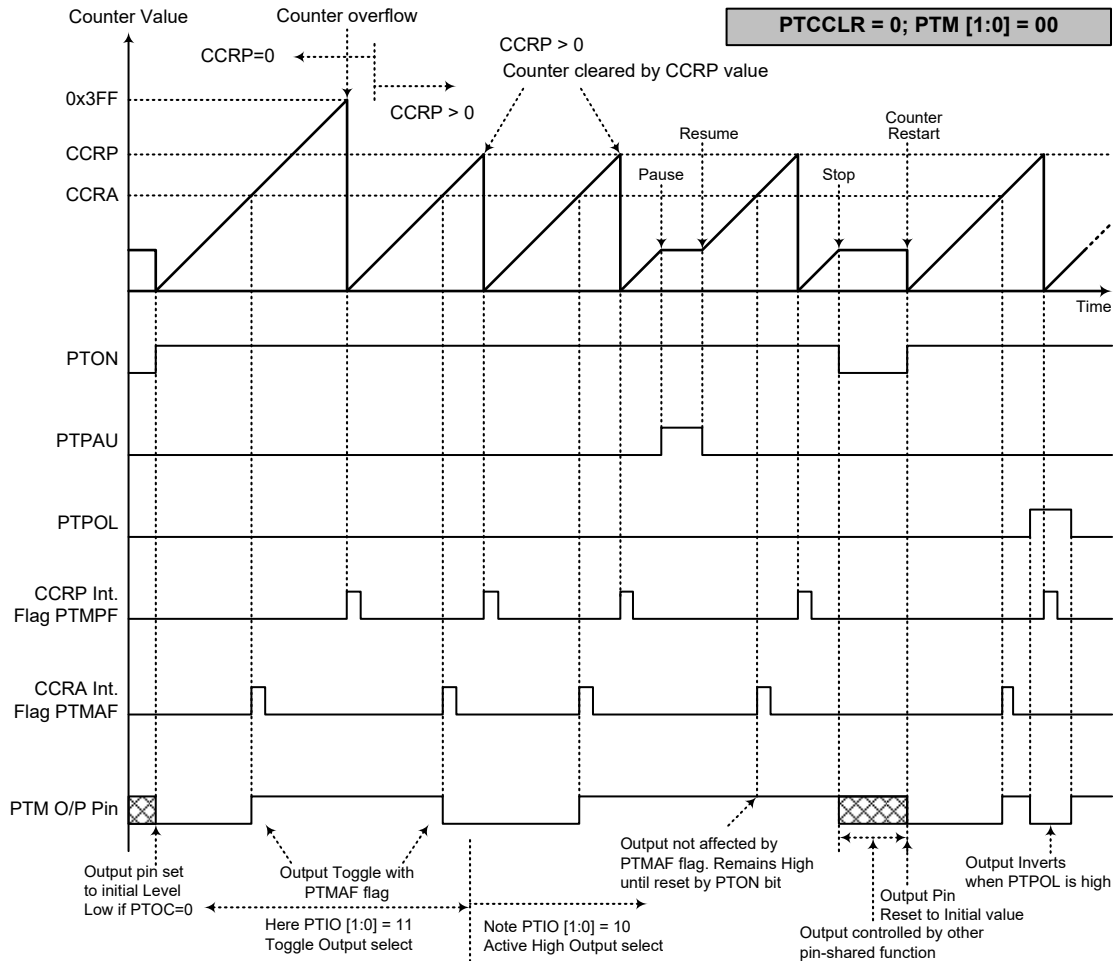
The Periodic Type TM can operate in one of five operating modes, Compare Match Output Mode, PWM Output Mode, Single Pulse Output Mode, Capture Input Mode or Timer/Counter Mode. The operating mode is selected using the PTM1 and PTM0 bits in the PTMC1 register.

Compare Match Output Mode

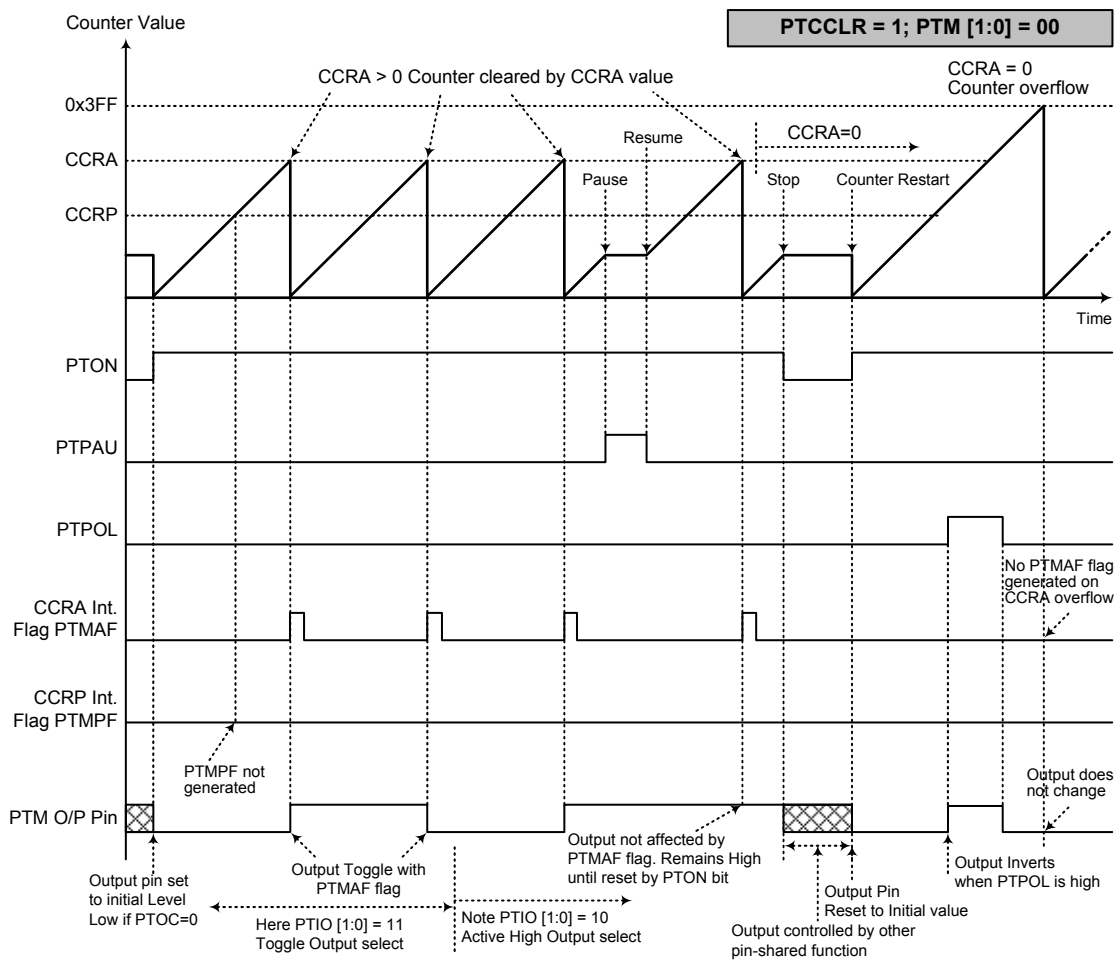
To select this mode, bits PTM1 and PTM0 in the PTMC1 register, should be set to “00” respectively. In this mode once the counter is enabled and running it can be cleared by three methods. These are a counter overflow, a compare match from Comparator A and a compare match from Comparator P. When the PTCCLR bit is low, there are two ways in which the counter can be cleared. One is when a compare match from Comparator P, the other is when the CCRP bits are all zero which allows the counter to overflow. Here both PTMAF and PTMPF interrupt request flags for Comparator A and Comparator P respectively, will both be generated.

If the PTCCLR bit in the PTMC1 register is high then the counter will be cleared when a compare match occurs from Comparator A. However, here only the PTMAF interrupt request flag will be generated even if the value of the CCRP bits is less than that of the CCRA registers. Therefore when PTCCLR is high no PTMPF interrupt request flag will be generated. In the Compare Match Output Mode, the CCRA cannot be cleared to zero. If the CCRA bits are all zero, the counter will overflow when it reaches its maximum 10-bit, 3FF Hex, value, however here the PTMAF interrupt request flag will not be generated.

As the name of the mode suggests, after a comparison is made, the PTM output pin will change state. The PTM output pin condition however only changes state when a PTMAF interrupt request flag is generated after a compare match occurs from Comparator A. The PTMPF interrupt request flag, generated from a compare match occurs from Comparator P, will have no effect on the PTM output pin. The way in which the PTM output pin changes state are determined by the condition of the PTIO1 and PTIO0 bits in the PTMC1 register. The PTM output pin can be selected using the PTIO1 and PTIO0 bits to go high, to go low or to toggle from its present condition when a compare match occurs from Comparator A. The initial condition of the PTM output pin, which is set after the PTON bit changes from low to high, is set using the PTOC bit. Note that if the PTIO1 and PTIO0 bits are zero then no pin change will take place.



- Note: 1. With $PTCCLR=0$, a Comparator P match will clear the counter
2. The PTM output pin is controlled only by the PTMAF flag
3. The output pin is reset to its initial state by a PTON bit rising edge



Compare Match Output Mode – PTCCLR=1

- Note: 1. With PTCCLR=1, a Comparator A match will clear the counter
2. The PTM output pin is controlled only by the PTMAF flag
3. The output pin is reset to its initial state by a PTON bit rising edge
4. A PTMPF flag is not generated when PTCCLR=1

Timer/Counter Mode

To select this mode, bits PTM1 and PTM0 in the PTMC1 register should be set to “11” respectively. The Timer/Counter Mode operates in an identical way to the Compare Match Output Mode generating the same interrupt flags. The exception is that in the Timer/Counter Mode the PTM output pin is not used. Therefore the above description and Timing Diagrams for the Compare Match Output Mode can be used to understand its function. As the PTM output pin is not used in this mode, the pin can be used as a normal I/O pin or other pin-shared functions.

PWM Output Mode

To select this mode, bits PTM1 and PTM0 in the PTMC1 register should be set to “10” respectively and also the PTIO1 and PTIO0 bits should be set to “10” respectively. The PWM function within the PTM is useful for applications which require functions such as motor control, heating control, illumination control, etc. By providing a signal of fixed frequency but of varying duty cycle on the PTM output pin, a square wave AC waveform can be generated with varying equivalent DC RMS values.

As both the period and duty cycle of the PWM waveform can be controlled, the choice of generated waveform is extremely flexible. In the PWM Output Mode, the PTCCLR bit has no effect as the PWM period. Both of the CCRP and CCRA registers are used to generate the PWM waveform, one register is used to clear the internal counter and thus control the PWM waveform frequency, while the other one is used to control the duty cycle. The PWM waveform frequency and duty cycle can therefore be controlled by the values in the CCRA and CCRP registers.

An interrupt flag, one for each of the CCRA and CCRP, will be generated when a compare match occurs from either Comparator A or Comparator P. The PTOC bit in the PTMC1 register is used to select the required polarity of the PWM waveform while the two PTIO1 and PTIO0 bits are used to enable the PWM output or to force the PTM output pin to a fixed high or low level. The PTPOL bit is used to reverse the polarity of the PWM output waveform.

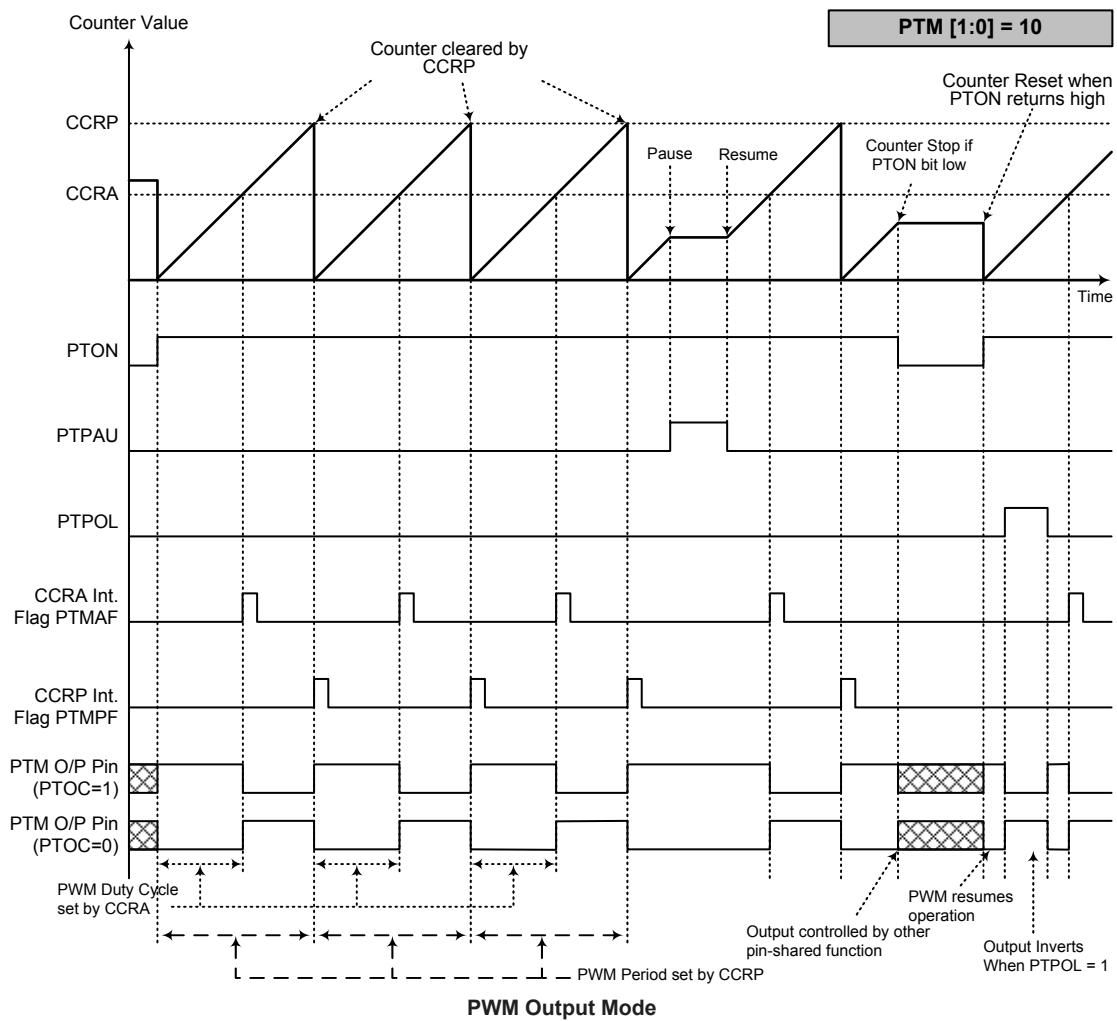
• 10-bit PTM, PWM Output Mode, Edge-aligned Mode

CCRP	1~1023	0
Period	1~1023	1024
Duty	CCRA	

If $f_{SYS}=16\text{MHz}$, TM clock source select $f_{SYS}/4$, CCRP=512 and CCRA=128,

The PTM PWM output frequency= $(f_{SYS}/4)/512=f_{SYS}/2048=8\text{kHz}$, duty=128/512=25%,

If the Duty value defined by the CCRA register is equal to or greater than the Period value, then the PWM output duty is 100%.



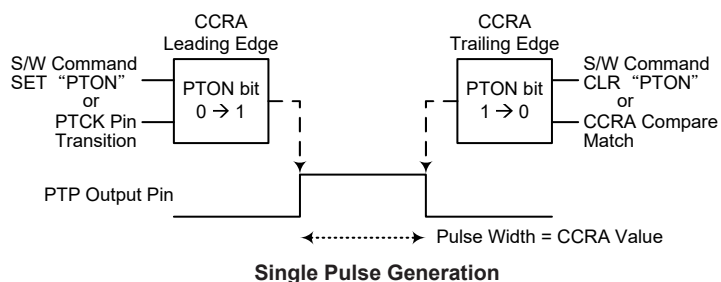
- Note:
1. The counter is cleared by CCRP
 2. A counter clear sets the PWM Period
 3. The internal PWM function continues running even when PTIO[1:0]=00 or 01
 4. The PTCCLR bit has no influence on PWM operation

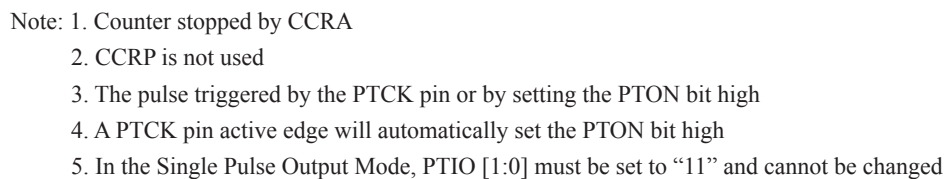
Single Pulse Output Mode

To select this mode, bits PTM1 and PTM0 in the PTMC1 register should be set to “10” respectively and also the PTIO1 and PTIO0 bits should be set to “11” respectively. The Single Pulse Output Mode, as the name suggests, will generate a single shot pulse on the PTM output pin.

The trigger for the pulse output leading edge is a low to high transition of the PTON bit, which can be implemented using the application program. However in the Single Pulse Output Mode, the PTON bit can also be made to automatically change from low to high using the external PTCK pin, which will in turn initiate the Single Pulse output. When the PTON bit transitions to a high level, the counter will start running and the pulse leading edge will be generated. The PTON bit should remain high when the pulse is in its active state. The generated pulse trailing edge will be generated when the PTON bit is cleared to zero, which can be implemented using the application program or when a compare match occurs from Comparator A.

However a compare match from Comparator A will also automatically clear the PTON bit and thus generate the Single Pulse output trailing edge. In this way the CCRA value can be used to control the pulse width. A compare match from Comparator A will also generate a PTM interrupt. The counter can only be reset back to zero when the PTON bit changes from low to high when the counter restarts. In the Single Pulse Output Mode CCRP is not used. The PTCCLR is not used in this Mode.



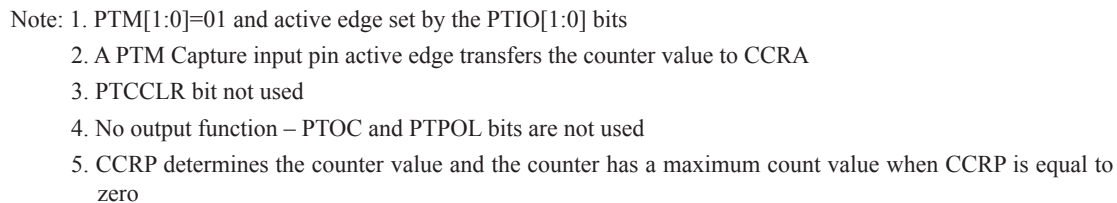


Capture Input Mode

To select this mode bits PTM1 and PTM0 in the PTMC1 register should be set to “01” respectively. This mode enables external signals to capture and store the present value of the internal counter and can therefore be used for applications such as pulse width measurements. The external signal is supplied on the PTPI or PTCK pin, selected by the PTCAPTS bit in the PTMC1 register. The input pin active edge can be either a rising edge, a falling edge or both rising and falling edges; the active edge transition type is selected using the PTIO1 and PTIO0 bits in the PTMC1 register. The counter is started when the PTON bit changes from low to high which is initiated using the application program.

When the required edge transition appears on the PTPI or PTCK pin the present value in the counter will be latched into the CCRA registers and a PTM interrupt generated. Irrespective of what events occur on the PTPI or PTCK pin the counter will continue to free run until the PTON bit changes from high to low. When a CCRP compare match occurs the counter will reset back to zero; in this way the CCRP value can be used to control the maximum counter value. When a CCRP compare match occurs from Comparator P, a PTM interrupt will also be generated. Counting the number of overflow interrupt signals from the CCRP can be a useful method in measuring long pulse widths. The PTIO1 and PTIO0 bits can select the active trigger edge on the PTPI or PTCK pin to be a rising edge, falling edge or both edge types. If the PTIO1 and PTIO0 bits are both set high, then no capture operation will take place irrespective of what happens on the PTPI or PTCK pin, however it must be noted that the counter will continue to run.

As the PTPI or PTCK pin is pin shared with other functions, care must be taken if the PTM is in the Input Capture Mode. This is because if the pin is set as an output, then any transitions on this pin may cause an input capture operation to be executed. The PTCCLR, PTOC and PTPOL bits are not used in this Mode.



Analog to Digital Converter

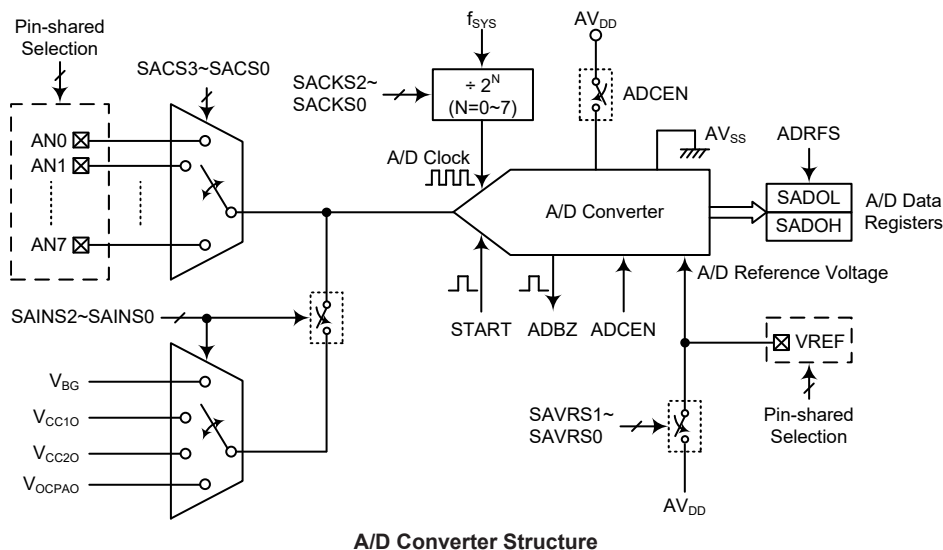
The need to interface to real world analog signals is a common requirement for many electronic systems. However, to properly process these signals by a microcontroller, they must first be converted into digital signals by A/D converters. By integrating the A/D conversion electronic circuitry into the microcontroller, the need for external components is reduced significantly with the corresponding follow-on benefits of lower costs and reduced component space requirements.

A/D Overview

The device contains a multi-channel analog to digital converter which can directly interface to external analog signals, such as that from sensors or other control signals and convert these signals directly into a 12-bit digital value. It also can convert the internal signals, such as the bandgap reference voltage V_{BG} , high voltage power V_{CC1} divided voltage V_{CC1O} , high voltage power V_{CC2} divided voltage V_{CC2O} and OCP operational amplifier output voltage V_{OCPAO} , into a 12-bit digital value. The external or internal analog signal to be converted is determined by the SAINS and SACS bit fields. When the external analog signal is to be converted, the corresponding external channel input pin function should first be properly configured and then the desired external channel input should be selected using the SAINS and SACS fields. Note that when the internal analog signal is selected to be converted, the SAINS and SACS fields should also be properly configured. More detailed information about the A/D input signal selection will be described in the “A/D converter Control Registers” and “A/D Converter Input Signals” section respectively.

External Input Channels	Internal Signals	A/D Signal Select Bits
8: AN0~AN7	4: V_{BG} , V_{CC1O} , V_{CC2O} , V_{OCPAO}	SAINS2~SAINS0, SACS3~SACS0

The accompanying block diagram shows the internal structure of the A/D converter together with its associated registers and control bits.



A/D Converter Register Description

Overall operation of the A/D converter is controlled using five registers. A read only register pair exists to store the A/D Converter data 12-bit value. The remaining two registers, SADC0 and SADC1, are control registers which set the operating conditions and control function of the A/D converter.

Register Name	Bit							
	7	6	5	4	3	2	1	0
SADOL (ADRF5=0)	D3	D2	D1	D0	—	—	—	—
SADOL (ADRF5=1)	D7	D6	D5	D4	D3	D2	D1	D0
SADOH (ADRF5=0)	D11	D10	D9	D8	D7	D6	D5	D4
SADOH (ADRF5=1)	—	—	—	—	D11	D10	D9	D8
SADC0	START	ADBZ	ADCEN	ADRF5	SACS3	SACS2	SACS1	SACS0
SADC1	SAINS2	SAINS1	SAINS0	SAVRS1	SAVRS0	SACKS2	SACKS1	SACKS0

A/D Converter Register List

A/D Converter Data Registers – SADOL, SADOH

As the device contains an internal 12-bit A/D converter, it requires two data registers to store the converted value. These are a high byte register, known as SADOH, and a low byte register, known as SADOL. After the conversion process takes place, these registers can be directly read by the microcontroller to obtain the digitised conversion value. As only 12 bits of the 16-bit register space is utilised, the format in which the data is stored is controlled by the ADRF5 bit in the SADC0 register as shown in the accompanying table. D0~D11 are the A/D conversion result data bits. Any unused bits will be read as zero. The A/D data registers contents will be unchanged if the A/D converter is disabled.

ADRF5	SADOH								SADOL							
	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
0	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	0	0	0	0
1	0	0	0	0	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0

A/D Converter Data Registers

A/D Converter Control Registers – SADC0, SADC1

To control the function and operation of the A/D converter, two control registers known as SADC0 and SADC1 are provided. These 8-bit registers define functions such as the selection of which analog signal is connected to the internal A/D converter, the digitised data format, the A/D clock source as well as controlling the start function and monitoring the A/D converter busy status. As the device contains only one actual analog to digital converter hardware circuit, each of the external and internal analog signals must be routed to the converter. The SACS3~SACS0 bits in the SADC0 register are used to determine which external channel input is selected to be converted. The SAINS2~SAINS0 bits in the SADC1 register are used to determine that the analog signal to be converted comes from the internal analog signal or external analog channel input.

The relevant pin-shared function selection bits determine which pins on I/O Ports are used as analog inputs for the A/D converter input and which pins are not to be used as the A/D converter input. When the pin is selected to be an A/D input, its original function whether it is an I/O or other pin-shared function will be removed. In addition, any internal pull-high resistor connected to the pin will be automatically removed if the pin is selected to be an A/D converter input.

• **SADC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	START	ADBZ	ADCEN	ADRF5	SACS3	SACS2	SACS1	SACS0
R/W	R/W	R	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7 **START**: Start the A/D conversion
0→1→0: Start
This bit is used to initiate an A/D conversion process. The bit is normally low but if set high and then cleared low again, the A/D converter will initiate a conversion process.
- Bit 6 **ADBZ**: A/D converter busy flag
0: No A/D conversion is in progress
1: A/D conversion is in progress
This read only flag is used to indicate whether the A/D conversion is in progress or not. When the START bit is set from low to high and then to low again, the ADBZ flag will be set to 1 to indicate that the A/D conversion is initiated. The ADBZ flag will be cleared to 0 after the A/D conversion is complete.
- Bit 5 **ADCEN**: A/D converter function enable control
0: Disable
1: Enable
This bit controls the A/D internal function. This bit should be set to one to enable the A/D converter. If the bit is set low, then the A/D converter will be switched off reducing the device power consumption. When the A/D converter function is disabled, the contents of the A/D data register pair known as SADOH and SADOL will be unchanged.
- Bit 4 **ADRF5**: A/D conversion data format selection
0: A/D converter data format → SADOH=D[11:4]; SADOL=D[3:0]
1: A/D converter data format → SADOH=D[11:8]; SADOL=D[7:0]
This bit controls the format of the 12-bit converted A/D value in the two A/D data registers. Details are provided in the A/D converter data register section.
- Bit 3~0 **SACS3~SACS0**: A/D converter external analog input channel selection
0000: AN0
0001: AN1
0010: AN2
0011: AN3
0100: AN4
0101: AN5
0110: AN6
0111: AN7
1xxx: Non-existed channel, input floating if selected
These bits are used to select which external analog input channel is to be converted. When the external analog input channel is selected, the SAINS bit field must be set to “000”, “101” or “11x”. Details are summarized in the “A/D Converter Input Signal Selection” table.

• **SADC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	SAINS2	SAINS1	SAINS0	SAVRS1	SAVRS0	SACKS2	SACKS1	SACKS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7~5 SAINS2~SAINS0:** A/D converter input signal selection
 000: External source – External analog channel input, ANn
 001: Internal source – Internal bandgap reference voltage, V_{BG}
 010: Internal source – Internal high voltage power VCC1 divided voltage, V_{CC1O}
 011: Internal source – Internal high voltage power VCC2 divided voltage, V_{CC2O}
 100: Internal source – Internal OCP operational amplifier output voltage, V_{OCPAO}
 101~111: External source – External analog channel input, ANn
 Care must be taken if the SAINS field is set to “001~100” to select the internal analog signal to be converted. When the internal analog signal is selected to be converted, the external channel input pin must never be selected as the A/D input signal by properly setting the SACS bit field with a value of “1000~1111”. Otherwise, the external channel input will be connected together with the internal analog signal. This will result in unpredictable situations such as an irreversible damage.
- Bit 4~3 SAVRS1~SAVRS0:** A/D converter reference voltage selection
 00: External VREF pin
 01: Internal A/D converter power, AV_{DD}
 1x: External VREF pin
 These bits are used to select the A/D converter reference voltage source. Care must be taken if the SAVRS field is set to “01” to select the internal A/D converter power voltage as the reference voltage source. When the internal reference voltage source is selected, the external VREF pin cannot be configured as the reference voltage input by properly configuring the relevant pin-shared control bits. Otherwise, the external input voltage on the VREF pin will be connected to the internal A/D converter power. This will result in unpredictable situations.
- Bit 2~0 SACKS2~SACKS0:** A/D conversion clock source selection
 000: f_{SYS}
 001: $f_{SYS}/2$
 010: $f_{SYS}/4$
 011: $f_{SYS}/8$
 100: $f_{SYS}/16$
 101: $f_{SYS}/32$
 110: $f_{SYS}/64$
 111: $f_{SYS}/128$
 These bits are used to select the clock source for the A/D converter.

A/D Converter Reference Voltage

The actual reference voltage supply to the A/D converter can be supplied from the positive power supply, AV_{DD} , or an external reference source supplied on pin VREF determined by the SAVRS bit field in the SADC1 register. When the SAVRS field is set to “01”, the A/D converter reference voltage will come from the AV_{DD} . Otherwise, the A/D converter reference voltage will come from the VREF pin if the SAVRS field is set to any other value except “01”. When the VREF pin is selected as the reference voltage supply pin, the relevant pin-shared control bits should first be properly configured to enable the VREF pin function as it is pin-shared with other functions. However, if the internal A/D converter power is selected as the reference voltage, the external VREF pin must not be configured as the reference voltage input function to avoid the internal connection between the VREF pin to the internal A/D converter power. Note that the analog input values must not be allowed to exceed the value of the selected reference voltage.

A/D Converter Input Signals

All of the external A/D analog input pins are pin-shared with the I/O pins as well as other functions. The corresponding pin-shared function control bits in the PxS0 and PxS1 registers determine whether the external input pins are set as A/D converter analog channel inputs or whether they have other functions. If the corresponding pin is set to be an A/D converter analog channel input, the original pin function will be disabled. In this way, pins can be changed under program control to change their function between A/D inputs and other functions. All pull-high resistors, which are set through register programming, will be automatically disconnected if the pins are set as A/D inputs. Note that it is not necessary to first set the A/D pin as an input in the port control register to enable the A/D input as when the relevant A/D input function selection bits enable an A/D input, the status of the port control register will be overridden.

The A/D converter also has four internal analog input options, which are the bandgap reference voltage, high voltage power VCC1 divided voltage, high voltage power VCC2 divided voltage and OCP operational amplifier output signal V_{OCPAO} . The internal analog input signal is selected by setting the SAINS2~SAINS0 bits. If the SAINS2~SAINS0 bits are set to “000”, “101” or “11x”, the external channel input will be selected to be converted and the SACS3~SACS0 bits can determine which external channel is selected. If the internal analog signal is selected to be converted, the external channel signal input must be switched off by setting the SACS field to a value of “1xxx”. Otherwise, the internal analog signal will be connected together with the external channel input. This will result in unpredictable situations.

SAINS[2:0]	SACS[3:0]	Input Signals	Description
000, 101, 11x	0000~0111	AN0~AN7	External channel analog input ANn
	1xxx	—	Floating, no external channel is selected
001	1xxx	V_{BG}	Internal Bandgap reference voltage
010	1xxx	V_{CC1O}	Internal high voltage power VCC1 divided voltage
011	1xxx	V_{CC2O}	Internal high voltage power VCC2 divided voltage
100	1xxx	V_{OCPAO}	Internal OCP operational amplifier output

A/D Converter Input Signal Selection

A/D Converter Operation

The START bit in the SADC0 register is used to start the AD conversion. When the microcontroller sets this bit from low to high and then low again, an analog to digital conversion cycle will be initiated.

The ADBZ bit in the SADC0 register is used to indicate whether the analog to digital conversion process is in progress or not. This bit will be automatically set to 1 by the microcontroller after an A/D conversion is successfully initiated. When the A/D conversion is complete, the ADBZ bit will be cleared to 0. In addition, the corresponding A/D interrupt request flag will be set in the interrupt control register, and if the A/D interrupt is enabled, an internal A/D interrupt signal will be generated. This A/D internal interrupt signal will direct the program flow to the associated A/D internal interrupt address for processing. If the A/D internal interrupt is disabled, the microcontroller can poll the ADBZ bit in the SADC0 register to check whether it has been cleared as an alternative method of detecting the end of an A/D conversion cycle.

The clock source for the A/D converter, which originates from the system clock f_{SYS} , can be chosen to be either f_{SYS} or a subdivided version of f_{SYS} . The division ratio value is determined by the SACKS bit field in the SADC1 register. Although the A/D clock source is determined by the system clock f_{SYS} and by bits SACKS2~SACKS0, there are some limitations on the A/D clock source speed

that can be selected. As the recommended range of permissible A/D clock period, t_{ADCK} , is from 0.5 μ s to 10 μ s, care must be taken for system clock frequencies. For example, if the system clock operates at a frequency of 8MHz, the SACKS2~SACKS0 bits should not be set to 000, 001 or 111. Doing so will give A/D clock periods that are less than the minimum A/D clock period or greater than the maximum A/D clock period which may result in inaccurate A/D conversion values. Refer to the following table for examples, where values marked with an asterisk * show where, special care must be taken, as the values may be beyond the specified A/D Clock Period range.

f_{SYS}	A/D Clock Period (t_{ADCK})							
	SACKS[2:0] = 000 (f_{SYS})	SACKS[2:0] = 001 ($f_{SYS}/2$)	SACKS[2:0] = 010 ($f_{SYS}/4$)	SACKS[2:0] = 011 ($f_{SYS}/8$)	SACKS[2:0] = 100 ($f_{SYS}/16$)	SACKS[2:0] = 101 ($f_{SYS}/32$)	SACKS[2:0] = 110 ($f_{SYS}/64$)	SACKS[2:0] = 111 ($f_{SYS}/128$)
1MHz	1 μ s	2 μ s	4 μ s	8 μ s	16 μ s *	32 μ s *	64 μ s *	128 μ s *
2MHz	500ns	1 μ s	2 μ s	4 μ s	8 μ s	16 μ s *	32 μ s *	64 μ s *
4MHz	250ns *	500ns	1 μ s	2 μ s	4 μ s	8 μ s	16 μ s *	32 μ s *
8MHz	125ns *	250ns *	500ns	1 μ s	2 μ s	4 μ s	8 μ s	16 μ s *
12MHz	83ns*	167ns*	333ns*	667ns	1.33 μ s	2.67 μ s	5.33 μ s	10.67 μ s*
16MHz	62.5ns*	125ns*	250ns*	500ns	1 μ s	2 μ s	4 μ s	8 μ s

A/D Clock Period Examples

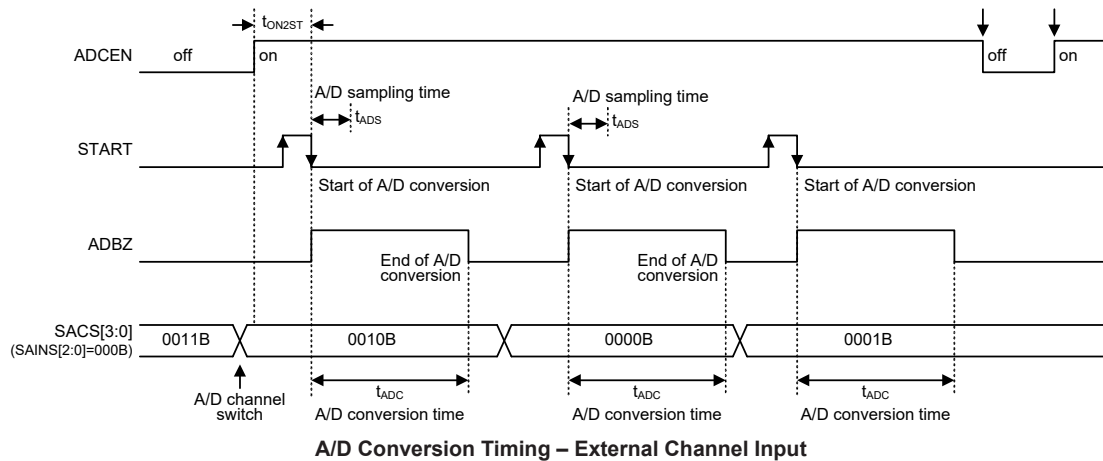
Controlling the power on/off function of the A/D converter circuitry is implemented using the ADCEN bit in the SADC0 register. This bit must be set high to power on the A/D converter. When the ADCEN bit is set high to power on the A/D converter internal circuitry, a certain delay, as indicated in the timing diagram, must be allowed before an A/D conversion is initiated. Even if no pins are selected for use as A/D inputs by configuring the relevant pin-shared control bits, if the ADCEN bit is high, then some power will still be consumed. In power conscious applications it is therefore recommended that the ADCEN is set low to reduce power consumption when the A/D converter function is not being used.

Conversion Rate and Timing Diagram

A complete A/D conversion contains two parts, data sampling and data conversion. The data sampling which is defined as t_{ADS} takes 4 A/D clock cycles and the data conversion takes 12 A/D clock cycles. Therefore a total of 16 A/D clock cycles for an analog signal A/D conversion which is defined as t_{ADC} are necessary.

$$\text{Maximum single A/D conversion rate} = \text{A/D clock period} / 16$$

The accompanying diagram shows graphically the various stages involved in an external channel input signal analog to digital conversion process and its associated timing. After an A/D conversion process has been initiated by the application program, the microcontroller internal hardware will begin to carry out the conversion, during which time the program can continue with other functions. The time taken for the A/D conversion is 16 t_{ADCK} clock cycles where t_{ADCK} is equal to the A/D clock period.



Summary of A/D Conversion Steps

The following summarises the individual steps that should be executed in order to implement an A/D conversion process.

- Step 1
Select the required A/D conversion clock by properly programming the SACKS2~SACKS0 bits in the SADC1 register.
- Step 2
Enable the A/D converter by setting the ADCEN bit in the SADC0 register to one.
- Step 3
Select which signal is to be connected to the internal A/D converter by correctly configuring the SAINS bit field.
Selecting the external channel input to be converted, go to Step 4.
Selecting the internal analog signal to be converted, go to Step 5.
- Step 4
If the SAINS field is set to select the external channel input, the corresponding pin should be configured as an A/D input function by selecting the relevant function control bits. Then the desired external channel input is selected by configuring the SACS field. Then go to Step 6.
- Step 5
Before the A/D input signal is selected to come from the internal analog signal by configuring the SAINS field, the SACS field must be set to “1xxx” to select a non-existed channel input. After this, the desired internal analog signal is selected by configuring the SAINS bit field. Then go to Step 6.
- Step 6
Select the A/D converter reference voltage source by configuring the SAVRS bit field in the SADC1 register. If the internal reference voltage is selected, the external reference input pin function must be disabled by properly configuring the relevant pin-shared control bits.
- Step 7
Select the A/D converter output data format by configuring the ADRFS bit in the SADC0 register.
- Step 8
If A/D conversion interrupt is used, the interrupt control registers must be correctly configured to ensure the A/D interrupt function is active. The master interrupt control bit, EMI, and the A/D conversion interrupt control bit, ADE, must both be set high in advance.

- Step 9
The A/D conversion procedure can now be initialized by setting the START bit from low to high and then low again.
- Step 10
If A/D conversion is in progress, the ADBZ flag will be set high. After the A/D conversion process is complete, the ADBZ flag will go low and then the output data can be read from SADOH and SADOL registers.

Note: When checking for the end of the conversion process, if the method of polling the ADBZ bit in the SADC0 register is used, the interrupt enable step above can be omitted.

Programming Considerations

During microcontroller operations where the A/D converter is not being used, the A/D internal circuitry can be switched off to reduce power consumption, by clearing bit ADCEN to zero in the SADC0 register. When this happens, the internal A/D converter circuits will not consume power irrespective of what analog voltage is applied to their input lines. If the A/D converter input lines are used as normal I/Os, then care must be taken as if the input voltage is not at a valid logic level, then this may lead to some increase in power consumption.

A/D Transfer Function

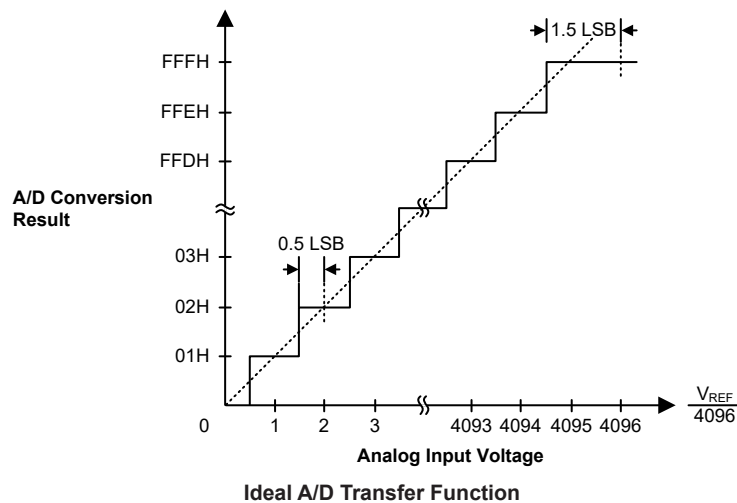
As the device contains a 12-bit A/D converter, its full-scale converted digitised value is equal to FFFH. Since the full-scale analog input value is equal to the actual A/D converter reference voltage, V_{REF} , this gives a single bit analog input value of reference voltage value divided by 4096.

$$1 \text{ LSB} = V_{REF} \div 4096$$

The A/D Converter input voltage value can be calculated using the following equation:

$$\text{A/D input voltage} = \text{A/D output digital value} \times (V_{REF} \div 4096)$$

The diagram shows the ideal transfer function between the analog input value and the digitised output value for the A/D converter. Except for the digitised zero value, the subsequent digitised values will change at a point 0.5 LSB below where they would change without the offset, and the last full scale digitised value will change at a point 1.5 LSB below the V_{REF} level. Note that here the V_{REF} voltage is the actual A/D converter reference voltage source determined by the SAVRS field.



A/D Programming Examples

The following two programming examples illustrate how to set and implement an A/D conversion. In the first example, the method of polling the ADBZ bit in the SADC0 register is used to detect when the conversion cycle is complete, whereas in the second example, the A/D interrupt is used to determine when the conversion is complete.

Example: using an ADBZ polling method to detect the end of conversion

```

clr ADE                ; disable ADC interrupt
mov a,03H              ; select fsys/8 as A/D clock
mov SADC1,a            ; A/D input signal comes from external channel
                        ; select VREF pin as A/D reference voltage source

mov a,80h
mov PAS0,a            ; set PAS0 to configure pin VREF
mov a,02h
mov PAS1,a            ; set PAS1 to configure pin AN0
mov a,20h
mov SADC0,a           ; enable the A/D converter and connect AN0 channel to A/D converter
:
start_conversion:
clr START              ; high pulse on start bit to initiate conversion
set START              ; reset A/D
clr START              ; start A/D
polling_EOC:
sz ADBZ                ; poll the SADC0 register ADBZ bit to detect end of A/D conversion
jmp polling_EOC        ; continue polling
mov a,SADOL             ; read low byte conversion result value
mov SADOL_buffer,a     ; save result to user defined register
mov a,SADOH             ; read high byte conversion result value
mov SADOH_buffer,a     ; save result to user defined register
:
:
jmp start_conversion   ; start next A/D conversion

```

Example: using the interrupt method to detect the end of conversion

```
clr ADE                ; disable ADC interrupt
mov a,03H              ; select fsys/8 as A/D clock
mov SADC1,a            ; A/D input signal comes from external channel
                        ; select VREF pin as A/D reference voltage source

mov a,80h
mov PAS0,a             ; set PAS0 to configure pin VREF
mov a,02h
mov PAS1,a             ; set PAS1 to configure pin AN0
mov a,20h
mov SADC0,a            ; enable the A/D converter and connect AN0 channel to A/D converter
:
start_conversion:
clr START              ; high pulse on START bit to initiate conversion
set START              ; reset A/D
clr START              ; start A/D
clr ADF                ; clear ADC interrupt request flag
set ADE                ; enable ADC interrupt
set EMI                ; enable global interrupt
:
:
; ADC interrupt service routine
ADC_ISR:
mov acc_stack,a        ; save ACC to user defined memory
mov a,STATUS
mov status_stack,a     ; save STATUS to user defined memory
:
:
mov a,SADOL             ; read low byte conversion result value
mov SADOL_buffer,a     ; save result to user defined register
mov a,SADOH             ; read high byte conversion result value
mov SADOH_buffer,a     ; save result to user defined register
:
:
EXIT_INT_ISR:
mov a,status_stack
mov STATUS,a           ; restore STATUS from user defined memory
mov a,acc_stack        ; restore ACC from user defined memory
reti
```

Touch Key Function

The device provides multiple touch key functions. The touch key function is fully integrated and requires no external components, allowing touch key functions to be implemented by the simple manipulation of internal registers.

Touch Key Structure

The touch keys are pin-shared with the I/O pins, with the desired function chosen via the corresponding selection register bits. Keys are organised into several groups, with each group known as a module and having a module number, M0 to M2. Each module is a fully independent set of four Touch Keys and each Touch Key has its own oscillator. Each module contains its own control logic circuits and register set. Examination of the register names will reveal the module number it is referring to.

Total Key Number	Touch Key Module	Touch Key
12	Mn (n=0~2)	M0
		M1
		M2
		KEY1~KEY4
		KEY5~KEY8
		KEY9~KEY12

Touch Key Structure

Touch Key Register Definition

Each touch key module, which contains four touch key functions, has its own suite registers. The following table shows the register set for each touch key module. The Mn within the register name refers to the Touch Key module number. The device has three Touch Key Modules.

Register Name	Description
TKTMR	Touch key time slot 8-bit counter preload register
TKC0	Touch key function Control register 0
TKC1	Touch key function Control register 1
TK16DL	Touch key function 16-bit counter low byte
TK16DH	Touch key function 16-bit counter high byte
TKMn16DL	Touch key module n 16-bit C/F counter low byte
TKMn16DH	Touch key module n 16-bit C/F counter high byte
TKMnROL	Touch key module n reference oscillator capacitor select low byte
TKMnROH	Touch key module n reference oscillator capacitor select high byte
TKMnC0	Touch key module n Control register 0
TKMnC1	Touch key module n Control register 1

Touch Key Function Register Definition (n=0~2)

Register Name	Bit							
	7	6	5	4	3	2	1	0
TKTMR	D7	D6	D5	D4	D3	D2	D1	D0
TKC0	—	TKRCOV	TKST	TKCFOV	TK16OV	TSCS	TK16S1	TK16S0
TKC1	—	—	—	—	—	—	TKFS1	TKFS0
TK16DL	D7	D6	D5	D4	D3	D2	D1	D0
TK16DH	D15	D14	D13	D12	D11	D10	D9	D8
TKMn16DL	D7	D6	D5	D4	D3	D2	D1	D0
TKMn16DH	D15	D14	D13	D12	D11	D10	D9	D8
TKMnROL	D7	D6	D5	D4	D3	D2	D1	D0
TKMnROH	—	—	—	—	—	—	D9	D8
TKMnC0	MnMXS1	MnMXS0	MnDFEN	MnFILEN	MnSOFC	MnSOF2	MnSOF1	MnSOF0
TKMnC1	MnTSS	—	MnROEN	MnKOEN	MnK4EN	MnK3EN	MnK2EN	MnK1EN

Touch Key Function Register List (n=0~2)

• **TKTMR Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0

D7~D0: Touch key time slot 8-bit counter preload register

The touch key time slot counter preload register is used to determine the touch key time slot overflow time. The time slot unit period is obtained by a 5-bit counter and is equal to 32 time slot clock cycles. Therefore, the time slot counter overflow time is equal to the following equation shown.

Time slot counter overflow time = $(256 - \text{TKTMR}[7:0]) \times 32 \text{ } t_{\text{TSC}}$, where t_{TSC} is the time slot counter clock period.

• **TKC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	TKRCOV	TKST	TKCFOV	TK16OV	TSCS	TK16S1	TK16S0
R/W	—	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	0	0	0	0	0	0	0

Bit 7

Unimplemented, read as “0”

Bit 6

TKRCOV: Touch key time slot counter overflow flag

0: No overflow occurs

1: Overflow occurs

This bit can be accessed by application program. When this bit is set by touch key time slot counter overflow, the corresponding touch key interrupt request flag will be set. However, if this bit is set by application program, the touch key interrupt request flag will not be affected. Therefore, this bit cannot be set by application program but must be cleared to 0 by application program.

If the module 0 time slot counter or the individual module time slot counter, selected by the TSCS bit, overflows, the TKRCOV bit and the Touch Key Interrupt request flag, TKMF, will be set and all module key oscillators and reference oscillators will automatically stop. The touch key module 16-bit C/F counter, touch key function 16-bit counter, 5-bit time slot unit period counter and 8-bit time slot counter will be automatically switched off.

- Bit 5 **TKST**: Touch key detection Start control
 0: Stopped or no operation
 0→1: Start detection
 In all modules the touch key module 16-bit C/F counter, touch key function 16-bit counter and 5-bit time slot unit period counter will automatically be cleared when this bit is cleared to zero. However, the 8-bit programmable time slot counter will not be cleared. When this bit is changed from low to high, the touch key module 16-bit C/F counter, touch key function 16-bit counter, 5-bit time slot unit period counter and 8-bit time slot counter will be switched on together with the key and reference oscillators to drive the corresponding counters.
- Bit 4 **TKCFOV**: Touch key module 16-bit C/F counter overflow flag
 0: No overflow occurs
 1: Overflow occurs
 This bit is set high by the touch key module 16-bit C/F counter overflow and must be cleared to 0 by application programs.
- Bit 3 **TK16OV**: Touch key function 16-bit counter overflow flag
 0: No overflow occurs
 1: Overflow occurs
 This bit is set high by the touch key function 16-bit counter overflow and must be cleared to 0 by application programs.
- Bit 2 **TSCS**: Touch key time slot counter select
 0: Each touch key module uses its own time slot counter
 1: All touch key modules use Module 0 time slot counter
- Bit 1~0 **TK16S1~TK16S0**: Touch key function 16-bit counter clock source select
 00: f_{sys}
 01: $f_{sys}/2$
 10: $f_{sys}/4$
 11: $f_{sys}/8$

• **TKC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	TKFS1	TKFS0
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	1	1

Bit 7~2 Unimplemented, read as “0”

Bit 1~0 **TKFS1~TKFS0**: Touch Key oscillator and Reference oscillator frequency selection
 00: 1MHz
 01: 3MHz
 10: 7MHz
 11: 11MHz

• **TK16DH/TK16DL – Touch Key Function 16-bit Counter Register Pair**

Register	TK16DH								TK16DL							
Bit	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
Name	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R	R	R	R	R	R	R	R	R	R	R	R	R	R	R	R
POR	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0

This register pair is used to store the touch key function 16-bit counter value. This 16-bit counter can be used to calibrate the reference or key oscillator frequency. When the touch key time slot counter overflows, this 16-bit counter will be stopped and the counter content will be unchanged. This register pair will be cleared to zero when the TKST bit is set low.

• **TKMn16DH/TKMn16DL – Touch Key Module n 16-bit C/F Counter Register Pair**

Register	TKMn16DH								TKMn16DL							
Bit	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
Name	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R	R	R	R	R	R	R	R	R	R	R	R	R	R	R	R
POR	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0

This register pair is used to store the touch key module n 16-bit C/F counter value. This 16-bit C/F counter will be stopped and the counter content will be kept unchanged when the touch key time slot counter overflows. This register pair will be cleared to zero when the TKST bit is set low.

• **TKMnROH/TKMnROL – Touch Key Module n Reference Oscillator Capacitor Select Register Pair**

Register	TKMnROH								TKMnROL							
Bit	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
R/W	—	—	—	—	—	—	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	—	—	—	—	—	0	0	0	0	0	0	0	0	0	0

This register pair is used to store the touch key module n reference oscillator capacitor value.

The reference oscillator internal capacitor value=(TKMnRO[9:0]×50pF)/1024

• **TKMnC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	MnMXS1	MnMXS0	MnDFEN	MnFILEN	MnSOFC	MnSOF2	MnSOF1	MnSOF0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~6 **MnMXS1~MnMXS0**: Multiplexer Key Selection

Bit	Touch Key Module Number		
MnMXS[1:0]	M0	M1	M2
00	KEY1	KEY5	KEY9
01	KEY2	KEY6	KEY10
10	KEY3	KEY7	KEY11
11	KEY4	KEY8	KEY12

Bit 5 **MnDFEN**: Touch key module n multi-frequency control

0: Disable

1: Enable

This bit is used to control the touch key oscillator frequency doubling function. When this bit is set to 1, the key oscillator frequency will be doubled.

Bit 4 **MnFILEN**: Touch key module n filter function control

0: Disable

1: Enable

Bit 3 **MnSOFC**: Touch key module n C/F oscillator frequency hopping function control select

0: Controlled by the MnSOF2~MnSOF0

1: Controlled by hardware circuit

This bit is used to select the touch key oscillator frequency hopping function control method. When this bit is set to 1, the key oscillator frequency hopping function is controlled by the hardware circuit regardless of the MnSOF2~MnSOF0 bits value.

Bit 2~0 **MnSOF2~MnSOF0**: Touch key module n Reference and Key oscillators hopping frequency selection (MnSOFC=0)

000: 1.020MHz
001: 1.040MHz
010: 1.059MHz
011: 1.074MHz
100: 1.085MHz
101: 1.099MHz
110: 1.111MHz
111: 1.125MHz

These bits are used to select the touch key oscillator frequency for the hopping function. Note that these bits are only available when the MnSOFC bit is cleared to 0.

The frequency mentioned here will be changed when the external or internal capacitor has different values. If the touch key operates at 1MHz frequency, users can adjust the frequency in scale when any other frequency is selected.

• **TKMnC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	MnTSS	—	MnROEN	MnKOEN	MnK4EN	MnK3EN	MnK2EN	MnK1EN
R/W	R/W	—	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	—	0	0	0	0	0	0

Bit 7 **MnTSS**: Touch key module n time slot counter clock source selection
0: Touch key module n reference oscillator
1: $f_{SYS}/4$

Bit 6 Unimplemented, read as “0”

Bit 5 **MnROEN**: Touch key module n Reference oscillator enable control
0: Disable
1: Enable

Bit 4 **MnKOEN**: Touch key module n Key oscillator enable control
0: Disable
1: Enable

Bit 3 **MnK4EN**: Touch key module n Key 4 enable control

MnK4EN	Touch Key Module n – Mn		
	M0	M1	M2
0: Disable	I/O or other functions		
1: Enable	KEY4	KEY8	KEY12

Bit 2 **MnK3EN**: Touch key module n Key 3 enable control

MnK3EN	Touch Key Module n – Mn		
	M0	M1	M2
0: Disable	I/O or other functions		
1: Enable	KEY3	KEY7	KEY11

Bit 1 **MnK2EN**: Touch key module n Key 2 enable control

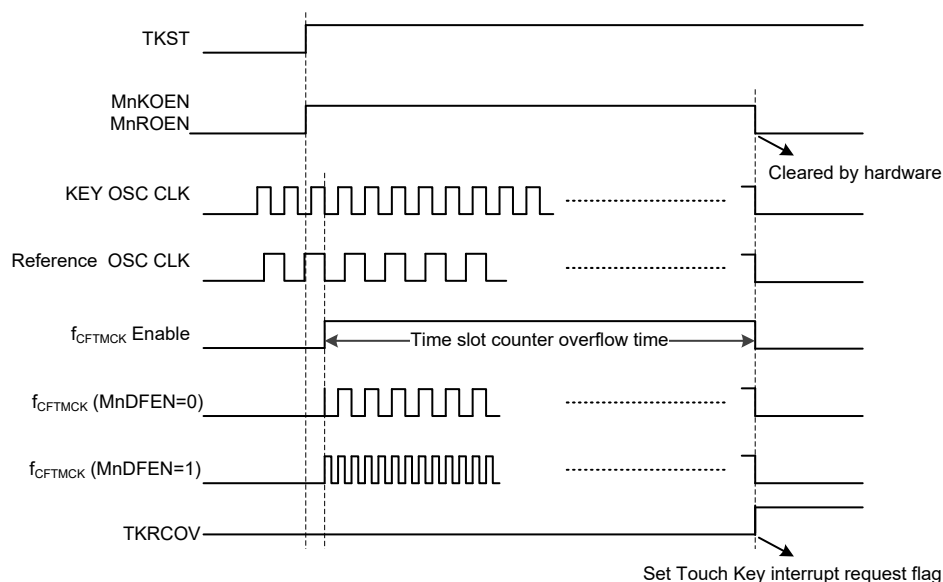
MnK2EN	Touch Key Module n – Mn		
	M0	M1	M2
0: Disable	I/O or other functions		
1: Enable	KEY2	KEY6	KEY10

Bit 0 **MnK1EN**: Touch key module n Key 1 enable control

MnK1EN	Touch Key Module n – Mn		
	M0	M1	M2
0: Disable	I/O or other functions		
1: Enable	KEY1	KEY5	KEY9

Touch Key Operation

When a finger touches or is in proximity to a touch pad, the capacitance of the pad will increase. By using this capacitance variation to change slightly the frequency of the internal sense oscillator, touch actions can be sensed by measuring these frequency changes. Using an internal programmable divider the reference clock is used to generate a fixed time period. By counting a number of generated clock cycles from the sense oscillator during this fixed time period touch key actions can be determined.



Touch Key Scan Mode Timing Diagram

Each touch key module contains four touch key inputs which are shared with logical I/O pins, and the desired function is selected using register bits. Each touch key has its own independent sense oscillator. Therefore, there are four sense oscillators within each touch key module.

During this reference clock fixed interval, the number of clock cycles generated by the sense oscillator is measured, and it is this value that is used to determine if a touch action has been made or not. At the end of the fixed reference clock time interval a Touch Key interrupt signal will be generated.

Using the TSCS bit in the TKC0 register can select the module 0 time slot counter as the time slot counter for all modules. All modules use the same started signal, TKST, in the TKC0 register. The touch key module 16-bit C/F counter, touch key function 16-bit counter, 5-bit time slot unit period counter in all modules will be automatically cleared when the TKST bit is cleared to zero, but the 8-bit programmable time slot counter will not be cleared. The overflow time is set by user. When the TKST bit changes from low to high, the 16-bit C/F counter, touch key function 16-bit counter, 5-bit time slot unit period counter and 8-bit time slot timer counter will be automatically switched on.

The key oscillator and reference oscillator in all modules will be automatically stopped and the 16-bit C/F counter, touch key function 16-bit counter, 5-bit time slot unit period counter and 8-bit time slot

timer counter will be automatically switched off when the time slot counter overflows. The clock source for the time slot counter is sourced from the reference oscillator or $f_{SYS}/4$ which is selected using the MnTSS bit in the TKMnC1 register. The reference oscillator and key oscillator will be enabled by setting the MnROEN bit and MnKOEN bits in the TKMnC1 register.

When the time slot counter in all the touch key modules or in the touch key module 0 overflows, an actual touch key interrupt will take place. The touch keys mentioned here are the keys which are enabled.

Each touch key module consists of four touch keys, KEY1~KEY4 are contained in module 0, KEY5~KEY8 are contained in module 1, KEY9~KEY12 are contained in module 2. Each touch key module has an identical structure.

Touch Key Interrupt

The touch key only has a single interrupt, when the time slot counter in all the touch key modules or in the touch key module 0 overflows, an actual touch key interrupt will take place. The touch keys mentioned here are the keys which are enabled. The 16-bit C/F counter, 16-bit counter, 5-bit time slot unit period counter and 8-bit time slot counter in all modules will be automatically cleared. More details regarding the touch key interrupt is located in the interrupt section of the datasheet.

Programming Considerations

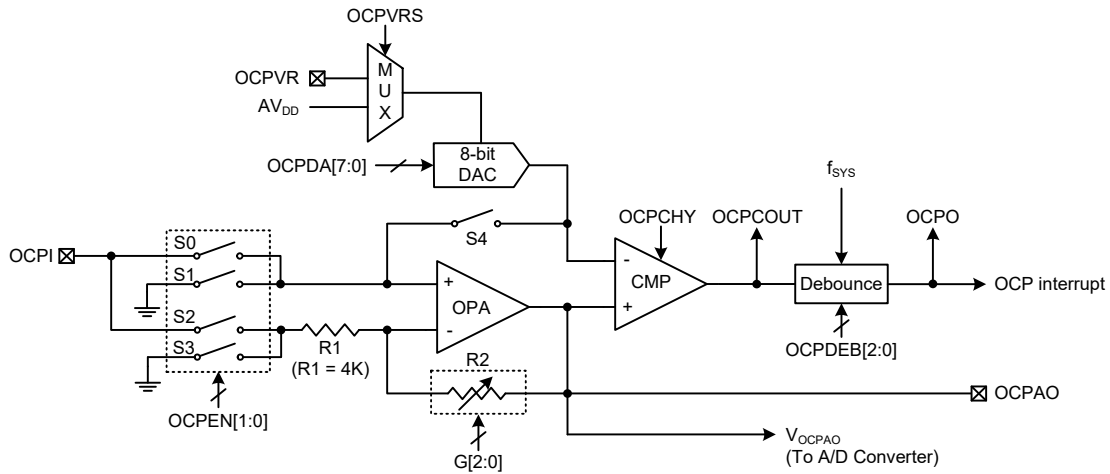
After the relevant registers are set, the touch key detection process is initiated by changing the TKST bit from low to high. This will enable and synchronise all relevant oscillators. The TKRCOV flag which is the time slot counter flag will go high when the counter overflows. When this happens an interrupt signal will be generated. As the TKRCOV flag will not be automatically cleared, it has to be cleared by the application program.

The TKCFOV flag which is the 16-bit C/F counter overflow flag will go high when any of the Touch Key Module 16-bit C/F counter overflows. As this flag will not be automatically cleared, it has to be cleared by the application program. The TK16OV flag which is the 16-bit counter overflow flag will go high when the 16-bit counter overflows. As this flag will not be automatically cleared, it has to be cleared by the application program.

When the external touch key size and layout are defined, their related capacitances will then determine the sensor oscillator frequency.

Over Current Protection – OCP

The device includes an over current protection function which provides a protection mechanism for applications. The current on the OCPI pin is converted to a relevant voltage level according to the current value using the OCP operational amplifier. It is then compared with a reference voltage generated by an 8-bit D/A converter. When an over current event occurs, an OCP interrupt will be generated if the corresponding interrupt control is enabled.



Note: As the OCP function relevant external pins are pin-shared with general I/O or other functions, before using the OCP function, make sure the corresponding pin-shared function registers be set properly.

Over Current Protection Circuit

Over Current Protection Operation

The OCP circuit is used to prevent the input current from exceeding a specific level. The current on the OCPI pin is converted to a voltage and then amplified by the OCP operational amplifier with a programmable gain from 1 to 50 selected by the G2~G0 bits in the OCPC1 register. This is known as a Programmable Gain Amplifier or PGA. This PGA can also be configured to operate in the non-inverting, inverting or input offset calibration mode determined by the OCPEN1~OCPEN0 bits in the OCPC0 register. After the current is converted and amplified to a specific voltage level, it will be compared with a reference voltage provided by an 8-bit D/A converter. The 8-bit D/A converter reference voltage can be supplied by AVDD or external OCPVR pin, which is selected by the OCPVRS bit in the OCPC0 register. The comparator output, OCPCHY, will first be filtered with a certain de-bounce time period selected by the OCPDEB2~OCPDEB0 bits in the OCPC1 register. Then a filtered OCP digital comparator output, OCPO, is obtained to indicate whether an over current condition occurs or not. The OCPO bit will be set to 1 if an over current condition occurs. Otherwise, the OCPO bit is zero. Once an over current event occurs, i.e., the converted voltage of the OCP input current is greater than the reference voltage, the corresponding interrupt will be generated if the relevant interrupt control bit is enabled.

Note that the debounce clock, f_{DEB}, comes from the system clock, f_{sys}. The operational amplifier output voltage can be directly output on the OCPAO pin, and also can be read out by the A/D converter through an A/D internal input channel. The D/A converter output voltage is controlled by the OCPDA register and the D/A converter output is defined as below:

$$\text{DAC } V_{\text{OUT}} = (\text{D/A converter reference voltage} / 256) \times \text{OCPDA}[7:0]$$

Over Current Protection Registers

Overall operation of the over current protection is controlled using several registers. The OCPDA register is used to provide the reference voltage for the over current protection circuit. The OCPOCAL and OCPCCAL registers are used to cancel out the operational amplifier and comparator input offset. The OCPC0 and OCPC1 registers are control registers which control the OCP function, D/A converter reference voltage selection, PGA gain selection, comparator de-bounce time together with the hysteresis function.

Register Name	Bit							
	7	6	5	4	3	2	1	0
OCPC0	OCPEN1	OCPEN0	OCPVRS	OCPCY	—	—	—	OCPO
OCPC1	—	—	G2	G1	G0	OCPDEB2	OCPDEB1	OCPDEB0
OCPCA	D7	D6	D5	D4	D3	D2	D1	D0
OCPOCAL	OCPOOFM	OCPORSP	OCPOOF5	OCPOOF4	OCPOOF3	OCPOOF2	OCPOOF1	OCPOOF0
OCPCCAL	OCPCOUT	OCPCOFM	OCPCRS	OCPCOF4	OCPCOF3	OCPCOF2	OCPCOF1	OCPCOF0

OCP Register List

• OCPC0 Register

Bit	7	6	5	4	3	2	1	0
Name	OCPEN1	OCPEN0	OCPVRS	OCPCY	—	—	—	OCPO
R/W	R/W	R/W	R/W	R/W	—	—	—	R
POR	0	0	0	0	—	—	—	0

- Bit 7~6 **OCPEN1~OCPEN0**: OCP function operating mode selection
 00: OCP function is disabled, S1 and S3 on, S0 and S2 off
 01: Non-inverting mode, S0 and S3 on, S1 and S2 off
 10: Inverting mode, S1 and S2 on, S0 and S3 off
 11: Calibration mode, S1 and S3 on, S0 and S2 off
- Bit 5 **OCPVRS**: OCP D/A converter reference voltage selection
 0: From AV_{DD}
 1: From OCPVR pin
- Bit 4 **OCPCY**: OCP comparator hysteresis function control
 0: Disable
 1: Enable
- Bit 3~1 Unimplemented, read as “0”
- Bit 0 **OCPO**: OCP digital output bit
 0: No over current situation occurred
 1: Over current situation occurred

• **OCPC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	G2	G1	G0	OCPDEB2	OCPDEB1	OCPDEB0
R/W	—	—	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	—	0	0	0	0	0	0

Bit 7~6 Unimplemented, read as “0”

Bit 5~3 **G2~G0**: PGA R2/R1 ratio selection

000: Unity gain buffer (non-inverting mode) or R2/R1=1 (inverting mode)

001: R2/R1=5

010: R2/R1=10

011: R2/R1=15

100: R2/R1=20

101: R2/R1=30

110: R2/R1=40

111: R2/R1=50

These bits are used to select the R2/R1 ratio to obtain various gain values for inverting and non-inverting mode. The calculating formula of the PGA gain for the inverting and non-inverting mode is described in the “Input Voltage Range” section.

Bit 2~0 **OCPDEB2~OCPDEB0**: OCP output filter debounce time selection

000: Bypass, without debounce

001: (1~2)×t_{DEB}

010: (3~4)×t_{DEB}

011: (7~8)×t_{DEB}

100: (15~16)×t_{DEB}

101: (31~32)×t_{DEB}

110: (63~64)×t_{DEB}

111: (127~128)×t_{DEB}

Note: t_{DEB}=1/f_{SYS}.

• **OCPCA Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7~0 **D7~D0**: OCP D/A converter output voltage control bits

DAC V_{OUT}=(D/A converter reference voltage/256)×OCPCA[7:0]

• **OCPOCAL Register**

Bit	7	6	5	4	3	2	1	0
Name	OCPOOFM	OCPORSP	OCPOOF5	OCPOOF4	OCPOOF3	OCPOOF2	OCPOOF1	OCPOOF0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	1	0	0	0	0	0

Bit 7 **OCPOOFM**: OCP Operational Amplifier Input Offset Calibration Mode enable control

0: Input Offset Calibration Mode disabled

1: Input Offset Calibration Mode enabled

This bit is used to control the OCP operational amplifier input offset Calibration function. The OCPEN1 and OCPEN0 bits must first be set to “11” and then the OCPOOFM bit must be set to 1 followed by the OCPCOFM bit being setting to 0, then the operational amplifier input offset Calibration mode will be enabled. Refer to the “Operational Amplifier Input Offset Calibration” section for the detailed offset Calibration procedures.

- Bit 6 **OCPORSP**: OCP Operational Amplifier Input Offset Voltage Calibration Reference selection
 0: Select negative input as the reference input
 1: Select positive input as the reference input
- Bit 5~0 **OCPOOF5~OCPOOF0**: OCP Operational Amplifier Input Offset Voltage Calibration value
 This 6-bit field is used to perform the operational amplifier input offset Calibration operation and the value for the OCP operational amplifier input offset Calibration can be restored into this bit field. More detailed information is described in the “Operational Amplifier Input Offset Calibration” section.

• **OCPCCAL Register**

Bit	7	6	5	4	3	2	1	0
Name	OCPCOUT	OCPCOFM	OCPCRSP	OCPCOF4	OCPCOF3	OCPCOF2	OCPCOF1	OCPCOF0
R/W	R	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	1	0	0	0	0

- Bit 7 **OCPCOUT**: OCP Comparator Output, positive logic (read only)
 0: Positive input voltage < Negative input voltage
 1: Positive input voltage > Negative input voltage
 This bit is used to indicate whether the positive input voltage is greater than the negative input voltage when the OCP operates in the input offset Calibration mode. If the OCPCOUT is set to 1, the positive input voltage is greater than the negative input voltage. Otherwise, the positive input voltage is less than the negative input voltage.
- Bit 6 **OCPCOFM**: OCP Comparator Input Offset Calibration Mode enable control
 0: Input Offset Calibration Mode disabled
 1: Input Offset Calibration Mode enabled
 This bit is used to control the OCP comparator input offset Calibration function. The OCPEN1 and OCPEN0 bits must first be set to “11” and then the OCPCOFM bit must be set to 1 followed by the OCPOOFM bit being setting to 0, then the comparator input offset calibration mode will be enabled. Refer to the “Comparator Input Offset Calibration” section for the detailed offset calibration procedures.
- Bit 5 **OCPCRSP**: OCP Comparator Input Offset Calibration Reference Input select
 0: Select negative input as the reference input
 1: Select positive input as the reference input
- Bit 4~0 **OCPCOF4~OCPCOF0**: OCP Comparator Input Offset Calibration value
 This 5-bit field is used to perform the comparator input offset calibration operation and the value for the OCP comparator input offset calibration can be restored into this bit field. More detailed information is described in the “Comparator Input Offset Calibration” section.

Input Voltage Range

Together with different PGA operating modes, the input voltage on the OCP pin can be positive or negative for flexible operation. The PGA output for the positive or negative input voltage is calculated based on different formulas and described by the following.

- For input voltages $V_{IN} > 0$, the PGA operates in the non-inverting mode and the PGA output is obtained using the formula below:

$$V_{OUT} = (1 + R_2/R_1) \times V_{IN}$$

- When the PGA operates in the non-inverting mode by setting the OCPEN[1:0] to “01” with unity gain select by setting the G[2:0] to “000”, the PGA will act as a unit-gain buffer whose output is equal to V_{IN} .

$$V_{OUT} = V_{IN}$$

- For input voltages $0 > V_{IN} > -0.2V$, the PGA operates in the inverting mode and the PGA output is obtained using the formula below. Note that if the input voltage is negative, it cannot be lower than $-0.2V$ which will result in current leakage.

$$V_{OUT} = -(R2/R1) \times V_{IN}$$

Input Offset Calibration

The OCP circuit has four operating modes controlled by the OCPEN[1:0] bit field, one of them is calibration mode. In calibration mode, operational amplifier and comparator offset can be calibrated.

Operational Amplifier Input Offset Calibration

- Step1. Set OCPEN[1:0]=11, OCPOOFM=1, OCPCOFM=0 and OCPORSP=1, the OCP will operate in the operational amplifier input offset calibration mode.
- Step2. Set OCPOOF[5:0]=000000 and then read the OCPCOUT bit.
- Step3. Increase the OCPOOF[5:0] value by 1 and then read the OCPCOUT bit.
- If the OCPCOUT bit state has not changed, then repeat Step 3 until the OCPCOUT bit state has changed.
- If the OCPCOUT bit state has changed, record the OCPOOF value as V_{OOS1} and then go to Step 4.
- Step4. Set OCPOOF[5:0]=111111 and read the OCPCOUT bit.
- Decrease the OCPOOF[5:0] value by 1 and then read the OCPCOUT bit.
- If the OCPCOUT bit state has not changed, then repeat Step 5 until the OCPCOUT bit state has changed.
- If the OCPCOUT bit state has changed, record the OCPOOF value as V_{OOS2} and then go to Step 6.
- Step5. Restore the operational amplifier input offset calibration value V_{OOS} into the OCPOOF[5:0] bit field. The offset Calibration procedure is now finished.

$$\text{Where } V_{OOS} = (V_{OOS1} + V_{OOS2}) / 2$$

Note: S4 is off. In this mode, the operational amplifier outputs to OCPCOUT bypassing the comparator.

Comparator Input Offset Calibration

- Step1. Set OCPEN[1:0]=11, OCPCOFM=1 and OCPOOFM=0, the OCP will now operate in the comparator input offset calibration mode.
- Step2. Set OCPCOF[4:0]=00000 and read the OCPCOUT bit.
- Step3. Increase the OCPCOF[4:0] value by 1 and then read the OCPCOUT bit.
- If the OCPCOUT bit state has not changed, then repeat Step 3 until the OCPCOUT bit state has changed.
- If the OCPCOUT bit state has changed, record the OCPCOF value as V_{COS1} and then go to Step 4.
- Step4. Set OCPCOF[4:0]=11111 and then read the OCPCOUT bit.
- Step5. Decrease the OCPCOF[4:0] value by 1 and then read the OCPCOUT bit.
- If the OCPCOUT bit state has not changed, then repeat Step 5 until the OCPCOUT bit state has changed.
- If the OCPCOUT bit state has changed, record the OCPCOF value as V_{COS2} and then go to Step 6.

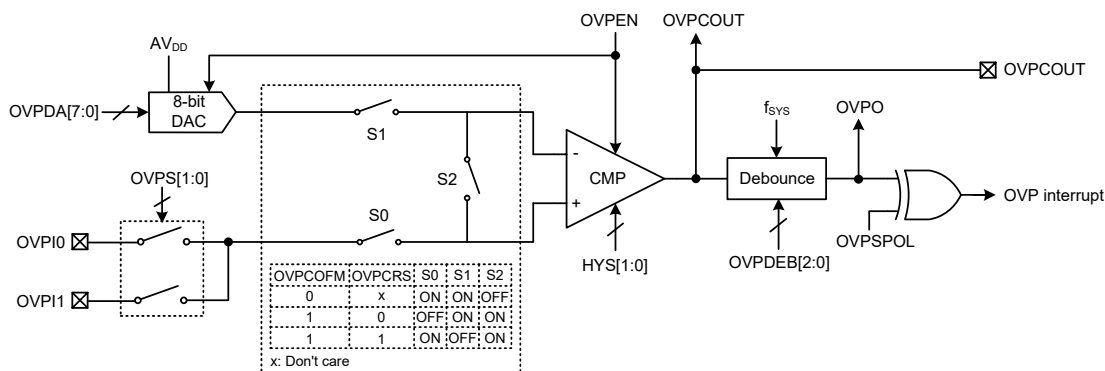
Step6. Restore the comparator input offset calibration value V_{COS} into the OCPCOF[4:0] bit field.
 The offset Calibration procedure is now finished.

$$\text{Where } V_{COS} = (V_{COS1} + V_{COS2}) / 2$$

Note: S4 is on and the D/A converter is off. This situation is only available for comparator calibration procedure. In the normal operation mode, S4 is off.

Over Voltage Protection – OVP

The device includes an over voltage protection function which provides a protection mechanism for applications. The input voltage on the OVPI0 or OVPI1 pin is compared with a reference voltage generated by an 8-bit D/A converter. When a preset voltage condition occurs, an OVP interrupt will be generated if the corresponding interrupt control is enabled.



Note: As the OVP function relevant external pins are pin-shared with general I/O or other functions, before using the OVP function, make sure the corresponding pin-shared function registers be set properly.

Over Voltage Protection Circuit

Over Voltage Protection Operation

The OVP circuit is used to prevent the input voltage from being in an unexpected level range. The voltage can be sourced from the OVPI0 or OVPI1 pin which is determined by the OVPS1~OVPS0 bits in the OVPC2 register. The selected OVP input voltage is compared with a reference voltage provided by an 8-bit D/A converter. The 8-bit D/A converter reference input signal is supplied by the positive power supply, AV_{DD} . The comparator output, OVPCOUT, will first be filtered with a certain de-bounce time period selected by the OVPDEB2~OVPDEB0 bits in the OVPC0 register. Then a filtered OVP digital comparator output, OVPO, is obtained to indicate whether a user-defined voltage condition occurs or not.

If the OVSPOL bit is cleared to 0 and the comparator inputs force the OVPO bit to change from 0 to 1, or if the OVSPOL bit is set to 1 and the comparator inputs force the OVPO bit changes from 1 to 0, the corresponding interrupt will be generated if the relevant interrupt control bit is enabled. Therefore, the OVSPOL bit must be properly configured according to user's application requirements. The comparator in the OVP circuit also has hysteresis function controlled by the HYS1~HYS0 bits.

Note that the debounce clock, f_{DEB} , comes from the system clock, f_{SYS} . The D/A converter output voltage is controlled by the OVPDA register, the D/A converter output is defined as below:

$$\text{DAC } V_{OUT} = (\text{D/A converter reference voltage} / 256) \times \text{OVPDA}[7:0]$$

Over Voltage Protection Registers

Overall operation of the OVP function is controlled using several registers. The OVPCA register is used to provide the reference voltage for the OVP circuit. The OVPC0~OVPC2 control registers are used to control the OVP function, D/A converter reference voltage selection, comparator de-bounce time selection, comparator hysteresis function and comparator output polarity control, etc.

Register Name	Bit							
	7	6	5	4	3	2	1	0
OVPC0	OVPO	OVSPOL	OVPEN	—	—	OVPEB2	OVPEB1	OVPEB0
OVPC1	OVPCOUT	OVPCOFM	OVPCRS	OVPCOF4	OVPCOF3	OVPCOF2	OVPCOF1	OVPCOF0
OVPC2	—	—	—	—	HYS1	HYS0	OVPS1	OVPS0
OVPCA	D7	D6	D5	D4	D3	D2	D1	D0

OVP Register List

• OVPC0 Register

Bit	7	6	5	4	3	2	1	0
Name	OVPO	OVSPOL	OVPEN	—	—	OVPEB2	OVPEB1	OVPEB0
R/W	R	R/W	R/W	—	—	R/W	R/W	R/W
POR	0	0	0	—	—	0	0	0

- Bit 7 **OVPO**: OVP comparator output bit after debounce
0: Positive input voltage < negative input voltage
1: Positive input voltage > negative input voltage
- Bit 6 **OVSPOL**: OVP debounced output signal polarity control
0: Non-invert
1: Invert
This bit will determine the OVP interrupt occurrence condition when the OVPO bit changes state from low to high or high to low.
- Bit 5 **OVPEN**: OVP function enable control
0: Disable
1: Enable
If the OVPEN bit is cleared to 0, the over voltage protection function is disabled and no power will be consumed. This results in the comparator and D/A converter of the OVP both being switched off.
- Bit 4~3 Unimplemented, read as “0”
- Bit 2~0 **OVPEB2~OVPEB0**: OVP comparator debounce time selection
000: No debounce
001: $(1\sim2) \times t_{DEB}$
010: $(3\sim4) \times t_{DEB}$
011: $(7\sim8) \times t_{DEB}$
100: $(15\sim16) \times t_{DEB}$
101: $(31\sim32) \times t_{DEB}$
110: $(63\sim64) \times t_{DEB}$
111: $(127\sim128) \times t_{DEB}$
Note: $t_{DEB} = 1/f_{SYS}$.

• **OVP C1 Register**

Bit	7	6	5	4	3	2	1	0
Name	OVP COUT	OVP C OFM	OVP CRS	OVP C OF4	OVP C OF3	OVP C OF2	OVP C OF1	OVP C OF0
R/W	R	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	1	0	0	0	0

- Bit 7 **OVP COUT**: OVP comparator output bit
0: Positive input voltage < negative input voltage
1: Positive input voltage > negative input voltage
- Bit 6 **OVP C OFM**: OVP comparator operating mode selection
0: Normal operation mode
1: Input offset voltage calibration mode
This bit is used to select the OVP comparator operating mode. To select the comparator input offset cancellation mode, the OVP C OFM bit must be set to 1. Refer to the “Comparator Input Offset Cancellation” section for the detailed offset cancellation procedures.
- Bit 5 **OVP CRS**: OVP comparator input offset voltage calibration reference selection
0: Input reference voltage comes from negative input
1: Input reference voltage comes from positive input
- Bit 4~0 **OVP C OF4~OVP C OF0**: OVP comparator input offset voltage calibration control
This 5-bit field is used to perform the comparator input offset cancellation operation and the value for the OVP comparator input offset cancellation can be restored into this bit field. More detailed information is described in the “Comparator Input Offset Cancellation” section.

• **OVP C2 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	HYS1	HYS0	OVPS1	OVPS0
R/W	—	—	—	—	R/W	R/W	R/W	R/W
POR	—	—	—	—	0	0	0	0

- Bit 7~4 Unimplemented, read as “0”
- Bit 3~2 **HYS1~HYS0**: OVP comparator hysteresis voltage window control
Refer to “Over Voltage Protection Electrical Characteristics” table for details.
- Bit 1~0 **OVPS1~OVPS0**: OVP input source selection
00: Not choose
01: OVPI0
10: OVPI1
11: Not choose

• **OVP D A Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7~0 **D7~D0**: OVP D/A converter output voltage control bits
 $DAC V_{OUT} = (D/A \text{ converter reference voltage} / 256) \times OVPDA[7:0]$

Comparator Input Offset Cancellation

As the OVP inputs are pin-shared with other pin functions, they should be configured as OVP input pin functions first by configuring the relevant pin-shared control bits. It should be noted that before offset calibration, the hysteresis voltage should be zero by clearing the HYS[1:0] bits to “00”. For comparator input offset calibration, the procedures are summarised as the following.

Step1: Set OVPCOFM=1, OVPCRS=1, the OVP is now in the comparator offset calibration mode, S0 and S2 are on. To make sure V_{OS} as minimised as possible after calibration, the input reference voltage in calibration mode should be the same as input DC operating voltage in normal mode operation.

Step2: Set OVPCOF[4:0]=00000 and then read the OVPCOUT bit status.

Step3: Increase the OVPCOF[4:0] value by 1 and then read the OVPCOUT bit status.

If the OVPCOUT state has not changed, repeat Step3 until the OVPCOUT bit state has changed.

If the OVPCOUT state has changed, record the OVPCOF[4:0] value as V_{OS1} and then go to Step4.

Step4: Set OVPCOF[4:0]=11111 and then read the OVPCOUT bit status.

Step5: Decrease the OVPCOF[4:0] value by 1 and then read the OVPCOUT bit status.

If the OVPCOUT state has not changed, repeat Step5 until the OVPCOUT bit state has changed.

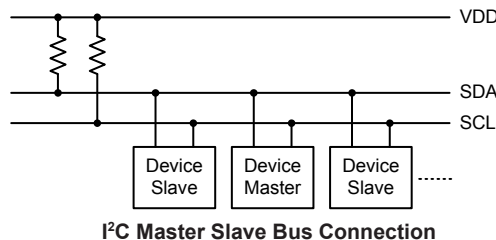
If the OVPCOUT state has changed, record the OVPCOF[4:0] value as V_{OS2} and then go to Step6.

Step6: Restore $V_{OS}=(V_{OS1}+V_{OS2})/2$ to the OVPCOF[4:0] bits. The calibration is finished.

If $(V_{OS1}+V_{OS2})/2$ is not integral, discard the decimal. Residue $V_{OS}=V_{OUT} - V_{IN}$.

I²C Interface

The I²C interface is used to communicate with external peripheral devices such as sensors, EEPROM memory etc. Originally developed by Philips, it is a two line low speed serial interface for synchronous serial data transfer. The advantage of only two lines for communication, relatively simple communication protocol and the ability to accommodate multiple devices on the same bus has made it an extremely popular interface type for many applications.

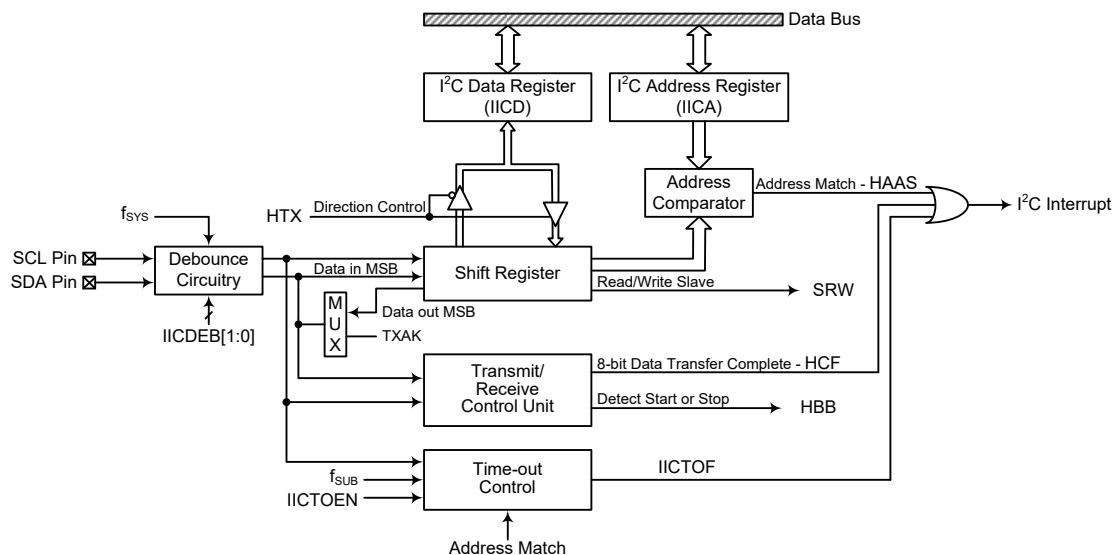


I²C Interface Operation

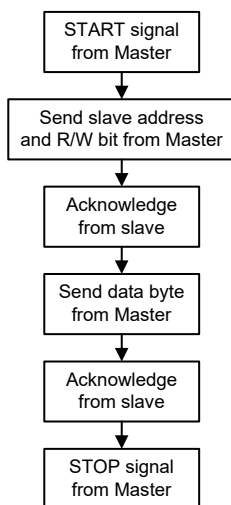
The I²C serial interface is a two line interface, a serial data line, SDA, and serial clock line, SCL. As many devices may be connected together on the same bus, their outputs are both open drain types. For this reason it is necessary that external pull-high resistors are connected to these outputs. Note that no chip select line exists, as each device on the I²C bus is identified by a unique address which will be transmitted and received on the I²C bus.

When two devices communicate with each other on the bidirectional I²C bus, one is known as the master device and one as the slave device. Both master and slave can transmit and receive data. However, it is the master device that has overall control of the bus. For this device, which only operate in slave mode, there are two methods of transferring data on the I²C bus, the slave transmit mode and the slave receive mode. The pull-high control function pin-shared with SCL/SDA pin is

still applicable even if the I²C device is activated and the related internal pull-high function could be controlled by its corresponding pull-high control register.



I²C Block Diagram



I²C Interface Operation

The IICDEB1 and IICDEB0 bits determine the debounce time of the I²C interface. This uses the internal clock to in effect add a debounce time to the external clock to reduce the possibility of glitches on the clock line causing erroneous operation. The debounce time, if selected, can be chosen to be either 2 or 4 system clocks. To achieve the required I²C data transfer speed, there exists a relationship between the system clock, f_{SYS} , and the I²C debounce time. For either the I²C Standard or Fast mode operation, users must take care of the selected system clock frequency and the configured debounce time to match the criterion shown in the following table.

I ² C Debounce Time Selection	I ² C Standard Mode (100kHz)	I ² C Fast Mode (400kHz)
No Debounce	$f_{\text{SYS}} > 2\text{MHz}$	$f_{\text{SYS}} > 5\text{MHz}$
2 system clock debounce	$f_{\text{SYS}} > 4\text{MHz}$	$f_{\text{SYS}} > 10\text{MHz}$
4 system clock debounce	$f_{\text{SYS}} > 8\text{MHz}$	$f_{\text{SYS}} > 20\text{MHz}$

I²C Minimum f_{SYS} Frequency Requirements

I²C Registers

There are three control registers associated with the I²C bus, IICC0, IICC1 and IICTOC, one address register IICA and one data register, IICD.

Register Name	Bit							
	7	6	5	4	3	2	1	0
IICC0	—	—	—	—	IICDEB1	IICDEB0	IICEN	—
IICC1	HCF	HAAS	HBB	HTX	TXAK	SRW	IAMWU	RXAK
IICD	D7	D6	D5	D4	D3	D2	D1	D0
IICA	IICA6	IICA5	IICA4	IICA3	IICA2	IICA1	IICA0	—
IICTOC	IICTOEN	IICTOF	IICTOS5	IICTOS4	IICTOS3	IICTOS2	IICTOS1	IICTOS0

I²C Register List

I²C Data Register

The IICD register is used to store the data being transmitted and received. Before the device writes data to the I²C bus, the actual data to be transmitted must be placed in the IICD register. After the data is received from the I²C bus, the device can read it from the IICD register. Any transmission or reception of data from the I²C bus must be made via the IICD register.

• IICD Register

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	x	x	x	x	x	x	x	x

“x”: unknown

Bit 7~0 **D7~D0**: I²C data register bit 7 ~ bit 0

I²C Address Register

The IICA register is the location where the 7-bit slave address of the slave device is stored. Bits 7~1 of the IICA register define the device slave address. Bit 0 is not defined. When a master device, which is connected to the I²C bus, sends out an address, which matches the slave address in the IICA register, the slave device will be selected.

• IICA Register

Bit	7	6	5	4	3	2	1	0
Name	IICA6	IICA5	IICA4	IICA3	IICA2	IICA1	IICA0	—
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	—
POR	0	0	0	0	0	0	0	—

Bit 7~1 **IICA6~IICA0**: I²C slave address
IICA6~IICA0 is the I²C slave address bit 6 ~ bit 0.

Bit 0 Unimplemented, read as “0”

I²C Control Registers

There are three control registers for the I²C interface, IICC0, IICC1 and IICTOC. The IICC0 register is used to control the enable/disable function and to set the data transmission clock frequency. The IICC1 register contains the relevant flags which are used to indicate the I²C communication status. Another register, IICTOC, is used to control the I²C time-out function and is described in the corresponding section.

• IICC0 Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	IICDEB1	IICDEB0	IICEN	—
R/W	—	—	—	—	R/W	R/W	R/W	—
POR	—	—	—	—	0	0	0	—

Bit 7~4 Unimplemented, read as “0”

Bit 3~2 **IICDEB1~IICDEB0**: I²C Debounce Time Selection

- 00: No debounce
- 01: 2 system clock debounce
- 1x: 4 system clock debounce

Note that the I²C debounce circuit will operate normally if the system clock, f_{sys} , is derived from the f_{H} clock or the IAMWU bit is equal to 0. Otherwise, the debounce circuit will have no effect and be bypassed.

Bit 1 **IICEN**: I²C Enable Control

- 0: Disable
- 1: Enable

The bit is the overall on/off control for the I²C interface. When the IICEN bit is cleared to zero to disable the I²C interface, the SDA and SCL lines will lose their I²C function and the I²C operating current will be reduced to a minimum value. When the bit is high the I²C interface is enabled. The I²C configuration option must have first enabled the I²C interface for this bit to be effective. If the IICEN bit changes from low to high, the contents of the I²C control bits such as HTX and TXAK will remain at the previous settings and should therefore be first initialised by the application program while the relevant I²C flags such as HCF, HAAS, HBB, SRW and RXAK will be set to their default states.

Bit 0 Unimplemented, read as “0”

• IICC1 Register

Bit	7	6	5	4	3	2	1	0
Name	HCF	HAAS	HBB	HTX	TXAK	SRW	IAMWU	RXAK
R/W	R	R	R	R/W	R/W	R	R/W	R
POR	1	0	0	0	0	0	0	1

Bit 7 **HCF**: I²C Bus data transfer completion flag

- 0: Data is being transferred
- 1: Completion of an 8-bit data transfer

The HCF flag is the data transfer flag. This flag will be zero when data is being transferred. Upon completion of an 8-bit data transfer the flag will go high and an interrupt will be generated.

Bit 6 **HAAS**: I²C Bus address match flag

- 0: Not address match
- 1: Address match

The HAAS flag is the address match flag. This flag is used to determine if the slave device address is the same as the master transmit address. If the addresses match then this bit will be high, if there is no match then the flag will be low.

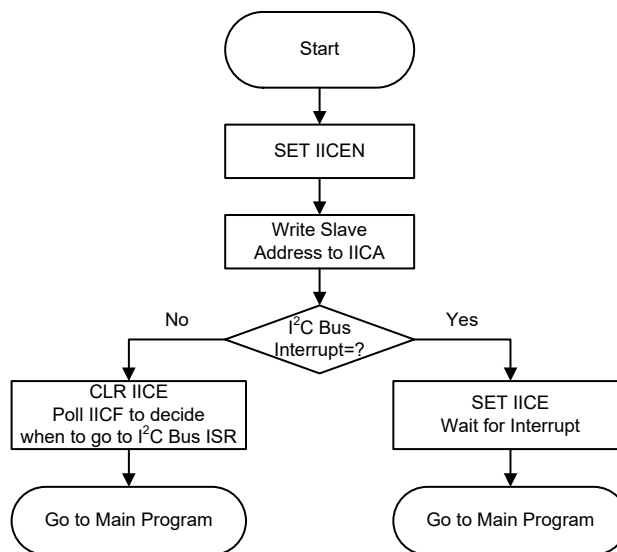
Bit 5	HBB: I²C Bus busy flag 0: I²C Bus is not busy 1: I²C Bus is busy The HBB flag is the I²C busy flag. This flag will be “1” when the I²C bus is busy which will occur when a START signal is detected. The flag will be set to “0” when the bus is free which will occur when a STOP signal is detected.
Bit 4	HTX: I²C slave device is transmitter or receiver selection 0: Slave device is the receiver 1: Slave device is the transmitter
Bit 3	TXAK: I²C Bus transmit acknowledge flag 0: Slave send acknowledge flag 1: Slave do not send acknowledge flag The TXAK bit is the transmit acknowledge flag. After the slave device receipt of 8 bits of data, this bit will be transmitted to the bus on the 9th clock from the slave device. The slave device must always set TXAK bit to “0” before further data is received.
Bit 2	SRW: I²C Slave Read/Write flag 0: Slave device should be in receive mode 1: Slave device should be in transmit mode The SRW flag is the I²C Slave Read/Write flag. This flag determines whether the master device wishes to transmit or receive data from the I²C bus. When the transmitted address and slave address is match, that is when the HAAS flag is set high, the slave device will check the SRW flag to determine whether it should be in transmit mode or receive mode. If the SRW flag is high, the master is requesting to read data from the bus, so the slave device should be in transmit mode. When the SRW flag is zero, the master will write data to the bus, therefore the slave device should be in receive mode to read this data.
Bit 1	IAMWU: I²C Address Match Wake-up control 0: Disable 1: Enable This bit should be set to 1 to enable the I²C address match wake up from the SLEEP or IDLE Mode. If the IAMWU bit has been set before entering either the SLEEP or IDLE mode to enable the I²C address match wake up, then this bit must be cleared by the application program after wake-up to ensure correction device operation.
Bit 0	RXAK: I²C Bus Receive acknowledge flag 0: Slave receive acknowledge flag 1: Slave does not receive acknowledge flag The RXAK flag is the receiver acknowledge flag. When the RXAK flag is “0”, it means that a acknowledge signal has been received at the 9th clock, after 8 bits of data have been transmitted. When the slave device in the transmit mode, the slave device checks the RXAK flag to determine if the master receiver wishes to receive the next byte. The slave transmitter will therefore continue sending out data until the RXAK flag is “1”. When this occurs, the slave transmitter will release the SDA line to allow the master to send a STOP signal to release the I²C Bus.

I²C Bus Communication

Communication on the I²C bus requires four separate steps, a START signal, a slave device address transmission, a data transmission and finally a STOP signal. When a START signal is placed on the I²C bus, all devices on the bus will receive this signal and be notified of the imminent arrival of data on the bus. The first seven bits of the data will be the slave address with the first bit being the MSB. If the address of the slave device matches that of the transmitted address, the HAAS bit in the IICC1 register will be set and an I²C interrupt will be generated. After entering the interrupt service routine, the slave device must first check the condition of the HAAS and IICTOF bits to determine whether the interrupt source originates from an address match or from the completion of an 8-bit

data transfer completion or from the I²C bus time-out occurrence. During a data transfer, note that after the 7-bit slave address has been transmitted, the following bit, which is the 8th bit, is the read/write bit whose value will be placed in the SRW bit. This bit will be checked by the slave device to determine whether to go into transmit or receive mode. Before any transfer of data to or from the I²C bus, the microcontroller must initialise the bus, the following are steps to achieve this:

- Step 1
Set the IICEN bit in the IICC0 register to “1” to enable the I²C bus.
- Step 2
Write the slave address of the device to the I²C bus address register IICA.
- Step 3
Set the IICE interrupt enable bit of the interrupt control register to enable the I²C interrupt.



I²C Bus Initialisation Flowchart

I²C Bus Start Signal

The START signal can only be generated by the master device connected to the I²C bus and not by the slave device. This START signal will be detected by all devices connected to the I²C bus. When detected, this indicates that the I²C bus is busy and therefore the HBB bit will be set. A START condition occurs when a high to low transition on the SDA line takes place when the SCL line remains high.

I²C Slave Address

The transmission of a START signal by the master will be detected by all devices on the I²C bus. To determine which slave device the master wishes to communicate with, the address of the slave device will be sent out immediately following the START signal. All slave devices, after receiving this 7-bit address data, will compare it with their own 7-bit slave address. If the address sent out by the master matches the internal address of the microcontroller slave device, then an internal I²C bus interrupt signal will be generated. The next bit following the address, which is the 8th bit, defines the read/write status and will be saved to the SRW bit of the IICC1 register. The slave device will then transmit an acknowledge bit, which is a low level, as the 9th bit. The slave device will also set the status flag HAAS when the addresses match.

As an I²C bus interrupt can come from three sources, when the program enters the interrupt subroutine, the HAAS and IICTOF bits should be examined to see whether the interrupt source has come from a matching slave address or from the completion of a data byte transfer or from the I²C bus time-out occurrence. When a slave address is matched, the device must be placed in either the transmit mode and then write data to the IICD register, or in the receive mode where it must implement a dummy read from the IICD register to release the SCL line.

I²C Bus Read/Write Signal

The SRW bit in the IICC1 register defines whether the master device wishes to read data from the I²C bus or write data to the I²C bus. The slave device should examine this bit to determine if it is to be a transmitter or a receiver. If the SRW flag is “1” then this indicates that the master device wishes to read data from the I²C bus, therefore the slave device must be set to send data to the I²C bus as a transmitter. If the SRW flag is “0” then this indicates that the master wishes to send data to the I²C bus, therefore the slave device must be set to read data from the I²C bus as a receiver.

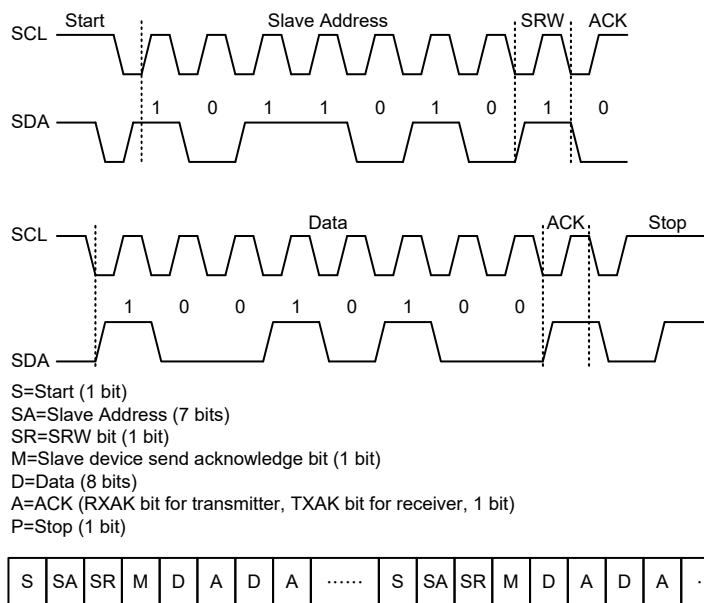
I²C Bus Slave Address Acknowledge Signal

After the master has transmitted a calling address, any slave device on the I²C bus, whose own internal address matches the calling address, must generate an acknowledge signal. The acknowledge signal will inform the master that a slave device has accepted its calling address. If no acknowledge signal is received by the master then a STOP signal must be transmitted by the master to end the communication. When the HAAS flag is high, the addresses have matched and the slave device must check the SRW flag to determine if it is to be a transmitter or a receiver. If the SRW flag is high, the slave device should be set to be a transmitter so the HTX bit in the IICC1 register should be set to “1”. If the SRW flag is low, then the microcontroller slave device should be set as a receiver and the HTX bit in the IICC1 register should be set to “0”.

I²C Bus Data and Acknowledge Signal

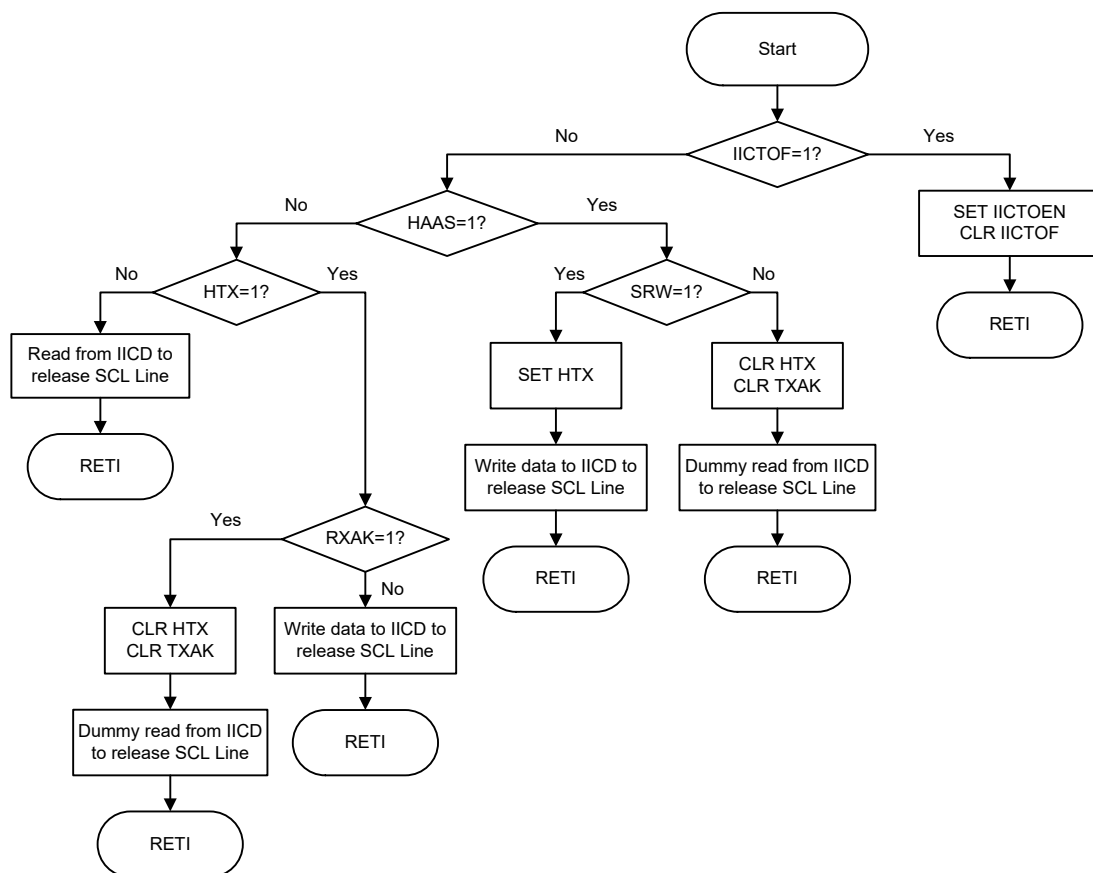
The transmitted data is 8-bit wide and is transmitted after the slave device has acknowledged receipt of its slave address. The order of serial bit transmission is the MSB first and the LSB last. After receipt of 8 bits of data, the receiver must transmit an acknowledge signal, level “0”, before it can receive the next data byte. If the slave transmitter does not receive an acknowledge bit signal from the master receiver, then the slave transmitter will release the SDA line to allow the master to send a STOP signal to release the I²C Bus. The corresponding data will be stored in the IICD register. If set as a transmitter, the slave device must first write the data to be transmitted into the IICD register. If set as a receiver, the slave device must read the transmitted data from the IICD register.

When the slave receiver receives the data byte, it must generate an acknowledge bit, known as TXAK, on the 9th clock. The slave device, which is set as a transmitter will check the RXAK bit in the IICC1 register to determine if it is to send another data byte, if not then it will release the SDA line and await the receipt of a STOP signal from the master.



I²C Communication Timing Diagram

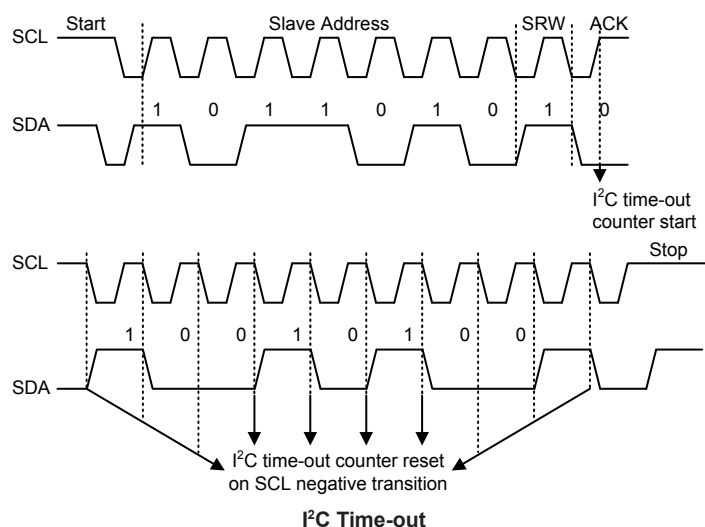
Note: When a slave address is matched, the device must be placed in either the transmit mode and then write data to the IICD register, or in the receive mode where it must implement a dummy read from the IICD register to release the SCL line.



I²C Bus ISR Flowchart

I²C Time-out Control

In order to reduce the problem of I²C lockup due to reception of erroneous clock sources, a time-out function is provided. If the clock source to the I²C is not received for a while, then the I²C circuitry and registers will be reset after a certain time-out period. The time-out counter starts counting on an I²C bus “START” & “address match” condition, and is cleared by an SCL falling edge. Before the next SCL falling edge arrives, if the time elapsed is greater than the time-out setup by the IICTOC register, then a time-out condition will occur. The time-out function will stop when an I²C “STOP” condition occurs.



When an I²C time-out counter overflow occurs, the counter will stop and the IICTOEN bit will be cleared to zero and the IICTOF bit will be set high to indicate that a time-out condition has occurred. The time-out condition will also generate an interrupt which uses the I²C interrupt vector. When an I²C time-out occurs, the I²C internal circuitry will be reset and the registers will be reset into the following condition:

Registers	After I ² C Time-out
IICD, IICA, IICC0	No change
IICC1	Reset to POR condition

I²C Registers after Time-out

The IICTOF flag can be cleared by the application program. There are 64 time-out periods which can be selected using IICTOS bit field in the IICTOC register. The time-out time is given by the formula: $((1 \sim 64) \times 32) / f_{\text{SUB}}$. This gives a time-out period which ranges from about 1ms to 64ms.

• IICTOC Register

Bit	7	6	5	4	3	2	1	0
Name	IICTOEN	IICTOF	IICTOS5	IICTOS4	IICTOS3	IICTOS2	IICTOS1	IICTOS0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **IICTOEN**: I²C Time-out control
 0: Disable
 1: Enable

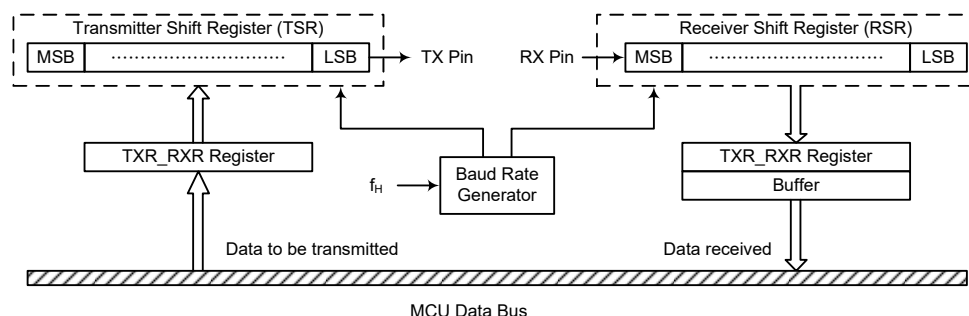
- Bit 6 **IICTOF**: I²C Time-out flag
 0: No time-out occurred
 1: Time-out occurred
- Bit 5~0 **IICTOSS~IICTOS0**: I²C Time-out period selection
 I²C time-out clock source is $f_{SUB}/32$.
 I²C time-out time is equal to $(IICTOS[5:0]+1) \times (32/f_{SUB})$.

UART Interface

The device contains an integrated full-duplex asynchronous serial communications UART interface that enables communication with external devices that contain a serial interface. The UART function has many features and can transmit and receive data serially by transferring a frame of data with eight or nine data bits per transmission as well as being able to detect errors when the data is overwritten or incorrectly framed. The UART function possesses its own internal interrupt which can be used to indicate when a reception occurs or when a transmission terminates.

The integrated UART function contains the following features:

- Full-duplex, asynchronous communication
- 8 or 9 bits character length
- Even, odd or no parity options
- One or two stop bits
- Baud rate generator with 8-bit prescaler
- Parity, framing, noise and overrun error detection
- Support for interrupt on address detect (last character bit=1)
- Separately enabled transmitter and receiver
- 2-byte Deep FIFO Receive Data Buffer
- RX pin wake-up function
- Transmit and receive interrupts
- Interrupts can be triggered by the following conditions:
 - ♦ Transmitter Empty
 - ♦ Transmitter Idle
 - ♦ Receiver Full
 - ♦ Receiver Overrun
 - ♦ Address Mode Detect



UART Data Transfer Block Diagram

UART External Pins

To communicate with an external serial interface, the internal UART has two external pins known as TX and RX. The TX and RX pins are the UART transmitter and receiver pins respectively. The TX and RX pin function should first be selected by the corresponding pin-shared function selection register before the UART function is used. Along with the UARTEN bit, the TXEN and RXEN bits, if set, will configure these pins to their respective TX output and RX input conditions and disable any pull-high resistor option which may exist on the TX and RX pins. When the TX or RX pin function is disabled by clearing the UARTEN, TXEN or RXEN bit, the TX or RX pin will be placed into a floating state. At this time whether the internal pull-high resistor is connected to the TX or RX pin or not is determined by the corresponding I/O pull-high function control bit.

UART Data Transfer Scheme

The above block diagram shows the overall data transfer structure arrangement for the UART. The actual data to be transmitted from the MCU is first transferred to the TXR_RXR register by the application program. The data will then be transferred to the Transmit Shift Register from where it will be shifted out, LSB first, onto the TX pin at a rate controlled by the Baud Rate Generator. Only the TXR_RXR register is mapped onto the MCU Data Memory, the Transmit Shift Register is not mapped and is therefore inaccessible to the application program.

Data to be received by the UART is accepted on the external RX pin, from where it is shifted in, LSB first, to the Receiver Shift Register at a rate controlled by the Baud Rate Generator. When the shift register is full, the data will then be transferred from the shift register to the internal TXR_RXR register, where it is buffered and can be manipulated by the application program. Only the TXR_RXR register is mapped onto the MCU Data Memory, the Receiver Shift Register is not mapped and is therefore inaccessible to the application program.

It should be noted that the actual register for data transmission and reception only exists as a single shared register in the Data Memory. This shared register known as the TXR_RXR register is used for both data transmission and data reception.

UART Status and Control Registers

There are five control registers associated with the UART function. The USR, UCR1 and UCR2 registers control the overall function of the UART, while the BRG register controls the Baud rate. The actual data to be transmitted and received on the serial interface is managed through the TXR_RXR data register.

Register Name	Bit							
	7	6	5	4	3	2	1	0
USR	PERR	NF	FERR	OERR	RIDLE	RXIF	TIDLE	TXIF
UCR1	UARTEN	BNO	PREN	PRT	STOPS	TXBRK	RX8	TX8
UCR2	TXEN	RXEN	BRGH	ADDEN	WAKE	RIE	TIIE	TEIE
TXR_RXR	D7	D6	D5	D4	D3	D2	D1	D0
BRG	D7	D6	D5	D4	D3	D2	D1	D0

UART Register List

• USR Register

The USR register is the status register for the UART, which can be read by the program to determine the present status of the UART. All flags within the USR register are read only. Further explanation on each of the flags is given below:

Bit	7	6	5	4	3	2	1	0
Name	PERR	NF	FERR	OERR	RIDLE	RXIF	TIDLE	TXIF
R/W	R	R	R	R	R	R	R	R
POR	0	0	0	0	1	0	1	1

- Bit 7 PERR:** Parity error flag
 0: No parity error is detected
 1: Parity error is detected
 The PERR flag is the parity error flag. When this read only flag is “0”, it indicates a parity error has not been detected. When the flag is “1”, it indicates that the parity of the received word is incorrect. This error flag is applicable only if Parity mode (odd or even) is selected. The flag can also be cleared by a software sequence which involves a read to the status register UnSR followed by an access to the TXR_RXR data register.
- Bit 6 NF:** Noise flag
 0: No noise is detected
 1: Noise is detected
 The NF flag is the noise flag. When this read only flag is “0”, it indicates no noise condition. When the flag is “1”, it indicates that the UART has detected noise on the receiver input. The NF flag is set during the same cycle as the RXIF flag but will not be set in the case of an overrun. The NF flag can be cleared by a software sequence which will involve a read to the status register USR followed by an access to the TXR_RXR data register.
- Bit 5 FERR:** Framing error flag
 0: No framing error is detected
 1: Framing error is detected
 The FERR flag is the framing error flag. When this read only flag is “0”, it indicates that there is no framing error. When the flag is “1”, it indicates that a framing error has been detected for the current character. The flag can also be cleared by a software sequence which will involve a read to the status register USR followed by an access to the TXR_RXR data register.
- Bit 4 OERR:** Overrun error flag
 0: No overrun error is detected
 1: Overrun error is detected
 The OERR flag is the overrun error flag which indicates when the receiver buffer has overflowed. When this read only flag is “0”, it indicates that there is no overrun error. When the flag is “1”, it indicates that an overrun error occurs which will inhibit further transfers to the TXR_RXR receive data register. The flag is cleared by a software sequence, which is a read to the status register USR followed by an access to the TXR_RXR data register.
- Bit 3 RIDLE:** Receiver status
 0: Data reception is in progress (Data being received)
 1: No data reception is in progress (Receiver is idle)
 The RIDLE flag is the receiver status flag. When this read only flag is “0”, it indicates that the receiver is between the initial detection of the start bit and the completion of the stop bit. When the flag is “1”, it indicates that the receiver is idle. Between the completion of the stop bit and the detection of the next start bit, the RIDLE bit is “1” indicating that the UART receiver is idle and the RX pin stays in logic high condition.

- Bit 2** **RXIF:** Receive TXR_RXR data register status
 0: TXR_RXR data register is empty
 1: TXR_RXR data register has available data
 The RXIF flag is the receive data register status flag. When this read only flag is “0”, it indicates that the TXR_RXR read data register is empty. When the flag is “1”, it indicates that the TXR_RXR read data register contains new data. When the contents of the shift register are transferred to the TXR_RXR register, an interrupt is generated if RIE=1 in the UCR2 register. If one or more errors are detected in the received word, the appropriate receive-related flags NF, FERR, and/or PERR are set within the same clock cycle. The RXIF flag will eventually be cleared when the USR register is read with RXIF set, followed by a read from the TXR_RXR register, and if the TXR_RXR register has no more new data available.
- Bit 1** **TIDLE:** Transmission idle
 0: Data transmission is in progress (Data being transmitted)
 1: No data transmission is in progress (Transmitter is idle)
 The TIDLE flag is known as the transmission complete flag. When this read only flag is “0”, it indicates that a transmission is in progress. This flag will be set high when the TXIF flag is “1” and when there is no transmit data or break character being transmitted. When TIDLE is equal to “1”, the TX pin becomes idle with the pin state in logic high condition. The TIDLE flag is cleared by reading the USR register with TIDLE set and then writing to the TXR_RXR register. The flag is not generated when a data character or a break is queued and ready to be sent.
- Bit 0** **TXIF:** Transmit TXR_RXR data register status
 0: Character is not transferred to the transmit shift register
 1: Character has transferred to the transmit shift register (TXR_RXR data register is empty)
 The TXIF flag is the transmit data register empty flag. When this read only flag is “0”, it indicates that the character is not transferred to the transmitter shift register. When the flag is “1”, it indicates that the transmitter shift register has received a character from the TXR_RXR data register. The TXIF flag is cleared by reading the UART status register (USR) with TXIF set and then writing to the TXR_RXR data register. Note that when the TXEN bit is set, the TXIF flag bit will also be set since the transmit data register is not yet full.

• UCR1 Register

The UCR1 register together with the UCR2 register are the two UART control registers that are used to set the various options for the UART function, such as overall on/off control, parity control, data transfer bit length etc. Further explanation on each of the bits is given below:

Bit	7	6	5	4	3	2	1	0
Name	UARTEN	BNO	PREN	PRT	STOPS	TXBRK	RX8	TX8
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R	W
POR	0	0	0	0	0	0	x	0

“x”: unknown

- Bit 7** **UARTEN:** UART function enable control
 0: Disable UART. TX and RX pins are in a floating state
 1: Enable UART. TX and RX pins function as UART pins
 The UARTEN bit is the UART enable bit. When this bit is equal to “0”, the UART will be disabled and the RX pin as well as the TX pin will be in a floating state. When the bit is equal to “1” the UART will be enabled and the TX and RX pins will function as defined by the TXEN and RXEN enable control bits.
 When the UART is disabled, it will empty the buffer so any character remaining in the buffer will be discarded. In addition, the value of the baud rate counter will be reset. If the UART is disabled, all error and status flags will be reset. Also the TXEN, RXEN, TXBRK, RXIF, OERR, FERR, PERR and NF bits will be cleared, while the

TIDLE, TXIF and RIDLE bits will be set. Other control bits in UCR1, UCR2 and BRG registers will remain unaffected. If the UART is active and the UARTEN bit is cleared, all pending transmissions and receptions will be terminated and the module will be reset as defined above. When the UART is re-enabled, it will restart in the same configuration.

- | | |
|-------|--|
| Bit 6 | <p>BNO: Number of data transfer bits selection</p> <p>0: 8-bit data transfer</p> <p>1: 9-bit data transfer</p> <p>This bit is used to select the data length format, which can have a choice of either 8-bit or 9-bit format. When this bit is equal to “1”, a 9-bit data length format will be selected. If the bit is equal to “0”, then an 8-bit data length format will be selected. If 9-bit data length format is selected, then bits RX8 and TX8 will be used to store the 9th bit of the received and transmitted data respectively.</p> |
| Bit 5 | <p>PREN: Parity function enable control</p> <p>0: Parity function is disabled</p> <p>1: Parity function is enabled</p> <p>This is the parity enable bit. When this bit is equal to “1”, the parity function will be enabled. If the bit is equal to “0”, then the parity function will be disabled.</p> |
| Bit 4 | <p>PRT: Parity type selection bit</p> <p>0: Even parity for parity generator</p> <p>1: Odd parity for parity generator</p> <p>This bit is the parity type selection bit. When this bit is equal to “1”, odd parity type will be selected. If the bit is equal to “0”, then even parity type will be selected.</p> |
| Bit 3 | <p>STOPS: Number of Stop bits selection</p> <p>0: One stop bit format is used</p> <p>1: Two stop bits format is used</p> <p>This bit determines if one or two stop bits are to be used. When this bit is equal to “1”, two stop bits are used. If this bit is equal to “0”, then only one stop bit is used.</p> |
| Bit 2 | <p>TXBRK: Transmit break character</p> <p>0: No break character is transmitted</p> <p>1: Break characters transmit</p> <p>The TXBRK bit is the Transmit Break Character bit. When this bit is “0”, there are no break characters and the TX pin operates normally. When the bit is “1”, there are transmit break characters and the transmitter will send logic zeros. When this bit is equal to “1”, after the buffered data has been transmitted, the transmitter output is held low for a minimum of a 13-bit length and until the TXBRK bit is reset.</p> |
| Bit 1 | <p>RX8: Receive data bit 8 for 9-bit data transfer format (read only)</p> <p>This bit is only used if 9-bit data transfers are used, in which case this bit location will store the 9th bit of the received data known as RX8. The BNO bit is used to determine whether data transfers are in 8-bit or 9-bit format.</p> |
| Bit 0 | <p>TX8: Transmit data bit 8 for 9-bit data transfer format (write only)</p> <p>This bit is only used if 9-bit data transfers are used, in which case this bit location will store the 9th bit of the transmitted data known as TX8. The BNO bit is used to determine whether data transfers are in 8-bit or 9-bit format.</p> |

• **UCR2 Register**

The UCR2 register is the second of the two UART control registers and serves several purposes. One of its main functions is to control the basic enable/disable operation of the UART Transmitter and Receiver as well as enabling the various UART interrupt sources. The register also serves to control the baud rate speed, receiver wake-up enable and the address detect enable. Further explanation on each of the bits is given below:

Bit	7	6	5	4	3	2	1	0
Name	TXEN	RXEN	BRGH	ADDEN	WAKE	RIE	TIIE	TEIE
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

Bit 7 **TXEN**: UART Transmitter enabled control

0: UART transmitter is disabled

1: UART transmitter is enabled

The bit named TXEN is the Transmitter Enable Bit. When this bit is equal to “0”, the transmitter will be disabled with any pending data transmissions being aborted. In addition the buffers will be reset. In this situation the TX pin will be in a floating state.

If the TXEN bit is equal to “1” and the UARTEN bit is also equal to “1”, the transmitter will be enabled and the TX pin will be controlled by the UART. Clearing the TXEN bit during a transmission will cause the data transmission to be aborted and will reset the transmitter. If this situation occurs, the TX pin will be in a floating state.

Bit 6 **RXEN**: UART Receiver enabled control

0: UART receiver is disabled

1: UART receiver is enabled

The bit named RXEN is the Receiver Enable Bit. When this bit is equal to “0”, the receiver will be disabled with any pending data receptions being aborted. In addition the receive buffers will be reset. In this situation the RX pin will be in a floating state. If the RXEN bit is equal to “1” and the UARTEN bit is also equal to “1”, the receiver will be enabled and the RX pin will be controlled by the UART. Clearing the RXEN bit during a reception will cause the data reception to be aborted and will reset the receiver. If this situation occurs, the RX pin will be in a floating state.

Bit 5 **BRGH**: Baud Rate speed selection

0: Low speed baud rate

1: High speed baud rate

The bit named BRGH selects the high or low speed mode of the Baud Rate Generator. This bit, together with the value placed in the baud rate register BRG, controls the Baud Rate of the UART. If this bit is equal to “1”, the high speed mode is selected. If the bit is equal to “0”, the low speed mode is selected.

Bit 4 **ADDEN**: Address detect function enable control

0: Address detect function is disabled

1: Address detect function is enabled

The bit named ADDEN is the address detect function enable control bit. When this bit is equal to “1”, the address detect function is enabled. When it occurs, if the 8th bit, which corresponds to RX7 if BNO=0 or the 9th bit, which corresponds to RX8 if BNO=1, has a value of “1”, then the received word will be identified as an address, rather than data. If the corresponding interrupt is enabled, an interrupt request will be generated each time the received word has the address bit set, which is the 8th or 9th bit depending on the value of BNO. If the address bit known as the 8th or 9th bit of the received word is “0” with the address detect function being enabled, an interrupt will not be generated and the received data will be discarded.

Bit 3 **WAKE**: RX pin wake-up UART function enable control

0: RX pin wake-up UART function is disabled

1: RX pin wake-up UART function is enabled

This bit is used to control the wake-up UART function when a falling edge on the RX pin occurs. Note that this bit is only available when the UART clock (f_{H1}) is switched

off. There will be no RX pin wake-up UART function if the UART clock (f_{H1}) exists. If the WAKE bit is set to 1 as the UART clock (f_{H1}) is switched off, a UART wake-up request will be initiated when a falling edge on the RX pin occurs. When this request happens and the corresponding interrupt is enabled, an RX pin wake-up UART interrupt will be generated to inform the MCU to wake up the UART function by switching on the UART clock (f_{H1}) via the application program. Otherwise, the UART function cannot resume even if there is a falling edge on the RX pin when the WAKE bit is cleared to 0.

- Bit 2 RIE:** Receiver interrupt enable control
 0: Receiver related interrupt is disabled
 1: Receiver related interrupt is enabled
 This bit enables or disables the receiver interrupt. If this bit is equal to “1” and when the receiver overrun flag OERR or receive data available flag RXIF is set, the UART interrupt request flag will be set. If this bit is equal to “0”, the UART interrupt request flag will not be influenced by the condition of the OERR or RXIF flags.
- Bit 1 TIE:** Transmitter Idle interrupt enable control
 0: Transmitter idle interrupt is disabled
 1: Transmitter idle interrupt is enabled
 This bit enables or disables the transmitter idle interrupt. If this bit is equal to “1” and when the transmitter idle flag TIDLE is set, due to a transmitter idle condition, the UART interrupt request flag will be set. If this bit is equal to “0”, the UART interrupt request flag will not be influenced by the condition of the TIDLE flag.
- Bit 0 TEIE:** Transmitter Empty interrupt enable control
 0: Transmitter empty interrupt is disabled
 1: Transmitter empty interrupt is enabled
 This bit enables or disables the transmitter empty interrupt. If this bit is equal to “1” and when the transmitter empty flag TXIF is set, due to a transmitter empty condition, the UART interrupt request flag will be set. If this bit is equal to “0”, the UART interrupt request flag will not be influenced by the condition of the TXIF flag.

• **TXR_RXR Register**

The TXR_RXR register is the data register which is used to store the data to be transmitted on the TX pin or being received from the RX pin.

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	x	x	x	x	x	x	x	x

“x”: unknown

Bit 7~0 **D7~D0:** UART Transmit/Receive Data bit 7 ~ bit 0

• **BRG Register**

Bit	7	6	5	4	3	2	1	0
Name	D7	D6	D5	D4	D3	D2	D1	D0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	x	x	x	x	x	x	x	x

“x”: unknown

Bit 7~0 **D7~D0:** Baud Rate values

By programming the BRGH bit in UCR2 Register which allows selection of the related formula described above and programming the required value in the BRG register, the required baud rate can be set.

Note: Baud rate= $f_{H1}/[64 \times (N+1)]$ if BRGH=0.

Baud rate= $f_{H1}/[16 \times (N+1)]$ if BRGH=1.

Baud Rate Generator

To set the speed of the serial data communication, the UART function contains its own dedicated baud rate generator. The baud rate is controlled by its own internal free running 8-bit timer, the period of which is determined by two factors. The first of these is the value placed in the baud rate register BRG and the second is the value of the BRGH bit with the control register UCR2. The BRGH bit decides if the baud rate generator is to be used in a high speed mode or low speed mode, which in turn determines the formula that is used to calculate the baud rate. The value N in the BRG register which is used in the following baud rate calculation formula determines the division factor. Note that N is the decimal value placed in the BRG register and has a range of between 0 and 255.

UCR2 BRGH Bit	0	1
Baud Rate (BR)	$f_H/[64 (N+1)]$	$f_H/[16 (N+1)]$

By programming the BRGH bit which allows selection of the related formula and programming the required value in the BRG register, the required baud rate can be set. Note that because the actual baud rate is determined using a discrete value, N, placed in the BRG register, there will be an error associated between the actual and requested value. The following example shows how the BRG register value N and the error value can be calculated.

Calculating the Baud Rate and Error Values

For a clock frequency of 4MHz, and with BRGH cleared to zero determine the BRG register value N, the actual baud rate and the error value for a desired baud rate of 4800.

From the above table the desired baud rate $BR = f_H/[64(N+1)]$

Re-arranging this equation gives $N = [f_H/(BR \times 64)] - 1$

Giving a value for $N = [4000000/(4800 \times 64)] - 1 = 12.0208$

To obtain the closest value, a decimal value of 12 should be placed into the BRG register. This gives an actual or calculated baud rate value of $BR = 4000000/[64 \times (12+1)] = 4808$

Therefore the error is equal to $(4808 - 4800)/4800 = 0.16\%$

UART Setup and Control

For data transfer, the UART function utilizes a non-return-to-zero, more commonly known as NRZ, format. This is composed of one start bit, eight or nine data bits, and one or two stop bits. Parity is supported by the UART hardware, and can be set to be even, odd or no parity. For the most common data format, 8 data bits along with no parity and one stop bit, denoted as 8, N, 1, is used as the default setting, which is the setting at power-on. The number of data bits and stop bits, along with the parity, are set by programming the corresponding BNO, PRT, PREN, and STOPS bits in the UCR1 register. The baud rate used to transmit and receive data is set using the internal 8-bit baud rate generator, while the data is transmitted and received LSB first. Although the UART transmitter and receiver are functionally independent, they both use the same data format and baud rate. In all cases stop bits will be used for data transmission.

Enabling/Disabling the UART Interface

The basic on/off function of the internal UART function is controlled using the UARTEN bit in the UCR1 register. If the UARTEN, TXEN and RXEN bits are set, then these two UART pins will act as normal TX output pin and RX input pin respectively. If no data is being transmitted on the TX pin, then it will default to a logic high value.

Clearing the UARTEN bit will disable the TX and RX pins and allow these two pins to be in a floating state. When the UART function is disabled the buffer will be reset to an empty condition, at the same time discarding any remaining residual data. Disabling the UART will also reset the error

and status flags with bits TXEN, RXEN, TXBRK, RXIF, OERR, FERR, PERR and NF being cleared while bits TIDLE, TXIF and RIDLE will be set. The remaining control bits in the UCR1, UCR2 and BRG registers will remain unaffected. If the UARTEN bit in the UCR1 register is cleared while the UART is active, then all pending transmissions and receptions will be immediately suspended and the UART will be reset to a condition as defined above. If the UART is then subsequently re-enabled, it will restart again in the same configuration.

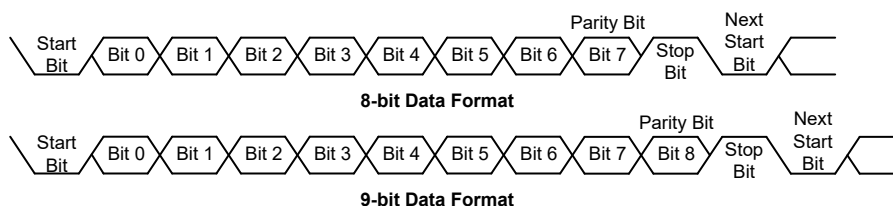
Data, Parity and Stop Bit Selection

The format of the data to be transferred is composed of various factors such as data bit length, parity on/off, parity type, address bits and the number of stop bits. These factors are determined by the setup of various bits within the UCR1 register. The BNO bit controls the number of data bits which can be set to either 8 or 9, the PRTn bit controls the choice of odd or even parity, the PREN bit controls the parity on/off function and the STOPS bit decides whether one or two stop bits are to be used. The following table shows various formats for data transmission. The address bit, which is the MSB of the data byte, identifies the frame as an address character or data if the address detect function is enabled. The number of stop bits, which can be either one or two, is independent of the data length and is only used for the transmitter. There is only one stop bit for the receiver.

Start Bit	Data Bits	Address Bit	Parity Bit	Stop Bit
Example of 8-bit Data Formats				
1	8	0	0	1
1	7	0	1	1
1	7	1	0	1
Example of 9-bit Data Formats				
1	9	0	0	1
1	8	0	1	1
1	8	1	0	1

Transmitter Receiver Data Format

The following diagram shows the transmit and receive waveforms for both 8-bit and 9-bit data formats.



UART Transmitter

Data word lengths of either 8 or 9 bits can be selected by programming the BNO bit in the UCR1 register. When BNO bit is set, the word length will be set to 9 bits. In this case the 9th bit, which is the MSB, needs to be stored in the TX8 bit in the UCR1 register. At the transmitter core lies the Transmitter Shift Register, more commonly known as the TSR, whose data is obtained from the transmit data register, which is known as the TXR_RXR register. The data to be transmitted is loaded into this TXR_RXR register by the application program. The TSR register is not written to with new data until the stop bit from the previous transmission has been sent out. As soon as this stop bit has been transmitted, the TSR can then be loaded with new data from the TXR_RXR register, if it is available. It should be noted that the TSR register, unlike many other registers, is not directly mapped into the Data Memory area and as such is not available to the application program for direct read/write operations. An actual transmission of data will normally be enabled when the

TXEN bit is set, but the data will not be transmitted until the TXR_RXR register has been loaded with data and the baud rate generator has defined a shift clock source. However, the transmission can also be initiated by first loading data into the TXR_RXR register, after which the TXEN bit can be set. When a transmission of data begins, the TSR is normally empty, in which case a transfer to the TXR_RXR register will result in an immediate transfer to the TSR. If during a transmission the TXEN bit is cleared, the transmission will immediately cease and the transmitter will be reset. The TX output pin can then be in a floating state.

Transmitting Data

When the UART is transmitting data, the data is shifted on the TX pin from the shift register, with the least significant bit first. In the transmit mode, the TXR_RXR register forms a buffer between the internal bus and the transmitter shift register. It should be noted that if 9-bit data format has been selected, then the MSB will be taken from the TX8 bit in the UCR1 register. The steps to initiate a data transfer can be summarized as follows:

- Make the correct selection of the BNO, PRT, PREN and STOPS bits to define the required word length, parity type and number of stop bits.
- Configure the BRG register to select the desired baud rate.
- Set the TXEN bit to ensure that the TX pin is used as a UART transmitter pin.
- Access the USR register and write the data that is to be transmitted into the TXR_RXR register. Note that this step will clear the TXIF bit.

This sequence of events can now be repeated to send additional data.

It should be noted that when TXIF=0, data will be inhibited from being written to the TXR_RXR register. Clearing the TXIF flag is always achieved using the following software sequence:

1. A USR register access
2. A TXR_RXR register write execution

The read-only TXIF flag is set by the UART hardware and if set indicates that the TXR_RXR register is empty and that other data can now be written into the TXR_RXR register without overwriting the previous data. If the TEIE bit is set then the TXIF flag will generate an interrupt.

During a data transmission, a write instruction to the TXR_RXR register will place the data into the TXR_RXR register, which will be copied to the shift register at the end of the present transmission. When there is no data transmission in progress, a write instruction to the TXR_RXR register will place the data directly into the shift register, resulting in the commencement of data transmission, and the TXIF bit being immediately set. When a frame transmission is complete, which happens after stop bits are sent or after the break frame, the TIDLE bit will be set. To clear the TIDLE bit the following software sequence is used:

1. A USR register access
2. A TXR_RXR register write execution

Note that both the TXIF and TIDLE bits are cleared by the same software sequence.

Transmit Break

If the TXBRK bit is set then break characters will be sent on the next transmission. Break character transmission consists of a start bit, followed by $13 \times N$ '0' bits and stop bits, where $N=1, 2$, etc. If a break character is to be transmitted then the TXBRK bit must be first set by the application program, and then cleared to generate the stop bits. Transmitting a break character will not generate a transmit interrupt. Note that a break condition length is at least 13 bits long. If the TXBRK bit is continually kept at a logic high level then the transmitter circuitry will transmit continuous break characters. After the application program has cleared the TXBRK bit, the transmitter will finish transmitting the

last break character and subsequently send out one or two stop bits. The automatic logic highs at the end of the last break character will ensure that the start bit of the next frame is recognized.

UART Receiver

The UART is capable of receiving word lengths of either 8 or 9 bits. If the BNO bit is set, the word length will be set to 9 bits with the MSB being stored in the RX8 bit of the UCR1 register. At the receiver core lies the Receive Serial Shift Register, commonly known as the RSR. The data which is received on the RX external input pin is sent to the data recovery block. The data recovery block operating speed is 16 times that of the baud rate, while the main receive serial shifter operates at the baud rate. After the RX pin is sampled for the stop bit, the received data in RSR is transferred to the receive data register, if the register is empty. The data which is received on the external RX input pin is sampled three times by a majority detect circuit to determine the logic level that has been placed onto the RX pin. It should be noted that the RSR register, unlike many other registers, is not directly mapped into the Data Memory area and as such is not available to the application program for direct read/write operations.

Receiving Data

When the UART receiver is receiving data, the data is serially shifted in on the external RX input pin, LSB first. In the read mode, the TXR_RXR register forms a buffer between the internal bus and the receiver shift register. The TXR_RXR register is a two byte deep FIFO data buffer, where two bytes can be held in the FIFO while a third byte can continue to be received. Note that the application program must ensure that the data is read from TXR_RXR before the third byte has been completely shifted in, otherwise this third byte will be discarded and an overrun error OERR will be subsequently indicated. The steps to initiate a data transfer can be summarized as follows:

- Make the correct selection of BNO, PRT and PREN bits to define the word length, parity type.
- Configure the BRG register to select the desired baud rate.
- Set the RXEN bit to ensure that the RX pin is used as a UART receiver pin.

At this point the receiver will be enabled which will begin to look for a start bit.

When a character is received the following sequence of events will occur:

- The RXIF bit in the USR register will be set when the TXR_RXR register has data available. There will be at most one more character available before an overrun error occurs.
- When the contents of the shift register have been transferred to the TXR_RXR register, then if the RIE bit is set, an interrupt will be generated.
- If during reception, a frame error, noise error, parity error, or an overrun error has been detected, then the error flags can be set.

The RXIF bit can be cleared using the following software sequence:

1. A USR register access
2. A TXR_RXR register read execution

Receive Break

Any break character received by the UART will be managed as a framing error. The receiver will count and expect a certain number of bit times as specified by the values programmed into the BNO bit plus one stop bit. If the break is much longer than 13 bit times, the reception will be considered as complete after the number of bit times specified by BNO plus one stop bit. The RXIF bit is set, FERR is set, zeros are loaded into the receive data register, interrupts are generated if appropriate and the RIDLE bit is set. A break is regarded as a character that contains only zeros with the FERR flag set. If a long break signal has been detected, the receiver will regard it as a data frame including

a start bit, data bits and the invalid stop bit and the FERR flag will be set. The receiver must wait for a valid stop bit before looking for the next start bit. The receiver will not make the assumption that the break condition on the line is the next start bit. The break character will be loaded into the buffer and no further data will be received until stop bits are received. It should be noted that the RIDLE read only flag will go high when the stop bits have not yet been received. The reception of a break character on the UART registers will result in the following:

- The framing error flag, FERR, will be set.
- The receive data register, TXR_RXR, will be cleared.
- The OERR, NF, PERR, RIDLE or RXIF flags will possibly be set.

Idle Status

When the receiver is reading data, which means it will be in between the detection of a start bit and the reading of a stop bit, the receiver status flag in the USR register, otherwise known as the RIDLE flag, will have a zero value. In between the reception of a stop bit and the detection of the next start bit, the RIDLE flag will have a high value, which indicates the receiver is in an idle condition.

Receiver Interrupt

The read only receive interrupt flag RXIF in the USR register is set by an edge generated by the receiver. An interrupt is generated if RIE=1, when a word is transferred from the Receive Shift Register, RSR, to the Receive Data Register, TXR_RXR. An overrun error can also generate an interrupt if RIE=1.

Managing Receiver Errors

Several types of reception errors can occur within the UART module, the following section describes the various types and how they are managed by the UART.

Overrun Error – OERR

The TXR_RXR register is composed of a two byte deep FIFO data buffer, where two bytes can be held in the FIFO register, while a third byte can continue to be received. Before this third byte has been entirely shifted in, the data should be read from the TXR_RXR register. If this is not done, the overrun error flag OERR will be consequently indicated.

In the event of an overrun error occurring, the following will happen:

- The OERR flag in the USR register will be set.
- The TXR_RXR contents will not be lost.
- The shift register will be overwritten.
- An interrupt will be generated if the RIE bit is set.

The OERR flag can be cleared by an access to the USR register followed by a read to the TXR_RXR register.

Noise Error – NF

Over-sampling is used for data recovery to identify valid incoming data and noise. If noise is detected within a frame the following will occur:

- The read only noise flag, NF, in the USR register will be set on the rising edge of the RXIF bit.
- Data will be transferred from the Shift register to the TXR_RXR register.
- No interrupt will be generated. However this bit rises at the same time as the RXIF bit which itself generates an interrupt.

Note that the NF flag is reset by an USR register read operation followed by a TXR_RXR register read operation.

Framing Error – FERR

The read only framing error flag, FERR, in the USR register, is set if a zero is detected instead of stop bits. If two stop bits are selected, both stop bits must be high; otherwise the FERR flag will be set. The FERR flag and the received data will be recorded in the USR and TXR_RXR registers respectively, and the flag is cleared in any reset.

Parity Error – PERR

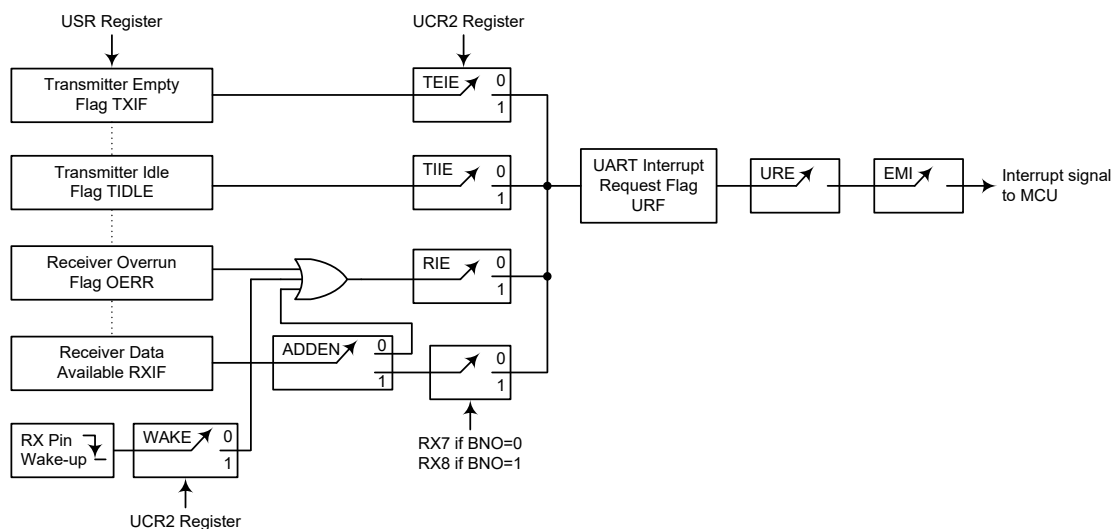
The read only parity error flag, PERR, in the USR register, is set if the parity of the received word is incorrect. This error flag is only applicable if the parity is enabled, PREN=1, and if the parity type, odd or even is selected. The read only PERR flag and the received data will be recorded in the USR and TXR_RXR registers respectively. It is cleared on any reset, it should be noted that the flags, FERR and PERR, in the USR register should first be read by the application program before reading the data word.

UART Interrupt Structure

Several individual UART conditions can generate a UART interrupt. When these conditions exist, a low pulse will be generated to get the attention of the microcontroller. These conditions are a transmitter data register empty, transmitter idle, receiver data available, receiver overrun, address detect and an RX pin wake-up. When any of these conditions are created, if the global interrupt enable bit and its corresponding interrupt control bit are enabled and the stack is not full, the program will jump to its corresponding interrupt vector where it can be serviced before returning to the main program. Four of these conditions have the corresponding USR register flags which will generate a UART interrupt if its associated interrupt enable control bit in the UCR2 register is set. The two transmitter interrupt conditions have their own corresponding enable control bits, while the two receiver interrupt conditions have a shared enable control bit. These enable bits can be used to mask out individual UART interrupt sources.

The address detect condition, which is also a UART interrupt source, does not have an associated flag, but will generate a UART interrupt when an address detect condition occurs if its function is enabled by setting the ADDEN bit in the UCR2 register. An RX pin wake-up, which is also a UART interrupt source, does not have an associated flag, but will generate a UART interrupt if the UART clock (f_{H}) source is switched off and the WAKE and RIE bits in the UCR2 register are set when a falling edge on the RX pin occurs.

Note that the USR register flags are read only and cannot be cleared or set by the application program, neither will they be cleared when the program jumps to the corresponding interrupt servicing routine, as is the case for some of the other interrupts. The flags will be cleared automatically when certain actions are taken by the UART, the details of which are given in the UART register section. The overall UART interrupt can be disabled or enabled by the related interrupt enable control bits in the interrupt control registers of the microcontroller to decide whether the interrupt requested by the UART module is masked out or allowed.



UART Interrupt Structure

Address Detect Mode

Setting the Address Detect Mode bit, ADDEN, in the UCR2 register, enables this special mode. If this bit is enabled then an additional qualifier will be placed on the generation of a Receiver Data Available interrupt, which is requested by the RXIF flag. If the ADDEN bit is enabled, then when data is available, an interrupt will only be generated, if the highest received bit has a high value. Note that the URE and EMI interrupt enable bits must also be enabled for correct interrupt generation. This highest address bit is the 9th bit if BNO=1 or the 8th bit if BNO=0. If this bit is high, then the received word will be defined as an address rather than data. A Data Available interrupt will be generated every time the last bit of the received word is set. If the ADDEN bit is not enabled, then a Receiver Data Available interrupt will be generated each time the RXIF flag is set, irrespective of the data last bit status. The address detect mode and parity enable are mutually exclusive functions. Therefore if the address detect mode is enabled, then to ensure correct operation, the parity function should be disabled by resetting the parity enable bit PREN to zero.

ADDEN	Bit 9 if BNO=1, Bit 8 if BNO=0	UART Interrupt Generated
0	0	√
	1	√
1	0	×
	1	√

ADDEN Bit Function

UART Power Down and Wake-up

When the UART clock (f_{H1}) is off, the UART will cease to function, and all clock sources to the module are shutdown. If the UART clock (f_{H1}) is off while a transmission is still in progress, then the transmission will be paused until the UART clock source derived from the microcontroller is activated. In a similar way, if the MCU enters the IDLE or SLEEP Mode while receiving data, then the reception of data will likewise be paused. When the MCU enters the IDLE or SLEEP Mode, note that the USR, UCR1, UCR2, transmit and receive registers, as well as the BRG register will not be affected. It is recommended to make sure first that the UART data transmission or reception has been finished before the microcontroller enters the IDLE or SLEEP mode.

The UART function contains a receiver RX pin wake-up function, which is enabled or disabled by the WAKE bit in the UCR2 register. If this bit, along with the UART enable bit, UARTEN, the receiver enable bit, RXEN and the receiver interrupt bit, RIE, are all set when the UART clock (f_{H1}) is off, then a falling edge on the RX pin will trigger an RX pin wake-up UART interrupt. Note that as it takes certain system clock cycles after a wake-up, before normal microcontroller operation resumes, any data received during this time on the RX pin will be ignored.

For a UART wake-up interrupt to occur, in addition to the bits for the wake-up being set, the global interrupt enable bit, EMI, and the UART interrupt enable bit, URE, must be set. If the EMI and URE bits are not set then only a wake up event will occur and no interrupt will be generated. Note also that as it takes certain system clock cycles after a wake-up before normal microcontroller resumes, the UART interrupt will not be generated until after this time has elapsed.

Low Voltage Detector – LVD

The device has a Low Voltage Detector function, also known as LVD. This enables the device to monitor the power supply voltage, V_{DD} , and provide a warning signal should it fall below a certain level. This function may be especially useful in battery applications where the supply voltage will gradually reduce as the battery ages, as it allows an early warning battery low signal to be generated. The Low Voltage Detector also has the capability of generating an interrupt signal.

LVD Register

The Low Voltage Detector function is controlled using a single register with the name LVDC. Three bits in this register, VLVD2~VLVD0, are used to select one of eight fixed voltages below which a low voltage condition will be determined. A low voltage condition is indicated when the LVDO bit is set. If the LVDO bit is low, this indicates that the V_{DD} voltage is above the preset low voltage value. The LVDEN bit is used to control the overall on/off function of the low voltage detector. Setting the bit high will enable the low voltage detector. Clearing the bit to zero will switch off the internal low voltage detector circuits. As the low voltage detector will consume a certain amount of power, it may be desirable to switch off the circuit when not in use, an important consideration in power sensitive battery powered applications.

• LVDC Register

Bit	7	6	5	4	3	2	1	0
Name	—	—	LVDO	LVDEN	VBGEN	VLVD2	VLVD1	VLVD0
R/W	—	—	R	R/W	R/W	R/W	R/W	R/W
POR	—	—	0	0	0	0	0	0

Bit 7~6 Unimplemented, read as “0”

Bit 5 **LVDO**: LVD Output Flag
0: No Low Voltage Detect
1: Low Voltage Detect

Bit 4 **LVDEN**: Low Voltage Detector Control
0: Disable
1: Enable

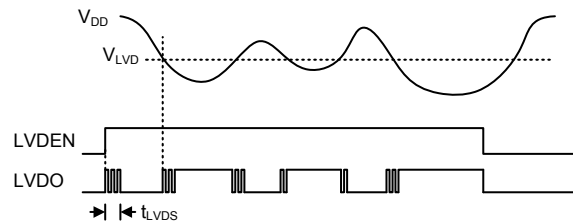
Bit 3 **VBGEN**: Bandgap Buffer Control
0: Disable
1: Enable

Note that the Bandgap circuit is enabled when the LVD or LVR function is enabled or when the VBGEN bit is set to 1.

Bit 2~0	VLVD2~VLVD0: Select LVD Voltage
	000: 2.0V
	001: 2.2V
	010: 2.4V
	011: 2.7V
	100: 3.0V
	101: 3.3V
	110: 3.6V
	111: 4.0V

LVD Operation

The Low Voltage Detector function operates by comparing the power supply voltage, V_{DD} , with a pre-specified voltage level stored in the LVDC register. This has a range of between 2.0V and 4.0V. When the power supply voltage, V_{DD} , falls below this pre-determined value, the LVDO bit will be set high indicating a low power supply voltage condition. When the device is in the SLEEP mode, the low voltage detector will be disabled even if the LVDEN bit is high. After enabling the Low Voltage Detector, a time delay t_{LVDS} should be allowed for the circuitry to stabilise before reading the LVDO bit. Note also that as the V_{DD} voltage may rise and fall rather slowly, at the voltage nears that of V_{LVD} , there may be multiple bit LVDO transitions.



LVD Operation

The Low Voltage Detector also has its own interrupt, providing an alternative means of low voltage detection, in addition to polling the LVDO bit. The interrupt will only be generated after a delay of t_{LVD} after the LVDO bit has been set high by a low voltage condition. In this case, the LVF interrupt request flag will be set, causing an interrupt to be generated if V_{DD} falls below the preset LVD voltage. This will cause the device to wake-up from the IDLE Mode, however if the Low Voltage Detector wake up function is not required then the LVF flag should be first set high before the device enters the IDLE Mode.

Interrupts

Interrupts are an important part of any microcontroller system. When an external event or an internal function such as a Timer Module or an A/D converter requires microcontroller attention, their corresponding interrupt will enforce a temporary suspension of the main program allowing the microcontroller to direct attention to their respective needs. The device contains an external interrupt and several internal interrupt functions. The external interrupt is generated by the action of the external INT pin, while the internal interrupts are generated by various internal functions such as the TMs, Time Base, LVD, EEPROM, UART and the A/D converter, etc.

Interrupt Registers

Overall interrupt control, which basically means the setting of request flags when certain microcontroller conditions occur and the setting of interrupt enable bits by the application program, is controlled by a series of registers, located in the Special Purpose Data Memory, as shown in the accompanying table. The number of registers depends upon the device chosen but fall into three categories. The first is the INTC0~INTC3 registers which configure the primary interrupts. The second is the MFI0~MFI1 registers which configure the Multi-function interrupts. Finally there is an INTEG register to configure the external interrupt trigger edge type.

Each register contains a number of enable bits to enable or disable individual interrupts as well as interrupt flags to indicate the presence of an interrupt request. The naming convention of these follows a specific pattern. First is listed an abbreviated interrupt type, then the (optional) number of that interrupt followed by either an “E” for enable/disable bit or “F” for request flag.

Function	Enable Bit	Request Flag	Notes
Global	EMI	—	—
INT Pin	INTE	INTF	—
Touch Key Module	TKME	TKMF	—
Time Base	TBE	TBF	—
Multi-function	MFnE	MFnF	n=0~1
I ² C	IICE	IICF	—
UART	URE	URF	—
LVD	LVE	LVF	—
A/D Converter	ADE	ADF	—
EEPROM write operation	DEE	DEF	—
Over voltage protection	OVPE	OVPF	—
Over circuit protection	OCPE	OCPF	—
High voltage short circuit	HVSCE	HVSCF	—
CTM	CTMnPE	CTMnPF	n=0~1
	CTMnAE	CTMnAF	
PTM	PTMPE	PTMPF	—
	PTMAE	PTMAF	

Interrupt Register Bit Naming Conventions

Register Name	Bit							
	7	6	5	4	3	2	1	0
INTEG	—	—	—	—	—	—	INTS1	INTS0
INTC0	—	TBF	TKMF	INTF	TBE	TKME	INTE	EMI
INTC1	URF	IICF	MF1F	MF0F	URE	IICE	MF1E	MF0E
INTC2	OVPF	DEF	ADF	LVF	OVPE	DEE	ADE	LVE
INTC3	—	—	HVSCF	OCPF	—	—	HVSCE	OCPE
MF10	PTMAF	PTMPF	CTM0AF	CTM0PF	PTMAE	PTMPE	CTM0AE	CTM0PE
MF11	—	—	CTM1AF	CTM1PF	—	—	CTM1AE	CTM1PE

Interrupt Register List

• **INTEG Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	INTS1	INTS0
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as “0”

Bit 1~0 **INTS1~INTS0**: Interrupt edge control for INT pin
 00: Disable
 01: Rising edge
 10: Falling edge
 11: Rising and falling edges

• **INTC0 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	TBF	TKMF	INTF	TBE	TKME	INTE	EMI
R/W	—	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	—	0	0	0	0	0	0	0

Bit 7 Unimplemented, read as “0”

Bit 6 **TBF**: Time Base interrupt request flag
 0: No request
 1: Interrupt request

Bit 5 **TKMF**: Touch Key Module interrupt request flag
 0: No request
 1: Interrupt request

Bit 4 **INTF**: INT interrupt request flag
 0: No request
 1: Interrupt request

Bit 3 **TBE**: Time Base interrupt control
 0: Disable
 1: Enable

Bit 2 **TKME**: Touch Key Module interrupt control
 0: Disable
 1: Enable

Bit 1 **INTE**: INT interrupt control
 0: Disable
 1: Enable

Bit 0 **EMI**: Global interrupt control
 0: Disable
 1: Enable

• **INTC1 Register**

Bit	7	6	5	4	3	2	1	0
Name	URF	IICF	MF1F	MF0F	URE	IICE	MF1E	MF0E
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7 **URF**: UART transfer interrupt request flag
0: No request
1: Interrupt request
- Bit 6 **IICF**: I²C interrupt request flag
0: No request
1: Interrupt request
- Bit 5 **MF1F**: Multi-function 1 interrupt request flag
0: No request
1: Interrupt request
- Bit 4 **MF0F**: Multi-function 0 interrupt request flag
0: No request
1: Interrupt request
- Bit 3 **URE**: UART transfer interrupt control
0: Disable
1: Enable
- Bit 2 **IICE**: I²C interrupt control
0: Disable
1: Enable
- Bit 1 **MF1E**: Multi-function 1 interrupt control
0: Disable
1: Enable
- Bit 0 **MF0E**: Multi-function 0 interrupt control
0: Disable
1: Enable

• **INTC2 Register**

Bit	7	6	5	4	3	2	1	0
Name	OVPF	DEF	ADF	LVF	OVPE	DEE	ADE	LVE
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7 **OVPF**: Over Voltage Protection interrupt request flag
0: No request
1: Interrupt request
- Bit 6 **DEF**: Data EEPROM interrupt request flag
0: No request
1: Interrupt request
- Bit 5 **ADF**: A/D Converter interrupt request flag
0: No request
1: Interrupt request
- Bit 4 **LVF**: LVD Interrupt request flag
0: No request
1: Interrupt request
- Bit 3 **OVPE**: Over Voltage Protection interrupt control
0: Disable
1: Enable

- Bit 2 **DEE**: Data EEPROM interrupt control
0: Disable
1: Enable
- Bit 1 **ADE**: A/D Converter interrupt control
0: Disable
1: Enable
- Bit 0 **LVE**: LVD Interrupt control
0: Disable
1: Enable

• **INTC3 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	HVSCF	OCPE	—	—	HVSCE	OCPE
R/W	—	—	R/W	R/W	—	—	R/W	R/W
POR	—	—	0	0	—	—	0	0

- Bit 7~6 Unimplemented, read as “0”
- Bit 5 **HVSCF**: High Voltage Short Circuit interrupt request flag
0: No request
1: Interrupt request
- Bit 4 **OCPE**: Over Circuit Protection interrupt request flag
0: No request
1: Interrupt request
- Bit 3~2 Unimplemented, read as “0”
- Bit 1 **HVSCE**: High Voltage Short Circuit interrupt control
0: Disable
1: Enable
- Bit 0 **OCPE**: Over Circuit Protection interrupt control
0: Disable
1: Enable

• **MFI0 Register**

Bit	7	6	5	4	3	2	1	0
Name	PTMAF	PTMPF	CTM0AF	CTM0PF	PTMAE	PTMPE	CTM0AE	CTM0PE
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
POR	0	0	0	0	0	0	0	0

- Bit 7 **PTMAF**: PTM Comparator A match interrupt request flag
0: No request
1: Interrupt request
- Bit 6 **PTMPF**: PTM Comparator P match interrupt request flag
0: No request
1: Interrupt request
- Bit 5 **CTM0AF**: CTM0 Comparator A match interrupt request flag
0: No request
1: Interrupt request
- Bit 4 **CTM0PF**: CTM0 Comparator P match interrupt request flag
0: No request
1: Interrupt request
- Bit 3 **PTMAE**: PTM Comparator A match interrupt control
0: Disable
1: Enable

- Bit 2 **PTMPE**: PTM Comparator P match interrupt control
0: Disable
1: Enable
- Bit 1 **CTM0AE**: CTM0 Comparator A match interrupt control
0: Disable
1: Enable
- Bit 0 **CTM0PE**: CTM0 Comparator P match interrupt control
0: Disable
1: Enable

• **MF11 Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	CTM1AF	CTM1PF	—	—	CTM1AE	CTM1PE
R/W	—	—	R/W	R/W	—	—	R/W	R/W
POR	—	—	0	0	—	—	0	0

- Bit 7~6 Unimplemented, read as “0”
- Bit 5 **CTM1AF**: CTM1 Comparator A match interrupt request flag
0: No request
1: Interrupt request
- Bit 4 **CTM1PF**: CTM1 Comparator P match interrupt request flag
0: No request
1: Interrupt request
- Bit 3~2 Unimplemented, read as “0”
- Bit 1 **CTM1AE**: CTM1 Comparator A match interrupt control
0: Disable
1: Enable
- Bit 0 **CTM1PE**: CTM1 Comparator P match interrupt control
0: Disable
1: Enable

Interrupt Operation

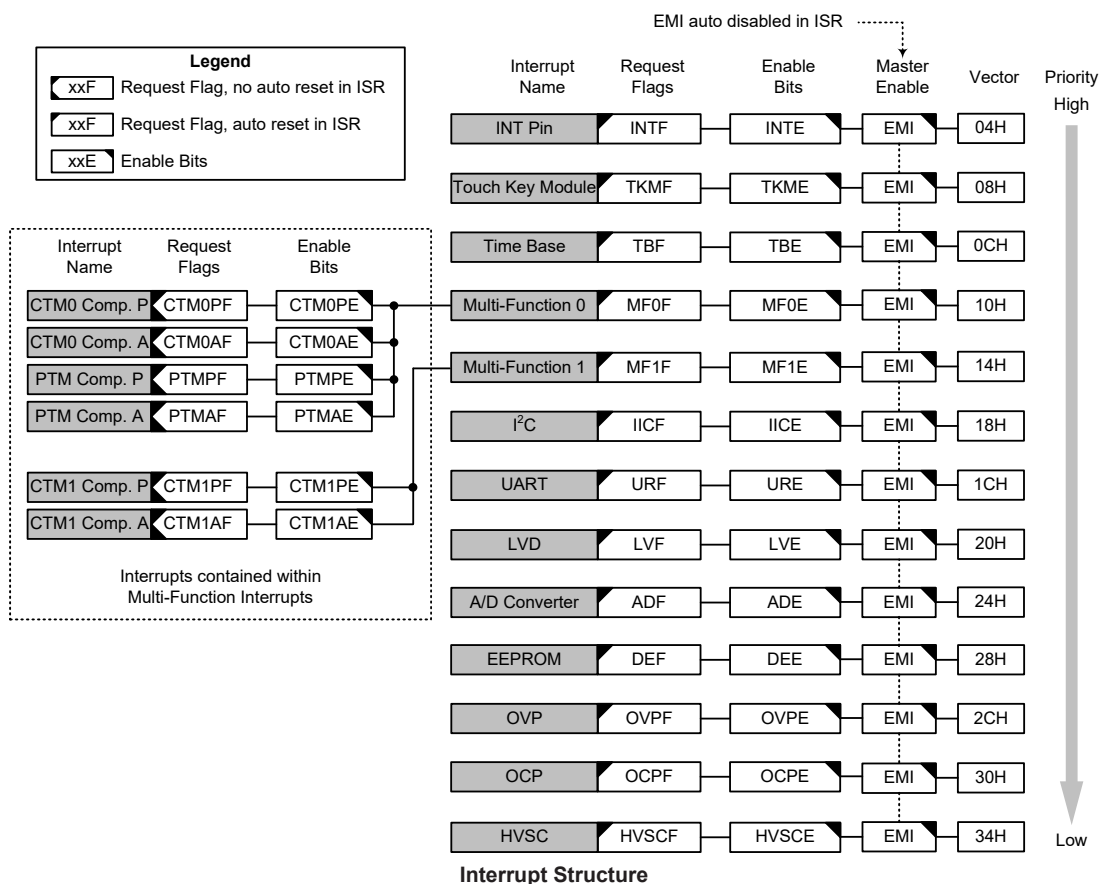
When the conditions for an interrupt event occur, such as a Touch Key Counter overflow, a TM Comparator P or Comparator A match or A/D conversion completion, etc., the relevant interrupt request flag will be set. Whether the request flag actually generates a program jump to the relevant interrupt vector is determined by the condition of the interrupt enable bit. If the enable bit is set high then the program will jump to its relevant vector; if the enable bit is zero then although the interrupt request flag is set an actual interrupt will not be generated and the program will not jump to the relevant interrupt vector. The global interrupt enable bit, if cleared to zero, will disable all interrupts.

When an interrupt is generated, the Program Counter, which stores the address of the next instruction to be executed, will be transferred onto the stack. The Program Counter will then be loaded with a new address which will be the value of the corresponding interrupt vector. The microcontroller will then fetch its next instruction from this interrupt vector. The instruction at this vector will usually be a JMP which will jump to another section of program which is known as the interrupt service routine. Here is located the code to control the appropriate interrupt. The interrupt service routine must be terminated with a RETI, which retrieves the original Program Counter address from the stack and allows the microcontroller to continue with normal execution at the point where the interrupt occurred.

The various interrupt enable bits, together with their associated request flags, are shown in the accompanying diagrams with their order of priority. Some interrupt sources have their own individual vector while others share the same multi-function interrupt vector. Once an interrupt

subroutine is serviced, all other interrupts will be blocked, as the global interrupt enable bit, EMI bit will be cleared automatically. This will prevent any further interrupt nesting from occurring. However, if other interrupt requests occur during this interval, although the interrupt will not be immediately serviced, the request flag will still be recorded.

If an interrupt requires immediate servicing while the program is already in another interrupt service routine, the EMI bit should be set after entering the routine to allow interrupt nesting. If the stack is full, the interrupt request will not be acknowledged, even if the related interrupt is enabled, until the Stack Pointer is decremented. If immediate service is desired, the stack must be prevented from becoming full. In case of simultaneous requests, the accompanying diagram shows the priority that is applied. All of the interrupt request flags when set will wake-up the device if it is in SLEEP or IDLE Mode, however to prevent a wake-up from occurring the corresponding flag should be set before the device is in SLEEP or IDLE Mode.



External Interrupt

The external interrupt is controlled by signal transitions on the INT pin. An external interrupt request will take place when the external interrupt request flag, INTF, is set, which will occur when a transition, whose type is chosen by the edge select bits, appears on the external interrupt pin. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, and respective external interrupt enable bit, INTE, must first be set. Additionally the correct interrupt edge type must be selected using the INTEG register to enable the external interrupt function and to choose the trigger edge type. As the external interrupt pin is pin-shared with I/O pins, they can only be configured as an external interrupt pin if the external interrupt enable bit in

the corresponding interrupt register has been set. The pin must also be set as an input by setting the corresponding bit in the port control register. When the interrupt is enabled, the stack is not full and the correct transition type appears on the external interrupt pin, a subroutine call to the external interrupt vector, will take place. When the interrupt is serviced, the external interrupt request flags, INTF, will be automatically reset and the EMI bit will be automatically cleared to disable other interrupts. Note that any pull-high resistor selections on the external interrupt pins will remain valid even if the pin is used as an external interrupt input.

The INTEG register is used to select the type of active edge that will trigger the external interrupt. A choice of either rising or falling or both edge types can be chosen to trigger an external interrupt. Note that the INTEG register can also be used to disable the external interrupt function.

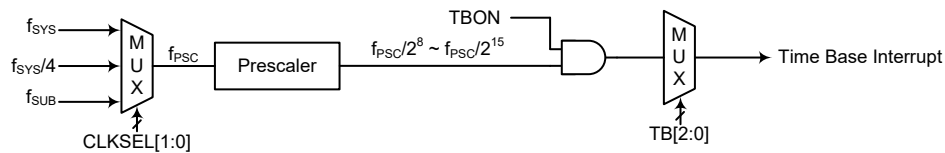
Touch Key Module Interrupt

For a Touch Key interrupt to occur, the global interrupt enable bit, EMI, and the Touch Key interrupt enable bit, TKME, must be first set. An actual Touch Key interrupt will take place when the Touch Key interrupt request flag, TMKF, is set, a situation that will occur when the time slot counter overflows. When the interrupt is enabled, the stack is not full and the Touch Key time slot counter overflow occurs, a subroutine call to the relevant interrupt vector, will take place. When the interrupt is serviced, the Touch Key interrupt request flag will be automatically reset and the EMI bit will also be automatically cleared to disable other interrupts.

Time Base Interrupt

The function of the Time Base Interrupt is to provide regular time signal in the form of an internal interrupt. It is controlled by the overflow signal from its internal timer. When this happens its interrupt request flag, TBF, will be set. To allow the program to branch to its respective interrupt vector addresses, the global interrupt enable bit, EMI and Time Base enable bit, TBE, must first be set. When the interrupt is enabled, the stack is not full and the Time Base overflows, a subroutine call to its respective vector location will take place. When the interrupt is serviced, the interrupt request flag, TBF, will be automatically reset and the EMI bit will be cleared to disable other interrupts.

The purpose of the Time Base Interrupt is to provide an interrupt signal at fixed time periods. Its clock source, f_{PSC} , originates from the internal clock source f_{SYS} , $f_{SYS}/4$ or f_{SUB} and then passes through a divider, the division ratio of which is selected by programming the appropriate bits in the TBC register to obtain longer interrupt periods whose value ranges. The clock source which in turn controls the Time Base interrupt period is selected using the CLKSEL[1:0] bits in the PSCR register.



Time Base Interrupt

• **PSCR Register**

Bit	7	6	5	4	3	2	1	0
Name	—	—	—	—	—	—	CLKSEL1	CLKSEL0
R/W	—	—	—	—	—	—	R/W	R/W
POR	—	—	—	—	—	—	0	0

Bit 7~2 Unimplemented, read as “0”

Bit 1~0 **CLKSEL1~CLKSEL0**: Prescaler clock source f_{PSC} selection
 00: f_{SYS}
 01: $f_{SYS}/4$
 1x: f_{SUB}

• **TBC Register**

Bit	7	6	5	4	3	2	1	0
Name	TBON	—	—	—	—	TB2	TB1	TB0
R/W	R/W	—	—	—	—	R/W	R/W	R/W
POR	0	—	—	—	—	0	0	0

Bit 7 **TBON**: Time Base Enable Control
 0: Disable
 1: Enable

Bit 6~4 Unimplemented, read as “0”

Bit 2~0 **TB2~TB0**: Time Base time-out period selection
 000: $2^8/f_{PSC}$
 001: $2^9/f_{PSC}$
 010: $2^{10}/f_{PSC}$
 011: $2^{11}/f_{PSC}$
 100: $2^{12}/f_{PSC}$
 101: $2^{13}/f_{PSC}$
 110: $2^{14}/f_{PSC}$
 111: $2^{15}/f_{PSC}$

Multi-function Interrupts

Within the device there are two Multi-function interrupts. Unlike the other independent interrupts, these interrupts have no independent source, but rather are formed from other existing interrupt sources, namely the TM interrupts.

A Multi-function interrupt request will take place when the Multi-function interrupt request flag MFnF is set. The Multi-function interrupt flag will be set when any of its included functions generate an interrupt request flag. To allow the program to branch to its respective interrupt vector address, when the Multi-function interrupt is enabled and the stack is not full, and one of the interrupts contained within the Multi-function interrupt occurs, a subroutine call to the Multi-function interrupt vector will take place. When the interrupt is serviced, the related Multi-Function request flag will be automatically reset and the EMI bit will be automatically cleared to disable other interrupts.

However, it must be noted that, although the Multi-function Interrupt request flag will be automatically reset when the interrupt is serviced, the request flags from the original source of the Multi-function interrupt will not be automatically reset and must be manually reset by the application program.

I²C Interrupt

An I²C interrupt request will take place when the I²C Interrupt request flag, IICF, is set, which occurs when a byte of data has been received or transmitted by the I²C interface, or an I²C slave address match occurs, or an I²C bus time-out occurs. To allow the program to branch to its respective

interrupt vector address, the global interrupt enable bit, EMI, and the Serial Interface Interrupt enable bit, IICE, must first be set. When the interrupt is enabled, the stack is not full and any of the above described situations occurs, a subroutine call to the respective Interrupt vector, will take place. When the interrupt is serviced, the Serial Interface Interrupt flag, IICF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

UART Transfer Interrupt

The UART Transfer Interrupt is controlled by several individual UART transfer conditions. When one of these conditions occurs, an interrupt pulse will be generated to get the attention of the microcontroller. These conditions are a transmitter data register empty, transmitter idle, receiver data available, receiver overrun, address detect and an RX pin wake-up. To allow the program to branch to the respective interrupt vector addresses, the global interrupt enable bit, EMI, and the UART interrupt enable bit, URE, must first be set. When the interrupt is enabled, the stack is not full and any of these conditions are created, a subroutine call to the UART Interrupt vector, will take place. When the UART Interrupt is serviced, the UART Interrupt flag, URF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts. However, the USR register flags will only be cleared when certain actions are taken by the UART, the details of which are given in the UART Interfaces chapter.

LVD Interrupt

The Low Voltage Detector Interrupt is an individual interrupt source with its own interrupt vector. An LVD Interrupt request will take place when the LVD Interrupt request flag, LVF, is set, which occurs when the Low Voltage Detector function detects a low power supply voltage. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, and Low Voltage Interrupt enable bit, LVE, must first be set. When the interrupt is enabled, the stack is not full and a low voltage condition occurs, a subroutine call to the relevant interrupt vector will take place. When the Low Voltage Interrupt is serviced, the LVF flag will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

A/D Converter Interrupt

The A/D Converter Interrupt is controlled by the termination of an A/D conversion process. An A/D Converter Interrupt request will take place when the A/D Converter Interrupt request flag, ADF, is set, which occurs when the A/D conversion process finishes. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, and A/D Interrupt enable bit, ADE, must first be set. When the interrupt is enabled, the stack is not full and the A/D conversion process has ended, a subroutine call to the A/D Converter Interrupt vector will take place. When the interrupt is serviced, the A/D Converter Interrupt flag, ADF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

EEPROM Interrupt

The EEPROM Write Interrupt is an individual interrupt source with its own interrupt vector. An EEPROM Write Interrupt request will take place when the EEPROM Write Interrupt request flag, DEF, is set, which occurs when an EEPROM Write cycle ends. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, and EEPROM Write Interrupt enable bit, DEE, must first be set. When the interrupt is enabled, the stack is not full and an EEPROM Write cycle ends, a subroutine call to the respective interrupt vector will take place. When the EEPROM Write Interrupt is serviced, the DEF flag will be automatically cleared and the EMI bit will also be automatically cleared to disable other interrupts.

Over Voltage Protection Interrupt

The OVP Interrupt is controlled by detecting the OVP input voltage. An OVP Interrupt request will take place when the OVP Interrupt request flag, OVPF, is set, which occurs when the Over Voltage Protection circuit detects an over voltage condition. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, and OVP Interrupt enable bit, OVPE, must first be set. When the interrupt is enabled, the stack is not full and an over voltage is detected, a subroutine call to the OVP Interrupt vector, will take place. When the interrupt is serviced, the OVP Interrupt flag, OVPF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

Over Current Protection Interrupt

The OCP Interrupt is controlled by detecting the OCP input current. An OCP Interrupt request will take place when the OCP Interrupt request flag, OCPF, is set, which occurs when the Over Current Protection circuit detects an over current condition. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, and OCP Interrupt enable bit, OCPE, must first be set. When the interrupt is enabled, the stack is not full and an over current is detected, a subroutine call to the OCP Interrupt vector, will take place. When the interrupt is serviced, the OCP Interrupt flag, OCPF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

High Voltage Short Circuit Interrupt

The High Voltage Short Circuit Interrupt is controlled by detecting the high voltage I/O port circuit condition. An High Voltage Short Circuit Interrupt request will take place when its interrupt request flag, HVSCF, is set, which occurs when a short circuit situation occurs on any one of the high voltage I/O pins. To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, and High Voltage Short Circuit Interrupt enable bit, HVSCE, must first be set. When the interrupt is enabled, the stack is not full and a short circuit condition is detected, a subroutine call to the High Voltage Short Circuit Interrupt vector, will take place. When the interrupt is serviced, the High Voltage Short Circuit Interrupt flag, HVSCF, will be automatically cleared. The EMI bit will also be automatically cleared to disable other interrupts.

TM Interrupts

The Compact and Periodic TMs each have two interrupts, one comes from the comparator A match situation and the other comes from the comparator P match situation. All of the TM interrupts are contained within the Multi-function Interrupts. For each of the Compact and Periodic TMs there are two interrupt request flags and two enable control bits. A TM interrupt request will take place when any of the TM request flags are set, a situation which occurs when a TM comparator P or A match situation happens.

To allow the program to branch to its respective interrupt vector address, the global interrupt enable bit, EMI, the respective TM Interrupt enable bit and the relevant Multi-function Interrupt enable bit, MFnE, must also be set. When the interrupt is enabled, the stack is not full and a TM comparator match situation occurs, a subroutine call to the relevant TM Interrupt vector locations will take place. When the TM interrupt is serviced, the EMI bit will be automatically cleared to disable other interrupts. However, only the related MFnF flag will be automatically cleared. As the TM interrupt request flags will not be automatically cleared, it has to be cleared by the application program.

Interrupt Wake-up Function

Each of the interrupt functions has the capability of waking up the microcontroller when in the SLEEP or IDLE Mode. A wake-up is generated when an interrupt request flag changes from low

to high and is independent of whether the interrupt is enabled or not. Therefore, even though this device is in the SLEEP or IDLE Mode and its system oscillator stopped, situations such as external edge transitions on the external interrupt pin or a low power supply voltage may cause their respective interrupt flag to be set high and consequently generate an interrupt. Care must therefore be taken if spurious wake-up situations are to be avoided. If an interrupt wake-up function is to be disabled then the corresponding interrupt request flag should be set high before the device enters the SLEEP or IDLE Mode. The interrupt enable bits have no effect on the interrupt wake-up function.

Programming Considerations

By disabling the relevant interrupt enable bits, a requested interrupt can be prevented from being serviced, however, once an interrupt request flag is set, it will remain in this condition in the interrupt register until the corresponding interrupt is serviced or until the request flag is cleared by the application program.

Where a certain interrupt is contained within a Multi-function interrupt, then when the interrupt service routine is executed, as only the Multi-function interrupt request flags, MF_nF, will be automatically cleared, the individual request flag for the function needs to be cleared by the application program.

It is recommended that programs do not use the “CALL” instruction within the interrupt service subroutine. Interrupts often occur in an unpredictable manner or need to be serviced immediately. If only one stack is left and the interrupt is not well controlled, the original control sequence will be damaged once a CALL subroutine is executed in the interrupt subroutine.

Every interrupt has the capability of waking up the microcontroller when it is in the SLEEP or IDLE Mode, the wake up being generated when the interrupt request flag changes from low to high. If it is required to prevent a certain interrupt from waking up the microcontroller then its respective request flag should be first set high before enter SLEEP or IDLE Mode.

As only the Program Counter is pushed onto the stack, then when the interrupt is serviced, if the contents of the accumulator, status register or other registers are altered by the interrupt service program, their contents should be saved to the memory at the beginning of the interrupt service routine.

To return from an interrupt subroutine, either a RET or RETI instruction may be executed. The RETI instruction in addition to executing a return to the main program also automatically sets the EMI bit high to allow further interrupts. The RET instruction however only executes a return to the main program leaving the EMI bit in its present zero state and therefore disabling the execution of further interrupts.

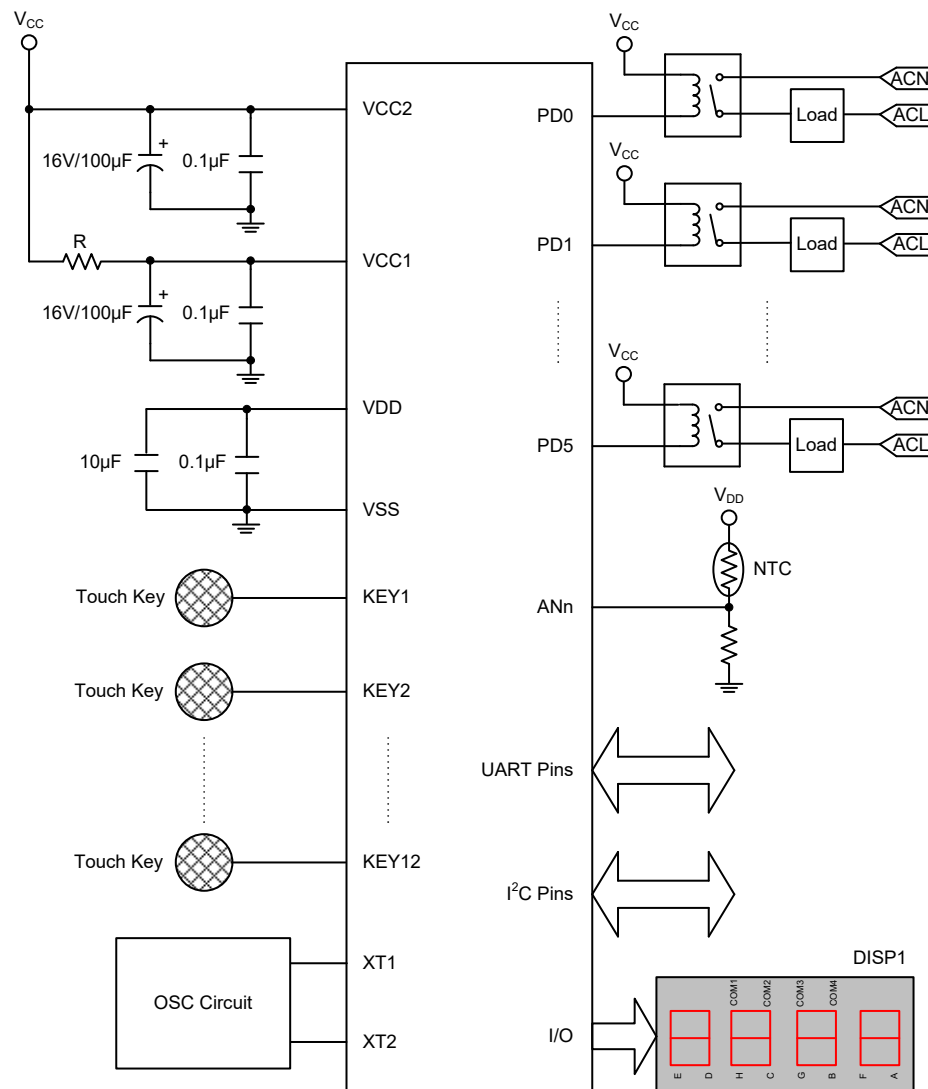
Configuration Options

Configuration options refer to certain options within the MCU that are programmed into the device during the programming process. During the development process, these options are selected using the HT-IDE software development tools. The option must be defined for proper system function, the details of which are shown in the table.

No.	Options
Oscillator Option	
1	HIRC frequency selection – f _{HIRC} : 8MHz, 12MHz or 16MHz

Note: When the HIRC has been configured at a frequency shown in this table, the HIRC1 and HIRC0 bits should also be set to select the same frequency to achieve the HIRC frequency accuracy specified in the A.C. Characteristics.

Application Circuits



Instruction Set

Introduction

Central to the successful operation of any microcontroller is its instruction set, which is a set of program instruction codes that directs the microcontroller to perform certain operations. In the case of Holtek microcontroller, a comprehensive and flexible set of over 60 instructions is provided to enable programmers to implement their application with the minimum of programming overheads.

For easier understanding of the various instruction codes, they have been subdivided into several functional groupings.

Instruction Timing

Most instructions are implemented within one instruction cycle. The exceptions to this are branch, call, or table read instructions where two instruction cycles are required. One instruction cycle is equal to 4 system clock cycles, therefore in the case of an 8MHz system oscillator, most instructions would be implemented within 0.5 μ s and branch or call instructions would be implemented within 1 μ s. Although instructions which require one more cycle to implement are generally limited to the JMP, CALL, RET, RETI and table read instructions, it is important to realize that any other instructions which involve manipulation of the Program Counter Low register or PCL will also take one more cycle to implement. As instructions which change the contents of the PCL will imply a direct jump to that new address, one more cycle will be required. Examples of such instructions would be "CLR PCL" or "MOV PCL, A". For the case of skip instructions, it must be noted that if the result of the comparison involves a skip operation then this will also take one more cycle, if no skip is involved then only one cycle is required.

Moving and Transferring Data

The transfer of data within the microcontroller program is one of the most frequently used operations. Making use of several kinds of MOV instructions, data can be transferred from registers to the Accumulator and vice-versa as well as being able to move specific immediate data directly into the Accumulator. One of the most important data transfer applications is to receive data from the input ports and transfer data to the output ports.

Arithmetic Operations

The ability to perform certain arithmetic operations and data manipulation is a necessary feature of most microcontroller applications. Within the Holtek microcontroller instruction set are a range of add and subtract instruction mnemonics to enable the necessary arithmetic to be carried out. Care must be taken to ensure correct handling of carry and borrow data when results exceed 255 for addition and less than 0 for subtraction. The increment and decrement instructions such as INC, INCA, DEC and DECA provide a simple means of increasing or decreasing by a value of one of the values in the destination specified.

Logical and Rotate Operation

The standard logical operations such as AND, OR, XOR and CPL all have their own instruction within the Holtek microcontroller instruction set. As with the case of most instructions involving data manipulation, data must pass through the Accumulator which may involve additional programming steps. In all logical data operations, the zero flag may be set if the result of the operation is zero. Another form of logical data manipulation comes from the rotate instructions such as RR, RL, RRC and RLC which provide a simple means of rotating one bit right or left. Different rotate instructions exist depending on program requirements. Rotate instructions are useful for serial port programming applications where data can be rotated from an internal register into the Carry bit from where it can be examined and the necessary serial bit set high or low. Another application which rotate data operations are used is to implement multiplication and division calculations.

Branches and Control Transfer

Program branching takes the form of either jumps to specified locations using the JMP instruction or to a subroutine using the CALL instruction. They differ in the sense that in the case of a subroutine call, the program must return to the instruction immediately when the subroutine has been carried out. This is done by placing a return instruction “RET” in the subroutine which will cause the program to jump back to the address right after the CALL instruction. In the case of a JMP instruction, the program simply jumps to the desired location. There is no requirement to jump back to the original jumping off point as in the case of the CALL instruction. One special and extremely useful set of branch instructions are the conditional branches. Here a decision is first made regarding the condition of a certain data memory or individual bits. Depending upon the conditions, the program will continue with the next instruction or skip over it and jump to the following instruction. These instructions are the key to decision making and branching within the program perhaps determined by the condition of certain input switches or by the condition of internal data bits.

Bit Operations

The ability to provide single bit operations on Data Memory is an extremely flexible feature of all Holtek microcontrollers. This feature is especially useful for output port bit programming where individual bits or port pins can be directly set high or low using either the “SET [m].i” or “CLR [m].i” instructions respectively. The feature removes the need for programmers to first read the 8-bit output port, manipulate the input data to ensure that other bits are not changed and then output the port with the correct new data. This read-modify-write process is taken care of automatically when these bit operation instructions are used.

Table Read Operations

Data storage is normally implemented by using registers. However, when working with large amounts of fixed data, the volume involved often makes it inconvenient to store the fixed data in the Data Memory. To overcome this problem, Holtek microcontrollers allow an area of Program Memory to be setup as a table where data can be directly stored. A set of easy to use instructions provides the means by which this fixed data can be referenced and retrieved from the Program Memory.

Other Operations

In addition to the above functional instructions, a range of other instructions also exist such as the “HALT” instruction for Power-down operations and instructions to control the operation of the Watchdog Timer for reliable program operations under extreme electric or electromagnetic environments. For their relevant operations, refer to the functional related sections.

Instruction Set Summary

The instructions related to the data memory access in the following table can be used when the desired data memory is located in Data Memory sector 0.

Table Conventions

x: Bits immediate data
m: Data Memory address
A: Accumulator
i: 0~7 number of bits
addr: Program memory address

Mnemonic	Description	Cycles	Flag Affected
Arithmetic			
ADD A,[m]	Add Data Memory to ACC	1	Z, C, AC, OV, SC
ADDM A,[m]	Add ACC to Data Memory	1 ^{Note}	Z, C, AC, OV, SC
ADD A,x	Add immediate data to ACC	1	Z, C, AC, OV, SC
ADC A,[m]	Add Data Memory to ACC with Carry	1	Z, C, AC, OV, SC
ADCM A,[m]	Add ACC to Data memory with Carry	1 ^{Note}	Z, C, AC, OV, SC
SUB A,x	Subtract immediate data from the ACC	1	Z, C, AC, OV, SC, CZ
SUB A,[m]	Subtract Data Memory from ACC	1	Z, C, AC, OV, SC, CZ
SUBM A,[m]	Subtract Data Memory from ACC with result in Data Memory	1 ^{Note}	Z, C, AC, OV, SC, CZ
SBC A,x	Subtract immediate data from ACC with Carry	1	Z, C, AC, OV, SC, CZ
SBC A,[m]	Subtract Data Memory from ACC with Carry	1	Z, C, AC, OV, SC, CZ
SBCM A,[m]	Subtract Data Memory from ACC with Carry, result in Data Memory	1 ^{Note}	Z, C, AC, OV, SC, CZ
DAA [m]	Decimal adjust ACC for Addition with result in Data Memory	1 ^{Note}	C
Logic Operation			
AND A,[m]	Logical AND Data Memory to ACC	1	Z
OR A,[m]	Logical OR Data Memory to ACC	1	Z
XOR A,[m]	Logical XOR Data Memory to ACC	1	Z
ANDM A,[m]	Logical AND ACC to Data Memory	1 ^{Note}	Z
ORM A,[m]	Logical OR ACC to Data Memory	1 ^{Note}	Z
XORM A,[m]	Logical XOR ACC to Data Memory	1 ^{Note}	Z
AND A,x	Logical AND immediate Data to ACC	1	Z
OR A,x	Logical OR immediate Data to ACC	1	Z
XOR A,x	Logical XOR immediate Data to ACC	1	Z
CPL [m]	Complement Data Memory	1 ^{Note}	Z
CPLA [m]	Complement Data Memory with result in ACC	1	Z
Increment & Decrement			
INCA [m]	Increment Data Memory with result in ACC	1	Z
INC [m]	Increment Data Memory	1 ^{Note}	Z
DECA [m]	Decrement Data Memory with result in ACC	1	Z
DEC [m]	Decrement Data Memory	1 ^{Note}	Z
Rotate			
RRA [m]	Rotate Data Memory right with result in ACC	1	None
RR [m]	Rotate Data Memory right	1 ^{Note}	None
RRCA [m]	Rotate Data Memory right through Carry with result in ACC	1	C
RRC [m]	Rotate Data Memory right through Carry	1 ^{Note}	C
RLA [m]	Rotate Data Memory left with result in ACC	1	None
RL [m]	Rotate Data Memory left	1 ^{Note}	None
RLCA [m]	Rotate Data Memory left through Carry with result in ACC	1	C
RLC [m]	Rotate Data Memory left through Carry	1 ^{Note}	C

Mnemonic	Description	Cycles	Flag Affected
Data Move			
MOV A,[m]	Move Data Memory to ACC	1	None
MOV [m],A	Move ACC to Data Memory	1 ^{Note}	None
MOV A,x	Move immediate data to ACC	1	None
Bit Operation			
CLR [m].i	Clear bit of Data Memory	1 ^{Note}	None
SET [m].i	Set bit of Data Memory	1 ^{Note}	None
Branch Operation			
JMP addr	Jump unconditionally	2	None
SZ [m]	Skip if Data Memory is zero	1 ^{Note}	None
SZA [m]	Skip if Data Memory is zero with data movement to ACC	1 ^{Note}	None
SZ [m].i	Skip if bit i of Data Memory is zero	1 ^{Note}	None
SNZ [m]	Skip if Data Memory is not zero	1 ^{Note}	None
SNZ [m].i	Skip if bit i of Data Memory is not zero	1 ^{Note}	None
SIZ [m]	Skip if increment Data Memory is zero	1 ^{Note}	None
SDZ [m]	Skip if decrement Data Memory is zero	1 ^{Note}	None
SIZA [m]	Skip if increment Data Memory is zero with result in ACC	1 ^{Note}	None
SDZA [m]	Skip if decrement Data Memory is zero with result in ACC	1 ^{Note}	None
CALL addr	Subroutine call	2	None
RET	Return from subroutine	2	None
RET A,x	Return from subroutine and load immediate data to ACC	2	None
RETI	Return from interrupt	2	None
Table Read Operation			
TABRD [m]	Read table (specific page) to TBLH and Data Memory	2 ^{Note}	None
TABRDL [m]	Read table (last page) to TBLH and Data Memory	2 ^{Note}	None
ITABRD [m]	Increment table pointer TBLP first and Read table to TBLH and Data Memory	2 ^{Note}	None
ITABRDL [m]	Increment table pointer TBLP first and Read table (last page) to TBLH and Data Memory	2 ^{Note}	None
Miscellaneous			
NOP	No operation	1	None
CLR [m]	Clear Data Memory	1 ^{Note}	None
SET [m]	Set Data Memory	1 ^{Note}	None
CLR WDT	Clear Watchdog Timer	1	TO, PDF
SWAP [m]	Swap nibbles of Data Memory	1 ^{Note}	None
SWAPA [m]	Swap nibbles of Data Memory with result in ACC	1	None
HALT	Enter power down mode	1	TO, PDF

Note: 1. For skip instructions, if the result of the comparison involves a skip then up to three cycles are required, if no skip takes place only one cycle is required.

2. Any instruction which changes the contents of the PCL will also require 2 cycles for execution.

3. For the “CLR WDT” instruction the TO and PDF flags may be affected by the execution status. The TO and PDF flags are cleared after the “CLR WDT” instructions is executed. Otherwise the TO and PDF flags remain unchanged.

Extended Instruction Set

The extended instructions are used to support the full range address access for the data memory. When the accessed data memory is located in any data memory sector except sector 0, the extended instruction can be used to directly access the data memory instead of using the indirect addressing access. This can not only reduce the use of Flash memory space but also improve the CPU execution efficiency.

Mnemonic	Description	Cycles	Flag Affected
Arithmetic			
LADD A,[m]	Add Data Memory to ACC	2	Z, C, AC, OV, SC
LADDM A,[m]	Add ACC to Data Memory	2 ^{Note}	Z, C, AC, OV, SC
LADC A,[m]	Add Data Memory to ACC with Carry	2	Z, C, AC, OV, SC
LADCM A,[m]	Add ACC to Data memory with Carry	2 ^{Note}	Z, C, AC, OV, SC
LSUB A,[m]	Subtract Data Memory from ACC	2	Z, C, AC, OV, SC, CZ
LSUBM A,[m]	Subtract Data Memory from ACC with result in Data Memory	2 ^{Note}	Z, C, AC, OV, SC, CZ
LSBC A,[m]	Subtract Data Memory from ACC with Carry	2	Z, C, AC, OV, SC, CZ
LSBCM A,[m]	Subtract Data Memory from ACC with Carry, result in Data Memory	2 ^{Note}	Z, C, AC, OV, SC, CZ
LDAA [m]	Decimal adjust ACC for Addition with result in Data Memory	2 ^{Note}	C
Logic Operation			
LAND A,[m]	Logical AND Data Memory to ACC	2	Z
LOR A,[m]	Logical OR Data Memory to ACC	2	Z
LXOR A,[m]	Logical XOR Data Memory to ACC	2	Z
LANDM A,[m]	Logical AND ACC to Data Memory	2 ^{Note}	Z
LORM A,[m]	Logical OR ACC to Data Memory	2 ^{Note}	Z
LXORM A,[m]	Logical XOR ACC to Data Memory	2 ^{Note}	Z
LCPL [m]	Complement Data Memory	2 ^{Note}	Z
LCPLA [m]	Complement Data Memory with result in ACC	2	Z
Increment & Decrement			
LINCA [m]	Increment Data Memory with result in ACC	2	Z
LINC [m]	Increment Data Memory	2 ^{Note}	Z
LDECA [m]	Decrement Data Memory with result in ACC	2	Z
LDEC [m]	Decrement Data Memory	2 ^{Note}	Z
Rotate			
LRRA [m]	Rotate Data Memory right with result in ACC	2	None
LRR [m]	Rotate Data Memory right	2 ^{Note}	None
LRRCA [m]	Rotate Data Memory right through Carry with result in ACC	2	C
LRRC [m]	Rotate Data Memory right through Carry	2 ^{Note}	C
LRLA [m]	Rotate Data Memory left with result in ACC	2	None
LRL [m]	Rotate Data Memory left	2 ^{Note}	None
LRLCA [m]	Rotate Data Memory left through Carry with result in ACC	2	C
LRLC [m]	Rotate Data Memory left through Carry	2 ^{Note}	C
Data Move			
LMOV A,[m]	Move Data Memory to ACC	2	None
LMOV [m],A	Move ACC to Data Memory	2 ^{Note}	None
Bit Operation			
LCLR [m].i	Clear bit of Data Memory	2 ^{Note}	None
LSET [m].i	Set bit of Data Memory	2 ^{Note}	None

Mnemonic	Description	Cycles	Flag Affected
Branch			
LSZ [m]	Skip if Data Memory is zero	2 ^{Note}	None
LSZA [m]	Skip if Data Memory is zero with data movement to ACC	2 ^{Note}	None
LSNZ [m]	Skip if Data Memory is not zero	2 ^{Note}	None
LSZ [m].i	Skip if bit i of Data Memory is zero	2 ^{Note}	None
LSNZ [m].i	Skip if bit i of Data Memory is not zero	2 ^{Note}	None
LSIZ [m]	Skip if increment Data Memory is zero	2 ^{Note}	None
LSDZ [m]	Skip if decrement Data Memory is zero	2 ^{Note}	None
LSIZA [m]	Skip if increment Data Memory is zero with result in ACC	2 ^{Note}	None
LSDZA [m]	Skip if decrement Data Memory is zero with result in ACC	2 ^{Note}	None
Table Read			
LTABRD [m]	Read table to TBLH and Data Memory	3 ^{Note}	None
LTABRDL [m]	Read table (last page) to TBLH and Data Memory	3 ^{Note}	None
LITABRD [m]	Increment table pointer TBLP first and Read table to TBLH and Data Memory	3 ^{Note}	None
LITABRDL [m]	Increment table pointer TBLP first and Read table (last page) to TBLH and Data Memory	3 ^{Note}	None
Miscellaneous			
LCLR [m]	Clear Data Memory	2 ^{Note}	None
LSET [m]	Set Data Memory	2 ^{Note}	None
LSWAP [m]	Swap nibbles of Data Memory	2 ^{Note}	None
LSWAPA [m]	Swap nibbles of Data Memory with result in ACC	2	None

Note: 1. For these extended skip instructions, if the result of the comparison involves a skip then up to four cycles are required, if no skip takes place two cycles is required.

2. Any extended instruction which changes the contents of the PCL register will also require three cycles for execution.

Instruction Definition

ADC A,[m]	Add Data Memory to ACC with Carry
Description	The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC + [m] + C$
Affected flag(s)	OV, Z, AC, C, SC
ADCM A,[m]	Add ACC to Data Memory with Carry
Description	The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the specified Data Memory.
Operation	$[m] \leftarrow ACC + [m] + C$
Affected flag(s)	OV, Z, AC, C, SC
ADD A,[m]	Add Data Memory to ACC
Description	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC + [m]$
Affected flag(s)	OV, Z, AC, C, SC
ADD A,x	Add immediate data to ACC
Description	The contents of the Accumulator and the specified immediate data are added. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC + x$
Affected flag(s)	OV, Z, AC, C, SC
ADDM A,[m]	Add ACC to Data Memory
Description	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory.
Operation	$[m] \leftarrow ACC + [m]$
Affected flag(s)	OV, Z, AC, C, SC
AND A,[m]	Logical AND Data Memory to ACC
Description	Data in the Accumulator and the specified Data Memory perform a bitwise logical AND operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC \text{ "AND" } [m]$
Affected flag(s)	Z
AND A,x	Logical AND immediate data to ACC
Description	Data in the Accumulator and the specified immediate data perform a bit wise logical AND operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC \text{ "AND" } x$
Affected flag(s)	Z
ANDM A,[m]	Logical AND ACC to Data Memory
Description	Data in the specified Data Memory and the Accumulator perform a bitwise logical AND operation. The result is stored in the Data Memory.
Operation	$[m] \leftarrow ACC \text{ "AND" } [m]$
Affected flag(s)	Z

CALL addr	Subroutine call
Description	Unconditionally calls a subroutine at the specified address. The Program Counter then increments by 1 to obtain the address of the next instruction which is then pushed onto the stack. The specified address is then loaded and the program continues execution from this new address. As this instruction requires an additional operation, it is a two cycle instruction.
Operation	Stack $\leftarrow$ Program Counter + 1 Program Counter $\leftarrow$ addr
Affected flag(s)	None
CLR [m]	Clear Data Memory
Description	Each bit of the specified Data Memory is cleared to 0.
Operation	[m] $\leftarrow$ 00H
Affected flag(s)	None
CLR [m].i	Clear bit of Data Memory
Description	Bit i of the specified Data Memory is cleared to 0.
Operation	[m].i $\leftarrow$ 0
Affected flag(s)	None
CLR WDT	Clear Watchdog Timer
Description	The TO, PDF flags and the WDT are all cleared.
Operation	WDT cleared TO $\leftarrow$ 0 PDF $\leftarrow$ 0
Affected flag(s)	TO, PDF
CPL [m]	Complement Data Memory
Description	Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa.
Operation	[m] $\leftarrow$ $\overline{[m]}$
Affected flag(s)	Z
CPLA [m]	Complement Data Memory with result in ACC
Description	Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa. The complemented result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	ACC $\leftarrow$ $\overline{[m]}$
Affected flag(s)	Z
DAA [m]	Decimal-Adjust ACC for addition with result in Data Memory
Description	Convert the contents of the Accumulator value to a BCD (Binary Coded Decimal) value resulting from the previous addition of two BCD variables. If the low nibble is greater than 9 or if AC flag is set, then a value of 6 will be added to the low nibble. Otherwise the low nibble remains unchanged. If the high nibble is greater than 9 or if the C flag is set, then a value of 6 will be added to the high nibble. Essentially, the decimal conversion is performed by adding 00H, 06H, 60H or 66H depending on the Accumulator and flag conditions. Only the C flag may be affected by this instruction which indicates that if the original BCD sum is greater than 100, it allows multiple precision decimal addition.
Operation	[m] $\leftarrow$ ACC + 00H or [m] $\leftarrow$ ACC + 06H or [m] $\leftarrow$ ACC + 60H or [m] $\leftarrow$ ACC + 66H
Affected flag(s)	C

DEC [m]	Decrement Data Memory
Description	Data in the specified Data Memory is decremented by 1.
Operation	$[m] \leftarrow [m] - 1$
Affected flag(s)	Z
DECA [m]	Decrement Data Memory with result in ACC
Description	Data in the specified Data Memory is decremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	$ACC \leftarrow [m] - 1$
Affected flag(s)	Z
HALT	Enter power down mode
Description	This instruction stops the program execution and turns off the system clock. The contents of the Data Memory and registers are retained. The WDT and prescaler are cleared. The power down flag PDF is set and the WDT time-out flag TO is cleared.
Operation	$TO \leftarrow 0$ $PDF \leftarrow 1$
Affected flag(s)	TO, PDF
INC [m]	Increment Data Memory
Description	Data in the specified Data Memory is incremented by 1.
Operation	$[m] \leftarrow [m] + 1$
Affected flag(s)	Z
INCA [m]	Increment Data Memory with result in ACC
Description	Data in the specified Data Memory is incremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	$ACC \leftarrow [m] + 1$
Affected flag(s)	Z
JMP addr	Jump unconditionally
Description	The contents of the Program Counter are replaced with the specified address. Program execution then continues from this new address. As this requires the insertion of a dummy instruction while the new address is loaded, it is a two cycle instruction.
Operation	$Program\ Counter \leftarrow addr$
Affected flag(s)	None
MOV A,[m]	Move Data Memory to ACC
Description	The contents of the specified Data Memory are copied to the Accumulator.
Operation	$ACC \leftarrow [m]$
Affected flag(s)	None
MOV A,x	Move immediate data to ACC
Description	The immediate data specified is loaded into the Accumulator.
Operation	$ACC \leftarrow x$
Affected flag(s)	None
MOV [m],A	Move ACC to Data Memory
Description	The contents of the Accumulator are copied to the specified Data Memory.
Operation	$[m] \leftarrow ACC$
Affected flag(s)	None

NOP	No operation
Description	No operation is performed. Execution continues with the next instruction.
Operation	No operation
Affected flag(s)	None
OR A,[m]	Logical OR Data Memory to ACC
Description	Data in the Accumulator and the specified Data Memory perform a bitwise logical OR operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC \text{ "OR" } [m]$
Affected flag(s)	Z
OR A,x	Logical OR immediate data to ACC
Description	Data in the Accumulator and the specified immediate data perform a bitwise logical OR operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC \text{ "OR" } x$
Affected flag(s)	Z
ORM A,[m]	Logical OR ACC to Data Memory
Description	Data in the specified Data Memory and the Accumulator perform a bitwise logical OR operation. The result is stored in the Data Memory.
Operation	$[m] \leftarrow ACC \text{ "OR" } [m]$
Affected flag(s)	Z
RET	Return from subroutine
Description	The Program Counter is restored from the stack. Program execution continues at the restored address.
Operation	Program Counter $\leftarrow$ Stack
Affected flag(s)	None
RET A,x	Return from subroutine and load immediate data to ACC
Description	The Program Counter is restored from the stack and the Accumulator loaded with the specified immediate data. Program execution continues at the restored address.
Operation	Program Counter $\leftarrow$ Stack $ACC \leftarrow x$
Affected flag(s)	None
RETI	Return from interrupt
Description	The Program Counter is restored from the stack and the interrupts are re-enabled by setting the EMI bit. EMI is the master interrupt global enable bit. If an interrupt was pending when the RETI instruction is executed, the pending Interrupt routine will be processed before returning to the main program.
Operation	Program Counter $\leftarrow$ Stack $EMI \leftarrow 1$
Affected flag(s)	None
RL [m]	Rotate Data Memory left
Description	The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0.
Operation	$[m].(i+1) \leftarrow [m].i; (i=0\sim6)$ $[m].0 \leftarrow [m].7$
Affected flag(s)	None

RLA [m]	Rotate Data Memory left with result in ACC
Description	The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.(i+1) \leftarrow [m].i; (i=0\sim6)$ $ACC.0 \leftarrow [m].7$
Affected flag(s)	None
RLC [m]	Rotate Data Memory left through Carry
Description	The contents of the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into bit 0.
Operation	$[m].(i+1) \leftarrow [m].i; (i=0\sim6)$ $[m].0 \leftarrow C$ $C \leftarrow [m].7$
Affected flag(s)	C
RLCA [m]	Rotate Data Memory left through Carry with result in ACC
Description	Data in the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into the bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.(i+1) \leftarrow [m].i; (i=0\sim6)$ $ACC.0 \leftarrow C$ $C \leftarrow [m].7$
Affected flag(s)	C
RR [m]	Rotate Data Memory right
Description	The contents of the specified Data Memory are rotated right by 1 bit with bit 0 rotated into bit 7.
Operation	$[m].i \leftarrow [m].(i+1); (i=0\sim6)$ $[m].7 \leftarrow [m].0$
Affected flag(s)	None
RRA [m]	Rotate Data Memory right with result in ACC
Description	Data in the specified Data Memory is rotated right by 1 bit with bit 0 rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.i \leftarrow [m].(i+1); (i=0\sim6)$ $ACC.7 \leftarrow [m].0$
Affected flag(s)	None
RRC [m]	Rotate Data Memory right through Carry
Description	The contents of the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7.
Operation	$[m].i \leftarrow [m].(i+1); (i=0\sim6)$ $[m].7 \leftarrow C$ $C \leftarrow [m].0$
Affected flag(s)	C

RRCA [m]	Rotate Data Memory right through Carry with result in ACC
Description	Data in the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.i \leftarrow [m].(i+1); (i=0\sim6)$ $ACC.7 \leftarrow C$ $C \leftarrow [m].0$
Affected flag(s)	C
SBC A,[m]	Subtract Data Memory from ACC with Carry
Description	The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - [m] - \overline{C}$
Affected flag(s)	OV, Z, AC, C, SC, CZ
SBC A, x	Subtract immediate data from ACC with Carry
Description	The immediate data and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - [m] - \overline{C}$
Affected flag(s)	OV, Z, AC, C, SC, CZ
SBCM A,[m]	Subtract Data Memory from ACC with Carry and result in Data Memory
Description	The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$[m] \leftarrow ACC - [m] - \overline{C}$
Affected flag(s)	OV, Z, AC, C, SC, CZ
SDZ [m]	Skip if decrement Data Memory is 0
Description	The contents of the specified Data Memory are first decremented by 1. If the result is 0 the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$[m] \leftarrow [m] - 1$ Skip if $[m]=0$
Affected flag(s)	None
SDZA [m]	Skip if decrement Data Memory is zero with result in ACC
Description	The contents of the specified Data Memory are first decremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0, the program proceeds with the following instruction.
Operation	$ACC \leftarrow [m] - 1$ Skip if $ACC=0$
Affected flag(s)	None

SET [m]	Set Data Memory
Description	Each bit of the specified Data Memory is set to 1.
Operation	$[m] \leftarrow FFH$
Affected flag(s)	None
SET [m].i	Set bit of Data Memory
Description	Bit i of the specified Data Memory is set to 1.
Operation	$[m].i \leftarrow 1$
Affected flag(s)	None
SIZ [m]	Skip if increment Data Memory is 0
Description	The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$[m] \leftarrow [m] + 1$ Skip if $[m]=0$
Affected flag(s)	None
SIZA [m]	Skip if increment Data Memory is zero with result in ACC
Description	The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$ACC \leftarrow [m] + 1$ Skip if $ACC=0$
Affected flag(s)	None
SNZ [m].i	Skip if Data Memory is not 0
Description	If the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is 0 the program proceeds with the following instruction.
Operation	Skip if $[m].i \neq 0$
Affected flag(s)	None
SNZ [m]	Skip if Data Memory is not 0
Description	If the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is 0 the program proceeds with the following instruction.
Operation	Skip if $[m] \neq 0$
Affected flag(s)	None
SUB A,[m]	Subtract Data Memory from ACC
Description	The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - [m]$
Affected flag(s)	OV, Z, AC, C, SC, CZ

SUBM A,[m]	Subtract Data Memory from ACC with result in Data Memory
Description	The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$[m] \leftarrow ACC - [m]$
Affected flag(s)	OV, Z, AC, C, SC, CZ
SUB A,x	Subtract immediate data from ACC
Description	The immediate data specified by the code is subtracted from the contents of the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - x$
Affected flag(s)	OV, Z, AC, C, SC, CZ
SWAP [m]	Swap nibbles of Data Memory
Description	The low-order and high-order nibbles of the specified Data Memory are interchanged.
Operation	$[m].3 \sim [m].0 \leftrightarrow [m].7 \sim [m].4$
Affected flag(s)	None
SWAPA [m]	Swap nibbles of Data Memory with result in ACC
Description	The low-order and high-order nibbles of the specified Data Memory are interchanged. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	$ACC.3 \sim ACC.0 \leftarrow [m].7 \sim [m].4$ $ACC.7 \sim ACC.4 \leftarrow [m].3 \sim [m].0$
Affected flag(s)	None
SZ [m]	Skip if Data Memory is 0
Description	If the contents of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	Skip if $[m]=0$
Affected flag(s)	None
SZA [m]	Skip if Data Memory is 0 with data movement to ACC
Description	The contents of the specified Data Memory are copied to the Accumulator. If the value is zero, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$ACC \leftarrow [m]$ Skip if $[m]=0$
Affected flag(s)	None
SZ [m].i	Skip if bit i of Data Memory is 0
Description	If bit i of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0, the program proceeds with the following instruction.
Operation	Skip if $[m].i=0$
Affected flag(s)	None

TABRD [m]	Read table (specific page) to TBLH and Data Memory
Description	The low byte of the program code (specific page) addressed by the table pointer pair (TBLP and TBHP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
TABRDL [m]	Read table (last page) to TBLH and Data Memory
Description	The low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
ITABRD [m]	Increment table pointer low byte first and read table to TBLH and Data Memory
Description	Increment table pointer low byte, TBLP, first and then the program code addressed by the table pointer (TBHP and TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
ITABRDL [m]	Increment table pointer low byte first and read table (last page) to TBLH and Data Memory
Description	Increment table pointer low byte, TBLP, first and then the low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
XOR A,[m]	Logical XOR Data Memory to ACC
Description	Data in the Accumulator and the specified Data Memory perform a bitwise logical XOR operation. The result is stored in the Accumulator.
Operation	ACC ← ACC "XOR" [m]
Affected flag(s)	Z
XORM A,[m]	Logical XOR ACC to Data Memory
Description	Data in the specified Data Memory and the Accumulator perform a bitwise logical XOR operation. The result is stored in the Data Memory.
Operation	[m] ← ACC "XOR" [m]
Affected flag(s)	Z
XOR A,x	Logical XOR immediate data to ACC
Description	Data in the Accumulator and the specified immediate data perform a bitwise logical XOR operation. The result is stored in the Accumulator.
Operation	ACC ← ACC "XOR" x
Affected flag(s)	Z

Extended Instruction Definition

The extended instructions are used to directly access the data stored in any data memory sections.

LADC A,[m]	Add Data Memory to ACC with Carry
Description	The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC + [m] + C$
Affected flag(s)	OV, Z, AC, C, SC
LADCM A,[m]	Add ACC to Data Memory with Carry
Description	The contents of the specified Data Memory, Accumulator and the carry flag are added. The result is stored in the specified Data Memory.
Operation	$[m] \leftarrow ACC + [m] + C$
Affected flag(s)	OV, Z, AC, C, SC
LADD A,[m]	Add Data Memory to ACC
Description	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC + [m]$
Affected flag(s)	OV, Z, AC, C, SC
LADDM A,[m]	Add ACC to Data Memory
Description	The contents of the specified Data Memory and the Accumulator are added. The result is stored in the specified Data Memory.
Operation	$[m] \leftarrow ACC + [m]$
Affected flag(s)	OV, Z, AC, C, SC
LAND A,[m]	Logical AND Data Memory to ACC
Description	Data in the Accumulator and the specified Data Memory perform a bitwise logical AND operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC \text{ "AND" } [m]$
Affected flag(s)	Z
LANDM A,[m]	Logical AND ACC to Data Memory
Description	Data in the specified Data Memory and the Accumulator perform a bitwise logical AND operation. The result is stored in the Data Memory.
Operation	$[m] \leftarrow ACC \text{ "AND" } [m]$
Affected flag(s)	Z
LCLR [m]	Clear Data Memory
Description	Each bit of the specified Data Memory is cleared to 0.
Operation	$[m] \leftarrow 00H$
Affected flag(s)	None
LCLR [m].i	Clear bit of Data Memory
Description	Bit i of the specified Data Memory is cleared to 0.
Operation	$[m].i \leftarrow 0$
Affected flag(s)	None

LCPL [m]	Complement Data Memory
Description	Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa.
Operation	$[m] \leftarrow \overline{[m]}$
Affected flag(s)	Z
LCPLA [m]	Complement Data Memory with result in ACC
Description	Each bit of the specified Data Memory is logically complemented (1's complement). Bits which previously contained a 1 are changed to 0 and vice versa. The complemented result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC \leftarrow \overline{[m]}$
Affected flag(s)	Z
LDAA [m]	Decimal-Adjust ACC for addition with result in Data Memory
Description	Convert the contents of the Accumulator value to a BCD (Binary Coded Decimal) value resulting from the previous addition of two BCD variables. If the low nibble is greater than 9 or if AC flag is set, then a value of 6 will be added to the low nibble. Otherwise the low nibble remains unchanged. If the high nibble is greater than 9 or if the C flag is set, then a value of 6 will be added to the high nibble. Essentially, the decimal conversion is performed by adding 00H, 06H, 60H or 66H depending on the Accumulator and flag conditions. Only the C flag may be affected by this instruction which indicates that if the original BCD sum is greater than 100, it allows multiple precision decimal addition.
Operation	$[m] \leftarrow ACC + 00H$ or $[m] \leftarrow ACC + 06H$ or $[m] \leftarrow ACC + 60H$ or $[m] \leftarrow ACC + 66H$
Affected flag(s)	C
LDEC [m]	Decrement Data Memory
Description	Data in the specified Data Memory is decremented by 1.
Operation	$[m] \leftarrow [m] - 1$
Affected flag(s)	Z
LDECA [m]	Decrement Data Memory with result in ACC
Description	Data in the specified Data Memory is decremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	$ACC \leftarrow [m] - 1$
Affected flag(s)	Z
LINC [m]	Increment Data Memory
Description	Data in the specified Data Memory is incremented by 1.
Operation	$[m] \leftarrow [m] + 1$
Affected flag(s)	Z
LINCA [m]	Increment Data Memory with result in ACC
Description	Data in the specified Data Memory is incremented by 1. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	$ACC \leftarrow [m] + 1$
Affected flag(s)	Z

LMOV A,[m]	Move Data Memory to ACC
Description	The contents of the specified Data Memory are copied to the Accumulator.
Operation	$ACC \leftarrow [m]$
Affected flag(s)	None
LMOV [m],A	Move ACC to Data Memory
Description	The contents of the Accumulator are copied to the specified Data Memory.
Operation	$[m] \leftarrow ACC$
Affected flag(s)	None
LOR A,[m]	Logical OR Data Memory to ACC
Description	Data in the Accumulator and the specified Data Memory perform a bitwise logical OR operation. The result is stored in the Accumulator.
Operation	$ACC \leftarrow ACC \text{ "OR" } [m]$
Affected flag(s)	Z
LORM A,[m]	Logical OR ACC to Data Memory
Description	Data in the specified Data Memory and the Accumulator perform a bitwise logical OR operation. The result is stored in the Data Memory.
Operation	$[m] \leftarrow ACC \text{ "OR" } [m]$
Affected flag(s)	Z
LRL [m]	Rotate Data Memory left
Description	The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0.
Operation	$[m].(i+1) \leftarrow [m].i; (i=0\sim6)$ $[m].0 \leftarrow [m].7$
Affected flag(s)	None
LRLA [m]	Rotate Data Memory left with result in ACC
Description	The contents of the specified Data Memory are rotated left by 1 bit with bit 7 rotated into bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.(i+1) \leftarrow [m].i; (i=0\sim6)$ $ACC.0 \leftarrow [m].7$
Affected flag(s)	None
LRLC [m]	Rotate Data Memory left through Carry
Description	The contents of the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into bit 0.
Operation	$[m].(i+1) \leftarrow [m].i; (i=0\sim6)$ $[m].0 \leftarrow C$ $C \leftarrow [m].7$
Affected flag(s)	C
LRLCA [m]	Rotate Data Memory left through Carry with result in ACC
Description	Data in the specified Data Memory and the carry flag are rotated left by 1 bit. Bit 7 replaces the Carry bit and the original carry flag is rotated into the bit 0. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.(i+1) \leftarrow [m].i; (i=0\sim6)$ $ACC.0 \leftarrow C$ $C \leftarrow [m].7$
Affected flag(s)	C

LRR [m]	Rotate Data Memory right
Description	The contents of the specified Data Memory are rotated right by 1 bit with bit 0 rotated into bit 7.
Operation	$[m].i \leftarrow [m].(i+1); (i=0\sim6)$ $[m].7 \leftarrow [m].0$
Affected flag(s)	None
LRRA [m]	Rotate Data Memory right with result in ACC
Description	Data in the specified Data Memory is rotated right by 1 bit with bit 0 rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.i \leftarrow [m].(i+1); (i=0\sim6)$ $ACC.7 \leftarrow [m].0$
Affected flag(s)	None
LRRC [m]	Rotate Data Memory right through Carry
Description	The contents of the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7.
Operation	$[m].i \leftarrow [m].(i+1); (i=0\sim6)$ $[m].7 \leftarrow C$ $C \leftarrow [m].0$
Affected flag(s)	C
LRRCA [m]	Rotate Data Memory right through Carry with result in ACC
Description	Data in the specified Data Memory and the carry flag are rotated right by 1 bit. Bit 0 replaces the Carry bit and the original carry flag is rotated into bit 7. The rotated result is stored in the Accumulator and the contents of the Data Memory remain unchanged.
Operation	$ACC.i \leftarrow [m].(i+1); (i=0\sim6)$ $ACC.7 \leftarrow C$ $C \leftarrow [m].0$
Affected flag(s)	C
LSBC A,[m]	Subtract Data Memory from ACC with Carry
Description	The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - [m] - \overline{C}$
Affected flag(s)	OV, Z, AC, C, SC, CZ
LSBCM A,[m]	Subtract Data Memory from ACC with Carry and result in Data Memory
Description	The contents of the specified Data Memory and the complement of the carry flag are subtracted from the Accumulator. The result is stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$[m] \leftarrow ACC - [m] - \overline{C}$
Affected flag(s)	OV, Z, AC, C, SC, CZ

LSDZ [m]	Skip if decrement Data Memory is 0
Description	The contents of the specified Data Memory are first decremented by 1. If the result is 0 the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$[m] \leftarrow [m] - 1$ Skip if $[m]=0$
Affected flag(s)	None
LSDZA [m]	Skip if decrement Data Memory is zero with result in ACC
Description	The contents of the specified Data Memory are first decremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0, the program proceeds with the following instruction.
Operation	$ACC \leftarrow [m] - 1$ Skip if $ACC=0$
Affected flag(s)	None
LSET [m]	Set Data Memory
Description	Each bit of the specified Data Memory is set to 1.
Operation	$[m] \leftarrow FFH$
Affected flag(s)	None
LSET [m].i	Set bit of Data Memory
Description	Bit i of the specified Data Memory is set to 1.
Operation	$[m].i \leftarrow 1$
Affected flag(s)	None
LSIZ [m]	Skip if increment Data Memory is 0
Description	The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$[m] \leftarrow [m] + 1$ Skip if $[m]=0$
Affected flag(s)	None
LSIZA [m]	Skip if increment Data Memory is zero with result in ACC
Description	The contents of the specified Data Memory are first incremented by 1. If the result is 0, the following instruction is skipped. The result is stored in the Accumulator but the specified Data Memory contents remain unchanged. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$ACC \leftarrow [m] + 1$ Skip if $ACC=0$
Affected flag(s)	None
LSNZ [m].i	Skip if Data Memory is not 0
Description	If the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is 0 the program proceeds with the following instruction.
Operation	Skip if $[m].i \neq 0$
Affected flag(s)	None

LSNZ [m]	Skip if Data Memory is not 0
Description	If the content of the specified Data Memory is not 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is 0 the program proceeds with the following instruction.
Operation	Skip if [m] $\neq$ 0
Affected flag(s)	None
LSUB A,[m]	Subtract Data Memory from ACC
Description	The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Accumulator. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$ACC \leftarrow ACC - [m]$
Affected flag(s)	OV, Z, AC, C, SC, CZ
LSUBM A,[m]	Subtract Data Memory from ACC with result in Data Memory
Description	The specified Data Memory is subtracted from the contents of the Accumulator. The result is stored in the Data Memory. Note that if the result of subtraction is negative, the C flag will be cleared to 0, otherwise if the result is positive or zero, the C flag will be set to 1.
Operation	$[m] \leftarrow ACC - [m]$
Affected flag(s)	OV, Z, AC, C, SC, CZ
LSWAP [m]	Swap nibbles of Data Memory
Description	The low-order and high-order nibbles of the specified Data Memory are interchanged.
Operation	$[m].3 \sim [m].0 \leftrightarrow [m].7 \sim [m].4$
Affected flag(s)	None
LSWAPA [m]	Swap nibbles of Data Memory with result in ACC
Description	The low-order and high-order nibbles of the specified Data Memory are interchanged. The result is stored in the Accumulator. The contents of the Data Memory remain unchanged.
Operation	$ACC.3 \sim ACC.0 \leftarrow [m].7 \sim [m].4$ $ACC.7 \sim ACC.4 \leftarrow [m].3 \sim [m].0$
Affected flag(s)	None
LSZ [m]	Skip if Data Memory is 0
Description	If the contents of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	Skip if [m]=0
Affected flag(s)	None
LSZA [m]	Skip if Data Memory is 0 with data movement to ACC
Description	The contents of the specified Data Memory are copied to the Accumulator. If the value is zero, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0 the program proceeds with the following instruction.
Operation	$ACC \leftarrow [m]$ Skip if [m]=0
Affected flag(s)	None

LSZ [m].i	Skip if bit i of Data Memory is 0
Description	If bit i of the specified Data Memory is 0, the following instruction is skipped. As this requires the insertion of a dummy instruction while the next instruction is fetched, it is a two cycle instruction. If the result is not 0, the program proceeds with the following instruction.
Operation	Skip if [m].i=0
Affected flag(s)	None
LTABRD [m]	Read table (current page) to TBLH and Data Memory
Description	The low byte of the program code (current page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
LTABRDL [m]	Read table (last page) to TBLH and Data Memory
Description	The low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
LITABRD [m]	Increment table pointer low byte first and read table to TBLH and Data Memory
Description	Increment table pointer low byte, TBLP, first and then the program code addressed by the table pointer (TBHP and TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
LITABRDL [m]	Increment table pointer low byte first and read table (last page) to TBLH and Data Memory
Description	Increment table pointer low byte, TBLP, first and then the low byte of the program code (last page) addressed by the table pointer (TBLP) is moved to the specified Data Memory and the high byte moved to TBLH.
Operation	[m] ← program code (low byte) TBLH ← program code (high byte)
Affected flag(s)	None
LXOR A,[m]	Logical XOR Data Memory to ACC
Description	Data in the Accumulator and the specified Data Memory perform a bitwise logical XOR operation. The result is stored in the Accumulator.
Operation	ACC ← ACC "XOR" [m]
Affected flag(s)	Z
LXORM A,[m]	Logical XOR ACC to Data Memory
Description	Data in the specified Data Memory and the Accumulator perform a bitwise logical XOR operation. The result is stored in the Data Memory.
Operation	[m] ← ACC "XOR" [m]
Affected flag(s)	Z

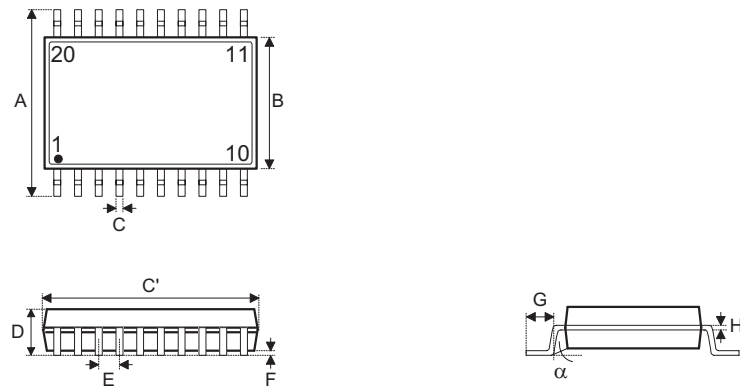
Package Information

Note that the package information provided here is for consultation purposes only. As this information may be updated at regular intervals users are reminded to consult the [Holtek website](#) for the latest version of the [Package/Carton Information](#).

Additional supplementary information with regard to packaging is listed below. Click on the relevant section to be transferred to the relevant website page.

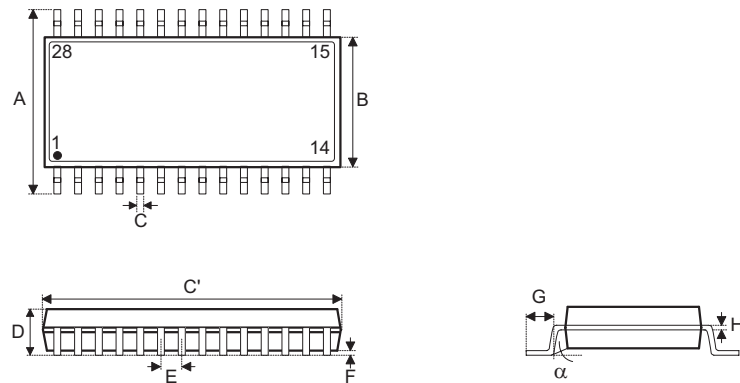
- [Package Information \(include Outline Dimensions, Product Tape and Reel Specifications\)](#)
- [The Operation Instruction of Packing Materials](#)
- [Carton information](#)

20-pin SOP (300mil) Outline Dimensions



Symbol	Dimensions in inch		
	Min.	Nom.	Max.
A	—	0.406 BSC	—
B	—	0.295 BSC	—
C	0.012	—	0.020
C'	—	0.504 BSC	—
D	—	—	0.104
E	—	0.050 BSC	—
F	0.004	—	0.012
G	0.016	—	0.050
H	0.008	—	0.013
α	0°	—	8°

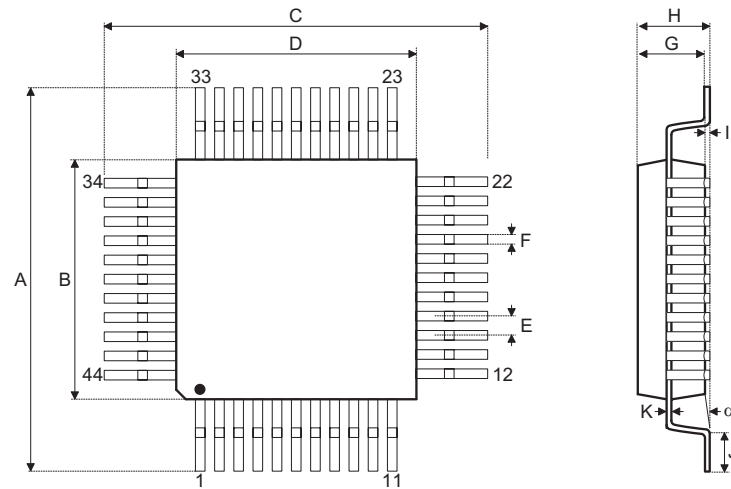
Symbol	Dimensions in mm		
	Min.	Nom.	Max.
A	—	10.30 BSC	—
B	—	7.50 BSC	—
C	0.31	—	0.51
C'	—	12.80 BSC	—
D	—	—	2.65
E	—	1.27 BSC	—
F	0.10	—	0.30
G	0.40	—	1.27
H	0.20	—	0.33
α	0°	—	8°

28-pin SOP (300mil) Outline Dimensions


Symbol	Dimensions in inch		
	Min.	Nom.	Max.
A	—	0.406 BSC	—
B	—	0.295 BSC	—
C	0.012	—	0.020
C'	—	0.705 BSC	—
D	—	—	0.104
E	—	0.050 BSC	—
F	0.004	—	0.012
G	0.016	—	0.050
H	0.008	—	0.013
α	0°	—	8°

Symbol	Dimensions in mm		
	Min.	Nom.	Max.
A	—	10.30 BSC	—
B	—	7.50 BSC	—
C	0.31	—	0.51
C'	—	17.90 BSC	—
D	—	—	2.65
E	—	1.27 BSC	—
F	0.10	—	0.30
G	0.40	—	1.27
H	0.20	—	0.33
α	0°	—	8°

44-pin LQFP (10mm×10mm) (FP2.0mm) Outline Dimensions



Symbol	Dimensions in inch		
	Min.	Nom.	Max.
A	—	0.472 BSC	—
B	—	0.394 BSC	—
C	—	0.472 BSC	—
D	—	0.394 BSC	—
E	—	0.032 BSC	—
F	0.012	0.015	0.018
G	0.053	0.055	0.057
H	—	—	0.063
I	0.002	—	0.006
J	0.018	0.024	0.030
K	0.004	—	0.008
α	0°	—	7°

Symbol	Dimensions in mm		
	Min.	Nom.	Max.
A	—	12.00 BSC	—
B	—	10.00 BSC	—
C	—	12.00 BSC	—
D	—	10.00 BSC	—
E	—	0.80 BSC	—
F	0.30	0.37	0.45
G	1.35	1.40	1.45
H	—	—	1.60
I	0.05	—	0.15
J	0.45	0.60	0.75
K	0.09	—	0.20
α	0°	—	7°

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